

(19) World Intellectual Property Organization
International Bureau



(43) International Publication Date
11 May 2006 (11.05.2006)

PCT

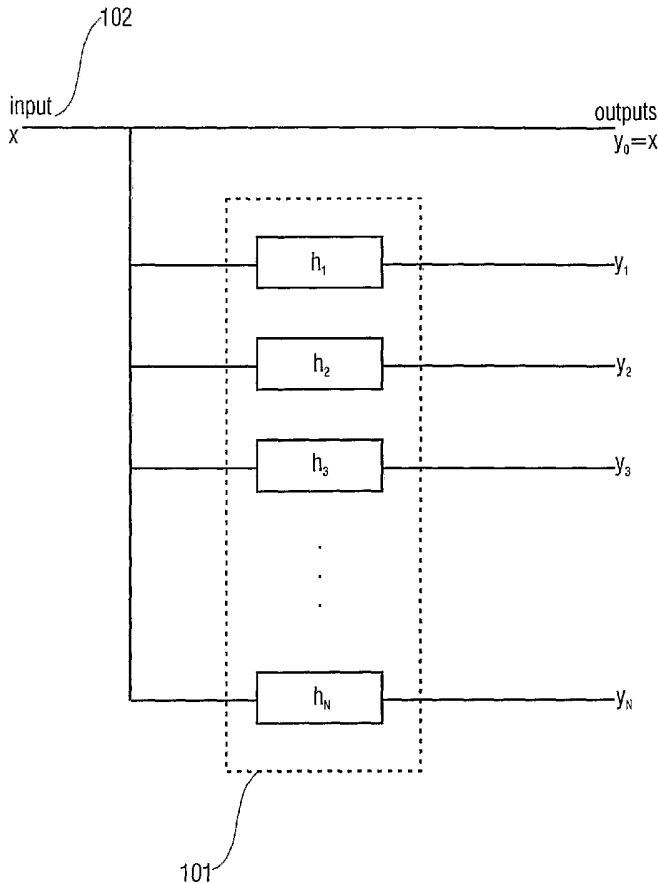
(10) International Publication Number
WO 2006/048227 A1

- (51) International Patent Classification:
H04S 5/02 (2006.01) *G10L 19/00* (2006.01)
- (21) International Application Number:
PCT/EP2005/011664
- (22) International Filing Date: 31 October 2005 (31.10.2005)
- (25) Filing Language: English
- (26) Publication Language: English
- (30) Priority Data:
0402649-8 2 November 2004 (02.11.2004) SE
- (71) Applicants (for all designated States except US): **COD-ING TECHNOLOGIES AB** [SE/SE]; Döbelnsgatan 64, S-113 52 Stockholm (SE). **KONINKLIJKE PHILIPS ELECTRONICS N.V.** [NL/NL]; Groenewoudseweg 1, NL-5621 BA Eindhoven (NL).
- (72) Inventors; and
- (75) Inventors/Applicants (for US only): **PUERNHAGEN, Heiko** [DE/SE]; Döbelnsgatan 64, S-113 52 Stockholm (SE). **ENGDEGARD, Jonas** [SE/SE]; Wenströmsvägen

- 6, S-113 52 Stockholm (SE). **BREEBAART, Jeroen** [NL/NL]; Groenewoudseweg 1, NL-5621 BA Eindhoven (NL). **SCHUIJERS, Erik** [NL/NL]; Groenewoudseweg 1, NL-5621 BA Eindhoven (NL).
- (74) Agents: **ZINKLER, Franz** et al.; Patentanwälte Schoppe, Zimmermann, Stöckeler & Zinkler, P.O. Box 246, 82043 Pullach (DE).
- (81) Designated States (unless otherwise indicated, for every kind of national protection available): AE, AG, AL, AM, AT, AU, AZ, BA, BB, BG, BR, BW, BY, BZ, CA, CH, CN, CO, CR, CU, CZ, DE, DK, DM, DZ, EC, EE, EG, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KM, KN, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, LY, MA, MD, MG, MK, MN, MW, MX, MZ, NA, NG, NI, NO, NZ, OM, PG, PH, PL, PT, RO, RU, SC, SD, SE, SG, SK, SL, SM, SY, TJ, TM, TN, TR, TT, TZ, UA, UG, US, UZ, VC, VN, YU, ZA, ZM, ZW.
- (84) Designated States (unless otherwise indicated, for every kind of regional protection available): ARIPO (BW, GH, GM, KE, LS, MW, MZ, NA, SD, SL, SZ, TZ, UG, ZM,

[Continued on next page]

(54) Title: MULTICHANNEL AUDIO SIGNAL DECODING USING DE-CORRELATED SIGNALS



(57) Abstract: A multi-channel audio signal having at least three channels can be reconstructed such, that the reconstructed channels are at least partly de-correlated from each other using a downmixed signal derived from an original multi-channel signal and a set of decorrelated signals provided by a de-correlator (101) that derives the set of de-correlated signals from the downmix signal, wherein the de-correlated signals within the set of de-correlated signals are mutually mostly orthogonal to each other, i.e. an orthogonality relation between channel pairs is satisfied within an orthogonality tolerance range.

WO 2006/048227 A1



ZW), Eurasian (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM),
European (AT, BE, BG, CH, CY, CZ, DE, DK, EE, ES, FI,
FR, GB, GR, HU, IE, IS, IT, LT, LU, LV, MC, NL, PL, PT,
RO, SE, SI, SK, TR), OAPI (BF, BJ, CF, CG, CI, CM, GA,
GN, GQ, GW, ML, MR, NE, SN, TD, TG).

— *before the expiration of the time limit for amending the
claims and to be republished in the event of receipt of
amendments*

Published:

— *with international search report*

For two-letter codes and other abbreviations, refer to the "Guidance Notes on Codes and Abbreviations" appearing at the beginning of each regular issue of the PCT Gazette.

MULTICHANNEL AUDIO SIGNAL DECODING USING DE-CORRELATED SIGNALS

Description

5

The present invention relates to coding of multi-channel audio signals using spatial parameters and in particular to new improved concepts for generating and using de-correlated signals.

10

Recently, multi-channel audio reproduction techniques are becoming more and more important. In the view of an efficient transmission of multi-channel audio signals having 5 or more separate audio channels, several ways of compressing a stereo or multi-channel signal have been developed. Recent approaches for the parametric coding of multi-channel audio signals (parametric stereo (PS), "Binaural Cue Coding" (BCC) etc.) represent a multi-channel audio signal by means of a down-mix signal (could be monophonic or comprise several channels) and parametric side information, also referred to as "spatial cues", characterizing its perceived spatial sound stage.

15

20

25

A multi-channel encoding device generally receives - as input - at least two channels, and outputs one or more carrier channels and parametric data. The parametric data is derived such that, in a decoder, an approximation of the original multi-channel signal can be calculated. Normally, the carrier channel (channels) will include sub-band samples, spectral coefficients, time domain samples, etc., which provide a comparatively fine representation of the underlying signal, while the parametric data do not include such samples of spectral coefficients but include control parameters for controlling a certain reconstruction algorithm instead. Such a reconstruction could comprise weighting by multiplication, time shifting, frequency shifting, phase shifting, etc. Thus,

30

35

the parametric data includes only a comparatively coarse representation of the signal or the associated channel.

The binaural cue coding (BCC) technique is described in a number of publications, as in "Binaural Cue Coding applied to Stereo and Multi-Channel Audio Compression", C. Faller, F. Baumgarte, AES convention paper 5574, May 2002, Munich, in the 2 ICASSP publications "Estimation of auditory spatial cues for binaural cue coding", and "Binaural cue coding: a normal and efficient representation of spatial audio", both authored by C. Faller, and F. Baumgarte, Orlando, FL, May 2002.

In BCC encoding, a number of audio input channels are converted to a spectral representation using a DFT (Discrete Fourier Transform) based transform with overlapping windows. The resulting uniform spectrum is then divided into non-overlapping partitions. Each partition has a bandwidth proportional to the equivalent rectangular bandwidth (ERB). Then, spatial parameters called ICLD (Inter-Channel Level Difference) and ICTD (Inter-Channel Time Difference) are estimated for each partition. The ICLD parameter describes a level difference between two channels and the ICTD parameter describes the time difference (phase shift) between two signals of different channels. The level differences and the time differences are normally given for each channel with respect to a reference channel. After the derivation of these parameters, the parameters are quantized and finally encoded for transmission.

30

Although ICLD and ICTD parameters represent the most important sound source localization parameters, a spatial representation using these parameters can be enhanced by introducing additional parameters.

35

A related technique, called "parametric stereo" describes the parametric coding of a two-channel stereo signal based on a

transmitted mono signal plus parameter side information. In this context, 3 types of spatial parameters, referred to as inter-channel intensity difference (IIDs), inter-channel phase differences (IPDs), and inter-channel coherence (ICC) are introduced. The extension of the spatial parameter set with a coherence parameter (correlation parameter) enables a parametrization of the perceived spatial "diffuseness" or spatial "compactness" of the sound stage. Parametric stereo is described in more detail in: "Parametric Coding of stereo audio", J. Breebaart, S. van de Par, A. Kohlrausch, E. Schuijers (2005) *Eurasip, J. Applied Signal Proc.* 9, pages 1305-1322", in "High-Quality Parametric Spatial Audio Coding at Low Bitrates", J. Breebaart, S. van de Par, A. Kohlrausch, E. Schuijers, AES 116th Convention, Preprint 6072, Berlin, May 2004, and in "Low Complexity Parametric Stereo Coding", E. Schuijers, J. Breebaart, H. Purnhagen, J. Engdegard, AES 116th Convention, Preprint 6073, Berlin, May 2004.

The present invention relates to parametric coding of the spatial properties of an audio signal. Parametric multi-channel audio decoders reconstruct N channels based on M transmitted channels, where $N > M$, and additional control data. The additional control data represents a significant lower data rate than transmitting all N channels, making the coding very efficient while at the same time ensuring compatibility with at least both M channel devices and N channel devices. Typical parameters used for describing spatial properties are inter-channel intensity differences (IID), inter-channel time differences (ITD), and inter-channel coherences (ICC). In order to reconstruct the spatial properties based on these parameters, a method is required that can reconstruct the correct level of correlation between two or more channels, according to the IC parameters. This is accomplished by means of a de-correlation method, i.e. a method to derive decorrelated signals from transmitted signals to combine decorrelated signals with transmitted signals within some upmixing process. Methods for upmixing

based on a transmitted signal, a decorrelated signal, and IID/ICC parameters is described in the references given above.

5 There are a couple of methods available for creation of de-correlated signals. Preferably, the decorrelated signals have similar or equal temporal and spectral envelopes as the original input signals. Ideally, a linear time invariant (LTI) function with all-pass frequency response is desired.
10 One obvious method for achieving this is by using a constant delay. However, using a delay, or any other LTI all-pass function, will result in non-all-pass response after addition of the non-processed signal. In the case of a delay, the result will be a typical comb-filter. The comb-filter
15 often gives an undesirable "metallic" sound that, even if the stereo widening effect can be efficient, reduces much naturalness of the original. The constant delay method and other prior art methods suffer from the inability to create more than one de-correlated signal while preserving quality
20 and mutual de-correlation.

The perceptual quality of a reconstructed multi-channel audio signal therefore depends strongly on an efficient concept that allows for the generation of a de-correlated signal from
25 a transmitted signal, wherein ideally the de-correlated signal is orthogonal to the signal from which it is derived, i.e. perfectly de-correlated. Even if a perfectly de-correlated signal is available, a multi-channel upmix in which the individual channels are mutually de-correlated cannot be derived using a single de-correlated signal. During
30 the upmixing a reconstructed audio channel is generated by combining a transmitted signal with the generated de-correlated signal, whereas the extent to which the de-correlated signal is mixed to the transmitted signal is typically controlled by a transmitted spatial audio parameter
35 (ICC). Mutually perfectly de-correlated signals can therefore

not be achieved, since every reconstructed audio channel has a fraction of the same de-correlated signal.

5 It is the object of the present invention to provide a more efficient concept for creation of highly de-correlated signals.

This object is achieved by an apparatus according to claim 1 or a method according to claim 15.

10

The present invention is based on the finding that a multi-channel signal having at least three channels can be reconstructed such that the reconstructed channels are at least partly de-correlated from each other using a downmixed signal derived from an original multi-channel signal and a set of decorrelated signals provided by a de-correlator that derives the set of de-correlated signals from the downmix signal, wherein the de-correlated signals within the set of de-correlated signals are mutually approximately orthogonal to each other, i.e. an orthogonality relation between channel pairs is satisfied within an orthogonality tolerance range.

15
20

An orthogonality tolerance range can for example be derived from the cross correlation coefficient that quantifies the degree of correlation between two signals. A cross correlation coefficient of 1 means perfect correlation, i.e. two identical signals. On the other and, a cross correlation coefficient of 0 means perfect anticorrelation or orthogonality of the signals. The orthogonality tolerance range, therefore, may be defined as interval of correlation coefficient values ranging from 0 to a specific upper limit.

25
30

Hence, the present invention relates to, and provides a solution to, the problem of efficiently generating one or more orthogonal signals while preserving impulse properties and perceived audio quality.

35

In one embodiment of the present invention an IIR lattice filter is implemented as a de-correlator having filter-coefficients derived from noise sequences, and the filtering is performed within a complex valued or real valued filter bank.
5

In one embodiment of the present invention, a method for reconstructing a multi-channel signal includes a method for creating several orthogonal or close to orthogonal signals by using a group of lattice IIR filters.
10

In a further embodiment of the present invention, the method for creating several orthogonal signals is having a method for choosing filter coefficients for achieving orthogonality or an approximation of orthogonality in a perceptually motivated way.
15

In a further embodiment of the present invention, a group of lattice IIR filters is used within a complex valued filterbank during the reconstruction of the multi-channel signal.
20

In a further embodiment of the present invention a method for creating one or more orthogonal or close to orthogonal signals is implemented, using one or more all-pass IIR filters based on lattice structure within in a spatial decoder.
25

In a further embodiment of the present invention, the embodiment described above is implemented such that the filter coefficients used for the IIR filtering are based on random noise sequences.
30

In a further embodiment of the present invention, additional time delays are added to the filters used.

35 In a further embodiment of the present invention, the filtering is processed in a filterbank domain.

In a further embodiment of the present invention, the filtering is processed in a complex valued filterbank.

5 In a further embodiment of the present invention, the orthogonal signals created by the filtering are mixed to form a set of output signals.

10 In a further embodiment of the present invention, the mixing of the orthogonal signals is depending on transmitted control data, additionally supplied to an inventive decoder.

15 In a further embodiment of the present invention, an inventive decoder or an inventive decoding method uses control data that contains at least one parameter indicating a desired cross-correlation of at least two of the output signals generated.

20 In a further embodiment of the present invention, a 5.1 channel surround signal is upmixed from a transmitted monophonic signal by deriving four de-correlated signals using the inventive concept. The monophonic downmixed signal and the four de-correlated signals are then mixed together according to some mixing rules to form the output 5.1 channel signal. Therefore the possibility is provided to generate output signals that are mutually de-correlated, since the signals used
25 for the upmix, i.e. the transmitted monophonic signal and the four generated de-correlated signals are mainly de-correlated due to their inventive generation.

30 In a further embodiment of the present invention, two individual channels are transmitted as a downmix of a 5.1 channel signal. In one implementation, two additional mutually de-correlated signals are derived using the inventive concept to provide four channels as basis for an upmix which are almost
35 perfectly de-correlated. In a modification of the embodiment described above a third de-correlated signal is derived and mixed with the other two de-correlated signals to provide a

further de-correlated signal available for the subsequent up-mixing. Using this feature, the perceptual quality can be further enhanced for individual channels, e.g. the center-channel of a 5.1 surround signal.

5

In a further embodiment of the present invention, five audio channels are upmixed from a monophonic transmitted channel prior to deriving, using the inventive concept, four de-correlated signals that are subsequently combined with four
10 of the five aforementioned upmixed channels, allowing for a creation of five output audio channels that are mutually mainly de-correlated.

15

In a further embodiment of the present invention, the audio signals are delayed prior to or after the application of the inventive IIR filter based filtering. The delay further enhances the de-correlation of the generated signals, and reduces colorization when mixing the generated de-correlated signals with the original downmixed signal.

20

In a further embodiment of the present invention, the generation of the de-correlated signals is performed in the subband domain of a (complex modulated) filterbank, wherein the filter coefficients used by the de-correlator are derived using
25 the specific filterbank index of the filterbank for which the de-correlated signals are derived.

30

In a further embodiment of the present invention, the de-correlated signals are derived using lattice IIR filters that perform a lattice IIR all-pass filtering of an audio signal. Using a lattice IIR filter has major advantages. An exponential decay of the response of such a filter, which is preferable for creating appropriate decorrelated signals, is an inherent property of such a filter. Furthermore, a desired long
35 decaying pulse response of a filter used to generate decorrelated signals can be achieved in an extremely memory and com-

putationally efficient (low complexity) manner by using a lattice filter structure.

In a modification of the previously described embodiment the filter coefficients (reflection coefficients) used are given by means of providing filter coefficients derived from noise sequences. In a modification, the reflection coefficients are individually calculated based on the sub-band index of a sub-band, in which the lattice filter is used to derive de-correlated signals.

In one embodiment of the present invention, the filtered signals and the unmodified input signal are combined by a mixing matrix D to form a set of output signals. The mixing matrix D defines the mutual correlations of the output signals, as well as the energy of each output signal. The entries (weights) of the mixing matrix D are preferably time-variable and dependent on transmitted control data. The control parameters preferably contain (desired) level differences between certain output signals and/or specific mutual correlation parameters.

In a further embodiment of the present invention, an inventive audio decoder is comprised within an audio receiver or playback device to enhance the perceptual quality of a reconstructed signal.

Preferred embodiments of the present invention are subsequently described by the following drawings, wherein:

30

Fig. 1 shows a block diagram of the inventive audio decoding concepts;

35

Fig. 2 shows a prior art decoder not implementing the inventive concepts;

Fig. 3 shows a 5.1 multi-channel audio decoder according to the present invention;

5 Fig. 4 shows a further 5.1 channel audio decoder according to the present invention;

Fig. 5 shows a further inventive audio decoder;

10 Fig. 6 shows a further embodiment of an inventive multi-channel audio decoder;

Fig. 7 shows schematically the generation of a de-correlated signal;

15 Fig. 8 shows a lattice IIR filter used for generating a de-correlated signal;

Fig. 9 shows a receiver or audio player having an inventive audio decoder; and

20

Fig. 10 shows a transmission having a receiver or playback device having an inventive audio decoder.

25 The embodiments described below are merely illustrative for the principles of the present invention for advanced methods for creating orthogonal signals. It is understood that modifications and variations of the arrangements and the details described herein will be apparent to those skilled in the art. It is the intent, therefore, to be limited only by the
30 scope of the impending patent claims and not by the specific details presented by way of description and explanation of the embodiments herein.

35 Fig. 1 illustrates an inventive apparatus for the de-correlation of signals as used in a parametric stereo or multi-channel system. The inventive apparatus includes means
101 for providing a plurality of orthogonal de-correlated

signals derived from an input signal 102. The providing means can be an array of de-correlation filters based on lattice IIR structures. The input signal 102 (x) can be a time-domain signal or a single sub-band domain signal as e.g. obtained from a complex QMF bank. The signals output by the means 101, y_1 - y_N are the resulting de-correlated signals that are all mutually orthogonal or close to orthogonal.

As it is vital for reconstructing the spatial properties of a parametric stereo or parametric multi-channel system to decrease the coherence between two or more channels in order to reconstruct the perceived wideness of the spatial image, the resulting de-correlated signal can be used to create a final upmix of a multi-channel signal. This can be done by adding filtered versions ($h_1(x)$) of the original signal (x) to the output channels. Hence, lowering the coherence between N signals using N different filters can be done according to:

$$\begin{aligned} y_1 &= a*x + b*h_1(x) \\ y_2 &= a*x + b*h_2(x) \\ &\dots \\ y_n &= a*x + b*h_n(x) \end{aligned}$$

where x is the original signal, y_1 to y_n are the resulting output signals, a and b are the gain factors controlling the amount of coherence and h_1 to h_n are the different decorrelation filters. In a more general sense, one can write the output signals y_i ($i=1\dots I$) as a linear combination of the input signal x and the input signal x filtered by filters h_n ($j=1\dots N$):

$$Y = \begin{pmatrix} y_1 \\ \vdots \\ y_I \end{pmatrix} = D \begin{pmatrix} x \\ h_1(x) \\ \vdots \\ h_N(x) \end{pmatrix}$$

Here, the mixing matrix D determines the mutual correlations and output levels of the output signals y_i .

In order to prevent changes in the timbre, the filter in question should preferably be of all-pass character. One successful approach is to use all-pass filters similar to those used for artificial reverberation processes. Artificial reverberation algorithms usually require a high time resolution to provide an impulse response that is satisfactory diffuse in time. One way of designing such all-pass filters is to use a random noise sequence as impulse response. The filter can then easily be implemented as an FIR filter. In order to achieve a sufficient degree of independence between the filtered outputs, the impulse response of the FIR filter should be relatively long, hence requiring a significant amount of computational effort to perform the convolution. An all-pass IIR filter is preferred for that purpose. The IIR structure has several advantages when it comes to designing de-correlation filters:

20

a) The natural exponential decay that is common for all natural reverberation is desired for a de-correlation filter. This is an inherent property of IIR filters.

25

b) For long decaying impulse responses of an IIR filter, the corresponding FIR filter is generally more expensive in terms of complexity and requires more memory.

However, designing IIR all-pass filters is less trivial than the FIR case where any random noise sequence qualifies as a coefficient vector. A design constraint when targeting multiple de-correlation filters is also the required ability to preserve the same decaying properties for all the filters while providing orthogonal outputs (i.e., a filter impulse responses that obey mutually substantially low correlation) of each filter output. Also as a basic requirement - stability has to be achieved.

The present invention shows a novel method to create multiple orthogonal all-pass filters by means of a lattice IIR filter structure. This approach has several advantages:

5

a) Lower complexity than FIR filters (given the required length of the impulse responses).

10

b) Stability constraints can be satisfied easily, as this is automatically achieved when absolute values of the magnitudes of all reflection coefficients are less than one.

15

c) Multiple orthogonal all-pass filters can be designed more easily with the same decaying properties based on random noise sequences.

20

d) High robustness against quantization errors due to finite word-length effects.

25

Although the reflection coefficients of the lattice IIR filter can be based on random noise sequences, for better performance those coefficients should also be sorted in more sophisticated ways or processed by non-random methods in order to achieve sufficient orthogonality and other important properties. A straightforward method is to generate a multitude of random reflection coefficient vectors, followed by a selection of a specific set based on certain criteria, such as a common decaying envelope, minimization of all mutual impulse response correlations of the selected set, and alike.

30

35

More specifically, one could start with a large set of random noise sequences. Each of these sequences is used as reflection coefficients in the allpass section. Subsequently, the impulse response of the resulting allpass section is computed for each random noise sequence. Finally, one selects those

noise sequences that give mutually decorrelated impulse responses.

5 There are great advantages in basing the de-correlation algorithm on a (complex) filter bank such as the complex valued QMF bank. This filter bank provides the flexibility to allow the properties of the de-correlator to be frequency selective in terms of for example equalization, decay time, impulse density and timbre. Note that many of these properties can be altered while preserving the all-pass characteristic. There is much knowledge related to auditory perception that guides the design of such lattice IIR filter. An important aspect is the length and shape of the decaying envelop of the impulse response. Also the need for an additional pre-delay, optionally frequency dependent, is important as this largely influences what kind of comb-filter characteristic will be obtained when mixing the de-correlated signal with the original one. For sufficient impulse density the noise based reflection coefficients in the lattice filter should preferably be different for the different filter bank channels. For even better impulse density fractional delay approximations can be used within the filter bank.

25 Fig. 2 shows a hierarchical decoding structure to derive a multi-channel signal for a transmitted monophonic downmix signal by subsequent parametric stereo boxes, using a single decorrelated signal. By shortly reviewing the prior art approach, the problem solved by the present invention shall again be motivated. The 1-to-3 channel decoder 110 shown in Fig. 2 comprises a de-correlator 112, a first parametric stereo upmixer 114 and a second parametric stereo upmixer 116.

35 A monophonic input signal 118 is input into the de-correlator 112 to derive a de-correlated signal 120. Only a single de-correlated signal is derived. The first parametric

stereo upmixer receives as an input the monophonic downmix signal 118 and the de-correlated signal 120. The first up-mixer 114 derives a center channel 122 and a combined channel 124 by mixing the monophonic downmix signal 118 and the
5 de-correlated signal 120 using a correlation parameter 126, that steers the mixing of the channels.

The combined channel 124 is then input into the second parametric stereo upmixer 116, building the second hierarchical
10 level of the audio decoder. The second parametric stereo up-mixer 116 is further receiving the de-correlated signal 120 as an input and derives a left channel 128 and a right channel 130 by mixing the combined channel 124 and the de-correlated signal 120.

15

It is principally feasible to generate a center channel 122 that is perfectly de-correlated from the combined channel 124, when the de-correlator 112 is able to derive a de-correlated signal which is fully orthogonal to the monophonic downmix signal 118. Almost perfect de-correlation
20 would be achieved when the steering information 126 indicates an upmix, in which each upmixed channel is mainly having a signal component coming from either the de-correlated signal 120 or from the monophonic downmix signal 118. Since,
25 however, the same de-correlated signal 120 is then used to derive the left channel 128 and the right channel 130, it is obvious, that this will result in a remaining correlation between the center channel 122 and one of the channels 128 or 130.

30

This becomes even more evident when examining the extreme case in which a completely de-correlated left channel 128 and right channel 130 shall be derived from a de-correlated signal 120 that is assumed to be perfectly orthogonal to the
35 monophonic downmix signal. Perfect decorrelation between the left channel 128 and the right channel 130 can be achieved, when the combined channel 124 holds information on the mono-

phonic downmix channel 118 only, which simultaneously means that the center channel 122 is mainly comprising the de-correlated signal 112. Therefore, a de-correlated left channel 128 and right channel 130 would mean that one of the channels does mainly comprise the information on the de-correlated signal 120 and the other channel would mainly comprise the combined signal 124, which then is identical to the monophonic downmix signal 118. Therefore the only way the left or the right channels are completely de-correlated forces an almost perfect correlation between the center channel 122 and one of the channels 128 or 130.

This most unwanted property can be successfully avoided by applying the inventive concept of generating different and mutually orthogonal de-correlated signals.

Fig. 3 shows an embodiment of an inventive multi-channel audio decoder 400 comprising a pre-de-correlator matrix 401, a de-correlator 402 and a mix-matrix 403. The inventive decoder 400 shows a 1-to-5 configuration, where five audio channels and a low-frequency enhancement channel are derived from a monophonic downmix signal 405 and additional spatial control data, such as ICC or ICLD parameters. These are not shown in the principle sketch in Fig. 3. The monophonic downmix signal 405 is input into the pre-de-correlator matrix 401 that derives four intermediate signals 406 which serve as an input for the de-correlator 402, that is comprising four inventive de-correlators h_1 - h_4 . These are supplying four mutually orthogonal de-correlated signals 408 at the output of the de-correlator 402.

The mix-matrix 403 receives as an input the four mutually orthogonal de-correlated signals 408 and in addition a downmix signal 410 derived from the monophonic downmix signal 405 by the pre-de-correlator matrix 401.

The mix-matrix 403 combines the monophonic signal 410 and the four de-correlated signals 408 to yield a 5.1 output signal 412 comprising a left-front channel 414a, a left-surround channel 414b, a right-front channel 414c, a right-surround channel 414d, a center channel 414e and a low-frequency enhancement channel 414f.

It is important to note that the generation of four mutually orthogonal de-correlated signals 408 enables the ability to derive five channels of the 5.1 channel signal that are at least partly de-correlated. In a preferred embodiment of the present invention, these are the channels 414a to 414e. The low-frequency enhancement channel 414f comprises low-frequency parts of the multi-channel signal, that are combined in one single low-frequency channel for all the surround channels 414a to 414e.

Fig. 4 shows an inventive 2-to-5 decoder to derive a 5.1 channel surround signal from two transmitted signals.

The multi-channel audio decoder 500 comprises a pre-decorrelator matrix 501, a de-correlator 502 and a mix-matrix 503. In the 2-to-5 setup, two transmitted channels, 505a and 505b are input into the pre-decorrelator matrix that derives an intermediate left channel 506a, an intermediate right channel 506b and an intermediate center channel 506c and two intermediate channels 506d from the submitted channels 505a and 505b, optionally also using additional control data such as ICC and ICLD parameters.

The intermediate channels 506d are used as input for the de-correlator 502 that derives two mutually orthogonal or nearly orthogonal de-correlated signals which are input into the mix-matrix 503 together with the intermediate left channel 506a, the intermediate right channel 506b and the intermediate center channel 506c.

The mix-matrix 503 derives the final 5.1 channel audio signal 508 from the previously mentioned signals, wherein the finally derived audio channels have the same advantageous properties as already described for the channels derived by
5 the 1-to-5 multi-channel audio decoder 400.

Fig. 5 shows a further embodiment of the present invention, that combines the features of multi-channel audio decoders 400 and 500. The multi-channel audio decoder 600 comprises a
10 pre-de-correlation matrix 601, a de-correlator 602 and a mix-matrix 603. The multi-channel audio decoder 600 is a flexible device allowing to operate in different modes depending on the configuration of input signals 605 input into the pre-de-correlator 601. Generally, the pre-de-correlator
15 derives intermediate signals 607 that serve as input for the de-correlator 602 and that are partially transmitted and altered to build input parameters 608. The input parameters 608 are the parameters input into the mix-matrix 603 that derives output channel configurations 610a or 610b depending
20 on the input channel configuration.

In a 1-to-5 configuration, a downmix signal and an optional residual signal is supplied to the pre-de-correlator matrix, that derives four intermediate signals (e_1 to e_4) that are
25 used as an input of the de-correlator, which derives four de-correlated signals (d_1 to d_4) that form the input parameters 608 together with a directly transmitted signal m derived from the input signal.

30 It may be noted, that in the case where an additional residual signal is supplied as input, the de-correlator 602 that is generally operative in a sub-band domain, may be operative to forward the residual signal instead of deriving a de-correlated signal. This may also be done in a selective
35 manner for certain frequency bands only.

In the 2-to-5 configuration the input signals 605 comprise a left channel, a right channel and optionally a residual signal. In that configuration, the pre-de-correlator matrix derives a left, a right and a center channel and in addition
5 two intermediate channels (e_1 , e_2). Hence, the input parameters to the mix-matrix 603 are formed by the left channel, the right channel, the center channel, and two de-correlated signals (d_1 and d_2). In a further modification, the pre-de-correlator matrix may derive an additional intermediate signal
10 (e_5) that is used as an input for a de-correlator (D_5) whose output is a combination of the de-correlated signal (d_5) derived from the signal (e_5) and the de-correlated signals (d_1 and d_2). In this case, an additional de-correlation can be guaranteed between the center channel and the left
15 and the right channel.

Fig. 6 shows a further embodiment of the present invention, in which de-correlated signals are combined with individual audio channels after the upmixing process. In this alternative embodiment, a monophonic audio channel 620 is upmixed
20 by an upmixer 624, wherein the upmixing may be controlled by additional control data 622. The upmix channels 630 comprise five audio channels that are correlated with each other, and commonly referred to as dry channels. Final channels 632 can
25 be derived by combining four of the dry channels 630 with de-correlated, mutually orthogonal signals. As a result, it is possible to provide five channels that are at least partly de-correlated from each other. With respect to Figure 3, this can be seen as a special case of a mix-matrix.

30

Fig. 7 shows a block diagram of an inventive de-correlator 700 for providing a de-correlated signal. The de-correlator 700 comprises a predelay unit 702 and a de-correlation unit
704.

35

An input signal 706 is input into the predelay unit 702 for delaying the signal 706 for a predetermined time. The output

from the predelay unit 702 is connected to the de-correlation unit 704 to derive a de-correlated signal 708 as an output of the de-correlator 700.

5 In a preferred embodiment of the present invention, the de-correlation unit 704 comprises a lattice IIR all-pass filter. In an optional variation of the de-correlator 700, the filter coefficients (reflection coefficients) are input to the de-correlation unit 704 by means of an provider of filter coef-
10 ficients 710. When the inventive de-correlator 700 is operated within a filtering sub-band (e.g. within a QMF filter-bank), the sub-band index of the currently processed sub-band signal may additionally be input into the de-correlation unit 704. In that case, in a further modification of the present
15 invention, different filter coefficients of the de-correlation unit 704 may be applied or calculated based on the sub-band index provided.

Fig. 8 shows a lattice IIR filter as preferably used to generate the de-correlated signals.
20

The IIR filter 800 shown in Fig. 8 receives as an input an audio signal 802 and derives as an output 804 a de-correlated version of the input signal. A big advantage using an IIR
25 lattice filter is, that the exponentially decaying impulse response required to derive an appropriate de-correlated signal comes at no additional costs, since this is an inherent property of the lattice IIR filter. It is to be noted, that it is necessary to have filter coefficients $k(0)$ to $k(M-1)$
30 whose absolute values are smaller than unity to achieve the required stability of the filter. Additionally, multiple orthogonal all-pass filters can be designed more easily based on lattice IIR filters which is a major advantage for the inventive concept of deriving multiple de-correlated signals
35 from a single input signal, wherein the different derived de-correlated signals shall be almost perfectly de-correlated or orthogonal to one another.

More details on the design and the properties of all-pass lattice filters may be found in "Adaptive Filter Theory", Simon Haykin, ISBN 0-13-090126-1, Prentice-Hall, 2002.

5

Fig. 9 shows an inventive receiver or audio player 900, having an inventive audio decoder 902, a bit stream input 904, and an audio output 906.

10 A bit stream can be input at the input 904 of the inventive receiver/audio player 900. The bit stream then is decoded by the decoder 902 and the decoded signal is output or played at the output 906 of the inventive receiver/audio player 900.

15 Fig. 10 shows a transmission system comprising a transmitter 908 and an inventive receiver 900.

The audio signal input at an input interface 910 of the transmitter 908 is encoded and transferred from the output of
20 the transmitter 908 to the input 904 of the receiver 900. The receiver decodes the audio signal and plays back or outputs the audio signal on its output 906.

The present invention relates to coding of multi-channel representations of audio signals using spatial parameters. The
25 present invention teaches new methods for de-correlating signals in order to lower the coherence between the output channels. It goes without saying that although the new concept to create multiple de-correlated signals is extremely advantageous in an inventive audio decoder, the inventive concept
30 may also be used in any other technical field that requires the efficient generation of such signals.

Although the present invention has been detailed within
35 multi-channel audio decoder that are performing an upmix in a single upmixing step, the present invention may of course also be incorporated in audio decoders that are based on a

hierarchical decoding structure, such as for example shown in Fig. 2.

Although the previously described embodiments mostly describe the derivation of decorrelated signals from a single downmix signal, it goes without saying that also more than one audio channel may be used as input for the decorrelators or the pre-decorrelation-matrix, i.e. that the downmix signal may comprise more than one downmixed audio channel.

10

Furthermore, the number of de-correlated signal derived from a single input signal is basically un-limited, since the filter order of lattice filters can be varied without limitation and, since it is possible to find a new set of filter coefficients deriving a de-correlated signal being orthogonal or mainly orthogonal to other signals in the set.

15

Depending on certain implementation requirements of the inventive methods, the inventive methods can be implemented in hardware or in software. The implementation can be performed using a digital storage medium, in particular a disk, DVD or a CD having electronically readable control signals stored thereon, which cooperate with a programmable computer system such that the inventive methods are performed. Generally, the present invention is, therefore, a computer program product with a program code stored on a machine readable carrier, the program code being operative for performing the inventive methods when the computer program product runs on a computer. In other words, the inventive methods are, therefore, a computer program having a program code for performing at least one of the inventive methods when the computer program runs on a computer.

20

25

30

While the foregoing has been particularly shown and described with reference to particular embodiments thereof, it will be understood by those skilled in the art that various other changes in the form and details may be made without departing

35

from the spirit and scope thereof. It is to be understood that various changes may be made in adapting to different embodiments without departing from the broader concepts disclosed herein and comprehended by the claims that follow.

Claims

1. Multi-channel decoder (400; 500; 600) for generating a
5 reconstruction of a multi-channel signal (412; 508;
610a; 610b; 630) using a downmix signal (405; 505a, b;
605; 620) derived from an original multi-channel signal,
the reconstruction of the multi-channel signal (412;
508; 610a; 610b; 630) having at least three channels,
10 comprising:
- a de-correlator (402; 502; 602; 700) for deriving a set
of de-correlated signals using a de-correlation rule,
wherein the de-correlation rule is such that a first de-
15 correlated signal and a second de-correlated signal are
derived using the downmix signal (405; 505a, b; 605;
620), and that the first de-correlated signal and the
second de-correlated signal are orthogonal to each other
within an orthogonality tolerance range; and
20
- an output channel calculator (403; 503; 603) for gener-
ating output channels using the downmix signal (405;
505a, b; 605; 620), the first and the second de-
correlated signals and upmix information so that the at
25 least three channels are at least partly de-correlated
from each other.
2. Multi-channel decoder (400; 500; 600) in accordance with
claim 1 in which the de-correlation rule is such that
30 the orthogonality tolerance range includes orthogonality
values < 0.5 when an orthogonality value of 0 indicates
perfect orthogonality and an orthogonality value of 1
indicates perfect correlation.
- 35 3. Multi-channel decoder (400; 500; 600) in accordance with
claim 1 or 2, in which the decoding rule is such that
the deriving of the first and second de-correlated sig-

nals comprises filtering of an audio channel (406; 506; 607) extracted from the downmix signal (405; 505a, b; 605; 620) by means of an IIR filter.

- 5 4. Multi-channel decoder (400; 500; 600) in accordance with claim 3, in which the IIR filter is a lattice filter (704; 800) based on a lattice structure having an all-pass filter characteristic.
- 10 5. Multi-channel decoder (400; 500; 600) in accordance with claim 3 or 4, in which the IIR filter (800) is having a first adder in a forward prediction path of the filter for adding an actual portion of the audio channel and a previous portion of the audio channel which is weighted with a first weighing factor ; and
- 15 a second adder in a backward prediction path for adding the previous portion of the audio channel to the actual portion which is weighted with a second weighing factor of the audio signal; and
- 20 wherein the absolute values of the first and the second weighting factors are equal.
- 25 6. Multi-channel decoder (400; 500; 600) in accordance with claim 5, in which the IIR filter (704; 800) is operative to use a first and a second weighting factor that are derived from random noise sequences.
- 30 7. Multi-channel decoder (400; 500; 600) in accordance with one of the previous claims, in which the de-correlation rule is such that the first de-correlated signal and the second de-correlated signal are derived using a time delayed version of the downmix signal (405; 505a, b; 605; 620).
- 35

8. Multi-channel decoder (400; 500; 600) in accordance with one of the claims above, in which the decoding rule is such that the first and the second de-correlated signals are derived using a portion of the downmix signal derived from the downmix signal (405; 505a, b; 605; 620) by a real or complex-valued filterbank.
- 5
9. Multi-channel decoder (400; 500; 600) in accordance with one of the claims 3 to 7, further comprising a channel decomposer (401; 501; 601) to derive the audio channel from the downmix signal (405; 505a, b; 605; 620) using a deriving rule.
- 10
10. Multi-channel decoder (400; 500; 600) in accordance with claim 9, in which the deriving rule is such that four channels are derived from the downmix signal (405; 505a, b; 605; 620), wherein the downmix signal is having information on one original channel.
- 15
11. Multi-channel decoder (400; 500; 600) in accordance with claim 9, in which the deriving rule is such that two channels are derived from the downmix signal (405; 505a, b; 605; 620), wherein the downmix signal is having information on two original channels.
- 20
12. Multi-channel decoder (400; 500; 600) in accordance with one of the claims above, in which the output channel calculator is operative to generate five output channels from a downmix signal (405; 505a, b; 605; 620) having information on one audio channel and from four de-correlated signals.
- 25
13. Multi-channel decoder (400; 500; 600) in accordance with one of the claims 1 to 11, in which the output channel calculator is operative to generate five output channels from the downmix signal (405; 505a, b; 605; 620) having
- 30
- 35

information on two audio channels and from two de-correlated signals.

5 14. Multi-channel decoder (400; 500; 600) in accordance with one of the claims above, in which the output channel calculator (403; 503; 603) is operative to use upmixed information comprising at least one parameter indicating a desired correlation of a first and a second output channel.

10

15 15. Method of generating a reconstruction of a multi-channel signal using a downmix signal derived from an original multi-channel signal, the reconstruction of the multi-channel signal having at least three channels, the method comprising:

15

20 deriving a set of de-correlated signals using a de-correlation rule, wherein the de-correlation rule is such that the first de-correlated signal and the second de-correlated signal are derived using the downmix signal and that the first de-correlated signal and the second de-correlated signal are orthogonal to each other within an orthogonality tolerance range; and

20

25

generating output channels using the downmix signal, the first and the second de-correlation signals and upmix information so that the at least three channels are at least partly de-correlated from each other.

30

16. Reconstructed multi-channel signal having at least three channels, the reconstructed multi-channel signal being reconstructed using a downmix signal derived from an original multi-channel signal and a first de-correlated signal and a second de-correlated signal derived using the downmix signal, wherein the first de-correlated signal and the second de-correlated signal are orthogonal to each other within an orthogonality tolerance range.

35

17. Computer-readable storage medium having stored thereon a reconstructed multi-channel signal in accordance with claim 16.
- 5
18. Receiver or audio player, the receiver or audio player having a Multi-channel decoder (400; 500; 600) in accordance with claim 1.
- 10
19. Method of receiving or audio playing, the method having a method for generating a reconstruction of a multi-channel signal in accordance with claim 15.
- 15
20. Computer program for performing, when running on a computer, a method in accordance with any of the method claims 15 or 19.

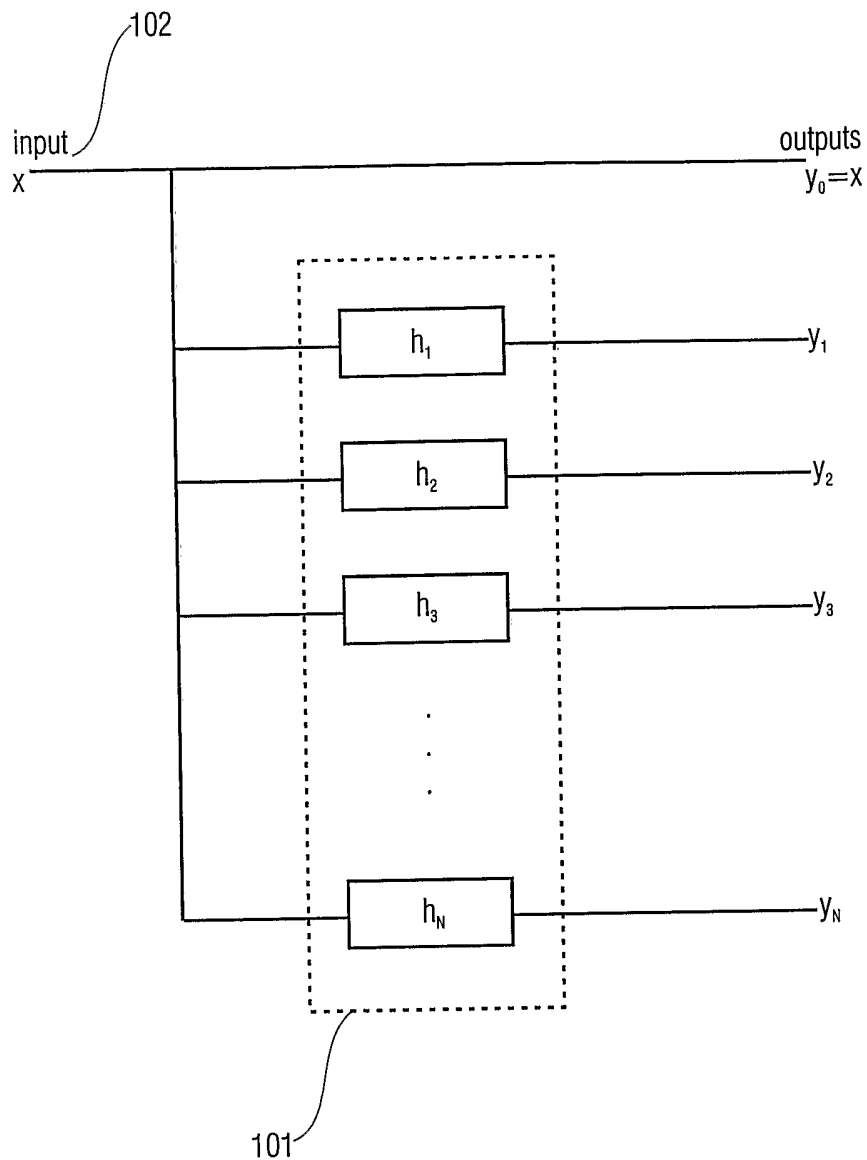


Fig. 1

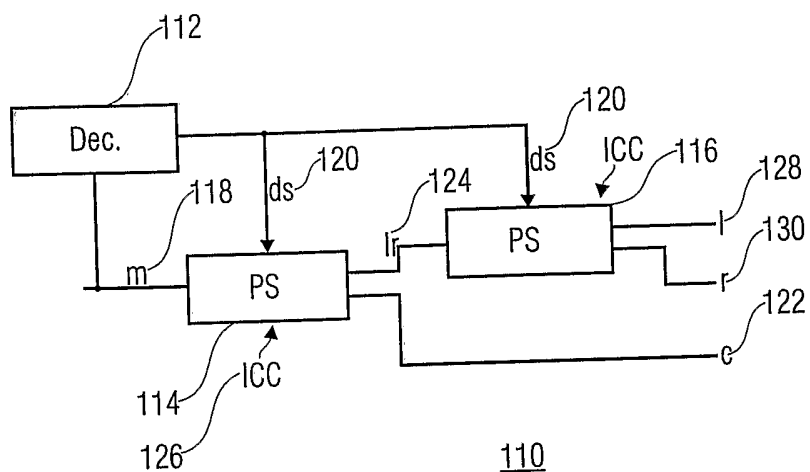


Fig. 2

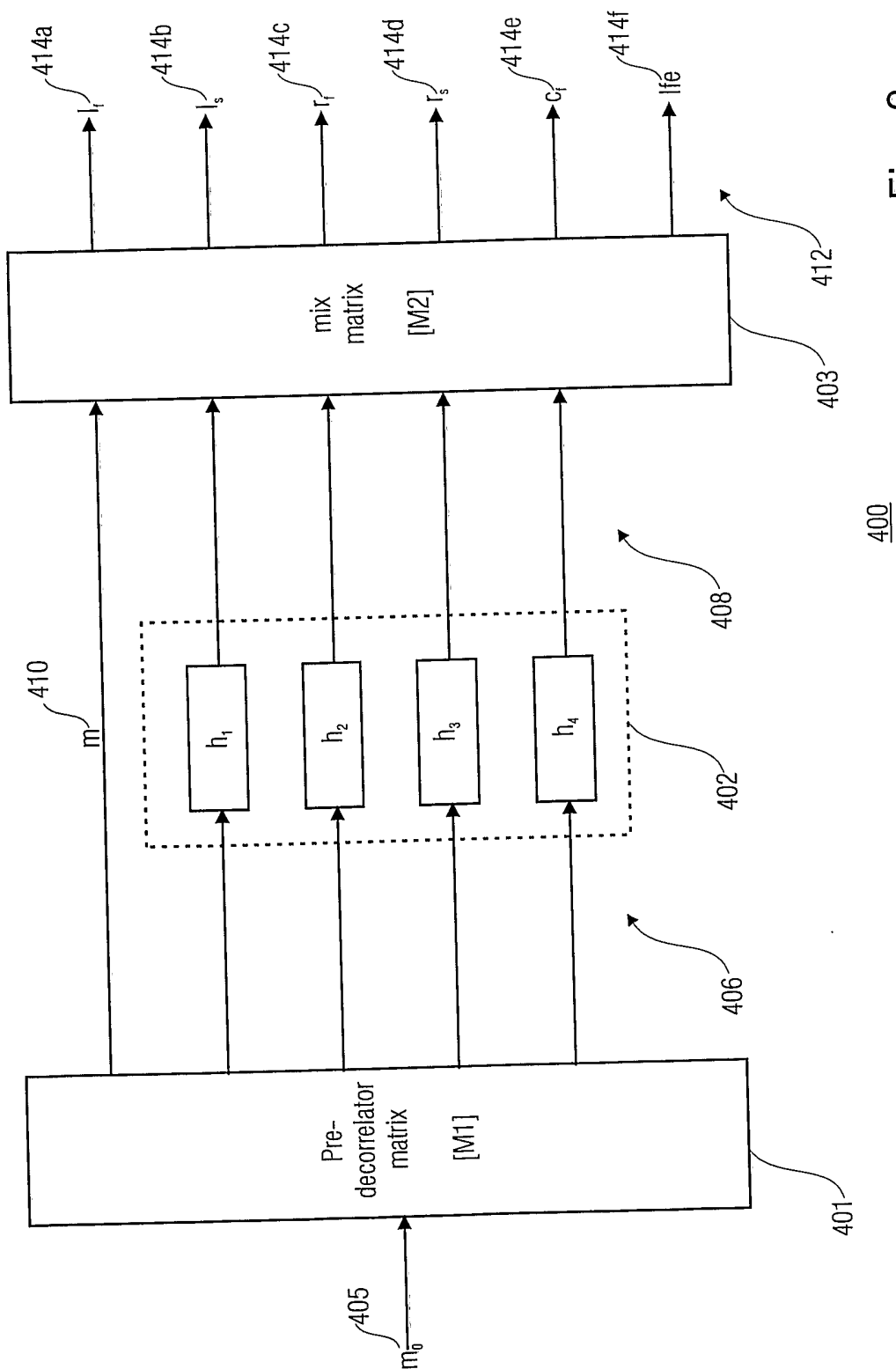


Fig. 3

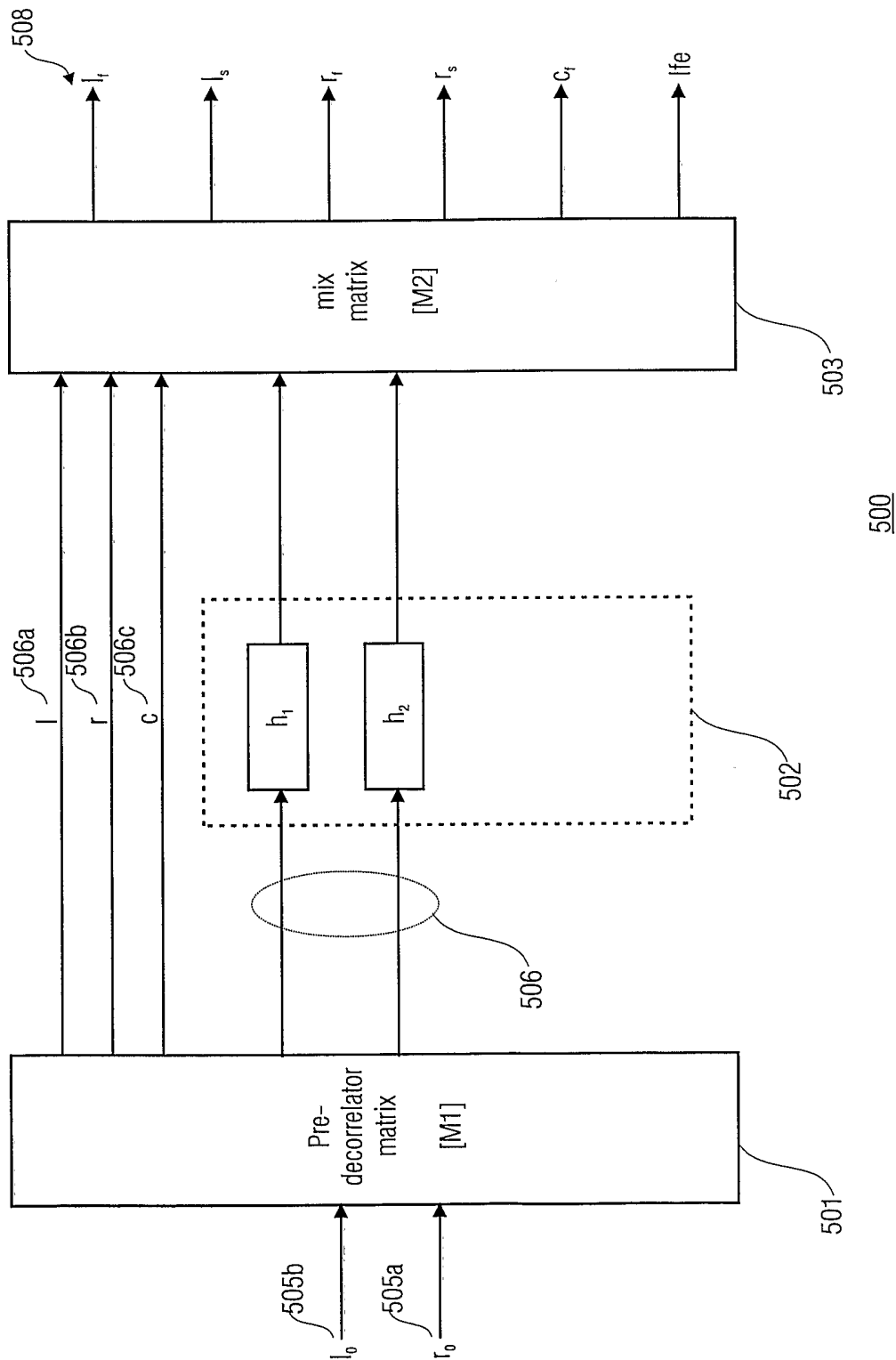


Fig. 4

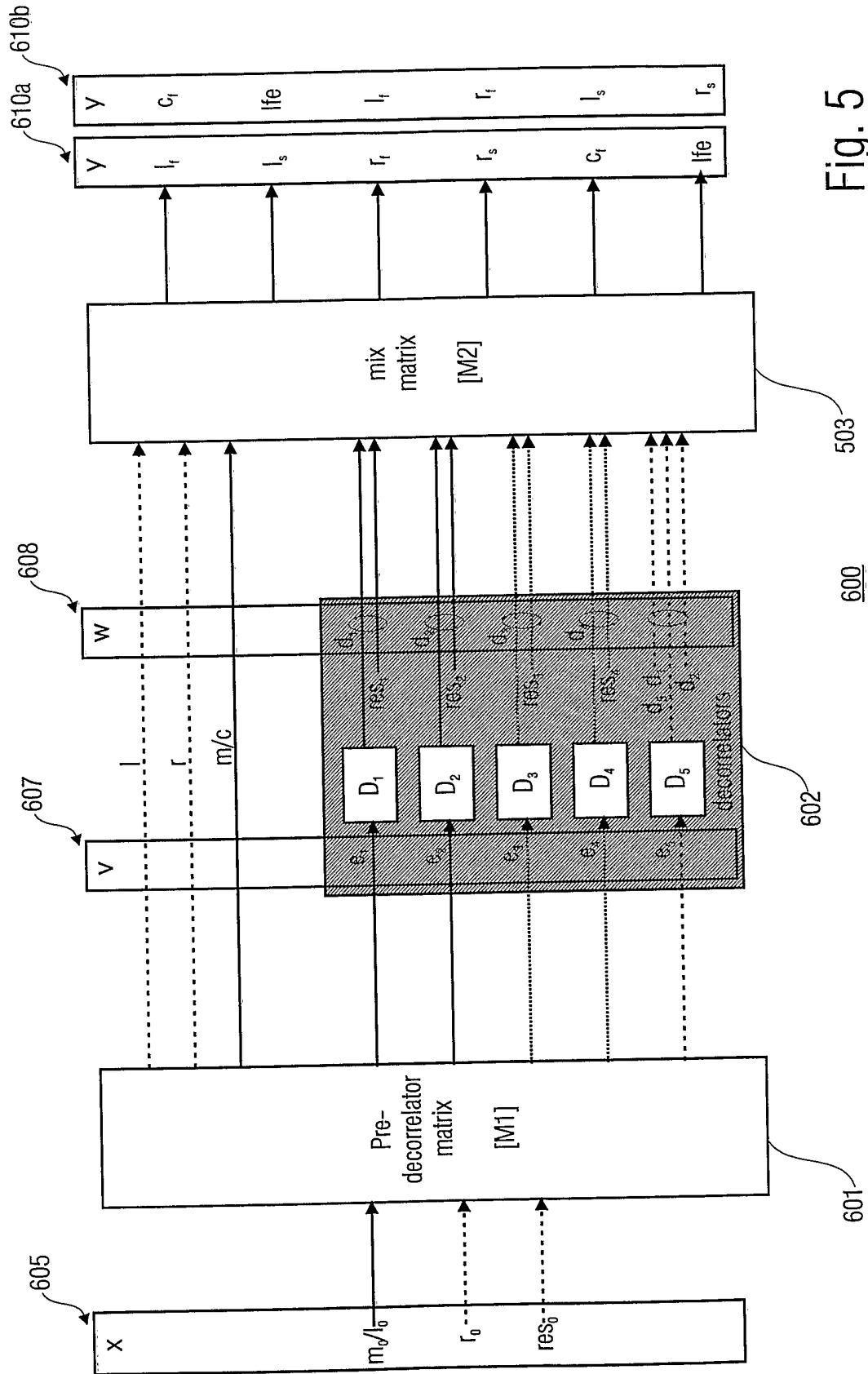


Fig. 5

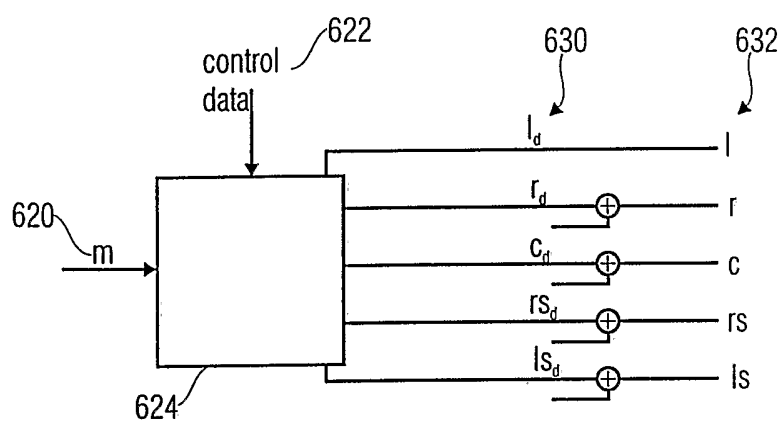


Fig. 6

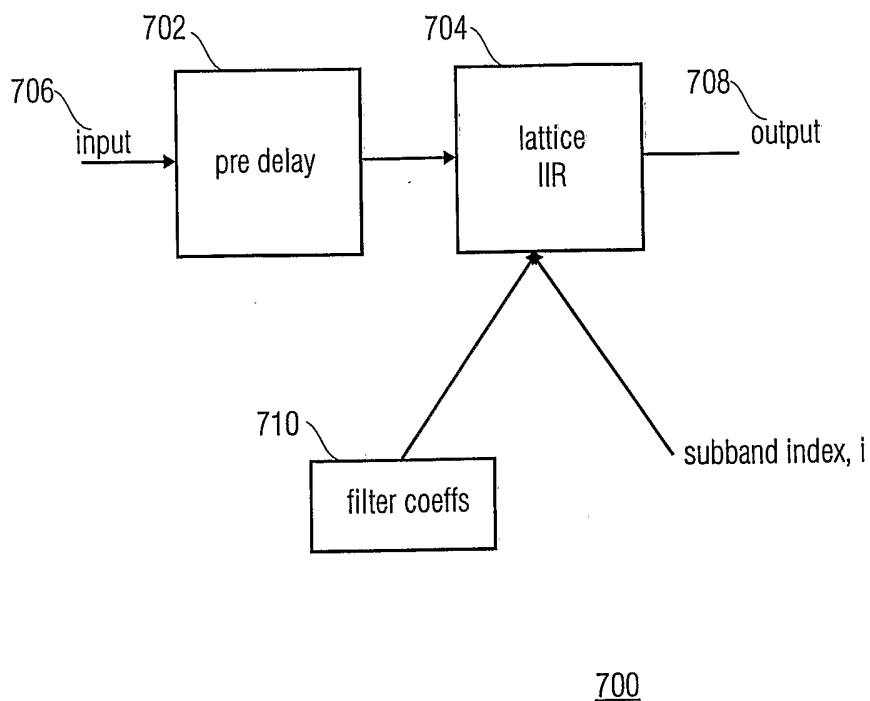


Fig. 7

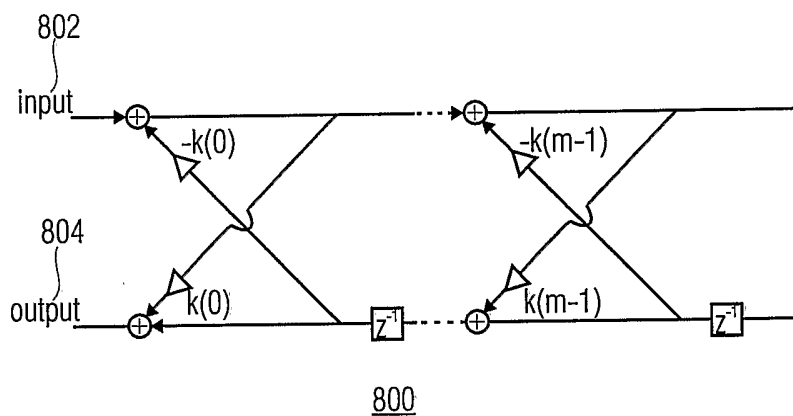


Fig. 8

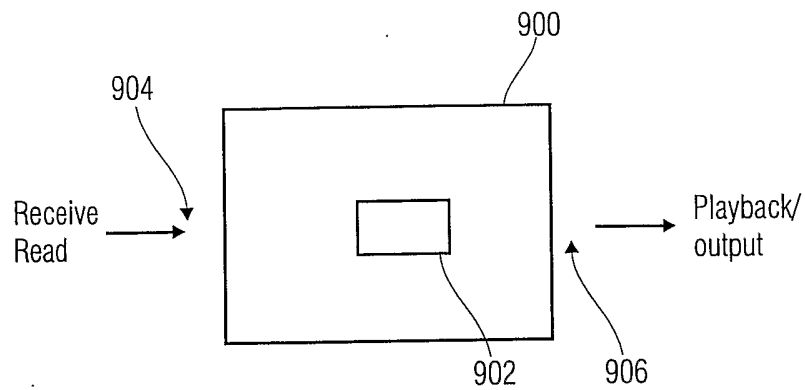


Fig. 9

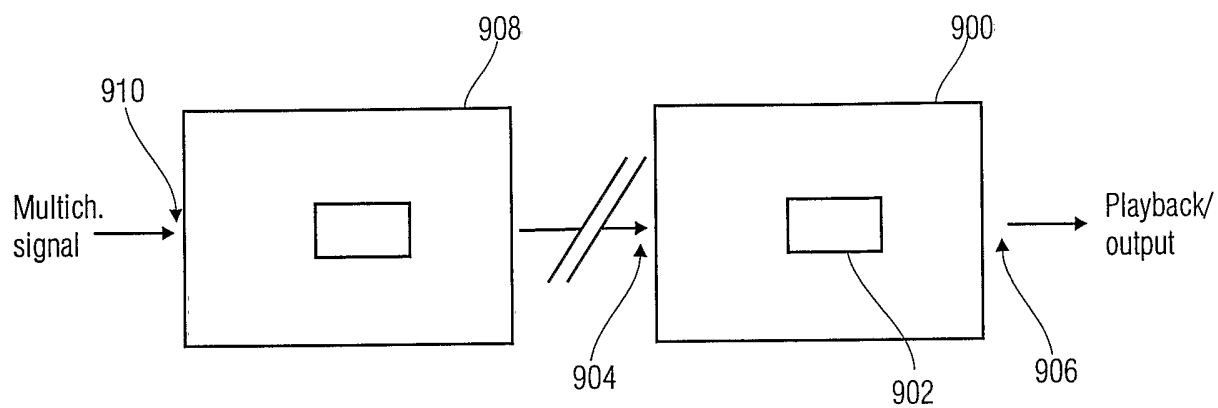


Fig. 10

INTERNATIONAL SEARCH REPORT

International application No
PCT/EP2005/011664A. CLASSIFICATION OF SUBJECT MATTER
H04S5/02 G10L19/00

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)
H04S G10L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data, PAJ, INSPEC

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
P,X	WO 2005/101370 A (CODING TECHNOLOGIES AB; PURNHAGEN, HEIKO; VILLEMOS, LARS; ENGDEGARD,) 27 October 2005 (2005-10-27) page 36, line 11 - page 43, line 14	1-20
A	G.POTARD, I.BURNETT: "Decorrelation techniques for the rendering of apparent sound source width in 3D audio displays" PROCEEDINGS OF THE 7TH INT. CONFERENCE ON DIGITAL AUDIO EFFECTS DAFX04, 'Online! 5 October 2004 (2004-10-05), pages 280-284, XP002369776 naples, italy Retrieved from the Internet: URL:http://dafx04.na.infn.it/WebProc/Proc/P_280.pdf> 'retrieved on 2006-02-23! paragraphs '3.1.1!, '3.1.2!, '3.1.3!; figure 2	1,15-17, 20

 Further documents are listed in the continuation of Box C. See patent family annex.

* Special categories of cited documents :

- *A* document defining the general state of the art which is not considered to be of particular relevance
- *E* earlier document but published on or after the international filing date
- *L* document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)
- *O* document referring to an oral disclosure, use, exhibition or other means
- *P* document published prior to the international filing date but later than the priority date claimed

- *T* later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention
- *X* document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone
- *Y* document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.
- *&* document member of the same patent family

Date of the actual completion of the international search

27 February 2006

Date of mailing of the international search report

13/03/2006

Name and mailing address of the ISA/
European Patent Office, P.B. 5818 Patentlaan 2
NL - 2280 HV Rijswijk
Tel. (+31-70) 340-2040, Tx. 31 651 epo nl,
Fax: (+31-70) 340-3016

Authorized officer

Krembel, L

INTERNATIONAL SEARCH REPORT

International application No
PCT/EP2005/011664

C(Continuation). DOCUMENTS CONSIDERED TO BE RELEVANT		
Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	<p>BREEBAART J ET AL: "High-quality parametric spatial audio coding at low bitrates" PREPRINTS OF PAPERS PRESENTED AT THE AES CONVENTION, XX, XX, 8 May 2004 (2004-05-08), pages 1-13, XP009042418 cited in the application paragraph '0005!</p> <p style="text-align: center;">-----</p>	1,15-17, 20
A	<p>SCHUIJERS E ET AL: "LOW COMPLEXITY PARAMETRIC STEREO CODING" PREPRINTS OF PAPERS PRESENTED AT THE AES CONVENTION, XX, XX, no. 6073, 8 May 2004 (2004-05-08), pages 1-11, XP008047510 paragraph '03.4!; figures 2,7</p> <p style="text-align: center;">-----</p>	1,15-17, 20
A	<p>KENDALL G S: "THE DECORRELATION OF AUDIO SIGNALS AND ITS IMPACT ON SPATIAL IMAGERY" COMPUTER MUSIC JOURNAL, CAMBRIDGE, MA, US, vol. 19, no. 4, 1995, pages 71-87, XP008026420 figure 5 * page 77 "Infinite-Impulse-Response (IIR) Filter Design" *</p> <p style="text-align: center;">-----</p>	1,15-17, 20

INTERNATIONAL SEARCH REPORT

Information on patent family members

International application No
PCT/EP2005/011664

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
WO 2005101370 A	27-10-2005	WO 2005101371 A1	27-10-2005