

(19) World Intellectual Property Organization  
International Bureau



(43) International Publication Date  
28 August 2008 (28.08.2008)

PCT

(10) International Publication Number  
**WO 2008/101762 A1**

(51) International Patent Classification:  
*H04L 27/26* (2006.01) *H04L 27/00* (2006.01)

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(21) International Application Number:  
PCT/EP2008/050939

(81) Designated States (unless otherwise indicated, for every kind of national protection available): AE, AG, AL, AM, AO, AT, AU, AZ, BA, BB, BG, BH, BR, BW, BY, BZ, CA, CH, CN, CO, CR, CU, CZ, DE, DK, DM, DO, DZ, EC, EE, EG, ES, FI, GB, GD, GE, GH, GM, GT, HN, HR, HU, ID, IL, IN, IS, JP, KE, KG, KM, KN, KP, KR, KZ, LA, LC, LK, LR, LS, LT, LU, LY, MA, MD, ME, MG, MK, MN, MW, MX, MY, MZ, NA, NG, NI, NO, NZ, OM, PG, PH, PL, PT, RO, RS, RU, SC, SD, SE, SG, SK, SL, SM, SV, SY, TJ, TM, TN, TR, TT, TZ, UA, UG, US, UZ, VC, VN, ZA, ZM, ZW.

(22) International Filing Date: 28 January 2008 (28.01.2008)

(25) Filing Language: English

(26) Publication Language: English

(30) Priority Data:  
60/891,265 23 February 2007 (23.02.2007) US  
11/943,440 20 November 2007 (20.11.2007) US

(84) Designated States (unless otherwise indicated, for every kind of regional protection available): ARIPO (BW, GH, GM, KE, LS, MW, MZ, NA, SD, SL, SZ, TZ, UG, ZM, ZW), Eurasian (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European (AT, BE, BG, CH, CY, CZ, DE, DK, EE, ES, FI, FR, GB, GR, HR, HU, IE, IS, IT, LT, LU, LV, MC, MT, NL, NO, PL, PT, RO, SE, SI, SK, TR), OAPI (BF, BJ, CF, CG, CI, CM, GA, GN, GQ, GW, ML, MR, NE, SN, TD, TG).

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**Published:**

- with international search report
- before the expiration of the time limit for amending the claims and to be republished in the event of receipt of amendments

(54) Title: SUBCARRIER SPACING IDENTIFICATION

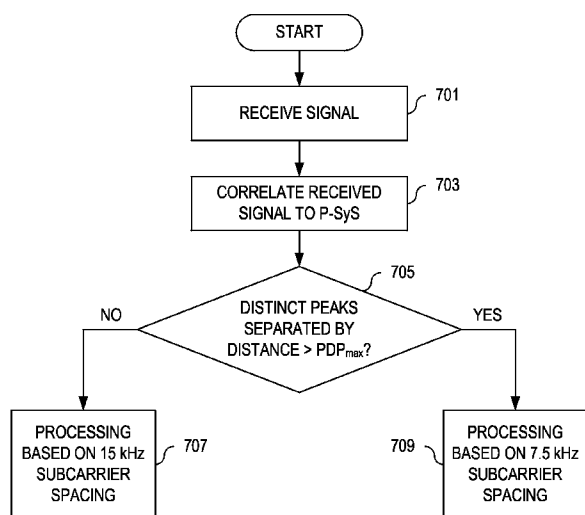


FIG. 7

(57) Abstract: A cellular communications system that utilizes OFDM in its radio interface is capable of utilizing either a first subcarrier spacing or a second subcarrier spacing. Which of these is presently in use is indicated by generating a first type of synchronization signal in response to the first subcarrier spacing presently being in use, and generating a second type of synchronization signal in response to the second subcarrier spacing presently being in use. Whichever of the first type of synchronization signal and the second type of synchronization signal was generated is transmitted. To distinguish between the first and second types of synchronization signals, a time domain representation of the second type of synchronization signal includes a plurality of instances of the first type of synchronization signal.

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## SUBCARRIER SPACING IDENTIFICATION

### BACKGROUND

5           The invention relates to subcarrier spacing in mobile communication systems that employ Orthogonal Frequency Division Multiplexing (OFDM), and more particularly to techniques that enable a user equipment operating in such a system to detect which of a number of possible subcarrier spacings are presently being used for communications.

10           In the forthcoming evolution of the mobile cellular standards like the Global System for Mobile Communication (GSM) and Wideband Code Division Multiple Access (WCDMA), new transmission techniques like OFDM are likely to occur. Furthermore, in order to have a smooth migration from the existing cellular systems to the new high capacity high data rate system in existing radio spectrum, the new  
15           system has to be able to operate in a dynamically changeable bandwidth. A proposal for such a new flexible cellular system is 3G Long Term Evolution (3G LTE, where “3G” stands for “Third Generation”) that can be seen as an evolution of the 3G WCDMA standard. OFDM will be used in this system in a technique called OFDM Multiple access (“OFDMA”) to enable multiple users to share access to the radio  
20           spectrum in the downlink. The system will be able to operate on bandwidths ranging from 1.25 MHz to 20 MHz. Furthermore, data rates up to 100 Mb/s will be supported on the largest bandwidth.

          Another important aspect of LTE is efficient support for broadcast and Multimedia Broadcast/Multicast Service (“MBMS”). In LTE, so called “Single  
25           Frequency Network” (SFN) operation is foreseen in which base stations are synchronized. Here, MBMS content is transmitted from several base stations using the same physical resources. The signals from these multiple transmissions are automatically “combined in the air”, so no additional receiver resources are required for this purpose. In order for this “over the air combining” to work, all participating  
30           base stations must be synchronized – both in the frequency domain and in the time-domain – down to the extent of a fraction of the length of a cyclic prefix. In order to ease time synchronization requirements a long cyclic prefix is advantageous. However, increasing the cyclic prefix without increasing the OFDM symbol duration increases overhead and is thus not attractive. One possible solution is to use a  
35           smaller subcarrier spacing (and corresponding bandwidth), thus increasing the OFDM

symbol duration (the OFDM symbol duration is inversely proportional to the subcarrier spacing). For example, halving the subcarrier spacing results in OFDM symbols that are twice as long, thereby enabling a cyclic prefix that is twice as long. In this manner, the amount of overhead is maintained constant. Therefore, in addition to support for  
5 15 kHz subcarrier spacing, LTE also supports the use of a 7.5 kHz subcarrier spacing for SFN operation.

The physical layer of a 3G LTE system includes a generic radio frame having a duration of 10ms. FIG. 1 illustrates one such frame 100. Each frame has 20 slots (numbered 0 through 19), each slot having a duration of 0.5 ms. A sub-frame is made  
10 up of two adjacent slots, and therefore has a duration of 1 ms.

One important aspect of LTE is the mobility function. Hence, synchronization symbols and cell search procedures are of major importance in order for the User Equipment (UE) to detect and synchronize with other cells. To facilitate cell search and synchronization procedures, defined signals include Primary and Secondary  
15 Synchronization Signals (P-SyS and S-SyS, respectively), which are transmitted on a Primary Synchronization Channel (P-SCH) and a Secondary Synchronization Channel (S-SCH), respectively. The P-SySs and S-SySs are each broadcast twice per frame: once in sub-frame 0, and again in sub-frame 5, as shown in FIG. 1.

The UE must detect, as soon as possible, whether it is connecting to a 7.5  
20 kHz/subcarrier cell or a 15 kHz/subcarrier cell, since subsequent procedures may be different for the two cases. One possibility, of course, is to have two completely different synchronization signal designs, each uniquely associated with one of the subcarrier spacing sizes. However, here the UE would be required to have both synchronization signal designs implemented, in which case it would either have to run  
25 search algorithms for both synchronization signals in parallel – thus increasing complexity – or sequentially – thus increasing cell search time.

It is therefore desired to have a technique that will enable a UE to detect what the subcarrier spacing is without having to have two different synchronization signal designs.

## **SUMMARY**

It should be emphasized that the terms "comprises" and "comprising", when used in this specification, are taken to specify the presence of stated features, integers, steps or components; but the use of these terms does not preclude the

presence or addition of one or more other features, integers, steps, components or groups thereof.

In accordance with one aspect of the present invention, the foregoing and other objects are achieved in apparatuses and methods that indicate which of a first subcarrier spacing (e.g., 15 kHz) and a second subcarrier spacing (e.g., 7.5 kHz) is presently in use in a cellular communication system. This includes generating a first type of synchronization signal in response to the first subcarrier spacing presently being in use, and generating a second type of synchronization signal in response to the second subcarrier spacing presently being in use. Whichever of the first type of synchronization signal and the second type of synchronization signal was generated is transmitted. A time domain representation of the second type of synchronization signal includes a plurality of instances of the first type of synchronization signal.

In another aspect, a radio interface of the cellular communication system utilizes Orthogonal Frequency Division Multiplexing (OFDM); generating the first type of synchronization signal comprises mapping a set of synchronization symbols,  $S_k$ , ( $0 \leq k \leq N-1$ ) onto a contiguous set of  $N$  subcarriers, each of the  $N$  subcarriers having the first subcarrier spacing; and generating the second type of synchronization signal comprises mapping the set of synchronization symbols,  $S_k$ , onto a set of  $n \cdot N$  subcarriers ( $n > 1$ ) such that every  $n^{\text{th}}$  one of the  $n \cdot N$  carries one of the  $N$  synchronization symbols, and remaining ones of the  $n \cdot N$  subcarriers carry zero values, wherein each of the  $n \cdot N$  subcarriers has the second subcarrier spacing.

In yet another aspect, the time domain representation of the second type of synchronization signal consists of two instances of the first type of synchronization signal.

In another aspect of embodiments consistent with the invention, apparatuses and methods detect which of a first subcarrier spacing (e.g., 15 kHz) and a second subcarrier spacing (e.g., 7.5 kHz) is presently in use in a cellular communication system. This includes processing a received signal to derive therefrom a synchronization signal. It is then detected whether a time domain representation of the synchronization signal comprises only one instance of a first type of synchronization signal. If the time domain representation of the synchronization signal comprises only one instance of the first type of synchronization signal, then the first subcarrier spacing is utilized in one or more subsequent communication operations. Otherwise, if the time domain representation of the synchronization signal comprises

more than one instance of the first type of synchronization signal, then the second subcarrier spacing is utilized in one or more subsequent communication operations.

In some embodiments, detecting whether the time domain representation of the synchronization signal comprises only one instance of the first type of synchronization  
5 signal comprises: generating correlation results by correlating the received signal against the first type of synchronization signal; and detecting whether the correlation results include distinct peaks separated in time by a predetermined amount.

In alternative embodiments, detecting whether the time domain representation of the synchronization signal comprises only one instance of the first type of  
10 synchronization signal comprises generating correlation results by correlating the received signal with a delayed copy of the received signal, wherein an amount of delay applied to generate the delayed copy of the received signal corresponds to a time period of first type of synchronization signal; and detecting whether the correlation results include a distinct peak.

In still other alternative embodiments, indicating which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system involves generating a first type of synchronization signal in response to the first subcarrier spacing presently being in use, and generating a second type of synchronization signal in response to the second subcarrier spacing  
20 presently being in use. Whichever one of the first type of synchronization signal and the second type of synchronization signal was generated is then transmitted. In such embodiments, a time domain representation of the first type of synchronization signal includes a first plurality of instances of a basic waveform; a time domain representation of the second type of synchronization signal includes a second plurality  
25 of instances of the basic waveform; and the first plurality is not equal to the second plurality.

In still other embodiments, detecting which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system involves processing a received signal to derive therefrom a synchronization signal,  
30 and detecting how many instances of a basic waveform are included in a time domain representation of the synchronization signal. If the time domain representation of the synchronization signal comprises a first plurality of instances of the basic waveform, then the first subcarrier spacing is utilized in one or more subsequent communication operations. If the time domain representation of the synchronization signal comprises

a second plurality of instances of the basic waveform, then the second subcarrier spacing is utilized in one or more subsequent communication operations, wherein the first plurality is not equal to the second plurality.

## 5 BRIEF DESCRIPTION OF THE DRAWINGS

The objects and advantages of the invention will be understood by reading the following detailed description in conjunction with the drawings in which:

FIG. 1 is a timing diagram of a generic radio frame utilized in an LTE system.

10 FIG. 2A is a diagram illustrating a first type of synchronization signal, which is used when a subcarrier spacing is a larger of two possible sizes.

FIG. 2B is a diagram illustrating a second type of synchronization signal, which is used when a subcarrier spacing is a smaller of two possible sizes.

15 FIG. 3A is a time-frequency diagram showing the mapping of the P-SyS<sub>MODE1</sub> signal, in which all  $N$  of the modulation symbols,  $S_k$ , are mapped to contiguous subcarriers.

FIG. 3B is a time-frequency diagram showing the exemplary mapping of the P-SyS<sub>MODE2</sub> signal, in which the same  $N$  modulation symbols,  $S_k$ , are mapped to every other one of a group of  $2N$  subcarriers, with remaining subcarriers set to zero, in accordance with an aspect of the invention.

20 FIG. 4A illustrates an exemplary correlation results waveform when the received signal includes a P-SyS<sub>MODE1</sub>.

FIG. 4B illustrates an exemplary correlation results waveform when the received signal includes a P-SyS<sub>MODE2</sub>, in accordance with an aspect of the invention.

25 FIG. 5 is a block diagram of relevant components in a transmitter for generating a P-SyS signal in accordance with aspects of the invention.

FIG. 6 is a block diagram of relevant components of an exemplary synchronization channel receiver in a UE that is capable of detecting the indication of subcarrier spacing in accordance with another aspect of embodiments consistent with the invention.

30 FIG. 7 is a flow chart of exemplary steps that can be carried out by circuitry in a UE that detects the indication of subcarrier spacing in accordance with another aspect of embodiments consistent with the invention.

## DETAILED DESCRIPTION

The various features of the invention will now be described with reference to the figures, in which like parts are identified with the same reference characters.

The various aspects of the invention will now be described in greater detail in connection with a number of exemplary embodiments. To facilitate an understanding of the invention, many aspects of the invention are described in terms of sequences of actions to be performed by elements of a computer system or other hardware capable of executing programmed instructions. It will be recognized that in each of the embodiments, the various actions could be performed by specialized circuits (e.g., discrete logic gates interconnected to perform a specialized function), by program instructions being executed by one or more processors, or by a combination of both. Moreover, the invention can additionally be considered to be embodied entirely within any form of computer readable carrier, such as solid-state memory, magnetic disk, optical disk containing an appropriate set of computer instructions that would cause a processor to carry out the techniques described herein. Thus, the various aspects of the invention may be embodied in many different forms, and all such forms are contemplated to be within the scope of the invention. For each of the various aspects of the invention, any such form of embodiments may be referred to herein as "logic configured to" perform a described action, or alternatively as "logic that" performs a described action.

In an aspect of the embodiments consistent with the invention, a Primary SyS ("P-SyS") is provided and utilized in a communication system, wherein a first type of synchronization signal, herein denoted  $P\text{-SyS}_{\text{MODE1}}$ , is utilized when the subcarrier spacing is the larger of two possible spacings (e.g., 15 kHz per subcarrier) and a second type of synchronization signal, herein denoted  $P\text{-SyS}_{\text{MODE2}}$ , is used when the subcarrier spacing is the smaller of the two possible spacings (e.g., 7.5 kHz per subcarrier), wherein the second type of synchronization signal is formed as a function of the first type of synchronization signal. In a more specific embodiment, the second type of synchronization signal comprises two or more instances of the first type of synchronization signal in immediate succession.

In another aspect of embodiments consistent with the invention, in use, if the detected synchronization signal contains two or more of the first type of synchronization signal in immediate succession, then the detected cell is ascertained to have a subcarrier spacing associated with the smaller of the two spacings (e.g., 7.5

kHz), whereas if no repetitions are detected, the detected cell is ascertained to have a subcarrier spacing associated with the larger of the two spacings (e.g., 15 kHz).

These and other aspects are described in greater detail in the following discussion.

The following description is set in the context of an LTE system. However, this is done merely to facilitate the reader's understanding of the various aspects (e.g., by utilizing terminology that will be readily understood by those having ordinary skill in the art). The use of LTE terminology, methods and apparatuses to illustrate the various aspects of the invention should not be construed to limit the scope of the invention, or to in any way suggest that the invention is limited only to LTE embodiments. To the contrary the various aspects of embodiments consistent with the invention are equally applicable to many other systems that share relevant characteristics with LTE systems (e.g., reliance on a synchronization signal to, for example, make an initial determination of things such as signal timing), but are different in other ways.

The P-SyS in LTE is constructed by mapping a sequence  $S_k$ ,  $k = 0, 1, \dots, N-1$ , onto the subcarrier used for the synchronization signal. After Inverse Fast Fourier Transform (IFFT) processing, this results in the time domain signal

$$s_{15\text{kHz}}(n) = \sum_{k=0}^{N-1} S_k e^{(j2\pi/N) \cdot k \cdot n}, \quad n = 0, 1, \dots, N-1 \quad (1)$$

where  $n$  is a sample number, and  $N$  is the number of subcarriers.

Some of the  $S_k$  values can actually be set to zero in order to confine the time domain signal,  $s(n)$ , to its bandwidth. The inclusion of zeros may also be necessary to avoid DC distortion.

In the exemplary embodiment, two modes of operation are presented, each having its own subcarrier spacing: a first mode which does not involve SFN (e.g., unicast transmissions), and a second mode which does involve SFN. In an aspect of the invention, the P-SyS for the second type of subcarrier spacing (P-SyS<sub>MODE2</sub>) is created such that – in the time domain – it is a replica of the P-SyS used in the first mode of operation (e.g., non-SFN use) (P-SyS<sub>MODE1</sub>) plus one or more immediate repetitions of that signal. FIGS. 2A and 2B are time domain signal diagrams that illustrate this point. Specifically, each of FIGS. 2A and 2B is a graph of the amplitude of transmitted signals plotted as a function of time. The shapes of the various illustrated signals are intended to merely represent the presence of the signals, and are not intended to represent actual waveforms.



FIG. 2A illustrates a first type of synchronization signal 201, which is used when a subcarrier spacing is a larger of two possible sizes (herein illustrated as  $P\text{-SyS}_{\text{MODE1}}$ ). The duration of the first type of synchronization signal 201 is herein denoted  $t_{\text{MODE1}}$ . The first type of synchronization signal 201 is generated in accordance with Equation (1), with the symbols,  $S_k$ , being mapped onto  $N$  contiguous subcarriers. In embodiments such as LTE, one of the subcarriers is DC, and for this reason is punctured. Accordingly, in such embodiments  $S_{\text{DC}}$  should be set equal to zero.

FIG. 2B is a diagram illustrating a second type of synchronization signal 203, which is used when a subcarrier spacing is a smaller of two possible sizes (herein illustrated as  $P\text{-SyS}_{\text{MODE2}}$ ). The second type of synchronization signal 203 comprises two (or in other embodiments, more than two) components (in this example denoted first component 203a and second component 203b) transmitted in immediate succession during two (or in other embodiments, more than two) contiguous time periods for a total duration of  $t_{\text{MODE2}}$ . In the time domain, the waveform of each of the first and second components 203a, 203b of the second type of synchronization signal 203 is identical to that of the first type of synchronization signal 201. Consequently, each of the first and second components 203a, 203b has a duration equal to  $t_{\text{MODE1}}$  and can be considered an instance of the first type of synchronization signal 201.

A  $P\text{-SyS}_{\text{MODE2}}$  having this property is easily constructed by using the sequence  $S_k$  associated with the  $P\text{-SyS}_{\text{MODE1}}$  (i.e., the  $P\text{-SyS}$  of the larger of the two possible sizes) and mapping it to every  $n$ -th one of  $n \cdot N$  subcarriers (where  $n > 1$ ). The remaining subcarriers within the bandwidth allocated for the  $P\text{-SyS}$  are then populated with zeros. (Subcarriers outside the  $P\text{-SyS}$  bandwidth should also be treated as zeros, but this can be achieved by filtering on the receiver side, so actual population with zeros is not essential in practice.) The resulting time domain waveform of the  $P\text{-SyS}_{\text{MODE2}}$  will have a waveform comprising  $n$  contiguous instances of the  $P\text{-SyS}_{\text{MODE1}}$  waveform. For the case in which the spacing for SFN operation is half the original subcarrier spacing, the original  $P\text{-SyS}$  sequence is preferably mapped to every second subcarrier (i.e.,  $n=2$ ), resulting in two successive (and contiguous) instances of the  $P\text{-SyS}_{\text{MODE1}}$  waveform in the time domain. This is illustrated in FIGS. 3A and 3B, wherein FIG. 3A is a time-frequency diagram showing the mapping of the  $P\text{-SyS}_{\text{MODE1}}$  signal (all  $N$  of the modulation symbols,  $S_k$ , are mapped to contiguous subcarriers), and FIG. 3B is a time-frequency diagram showing the exemplary mapping of the  $P\text{-}$

SyS<sub>MODE2</sub> signal (the same  $N$  modulation symbols,  $S_k$ , are in this case mapped to every other one of a group of  $2N$  subcarriers, with remaining subcarriers set to zero).

More generally, mapping should be in accordance with the ratio of the “original” subcarrier spacing divided by the new (smaller) subcarrier spacing. For example, if the subcarrier spacing were only one-third the size of the “original” subcarrier spacing, the  $N$  modulation symbols,  $S_k$ , would be mapped to every third one of a group of  $3N$  subcarriers, with remaining subcarriers set to zero.

With mapping to every other one of a group of  $2N$  subcarriers as described above, the size of the IFFT needed to convert the P-SyS<sub>MODE2</sub> is twice as large (i.e.,  $2N$  instead of  $N$ ), so the number of generated time-domain samples is also twice as large:  $2N$ . Assume that the sequence representing P-SyS<sub>MODE2</sub> is given by:

$$\hat{S}_k = \begin{cases} S_{k/2} & \text{for } k = \text{even} \\ 0 & \text{for } k = \text{odd} \end{cases} \quad (2)$$

where  $S_{k/2}$  is the sequence for P-SyS<sub>MODE1</sub>. Then the time domain signal obtained from the exemplary P-SyS<sub>MODE2</sub> after IFFT processing is

$$\begin{aligned} s_{7.5kHz}(n) &= \sum_{k=0}^{2N-1} \hat{S}_k e^{(j2\pi/(2N)) \cdot k \cdot n} \\ &= \sum_{\substack{k=0 \\ k \text{ even}}}^{2N-1} \hat{S}_k e^{(j2\pi/(2N)) \cdot k \cdot n} + \underbrace{\sum_{\substack{k=0 \\ k \text{ odd}}}^{2N-1} \hat{S}_k e^{(j2\pi/(2N)) \cdot k \cdot n}}_{0 \text{ since } \hat{S}_k \text{ is } 0 \text{ for odd } k} \\ &= \sum_{\substack{k=0 \\ k \text{ even}}}^{2N-1} \hat{S}_k e^{(j2\pi/(2N)) \cdot k \cdot n}, \quad n = 0, 1, \dots, 2N-1. \end{aligned} \quad (3)$$

Now, substituting  $k \rightarrow 2k'$  and letting  $k'$  run from 0 to  $N-1$  ( $k$  then runs from 0 to  $2N-1$  populating only even values), we obtain:

$$s_{7.5kHz}(n) = \sum_{k'=0}^{N-1} \hat{S}_{2k'} e^{(j2\pi/(2N)) \cdot 2k' \cdot n}, \quad n = 0, 1, \dots, 2N-1. \quad (4)$$

With  $\hat{S}_{2k'} = S_{k'}$  (because  $\hat{S}_{2k}$  is  $S_k$ , see equation (2)), we obtain:

$$\begin{aligned} s_{7.5kHz}(n) &= \sum_{k'=0}^{N-1} S_{k'} e^{(j2\pi/(2N)) \cdot 2k' \cdot n} \\ &= \sum_{k=0}^{N-1} S_k e^{(j2\pi/N) \cdot k \cdot n}, \quad n = 0, 1, \dots, 2N-1, \end{aligned} \quad (5)$$

which is exactly two instances of the original signal (i.e., P-SyS<sub>MODE1</sub>) generated in contiguous succession.

With this P-SyS structure the presence or absence of multiple instances of the P-SyS waveform indicates which of the subcarrier spacings is being used: only one instance means that the subcarrier spacing is 15 kHz, two (or in other embodiments, more) instances means that the subcarrier spacing is 7.5 kHz. Since the P-SyS waveform for the 7.5 kHz case is the same as the P-SyS waveform for the 15 kHz case but transmitted multiple times (preferably two times), the same synchronization circuitry in the UE is capable of detecting the P-SyS for both subcarrier spacings: In the case of a 15 kHz subcarrier spacing, the synchronization metric has one peak and in the case of a 7.5 kHz subcarrier spacing, the synchronization metric poses multiple peaks.

In practice, the synchronization metric peaks are typically made up of a number of clustered peaks due to the radio channel delay spread (i.e., the clustered peaks correspond to the entire power delay profile -- "PDP"). However, in a typical LTE system, the delay spread is on the order of a few microseconds (e.g., worst case up to about 15-20 microseconds), so the P-SyS<sub>MODE1</sub> waveform can be detected. An exemplary correlation results waveform 401 obtained when a received signal containing a P-SyS<sub>MODE1</sub> is correlated against the known P-SyS<sub>MODE1</sub> is illustrated in FIG. 4A. By contrast, the multiple peaks that occur when a received signal containing a P-SyS<sub>MODE2</sub> is correlated against the known P-SyS<sub>MODE1</sub> (i.e., the correlation results obtained in the 7.5 kHz subcarrier spacing case) are separated by a time distance that is on the order of 65-70 microseconds (corresponding to one OFDM symbol, in the 15 kHz subcarrier spacing case). The correlation results 403 obtained for the 7.5 kHz subcarrier spacing case are illustrated in FIG. 4B. Because of the relatively large separation between the multiple peaks associated with the 7.5 kHz subcarrier spacing, they are easily distinguishable from multiple peaks associated with typical delay spread.

FIG. 5 is a block diagram of relevant components 500 in a transmitter for generating a P-SyS signal in accordance with aspects of the invention. In this exemplary embodiment, it is assumed that  $n=2$  (i.e., to indicate the smaller (e.g., 7.5 kHz) subcarrier spacing, the synchronization signal sequence elements are mapped to every other subcarrier, with zeros mapped in-between. A synchronization signal generator 501 generates the sequence,  $S_k$ , and supplies these to the input of a switch. The switch 503 is controlled by a signal 505 that indicates whether the subcarrier spacing is the smaller (e.g., 7.5 kHz) or the larger (e.g., 15 kHz) of the two possible

values. When the smaller of the two possible spacings is indicated by the signal 505, the switch 503 routes the sequence,  $S_k$ , to logic 507 that inserts zeros in-between the  $S_k$  symbols. In this example, only one zero is inserted in-between two  $S_k$  symbols, but in other embodiments, the logic 507 that inserts zeros in-between the  $S_k$  symbols can  
5 insert more than one zero between  $S_k$  symbols, depending upon whether it is desired to generate more than two instances of the basic P-SyS waveform. The output of the logic 507 that inserts zeros in-between  $S_k$  symbols is supplied to those inputs of an Inverse Fast Fourier Transform (IFFT) 509 associated with the P-SyS.

If the signal 505 indicates that the larger of the two possible subcarrier spacings  
10 is in effect, then the switch 503 causes the  $S_k$  symbols to be routed directly to the corresponding inputs of the IFFT 509 -- the logic 507 that inserts zeros in-between  $S_k$  symbols is, in this instance, bypassed.

A data unit 511 generates other data that is supplied to other inputs of the IFFT 509. The signal 505 is supplied to a control input of each of the data unit 511 and the  
15 IFFT 509 to control, in the case of the data unit 511, how many symbols are generated, and, in the case of the IFFT 509, how many input symbols will be processed and the effective size of the IFFT 509.

The output of the IFFT 509 is the OFDM modulated signal that will be transmitted during the OFDM symbol time associated with the P-SyS.

FIG. 6 is a block diagram of relevant components of an exemplary  
20 synchronization channel receiver 600 in a UE that is capable of detecting the indication of subcarrier spacing in accordance with another aspect of embodiments consistent with the invention. It is assumed that the supplied signal,  $Y_t$ , is supplied by a filter that substantially eliminates frequency components lying outside the allocated spacing of the P-SyS. The synchronization channel receiver 600 includes a matched  
25 filter (correlator) 601 that is matched to one period of the P-SyS. The output ( $D_t$ ) of the matched filter 601 is supplied to a control unit 603 that analyzes the matched filter output signal and generates therefrom the 5 ms timing information and a signal that indicates whether the detected cell is using a 7.5 kHz or 15 kHz subcarrier spacing.  
30 As described earlier, the control unit 603 in this example bases this decision on whether only a single P-SyS response was detected (i.e.,  $P\text{-SyS}_{\text{MODE1}}$ ), or whether multiple (e.g., two) P-SyS power delay profiles (i.e.,  $P\text{-SyS}_{\text{MODE2}}$ ) were detected.

FIG. 7 is a flow chart of exemplary steps that can be carried out by circuitry (e.g., the control unit 603) in a UE that detects the indication of subcarrier spacing in

accordance with another aspect of embodiments consistent with the invention. A signal is received (step 701) and processed by a matched filter, which correlates the received signal to a known P-SyS<sub>MODE1</sub> (step 703). The results from this processing are then analyzed by, for example, a control unit that tests whether the correlation result includes distinct peaks separated by a distance greater than an expected delay spread associated with the radio channel ( $PDP_{max}$ ) (decision block 705). A value of  $PDP_{max}$  can, for example, be 20 microseconds. If distinct peaks are not detected ("NO" path out of decision block 705), then 15 kHz subcarrier spacings have been detected, and processing proceeds accordingly (step 707). Otherwise ("YES" path out of decision block 705), 7.5 kHz subcarrier spacings have been detected, and processing proceeds accordingly (step 709).

Alternative subcarrier spacing detection algorithms can also be implemented for use in a UE. In one such example, a differential correlator correlates the received signal with a delayed copy of itself. In this case, the delay corresponds to the time period of the P-SyS for the larger subcarrier spacing (i.e., P-SyS<sub>MODE1</sub>) (e.g.,  $1/15 \text{ kHz} = 66 \text{ } \mu\text{sec}$ ), and the integration time can be any duration larger than the longest cyclic prefix and up to the time period of the original P-SyS – e.g.,  $1/15 \text{ kHz}$ . When the time domain representation of the synchronization signal comprises more than one instance of the P-SyS<sub>MODE1</sub>, the correlator results will include a distinct peak. Such a detector is similar to a blind cyclic prefix estimator and therefore the smallest integration time should be larger than the longest cyclic prefix, otherwise a cyclic prefix might be detected rather than a repetition.

The proposed P-SyS structure enables a very simple structure and technique for detecting subcarrier spacing. No additional synchronization channel detection circuitry is needed for the smaller (e.g., 7.5 kHz) subcarrier spacing because the synchronization channel detection circuitry matched to the larger (e.g., 15 kHz) subcarrier spacing automatically also detects the P-SyS when 7.5 kHz subcarrier spacings are used.

The invention has been described with reference to particular embodiments. However, it will be readily apparent to those skilled in the art that it is possible to embody the invention in specific forms other than those of the embodiment described above.

For example, the exemplary embodiments described above assumed that only two different subcarrier spacings needed to be indicated and detected. However,

alternative embodiments could be designed, using the principles described above, in which more than two different subcarrier spacings are indicated and detected. In such cases, each of the possible subcarrier spacings can be uniquely associated with a different number of instances of the “original” P-SyS waveform, for example one  
5 instance of the waveform to indicate use of a first subcarrier spacing, two instances of the waveform to indicate use of a second subcarrier spacing, three instances of the waveform to indicate use of a third subcarrier spacing, and so on.

Also, the exemplary embodiments described above assumed that one of the subcarrier spacings (e.g., 15 kHz) was associated with the occurrence of only one  
10 instance of the “original” P-SyS waveform. However, this is not essential to the invention. To the contrary, the invention can more generally be defined as involving a number of different subcarrier spacings wherein each is uniquely associated with one of a like number of possible P-SyS waveforms, each of the possible P-SyS waveforms having a different number of instances of a “basic” waveform, thereby making them  
15 distinguishable from one another. For example, in an embodiment involving two possible subcarrier spacings (e.g., 15 kHz and 7.5 kHz), a first of these subcarrier spacings can be indicated by two occurrences of a “basic” waveform, and a second of these subcarrier spacings can be indicated by four occurrences of the “basic” waveform. This can be achieved by, for example, mapping the  $S_k$  symbols onto every  
20 other subcarrier (with zeros inserted in-between) for the first of the subcarrier spacings, and mapping the  $S_k$  symbols onto every fourth subcarrier (with zeros inserted in-between) for the second of the subcarrier spacings.

In yet another exemplary embodiment involving two possible subcarrier spacings (e.g., 15 kHz and 10 kHz), a first of these subcarrier spacings can be  
25 indicated by two occurrences of a “basic” waveform and a second of these subcarrier spacings can be indicated by three occurrences of the “basic” waveform. To implement this, the  $S_k$  symbols are mapped onto every second subcarrier (with zeros inserted in-between) for the case of 15 kHz subcarrier spacing, and are mapped onto every third subcarrier (with zeros inserted in-between) for the case of 10 kHz. This  
30 works out because in the 15 kHz subcarrier spacing mode, the OFDM symbol length is  $1/15 \text{ kHz} = 66.67 \mu\text{s}$ . When the  $S_k$  symbols are placed on every second subcarrier, the “effective” subcarrier spacing is  $2 \times 15 = 30 \text{ kHz}$ , and the duration of the “basic” P-SyS waveform is  $1/30 \text{ kHz} = 33.33 \mu\text{s}$ . Therefore, in one OFDM symbol ( $= 66.67 \mu\text{s}$ ) two basic waveforms are contained.

For the 10 kHz subcarrier spacing mode, the OFDM symbol length is  $1/10 \text{ kHz} = 100 \mu\text{s}$ . When the  $S_k$  symbols are placed on every third subcarrier (with zeros inserted in-between), the “effective” subcarrier spacing is  $3 \times 10 = 30 \text{ kHz}$ , and the duration of the “basic” P-SyS waveform is  $1/30 \text{ kHz} = 33.33 \mu\text{s}$ . Therefore, in one  
5 OFDM symbol ( $=100 \mu\text{s}$ ) three basic waveforms are contained. That three instances of the basic waveform will be produced by placing the  $S_k$  symbols onto every third subcarrier can be shown by equations analogous to equations 2 through 5.

Accordingly, the described embodiments are merely illustrative and should not be considered restrictive in any way. The scope of the invention is given by the  
10 appended claims, rather than the preceding description, and all variations and equivalents which fall within the range of the claims are intended to be embraced therein.

**CLAIMS**

1. A method of indicating which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system, the method comprising:

5 in response to the first subcarrier spacing presently being in use, generating a first type of synchronization signal;

in response to the second subcarrier spacing presently being in use, generating a second type of synchronization signal; and

10 transmitting whichever of the first type of synchronization signal and the second type of synchronization signal was generated,

wherein a time domain representation of the second type of synchronization signal includes a plurality of instances of the first type of synchronization signal.

2. The method of claim 1, wherein:

15 a radio interface of the cellular communication system utilizes Orthogonal Frequency Division Multiplexing (OFDM);

generating the first type of synchronization signal comprises mapping a set of synchronization symbols,  $S_k$ , ( $0 \leq k \leq N-1$ ) onto a contiguous set of  $N$  subcarriers, each of the  $N$  subcarriers having the first subcarrier spacing; and

20 generating the second type of synchronization signal comprises mapping the set of synchronization symbols,  $S_k$ , onto a set of  $n \cdot N$  subcarriers ( $n > 1$ ) such that every  $n^{th}$  one of the  $n \cdot N$  carries one of the  $N$  synchronization symbols, and remaining ones of the  $n \cdot N$  subcarriers carry zero values, wherein each of the  $n \cdot N$  subcarriers has the second subcarrier spacing.

25 3. The method of claim 2, wherein  $n=2$ .

4. The method of claim 1, wherein the time domain representation of the second type of synchronization signal consists of two instances of the first type of  
30 synchronization signal.

5. The method of claim 1, wherein:

the first subcarrier spacing is 15 kHz; and

the second subcarrier spacing is 7.5 kHz.



6. A method of detecting which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system, the method comprising:

5       processing a received signal to derive therefrom a synchronization signal;  
      detecting whether a time domain representation of the synchronization signal comprises only one instance of a first type of synchronization signal; and  
      if the time domain representation of the synchronization signal comprises only one instance of the first type of synchronization signal, then utilizing the first subcarrier  
10       spacing in one or more subsequent communication operations; and  
      if the time domain representation of the synchronization signal comprises more than one instance of the first type of synchronization signal, then utilizing the second subcarrier spacing in one or more subsequent communication operations.

15   7. The method of claim 6, wherein detecting whether the time domain representation of the synchronization signal comprises only one instance of the first type of synchronization signal comprises:

      generating correlation results by correlating the received signal against the first type of synchronization signal; and

20       detecting whether the correlation results include distinct peaks separated in time by a predetermined amount.

8. The method of claim 6, wherein detecting whether the time domain representation of the synchronization signal comprises only one instance of the first  
25       type of synchronization signal comprises:

      generating correlation results by correlating the received signal with a delayed copy of the received signal, wherein an amount of delay applied to generate the delayed copy of the received signal corresponds to a time period of first type of synchronization signal; and

30       detecting whether the correlation results include a distinct peak.

9. The method of claim 6, wherein:  
      the first subcarrier spacing is 15 kHz; and  
      the second subcarrier spacing is 7.5 kHz.

10. An apparatus for indicating which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system, the method comprising:

5 logic that responds to the first subcarrier spacing presently being in use by generating a first type of synchronization signal;

logic that responds to the second subcarrier spacing presently being in use by generating a second type of synchronization signal; and

10 logic that transmits whichever of the first type of synchronization signal and the second type of synchronization signal was generated,

wherein a time domain representation of the second type of synchronization signal includes a plurality of instances of the first type of synchronization signal.

11. The apparatus of claim 10, wherein:

15 a radio interface of the cellular communication system utilizes Orthogonal Frequency Division Multiplexing (OFDM);

the logic that generates the first type of synchronization signal comprises logic that maps a set of synchronization symbols,  $S_k$ , ( $0 \leq k \leq N-1$ ) onto a contiguous set of  $N$  subcarriers, each of the  $N$  subcarriers having the first subcarrier spacing; and

20 the logic that generates the second type of synchronization signal comprises logic that maps the set of synchronization symbols,  $S_k$ , onto a set of  $n \cdot N$  subcarriers ( $n > 1$ ) such that every  $n^{th}$  one of the  $n \cdot N$  carries one of the  $N$  synchronization symbols, and remaining ones of the  $n \cdot N$  subcarriers carry zero values, wherein each of the  $n \cdot N$  subcarriers has the second subcarrier spacing.

25 12. The apparatus of claim 11, wherein  $n=2$ .

13. The apparatus of claim 10, wherein the time domain representation of the second type of synchronization signal consists of two instances of the first type of synchronization signal.

14. The apparatus of claim 10, wherein:

the first subcarrier spacing is 15 kHz; and

the second subcarrier spacing is 7.5 kHz.

15. An apparatus for detecting which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system, the method comprising:

5 logic that processes a received signal to derive therefrom a synchronization signal;

logic that detects whether a time domain representation of the synchronization signal comprises only one instance of a first type of synchronization signal; and

10 logic that utilizes the first subcarrier spacing in one or more subsequent communication operations if the time domain representation of the synchronization signal comprises only one instance of the first type of synchronization signal; and

logic that utilizes the second subcarrier spacing in one or more subsequent communication operations if the time domain representation of the synchronization signal comprises more than one instance of the first type of synchronization signal.

15 16. The apparatus of claim 15, wherein the logic that detects whether the time domain representation of the synchronization signal comprises only one instance of the first type of synchronization signal comprises:

20 logic that generates correlation results by correlating the received signal against the first type of synchronization signal; and

logic that detects whether the correlation results include distinct peaks separated in time by a predetermined amount.

25 17. The apparatus of claim 15, wherein the logic that detects whether the time domain representation of the synchronization signal comprises only one instance of the first type of synchronization signal comprises:

30 logic that generates correlation results by correlating the received signal with a delayed copy of the received signal, wherein an amount of delay applied to generate the delayed copy of the received signal corresponds to a time period of first type of synchronization signal; and

logic that detects whether the correlation results include a distinct peak.

18. The apparatus of claim 15, wherein:  
the first subcarrier spacing is 15 kHz; and

the second subcarrier spacing is 7.5 kHz.

19. A method of indicating which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system, the method comprising:

in response to the first subcarrier spacing presently being in use, generating a first type of synchronization signal;

in response to the second subcarrier spacing presently being in use, generating a second type of synchronization signal; and

transmitting whichever of the first type of synchronization signal and the second type of synchronization signal was generated,

wherein:

a time domain representation of the first type of synchronization signal includes a first plurality of instances of a basic waveform;

a time domain representation of the second type of synchronization signal includes a second plurality of instances of the basic waveform; and

the first plurality is not equal to the second plurality.

20. An apparatus for indicating which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system, the apparatus comprising:

logic configured to generate a first type of synchronization signal in response to the first subcarrier spacing presently being in use;

logic configured to generate a second type of synchronization signal in response to the second subcarrier spacing presently being in use; and

logic configured to transmit whichever of the first type of synchronization signal and the second type of synchronization signal was generated,

wherein:

a time domain representation of the first type of synchronization signal includes a first plurality of instances of a basic waveform;

a time domain representation of the second type of synchronization signal includes a second plurality of instances of the basic waveform; and

the first plurality is not equal to the second plurality.

21. A method of detecting which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system, the method comprising:

processing a received signal to derive therefrom a synchronization signal;

5 detecting how many instances of a basic waveform are included in a time domain representation of the synchronization signal; and

if the time domain representation of the synchronization signal comprises a first plurality of instances of the basic waveform, then utilizing the first subcarrier spacing in one or more subsequent communication operations; and

10 if the time domain representation of the synchronization signal comprises a second plurality of instances of the basic waveform, then utilizing the second subcarrier spacing in one or more subsequent communication operations,

wherein the first plurality is not equal to the second plurality.

15 22. An apparatus for detecting which of a first subcarrier spacing and a second subcarrier spacing is presently in use in a cellular communication system, the apparatus comprising:

logic configured to process a received signal to derive therefrom a synchronization signal;

20 logic configured to detect how many instances of a basic waveform are included in a time domain representation of the synchronization signal; and

logic configured to utilize the first subcarrier spacing in one or more subsequent communication operations if the time domain representation of the synchronization signal comprises a first plurality of instances of the basic waveform; and

25 logic configured to utilize the second subcarrier spacing in one or more subsequent communication operations if the time domain representation of the synchronization signal comprises a second plurality of instances of the basic waveform,

wherein the first plurality is not equal to the second plurality.

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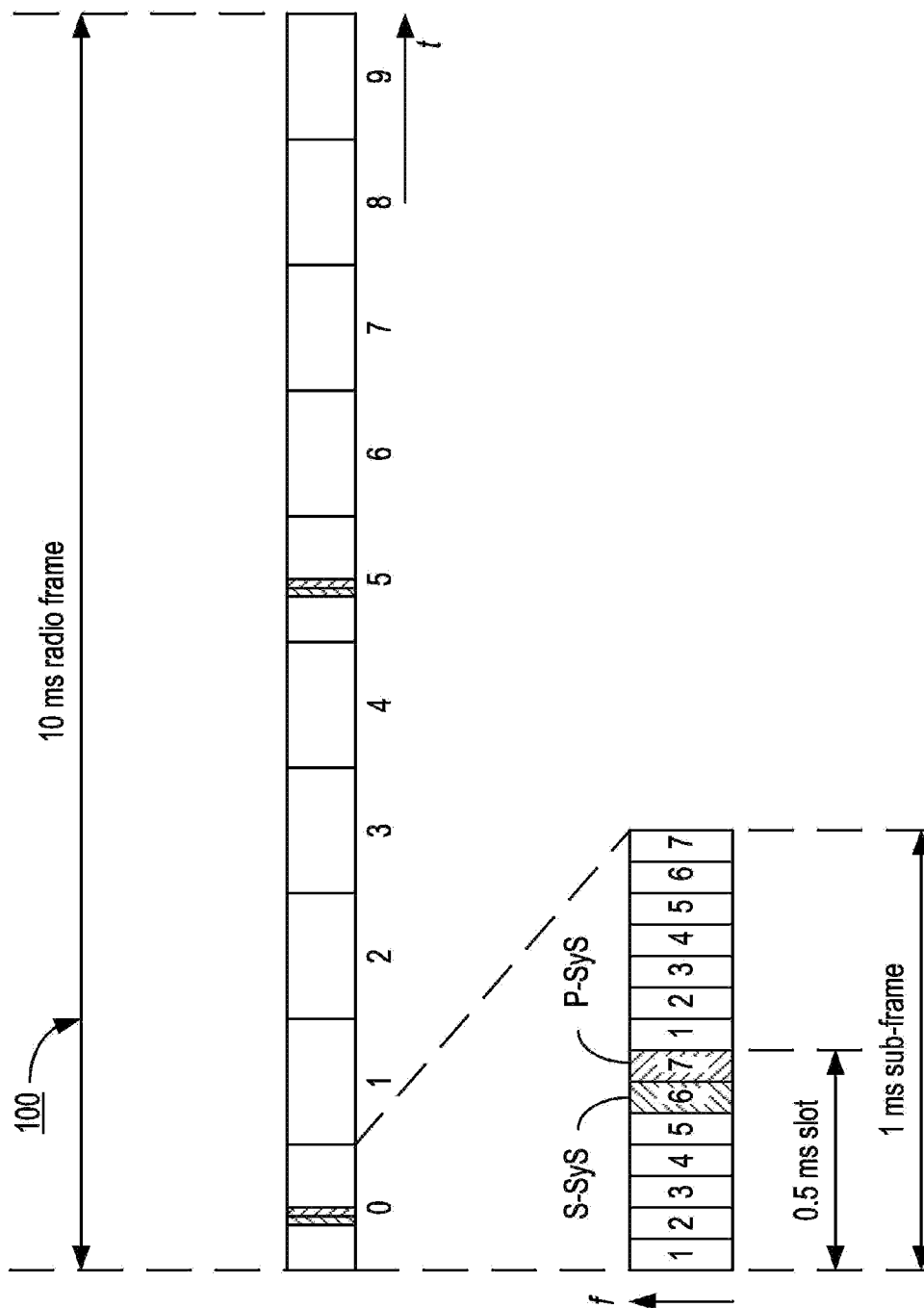


FIG.1

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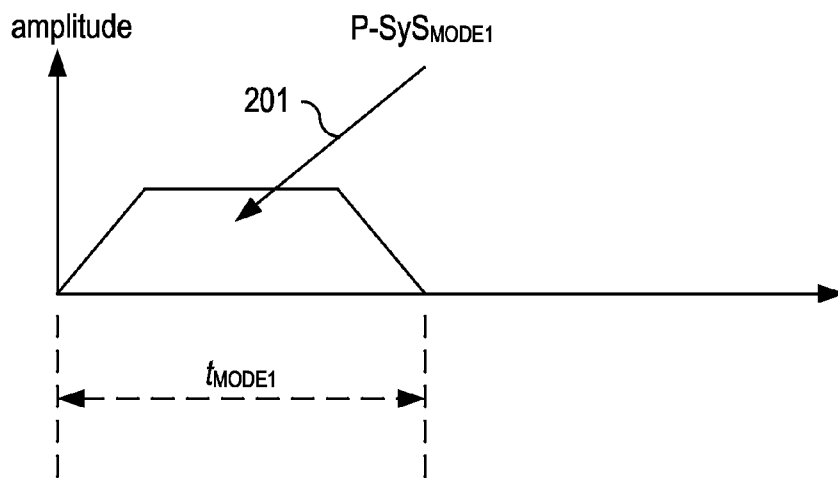


FIG. 2A

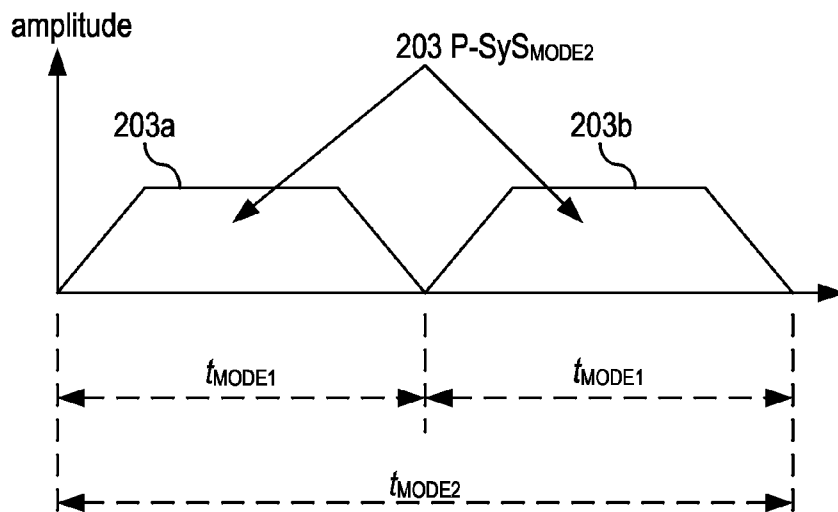


FIG. 2B

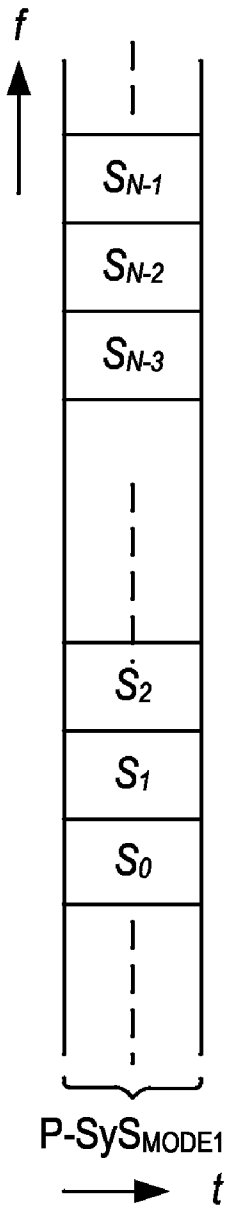


FIG. 3A

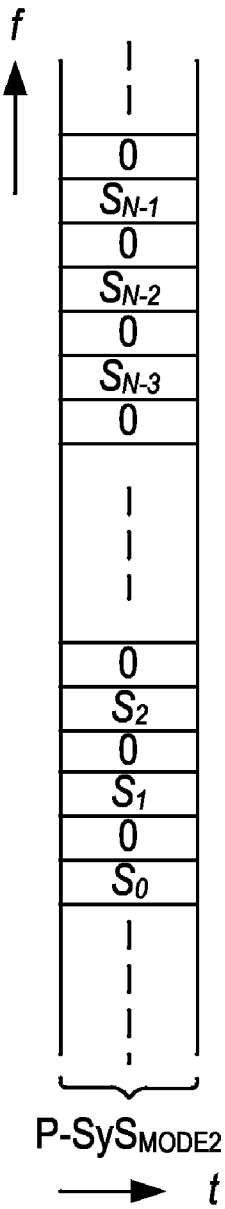
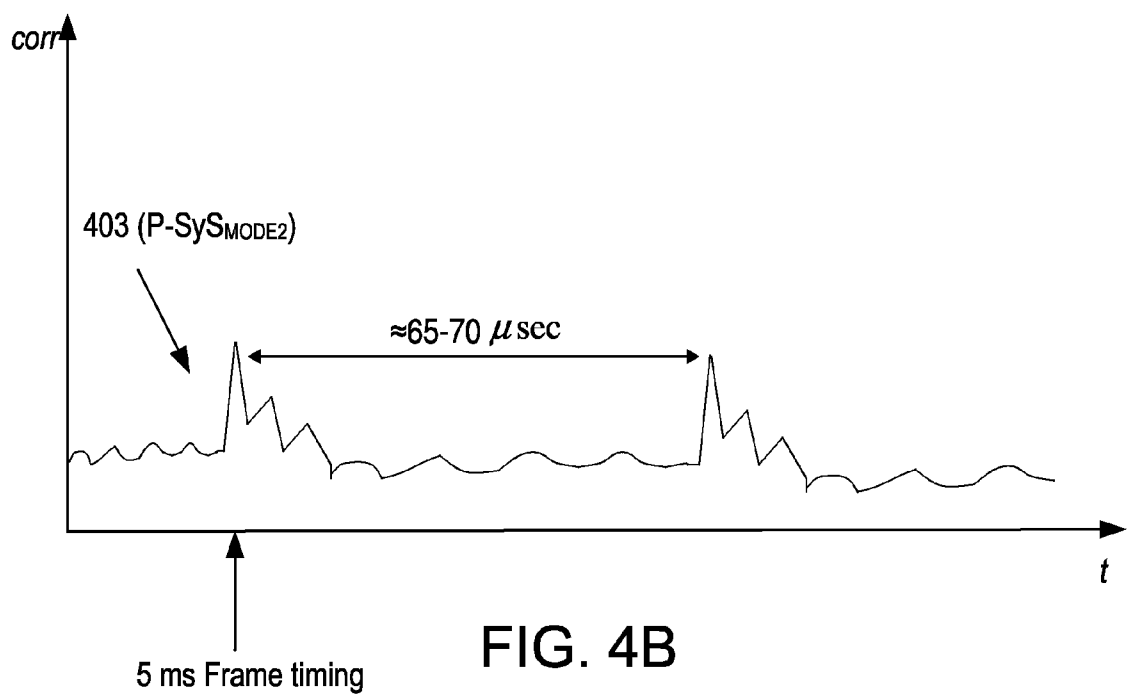
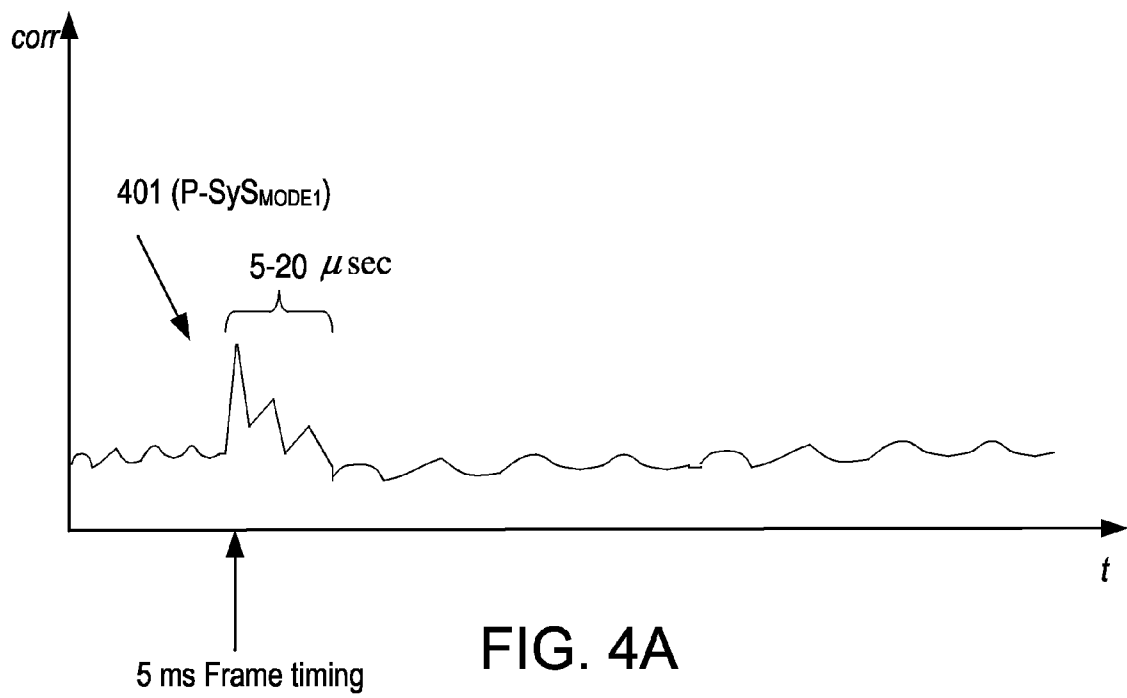


FIG. 3B



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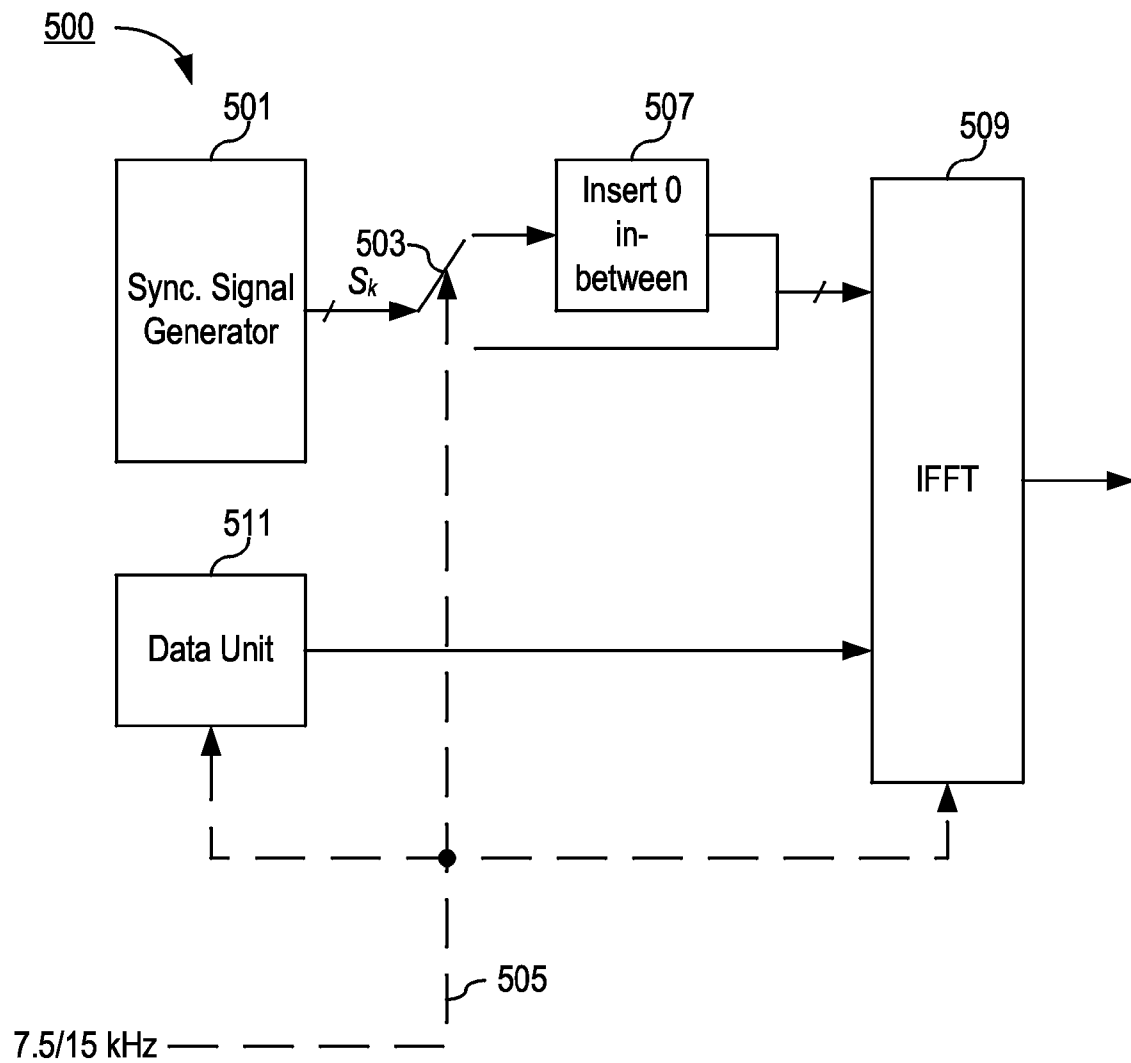


FIG. 5

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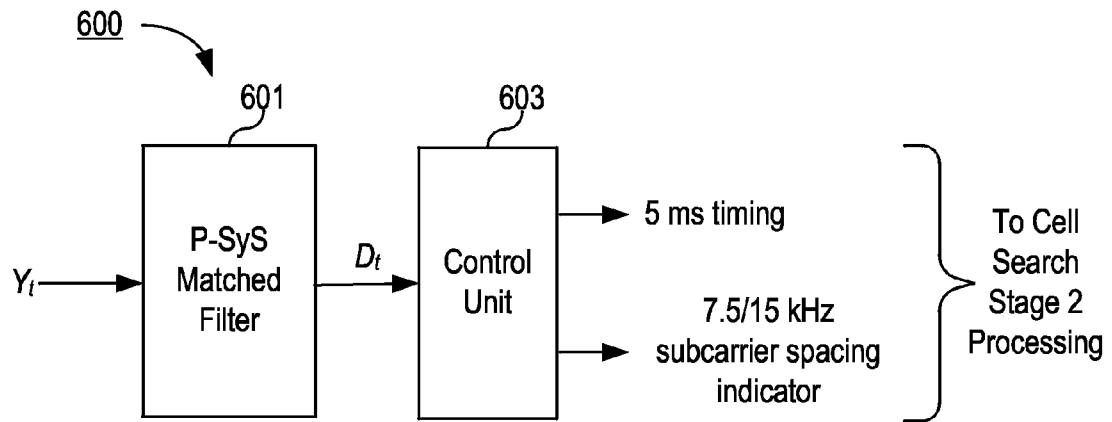


FIG. 6

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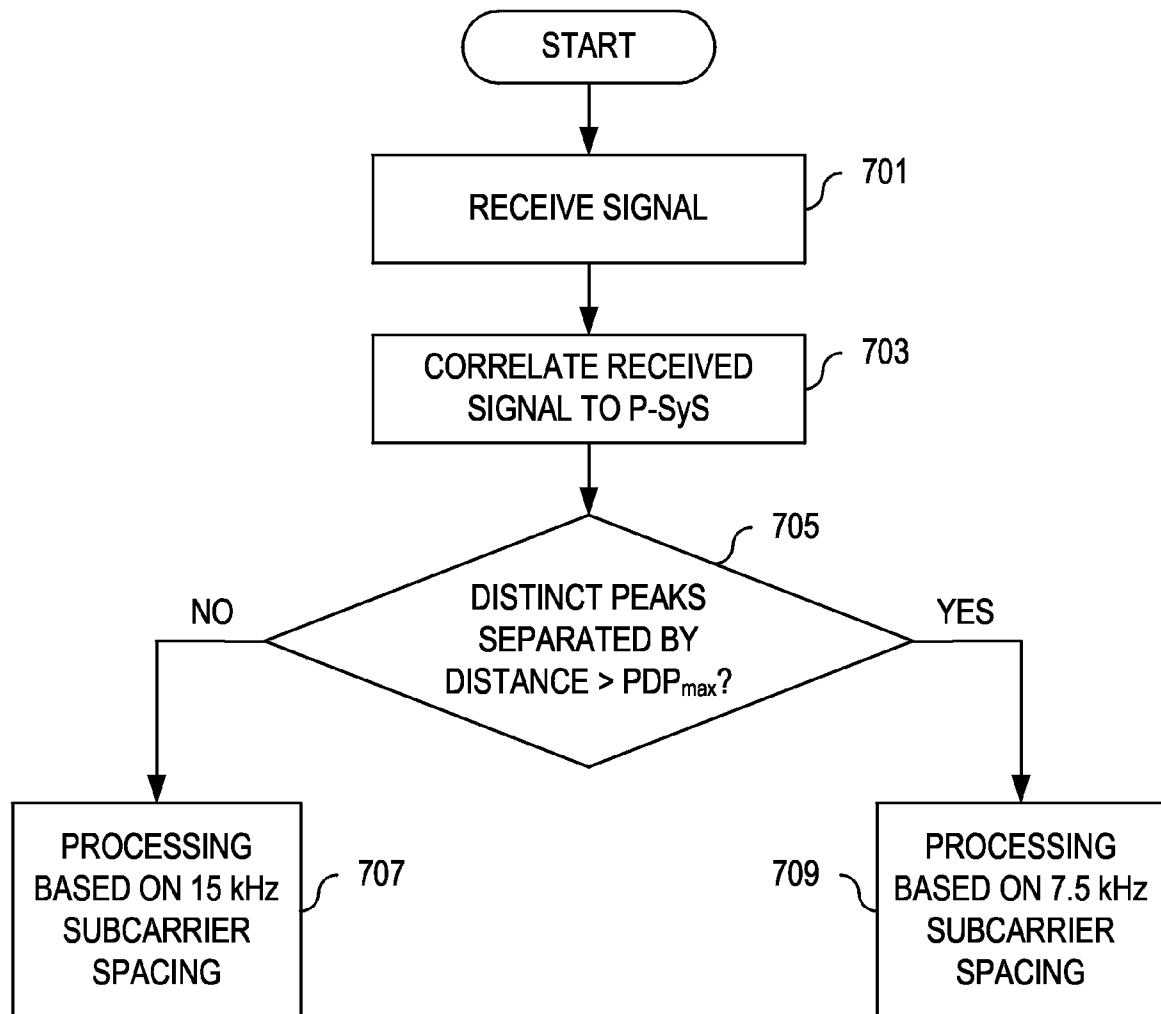


FIG. 7

# INTERNATIONAL SEARCH REPORT

International application No  
PCT/EP2008/050939

**A. CLASSIFICATION OF SUBJECT MATTER**  
INV. H04L27/26 H04L27/00

According to International Patent Classification (IPC) or to both national classification and IPC

**B. FIELDS SEARCHED**

Minimum documentation searched (classification system followed by classification symbols)  
H04L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data, INSPEC

**C. DOCUMENTS CONSIDERED TO BE RELEVANT**

Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
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A	US 2005/207482 A1 (REDFERN ARTHUR J [US]) 22 September 2005 (2005-09-22) abstract paragraph [0007] - paragraph [0009] paragraph [0033] paragraph [0036] paragraph [0042] claim 1 ----- -/--	1-22

☒ Further documents are listed in the continuation of Box C.

☒ See patent family annex.

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Date of the actual completion of the international search

14 July 2008

Date of mailing of the international search report

24/07/2008

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## INTERNATIONAL SEARCH REPORT

International application No  
PCT/EP2008/050939

C(Continuation). DOCUMENTS CONSIDERED TO BE RELEVANT

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Information on patent family members

International application No

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