



US006873294B1

(12) **United States Patent**
Anderson et al.

(10) **Patent No.:** **US 6,873,294 B1**
(45) **Date of Patent:** **Mar. 29, 2005**

(54) **ANTENNA ARRANGEMENT HAVING
MAGNETIC FIELD REDUCTION IN NEAR-
FIELD BY HIGH IMPEDANCE ELEMENT**

(75) Inventors: **Eric A. Anderson**, Crystal Lake, IL
(US); **Benjamin O. White**, Crystal
Lake, IL (US)

(73) Assignee: **Motorola, Inc.**, Schaumburg, IL (US)

(*) Notice: Subject to any disclaimer, the term of this
patent is extended or adjusted under 35
U.S.C. 154(b) by 48 days.

(21) Appl. No.: **10/658,183**

(22) Filed: **Sep. 9, 2003**

(51) **Int. Cl.**⁷ **H01Q 1/26; H01Q 1/38**

(52) **U.S. Cl.** **343/702; 343/700 MS**

(58) **Field of Search** **343/702, 700 MS,
343/745, 754, 815**

(56) **References Cited**

U.S. PATENT DOCUMENTS

4,367,474 A	*	1/1983	Schaubert et al. ...	343/700 MS
4,777,490 A	*	10/1988	Sharma et al.	343/754
5,585,810 A	*	12/1996	Tsuru et al.	343/745
5,668,560 A	*	9/1997	Evans et al.	343/702
6,034,636 A	*	3/2000	Saitoh	343/700 MS

6,255,994 B1	*	7/2001	Saito	343/700 MS
6,421,016 B1		7/2002	Phillips et al.	
6,466,170 B2	*	10/2002	Zhou	343/700 MS
2002/0140607 A1	*	10/2002	Zhou	343/700 MS
2003/0214443 A1	*	11/2003	Bauregger et al. ...	343/700 MS
2004/0041733 A1	*	3/2004	Milosavijevic	343/700 MS

* cited by examiner

Primary Examiner—Don Wong

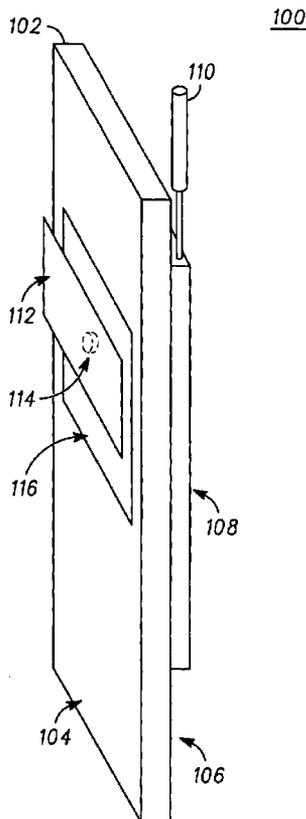
Assistant Examiner—Hung Tran Vy

(74) *Attorney, Agent, or Firm*—Hisashi D. Watanabe

(57) **ABSTRACT**

An apparatus and a method for reducing magnetic field in near-field by presenting high impedance at an operating frequency band are disclosed. A conducting element (112), which is suspended substantially parallel to a first side (104) of a printed circuit board (102) of a wireless portable communication device (100) over an electrically grounded conductor (116), forms a reactive element. The reactive element provides high impedance at operating frequencies of the wireless portable communication device and diverts radio frequency currents from the first side of the printed circuit board to the second side of the printed circuit board such that a magnetic field produced by the radio frequency currents on the first side of the printed circuit board is reduced in the near-field.

20 Claims, 6 Drawing Sheets



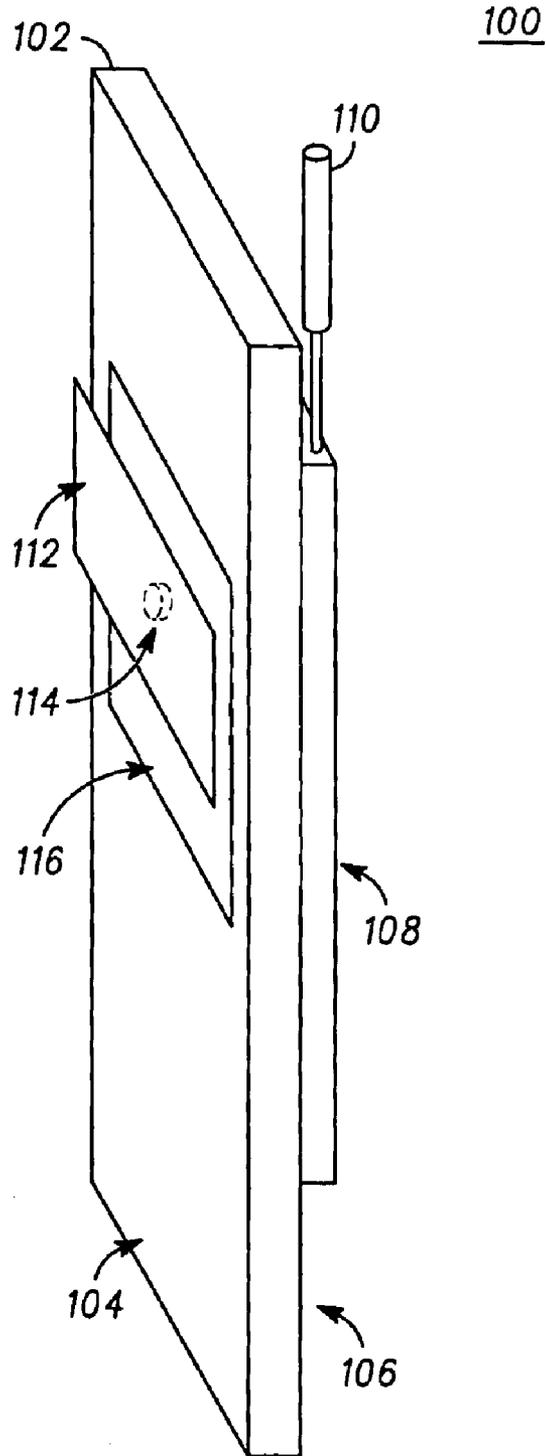


FIG. 1

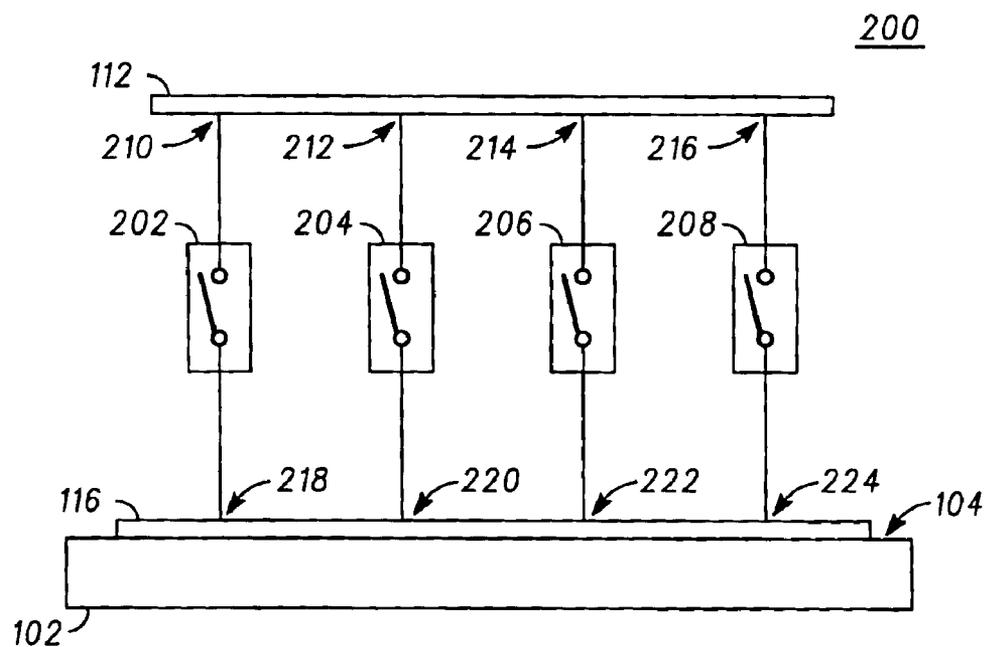


FIG. 2

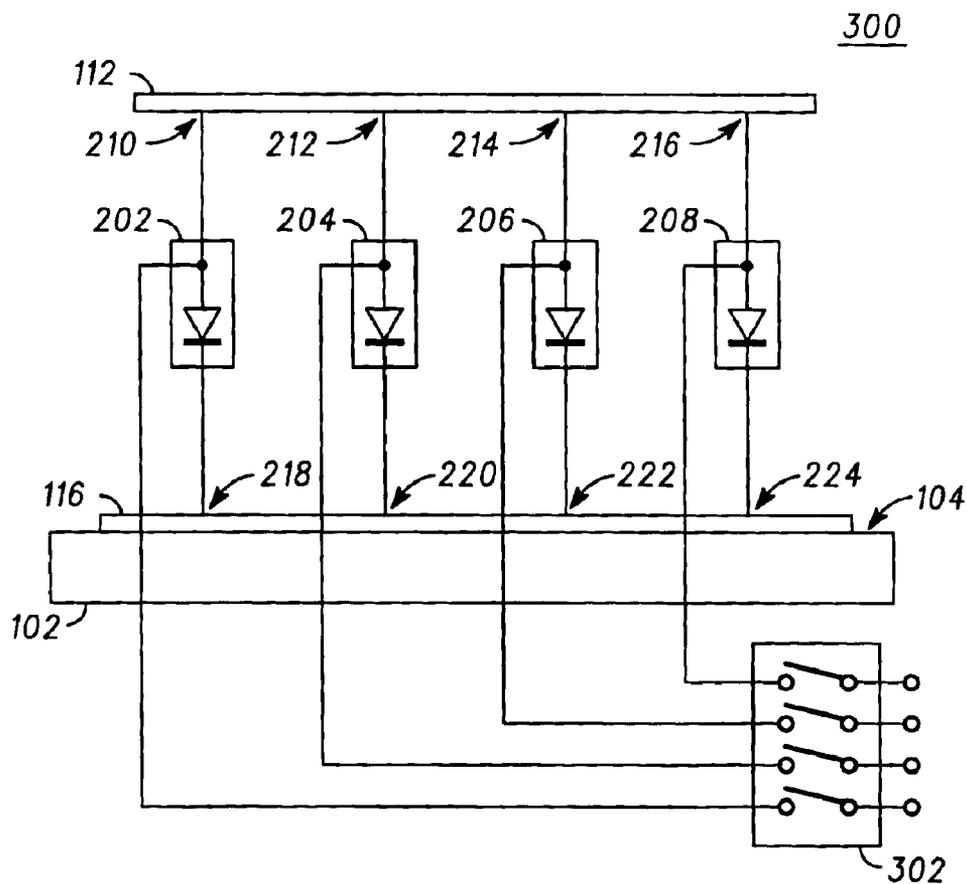


FIG. 3

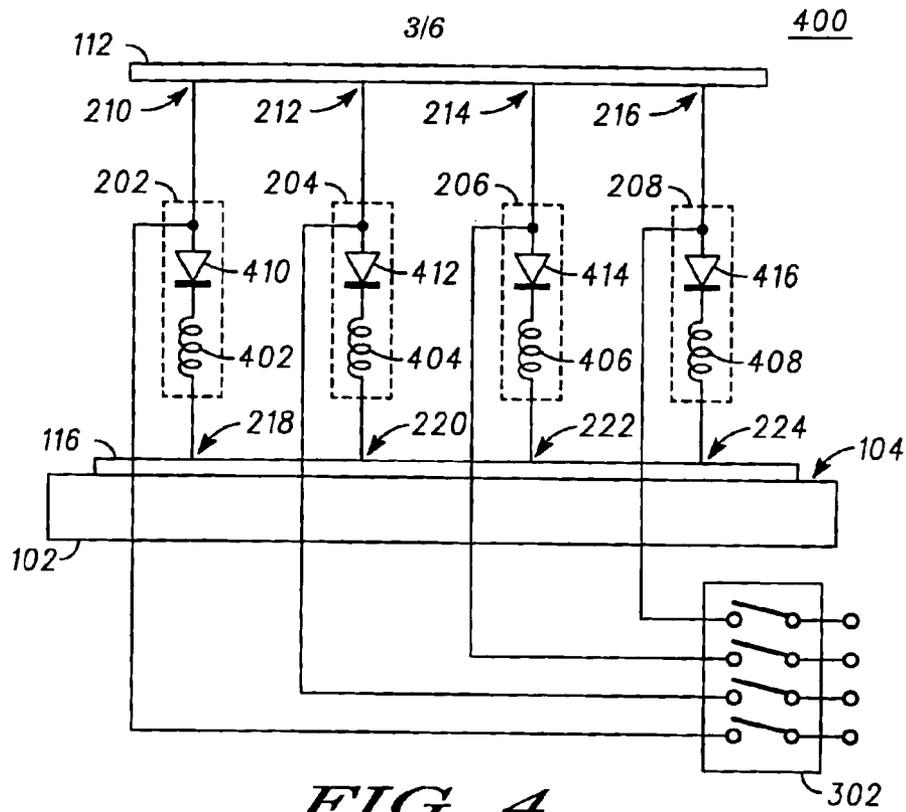


FIG. 4

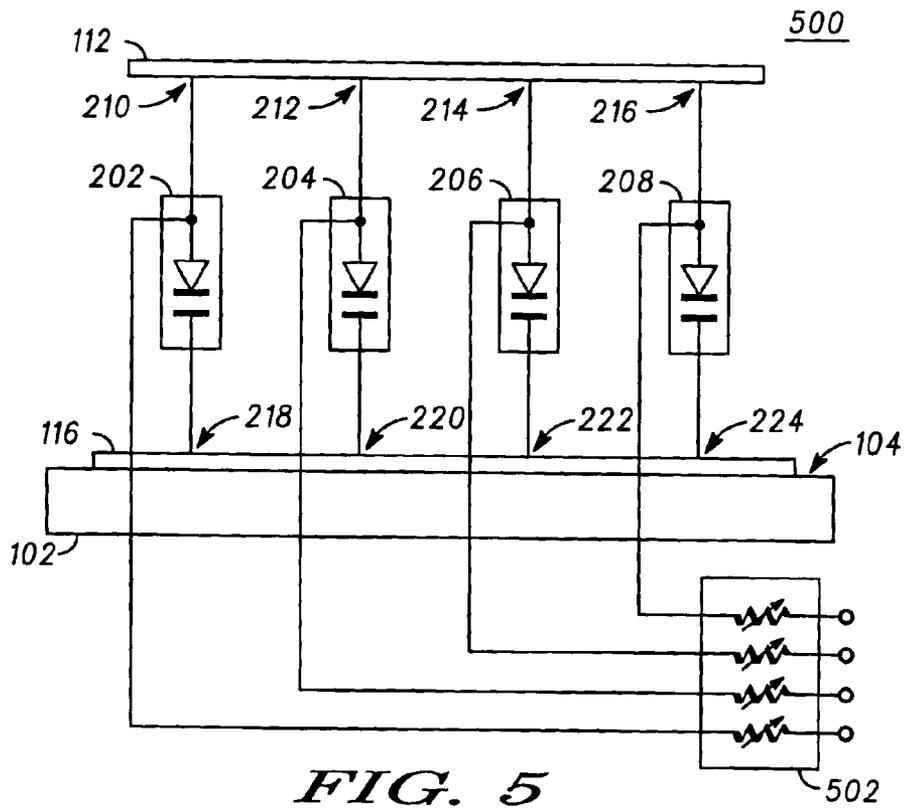


FIG. 5

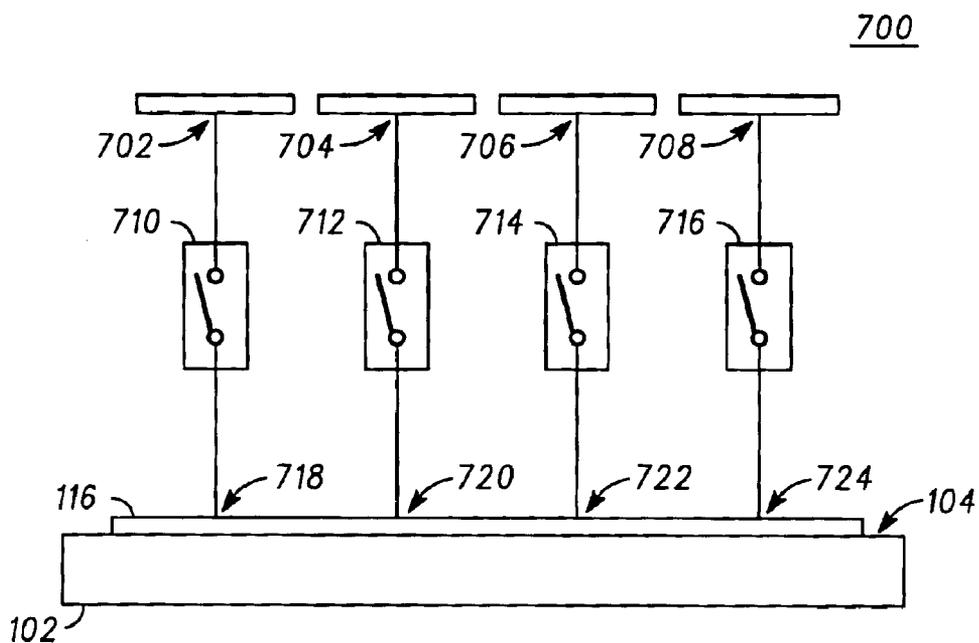


FIG. 7

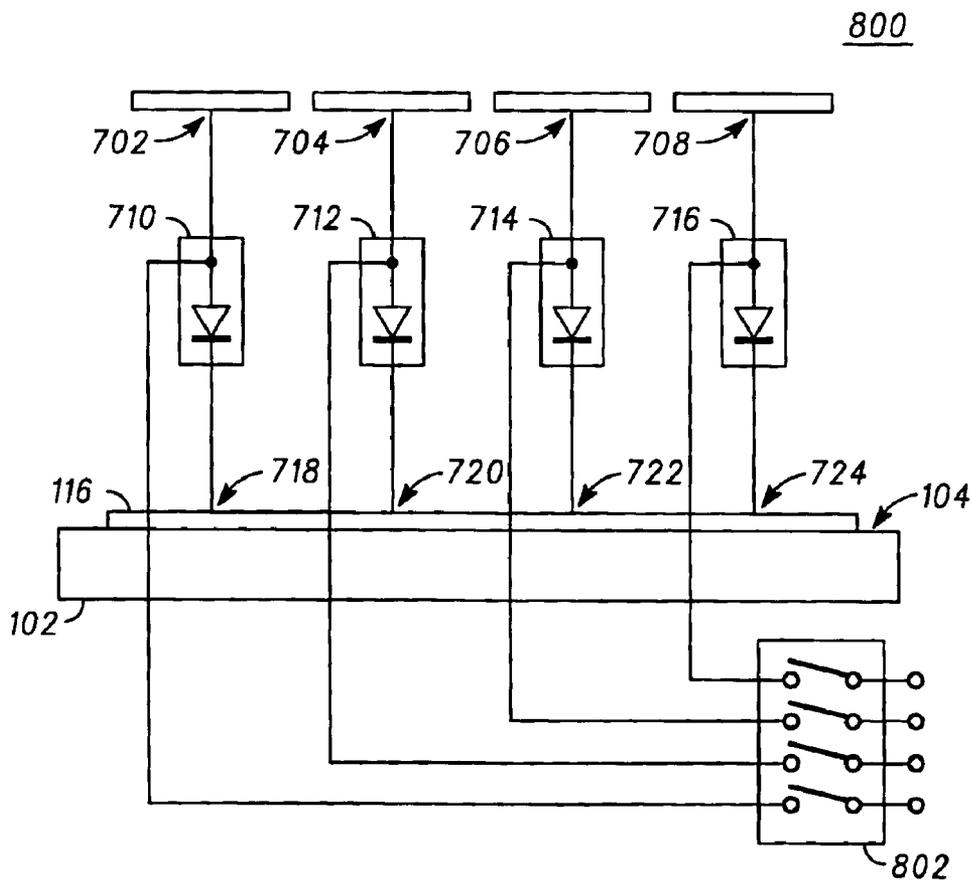


FIG. 8

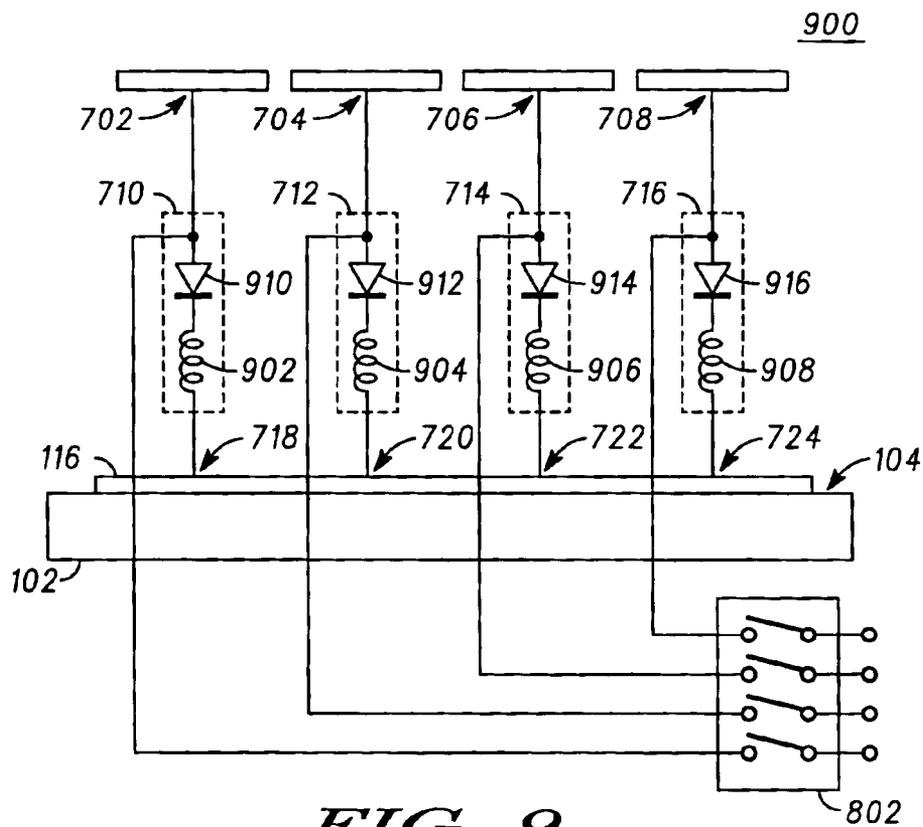


FIG. 9

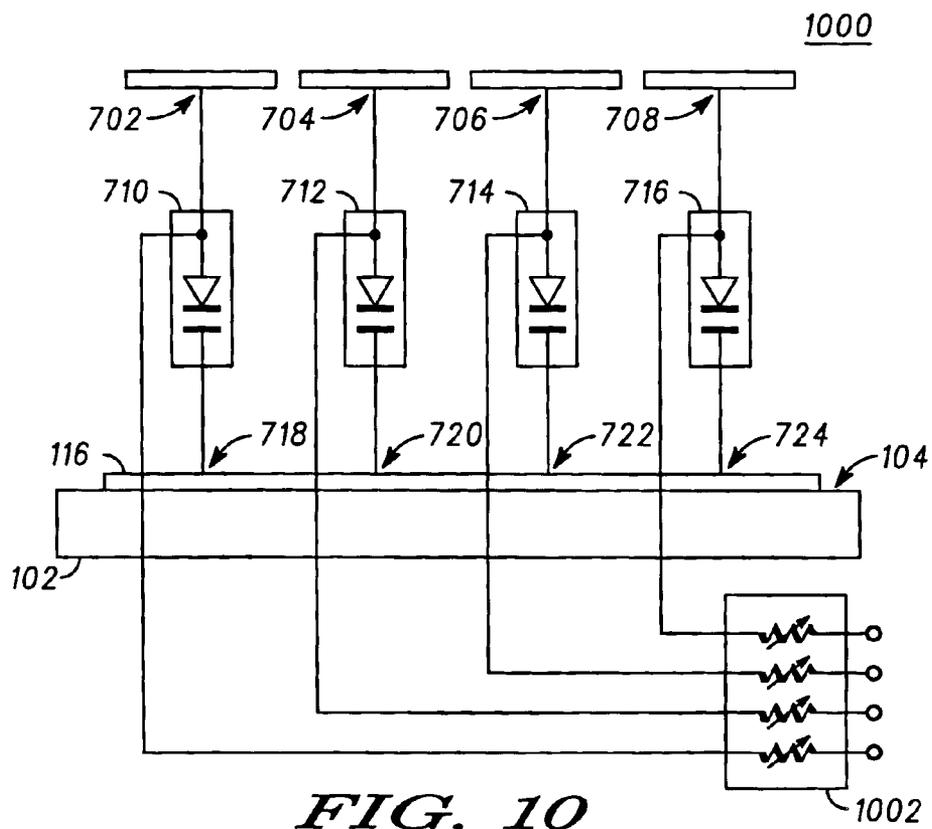


FIG. 10

1

ANTENNA ARRANGEMENT HAVING MAGNETIC FIELD REDUCTION IN NEAR- FIELD BY HIGH IMPEDANCE ELEMENT

FIELD OF THE INVENTION

The present invention generally relates to a method and an apparatus for an antenna arrangement, and more specifically to a method and an apparatus for an antenna arrangement reducing undesired magnetic field in a near-field.

BACKGROUND OF THE INVENTION

As wireless portable communication devices, such as cellular telephones, are made smaller, corresponding components including antennas used for those devices are also made smaller, and/or are sometimes at least partially integrated with other components. As a result, a printed circuit board ("PCB"), which is populated with electronic and mechanical components, is effectively used as part of radiating antenna elements of the wireless portable communication device. Radio frequency ("RF") current flows on the PCB and the PCB acts as an antenna. For cellular telephones operating in lower frequency band such as Global System for Mobile Communications ("GSM"), which covers the frequency band from about 880 MHz to 960 MHz, and Advanced Mobile Phone System ("AMPS"), which covers the frequency band from about 824 MHz to 894 MHz, the effect of the PCB radiation is more apparent compared to cellular telephones operating in higher frequency band such as Personal Communications Services ("PCS"), which covers the frequency band from about 1850 MHz to 1990 MHz. Because the size of an antenna is typically made to have an electrical length corresponding to the wavelength of the frequency used, the size of the antenna is generally larger for a lower frequency application.

By using a PCB of a cellular telephone as a radiating element, the radiating efficiency of the cellular telephone may be easily degraded due to the PCB being in close proximity to a user's body. For example, when a cellular telephone is used, it is typically held in a user's hand, which essentially covers one side of the PCB, and the other side of the PCB is held against the user's face. When the cellular telephone is carried in a user's pocket or is carried by a belt-clip, one side of the PCB faces the user's body. This presence of the user's body in close proximity to the PCB, which is being used as a radiating element, may significantly affect the radiation efficiency of the cellular telephone.

A cellular telephone may use a variety of types of antennas, such as a helical antenna and an internal antenna. The helical antenna may be viewed as a dipole-like structure comprising the antenna as a quarter-wave radiator and the PCB as another quarter-wave radiator. The internal antenna may be viewed as a matching network to the PCB for the 824-960 MHz bands of operation. Compared to the PCB, the antenna itself is generally much smaller in volume, and therefore contains more concentrated radiation energy than the PCB. To reduce the degradation in efficiency due to the presence of the user's body in close proximity, the antenna is typically located in the cellular telephone where it is kept away from the body of the user. However, the PCB, having RF currents flowing and emitting radiation, is still kept next to the user's body, and the radiation efficiency of the cellular telephone is still considerably susceptible to the proximity of the user's body to the PCB.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is an exemplary diagram of a first embodiment of an antenna arrangement for a wireless portable communication device in accordance with the present invention;

2

FIG. 2 is an exemplary block diagram of a second embodiment of the antenna arrangement having a plurality of switches in accordance with the present invention;

FIG. 3 is an exemplary block diagram of PIN diodes used as switches in accordance with the present invention;

FIG. 4 is an exemplary block diagram of the PIN diode switches coupled to reactive elements in accordance with the present invention;

FIG. 5 is an exemplary block diagram of varactor diodes used to provide variable capacitance in accordance with the present invention;

FIG. 6 is an exemplary illustration of a wireless portable communication device utilizing an antenna arrangement in accordance with the present invention;

FIG. 7 is an exemplary block diagram of a third embodiment of the antenna arrangement having a plurality of conducting elements in accordance with the present invention;

FIG. 8 is an exemplary block diagram of PIN diodes used as switches in accordance with the present invention;

FIG. 9 is an exemplary block diagram of the PIN diode switches coupled to reactive elements in accordance with the present invention; and

FIG. 10 is an exemplary block diagram of varactor diodes used to provide variable capacitance in accordance with the present invention.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

The present invention provides a method and an apparatus for an antenna arrangement suitable for a wireless portable communication device, such as a cellular telephone, that reduces undesired magnetic field in a near-field. The near-field is generally defined as an area or a volume defined by a distance of a few wavelengths of operating radio frequency from an origin of radio frequency radiation, such as an antenna of the cellular telephone. In a typical cellular telephone usage, a user may place the cellular telephone in a pocket, may clip it to his belt, or may hold in his hand and hold it against his face. Therefore, when the cellular telephone transmits a signal, part of the user's body is placed in the near-field created by the transmitted signal, and attenuates the transmitted signal, which results in reducing a usable transmitted power of the cellular telephone. It is generally desirable to direct the transmitted power away from the user. By reducing the transmitted power towards the user while maintaining the total transmitted power, the usable transmitted power is effectively increased. The transmitted power from the cellular telephone at radio frequency can be considered as a product of an electric field and a magnetic field produced by a transmitter of the cellular telephone and radiated from the antenna and a printed circuit board of the cellular telephone. Because the magnetic field is proportional to a current flow, the magnetic field may be reduced by reducing the current flow. Therefore, by reducing the current flow, which reduces the magnetic field, the transmitted power, as the product of the electric field and now the reduced-magnetic field, can be reduced. By reducing the current flow in an appropriate place, the resulting transmitted power towards the user can be reduced, without reducing the total transmitted power, effectively improving the usable transmitted power.

FIG. 1 is an exemplary diagram of a first embodiment of an antenna arrangement for a wireless portable communication device **100** in accordance with the present invention.

The wireless portable communication device **100** comprises a printed circuit board **102**, which has a first side **104** and a second side **106**, a transceiver **108**, which includes a transmitter, a receiver, and a controller, disposed on the second side of the printed circuit board **102**, an antenna **110** coupled to the transceiver **108**, and a conducting element **112** suspended parallel to the printed circuit board **102** over the first side **104**. The conducting element **112** is supported by a post **114**. An area on the first side **104** of the printed circuit board **102** under the conducting element **112** is substantially covered by a conductor **116**, which is electrically grounded, and the conducting element **112** and the conductor **116** form a capacitor. A capacitance of the capacitor formed is proportional to an area of the conducting element **112** and is also inversely proportional to the distance between the conducting element **112** and the conductor **116**. Mathematically, the capacitance, C in Farads ("F"), can be expressed as:

$$C = \frac{A\epsilon}{h},$$

where A, in square meters ("m²"), is the area of the conducting element **112**,

ϵ , in Farads per meter ("F/m"), is the dielectric constant of a material between the conducting element **112** and the conductor **116**, and

h, in meters ("m"), is the distance between the conducting element **112** and the conductor **116**.

The post **114**, which supports and electrically connects to the conducting element **112**, is also electrically grounded through the conductor **116**, and behaves as an inductor. An inductance of the inductor, L in Henries ("H"), is proportional to the length of the post **114**, and can be expressed mathematically as:

$$L = \alpha h,$$

where α in Henries per meter ("H/m") is a constant based upon a thickness of the post **114**, and

h, in meters, again is the distance between the conducting element **112** and the conductor **116**, which is the length of the post **114**.

A resonant frequency of the combination of the capacitor and the inductor, at which the impedance of the combination of the capacitor and the inductor approaches infinity, is inversely proportional to a square root of the product of the capacitance and the inductance. Mathematically, the resonant frequency, f, can be expressed as:

$$f = \frac{1}{2\pi\sqrt{LC}} = \frac{1}{2\pi\sqrt{\alpha h \frac{A\epsilon}{h}}} = \frac{1}{2\pi\sqrt{\alpha A\epsilon}}.$$

Therefore, the resonant frequency of the conducting element **112** can be adjusted to be equal to a desired frequency by varying any one or more of the terms present in the above equation.

When the wireless portable communication device **100** transmits a signal at a desired transmit frequency, radio frequency power associated with the signal is radiated from the antenna **110** as well as the printed circuit board **102** due to radio frequency currents generated by the transmitter of the transceiver **108** flowing on the printed circuit board **102**. However, by adjusting the resonant frequency of the con-

ducting element **112** to be the desired transmit frequency, the radio frequency currents flowing on the first side **104** of the printed circuit board **102** encounters high impedance at the conducting element **112**, and a portion of the radio frequency currents is diverted to the second side **106** of the printed circuit board **102**. Because the flow of the radio frequency currents on the first side **104** is reduced by diverting the radio frequency currents to the second side **106**, the magnetic field in the near-field produced by the radio frequency currents on the first side **104** is reduced, thereby reducing the radiated radio frequency power from the first side **104** of the printed circuit board **102**. By reducing the radiated radio frequency power from the first side **104** of the printed circuit board **102** without changing the total power transmitted by the transmitter of the transceiver **108**, the radiated radio frequency power from the second side **106** of the printed circuit board **102** is effectively increased, thereby increasing an effective, or usable, radiated radio frequency power of the transceiver **108**.

FIG. 2 is an exemplary block diagram of a second embodiment of an antenna arrangement **200** for the wireless portable communication device **100** in accordance with the present invention. FIG. 2 illustrates a side view of the conducting element **112** and the conductor **116**, which is electrically grounded, disposed on the first side **104** of the printed circuit board **102**. Instead of the post **114**, which provides a fixed inductance at a fixed location, in the second embodiment of the antenna arrangement **200** provides a plurality of switches (only four switches, **202**, **204**, **206**, and **208** are shown) configured to electrically couple from a plurality of element locations (only four element locations, **210**, **212**, **214**, and **216** are shown) of the conducting element **112** to a plurality of conductor locations (only four conductor locations, **218**, **220**, **222**, and **224** are shown) of the conductor **114**. By varying the location where the conducting element **112** is coupled to the conductor **114**, by varying the number of locations where the conducting element **112** is coupled to the conductor **114**, or by varying the locations and the number of locations where the conducting element **112** is coupled to the conductor **114**, a resulting reactance can be varied to achieve a desired effect of producing high impedance for the given desired transmit frequency to reduce radio frequency current flow on the first side **104** of the printed circuit board **102**. For multiple frequency band operations, different sets of switches may be activated with each set of switches corresponding to a specific band. For example, a dual band cellular telephone having a first band at GSM 900 MHz band and a second band at GSM 1900 MHz, the switches **202**, **204**, and **206** may be activated to achieve a desired effect of producing high impedance at the 900 MHz band while only the switch **208** may be activated to achieve a desired effect of producing high impedance at the 1900 MHz band.

FIG. 3 is an exemplary block diagram **300** of the switches **202**, **204**, **206**, and **208**. Each switch configured to couple the conducting element **112** to the conductor **114** is a PIN diode configured to be activated by a selector **302**. FIG. 4 is an exemplary block diagram **400** of the PIN diode switches **202**, **204**, **206**, and **208**, each of which further comprising a corresponding reactive element to increase or decrease reactance when it is activated to produce desired high impedance at a desired frequency band and is activated by the selector **302**. In this example, inductors **402**, **404**, **406**, and **408** as the reactive elements are shown to be connected in series with corresponding PIN diode **410**, **412**, **414**, and **416**, respectively, to form switches, **202**, **204**, **206**, and **208**, respectively. FIG. 5 is an exemplary block diagram **500** of

5

the switches **202**, **204**, **206**, and **208**. Each switch is a varactor diode configured to provide variable capacitance set by a controller **502**.

FIG. 6 is an exemplary illustration of a wireless portable communication device **600** utilizing an antenna arrangement in accordance with the present invention. In this example, the conducting element **112** is a metallic bezel, which holds in place a display **602** of the wireless portable communication device **600** to the printed circuit board **102**. The conducting element **112**, the metallic bezel, is coupled to the conductor **116**, which is electrically grounded, by four switches **202**, **204**, **206**, and **208**. As previously described, the switches **202**, **204**, **206**, and **208** may be activated individually or as a combination of any of the switches, and may comprise reactive elements or varactors.

FIG. 7 is an exemplary block diagram of a third embodiment of the antenna arrangement **700** having a plurality of conducting elements, of which only four conducting elements, **702**, **704**, **706**, and **708** are shown, in accordance with the present invention. The individual conducting elements **702**, **704**, **706**, and **708** are configured to couple to the conductor **116**, which is electrically grounded, by their corresponding switches, **710**, **712**, **714**, and **716**, respectively, at a plurality of conductor locations, **718**, **720**, **722**, and **724**. By coupling different conducting element or elements to the conductor **116**, a resulting reactance can be varied to achieve a desired effect of producing high impedance for a given desired transmit frequency to reduce radio frequency current flow on the first side **104** of the printed circuit board **102**. For multiple frequency band operations, different sets of switches may be activated to couple different set of conducting elements with each set of conducting elements corresponding to a specific band. For example, a dual band cellular telephone having a first band at GSM 900 MHz band and a second band at GSM 1900 MHz, the switches **710**, **712**, and **714** may be activated to couple the conducting elements **702**, **704**, and **706** achieving a desired effect of producing high impedance at the 900 MHz band while only the switch **716** may be activated to couple the conducting element **708** achieving a desired effect of producing high impedance at the 1900 MHz band.

FIG. 8 is an exemplary block diagram **800** of the switches **710**, **712**, **714**, and **716** where each switch is a PIN diode configured to couple the corresponding conducting element **702**, **704**, **706**, or **708** to the conductor **114** by a selector **802**. FIG. 9 is an exemplary block diagram **900** of the PIN diode switches **710**, **712**, **714**, and **716**, each of which further comprising a corresponding reactive element to increase or decrease reactance when it is activated to produce desired high impedance at a desired frequency band and is activated by the selector **802**. In this example, inductors **902**, **904**, **906**, and **908** as the reactive elements are shown to be connected in series with corresponding PIN diode **910**, **912**, **914**, and **916**, respectively, to form switches, **710**, **712**, **714**, and **716**, respectively. FIG. 10 is an exemplary block diagram **1000** of the switches **710**, **712**, **714**, and **716**. Each switch is a varactor diode configured to provide variable capacitance set by a controller **1002**.

While the preferred embodiments of the invention have been illustrated and described, it is to be understood that the invention is not so limited. Numerous modifications, changes, variations, substitutions and equivalents will occur to those skilled in the art without departing from the spirit and scope of the present invention as defined by the appended claims.

What is claimed is:

1. An antenna arrangement for a wireless portable communication device, the wireless portable communication

6

device having a transceiver disposed on a printed circuit board, the printed circuit board having first and second sides opposite of each other, the antenna arrangement comprising:

an antenna coupled to the transceiver; and

a conducting element suspended substantially parallel to the first side of the printed circuit board, the conducting element coupled to an electrical ground of the printed circuit board,

wherein the conducting element presents high impedance at operating frequencies of the wireless portable communication device and diverts radio frequency currents from the first side of the printed circuit board to the second side of the printed circuit board such that a magnetic field produced by the radio frequency currents on the first side of the printed circuit board is reduced in a near-field.

2. The antenna arrangement of claim 1, further comprising a PIN diode configured to couple the conducting element to the electrical ground of the printed circuit board.

3. The antenna arrangement of claim 1, further comprising a plurality of PIN diodes configured to couple the conducting element at a plurality of locations to the electrical ground of the printed circuit board.

4. The antenna arrangement of claim 3, wherein a predetermined set of PIN diodes of the plurality of PIN diodes are configured to couple the conducting element to the electrical ground of the printed circuit board based upon a predetermined range of the operating frequencies.

5. The antenna arrangement of claim 1, further comprising a reactive element configured to couple the conducting element to the electrical ground of the printed circuit board.

6. The antenna arrangement of claim 1, further comprising a plurality of reactive elements configured to couple the conducting element at a plurality of locations to the electrical ground of the printed circuit board.

7. The antenna arrangement of claim 6, wherein a predetermined set of reactive elements of the plurality of reactive elements are configured to couple the conducting element to the electrical ground of the printed circuit board based upon a predetermined range of the operating frequencies.

8. The antenna arrangement of claim 1, further comprising a varactor diode configured to couple the conducting element to the electrical ground of the printed circuit board, the varactor diode further configured to receive a control signal to provide variable capacitance based upon a predetermined range of the operating frequencies.

9. The antenna arrangement of claim 1, wherein the conducting element is a metallic bezel configured to retain a display of the wireless portable communication device.

10. An antenna arrangement for a wireless portable communication device, the wireless portable communication device having a printed circuit board, the printed circuit board having first and second sides opposite of each other, a transceiver disposed on a printed circuit board, the antenna arrangement comprising:

an antenna coupled to the transceiver; and

a plurality conducting elements suspended substantially parallel to the first side of the printed circuit board, at least one of the plurality of conducting elements coupled to an electrical ground of the printed circuit board,

wherein the at least conducting element of the plurality of conducting elements presents a high impedance at operating frequencies of the wireless portable communication device and diverts radio frequency currents from the first side of the printed circuit board to the

second side of the printed circuit board the such that magnetic field produced by the radio frequency currents on the first side of the printed circuit board is reduced in a near-field.

11. The antenna arrangement of claim 10, further comprising a plurality of PIN diodes, each of the plurality of PIN diode configured to couple a corresponding conducting element of the plurality of conducting elements to the electrical ground of the printed circuit board.

12. The antenna arrangement of claim 11, wherein a predetermined set of PIN diodes of the plurality of PIN diodes are configured to couple the corresponding conducting elements to the electrical ground of the printed circuit board based upon a predetermined range of the operating frequencies.

13. The antenna arrangement of claim 11, further comprising a plurality of reactive elements, each of the plurality of reactive elements configured to couple a corresponding conducting element of the plurality of conducting elements to the electrical ground of the printed circuit board.

14. The antenna arrangement of claim 13, wherein a predetermined set of reactive elements of the plurality of reactive elements are configured to couple the corresponding conducting elements to the electrical ground of the printed circuit board based upon a predetermined range of the operating frequencies.

15. A method in a wireless portable communication device for reducing a magnetic field in a near-field on a first side of a printed circuit board, the printed circuit board further having a second side opposite the first side, the method comprising:

- suspending a conducting element substantially parallel to the first side of the printed circuit board;
- coupling the conducting element to an electrical ground of the printed circuit board;

presenting a high impedance at operating frequencies of the wireless portable communication device with the conducting element; and

diverting radio frequency currents to the second side of the printed circuit board from the first side of the printed circuit board.

16. The method of claim 15, wherein coupling the conducting element to an electrical ground of the printed circuit board further comprises coupling the conducting element at a plurality of locations to the electrical ground of the printed circuit board.

17. The method of claim 16, wherein coupling the conducting element at a plurality of locations to the electrical ground of the printed circuit board further comprises coupling the conducting element at a predetermined set of locations of the plurality of locations to the electrical ground of the printed circuit board based upon a predetermined range of the operating frequencies.

18. The method of claim 17, further comprising activating a plurality of PIN diodes to couple the conducting element at the predetermined set of locations of the plurality of locations to the electrical ground of the printed circuit board.

19. The method of claim 15, wherein coupling the conducting element to an electrical ground of the printed circuit board further comprises coupling the conducting element to the electrical ground of the printed circuit board using a PIN diode.

20. The method of claim 15, wherein coupling the conducting element to an electrical ground of the printed circuit board further comprises coupling the conducting element to the electrical ground of the printed circuit board using a reactive element.

* * * * *