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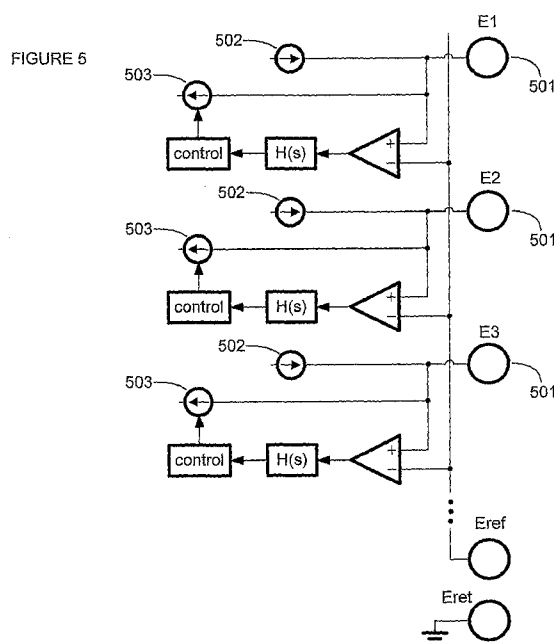
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(54) Title: ACTIVE ELECTRODE STATE CONTROL SYSTEM



(57) Abstract: A body stimulating device operatively adapted to provide electrical stimuli within a body, the device including stimulating electrodes, stimulus generator, and electrode voltage sensors, said electrode voltage sensors operatively measuring the DC/LF voltage of the electrodes, wherein if the sensors determine that the electrode voltage for an electrode is outside a predetermined range, then a compensating current is applied to that electrode, so as to reduce the voltage.

ACTIVE ELECTRODE STATE CONTROL SYSTEM

TECHNICAL FIELD

The present invention relates to systems, methods and devices for
5 controlling the states of stimulating electrodes in body stimulating devices, such
as cochlear implants and other neural stimulators.

BACKGROUND TO THE INVENTION

Electronic devices implanted within the body in order to stimulate nerve
tissue (e.g. cochlear implants) for perceptual or functional purposes generally use
10 platinum electrodes as the interface between the electronics and the body tissue.
In general terms, such electrodes are selectively driven with a current in order to
evoke a perception (for example sound) or a function (for example a limb
movement) in the user. Figure 1 illustrates this schematically. The platinum
electrodes 101 are connected to the implant via insulated wires 104 and driven by
15 the stimulating current 102, which passes through the tissue 100 and the nerve
cell 107, and returns to the implant (return current 103). At the surface of the
platinum electrodes 101, chemical reactions take place, changing the electron
current in the electronics to an ion current 105 in the tissue; charge 106 remains
on the electrode surface, causing an increase in voltage across the interface.

20 Under normal operation of the interface, these chemical reactions are
reversible and when the current direction is changed, the reactions are reversed,
leaving a neutral interface. It is usual for the stimuli to be structured as biphasic
pulses, in such a way that there is no net charge delivered to the tissue. If,
however, the current is allowed to flow in one direction for too long, toxic products
25 can escape the interface and damage or destroy the surrounding tissue.
Likewise, if the voltage across the interface is allowed to remain elevated for too
long, toxic species are irreversibly generated at the interface. To ensure, then,
that stimulation is safe, and that no toxic species escape the interface, it must be
ensured that the DC and low-frequency (LF) states of the electrodes, i.e. the
30 DC/LF interface voltages and the DC/LF interface currents, remain within certain
bounds. The usual target values are some hundreds of milli-volts, or some tens
of nano-amperes (for typical cochlear implant electrode areas of about 0.25mm²).

The FDA in the US requires that the magnitude of the current through an electrode is below 100nA measured over any 1ms period.

The use of charge-neutral pulses ensures, in principle, that the FDA requirement for the DC/LF current is met; in practice, however there will be a small error in the generated stimulation current. This requires a second measure to be taken to ensure low levels of DC/LF current at all times. This is particularly an issue when high stimulation rates and high current levels are used. Further, if the stimulation current source goes out of compliance, then significant charge errors can occur.

A number of approaches are currently employed to control the interface voltage and current.

One approach is to use DC blocking capacitors for each electrode to ensure zero DC currents through the electrodes. Figure 2 shows such an arrangement, whereby for each stimulating electrode 201, there is a DC blocking capacitor 202 disposed in the path for stimulation currents 203. A capacitor may also be disposed on the monopolar return electrode 204. In order for this approach to be effective, it is necessary to provide a capacitor with relatively high capacitance, in the nano-farad range, for each electrode. With current capacitor technology, this cannot be fabricated in an integrated circuit, and so discrete components are required to be used. It is difficult to have a capacitor per electrode when tens of electrodes are in use, as this requires tens of capacitors, which significantly increases the required space for the implant. As the number of electrodes increase in future devices, potentially to hundreds or thousands, DC blocking capacitors become even more impractical. This type of approach is known from, for example, US Patent No. 5, 324,316 to Schulman et al, US Patent No US 6,600,955 to Zierhofer et al, and US Patent No 6,219,580 to Faltys et al.

Another approach is to use periodic short-circuiting of all electrodes to ensure that the DC/LF electrode voltage does not drift out of the safe window. Figure 3 illustrates schematically such an arrangement, whereby a shorting switch 301 is provided for each electrode 201, and periodically the switches are closed. In some implementations, a series capacitor is used in the return electrode only. This approach allows for up-scaling of the number of electrodes. However, shorting all the electrodes requires the stimulation protocol to include

an inactive period when no stimulation takes place. This approach is disclosed in European Patent No. 0,241,101 to Cochlear Limited.

Another approach is to measure the differential residual voltage between electrodes during a dead period and adjust the duration or amplitude of the applied stimuli to compensate for the charge error. This approach is disclosed in
5 US Patent No. 5,674,264 to Cochlear Limited.

Using high-frequency asynchronous stimulation on many electrodes concurrently, and employing electrode arrays with a large number of electrodes, are considered by many as desirable in order to improve system performance in
10 cochlear implants, and other implant systems. When concurrent, asynchronous stimulation is used, there is no dead period available to carry out electrode shorting.

It is an object of the present invention to provide an improved arrangement for controlling the electrical states of stimulating electrodes for implanted devices.

15 SUMMARY OF THE INVENTION

In a broad form, the present invention provides a system in which the state of the stimulating electrodes is monitored in an ongoing way, and the states are actively corrected if they fall outside a predetermined window.

According to one aspect, the present invention provides a method of
20 controlling DC/LF voltages in a body stimulating electrode system, including the steps of:

(a) measuring the residual DC/LF voltages associated with each stimulating electrode; and

(b) if a residual voltage is outside a predetermined range, then applying a
25 compensating current of opposite polarity to reduce the residual voltage.

According to another aspect, the present invention provides a body stimulating device, operatively adapted to provide electrical stimuli within a body, the device including stimulating electrodes, a stimulus generator, and electrode voltage sensors, said electrode voltage sensors operatively measuring the DC/LF
30 voltage of the electrodes, wherein if the sensors determine that the electrode voltage for an electrode is outside a predetermined range, then a compensating current is applied to that electrode, so as to reduce the voltage.

Implementations of the present invention accordingly allow for stimulation without a periodic 'dead' period, to facilitate for example asynchronous stimulation on multiple electrodes. This approach also removes the requirement for a capacitor for each electrode.

5 **BRIEF DESCRIPTION OF DRAWINGS**

Illustrative implementations of the present invention will now be described with reference to the accompanying figures, in which:

Figure 1 is a conceptual drawing illustrating the passage of current using stimulating electrodes;

10 Figure 2 illustrates a prior art DC blocking capacitor approach;

Figure 3 illustrates a prior art shorting approach;

Figure 4 is a graph showing a typical relationship between DC current and DC voltage for tissue under stimulation;

15 Figure 5 is a schematic illustration of one implementation of the present invention;

Figure 5a is a schematic illustration of one implementation of a stimulation current source of Figure 5.

Figure 5b is a graph showing the experimental result of an implementation in accordance with Figure 5.

20 Figure 6 illustrates timing of currents and voltages during operation of one implementation of the present invention;

Figure 7 illustrates an implementation of a compensation current controller; and

25 Figure 8 is a schematic illustration of another implementation of the present invention.

DESCRIPTION

It will be appreciated that while the present invention will be described principally in the context of a particular implementation for a cochlear implant, it is applicable wherever electrical stimulation is to be delivered within the body, either
30 from an implanted device, or from an externally disposed device. Whilst it is most clearly applicable to neural stimulation, the present invention can be applied to any other form of electrical stimulation applied to the body.

Figure 5 illustrates one implementation of the present invention. Electrodes 501 are generally designated as E. A stimulation current from a stimulation current source 550, shown in Figure 5a, is (selectively) applied to each stimulating electrode 501. At each electrode, the electrode voltage is compared to a reference electrode E_{ref} which does not carry any current, using a suitable differential amplifier. The same reference electrode E_{ref} may be used for all the stimulating electrodes. The amplifier output passes through a filter $H(s)$ (the operation of which will be described below) and then into a control circuit. Similarly to the electrode shorting approach, the electrode DC/LF voltage for each electrode is controlled — however, instead of using a passive shorting action during an inactive time-slot, an active feed-back system using stimulating currents 502 and compensation currents 503 controls the electrode voltage. No specific time slot is required for the voltage control step.

Figure 5b shows an experimental result of the embodiment according to Figure 5 with the following parameters:

- Linear 12dB/octave roll-off low-pass filters with corner frequency of 15Hz.
- Compensation currents in fixed times of 1ms delivered (asynchronously with the stimulation currents) in the appropriate direction when the low-pass filtered electrode voltages drift outside the set compensation voltage thresholds.

Experiments were carried out using a Cochlear electrode array submerged in physiological saline. Figure 5b shows traces of the electrode voltages over a 200ms period when:

- One pair of electrodes is driven with 1mA unbalanced biphasic pulses of 10 μ s/phase and a 1ms repetition rate, having an equivalent net DC current of 10 μ A; the stimulation is carried out for 100ms.
- The compensation current magnitude is set to 20 μ A, while the compensation voltage threshold is set to 100mV.

As shown, the compensation feed-back loop keeps the low-frequency voltages on all electrodes within a window about 400mV.

Figure 6 is a timing diagram illustrating the process whereby three stimulating electrodes (E1, E2 and E3) are simultaneously being driven with

typical bi-phasic current pulses, but the pulse on E1 has a DC component (non-zero net charge); this causes the voltage on the electrode V (E1) to build up. This build up is detected by the control system and a compensation pulse on E1 is applied to bring back the voltage. During stimulation large (up to tens of volts) voltages appear on the electrodes; such voltages are a necessary part of the stimulation and must be removed before the electrode voltage is passed to the feedback control system. To this end, an appropriate filter ($H(s)$) is used as seen in figure 5. The implementation shown in figure 5 uses local controls for each electrode.

The filter transfer function, $H(s)$, needs to be chosen such that stimulation artefacts are suppressed while voltage drift is detected before it can cause harm. The simplest approach is to use a linear, time-invariant low-pass filter. While the stimulation current pulse is typically in the $10\mu s$ - $100\mu s$ range for cochlear implants, the duration of the electrode voltage response to the stimulation current is in the $1ms$ range. Non-zero voltages should not persist on the interface for more than about $100ms$; as such, the filter cut-off frequency, f_{LP} should lie between $1kHz$ and $10Hz$. A second or higher order (NLP) filter is therefore preferred with a (programmable) cut-off frequency of about $100Hz$.

It will be understood that these values are appropriate for a cochlear implant of standard design – other devices or differing designs may require different values as will be apparent to those skilled in the art. It will also be understood that there are many different implementations and designs for filters which can be selected as appropriate for the device in question.

When performing simultaneous stimulation on many electrodes, it is in practice, due to mismatch in the current sources, impossible to ensure an exact net zero current being delivered to the tissue. The use of a common return electrode provides a current path for the residual current (see Figure 5). The return electrode can also be used as a return current path for deliberate non-zero net stimulation currents (i.e. monopolar stimulations in the case of only one stimulating electrode) to enhance current flow across the nerves. For cochlear implants, the preferred form of the return electrode is a relatively large extra-cochlear electrode (typically in the range of $1mm^2$ to $10mm^2$). Another function of the return electrode is to define the DC potential between the implant and the

tissue (the implant ground); without such a measurement, this value would be undefined, as all other connections between the implant and tissue are high-impedance. Such connections create compliance issues on sinking or sourcing stimulation currents (see Figure 4 and previous discussion in relation to Figs 5 and 6). Note that even though significant currents may pass through the return electrode, it does not need a compensation current circuit: if all other stimulating electrodes are monitored, having near to net zero DC currents, the DC current in the return electrode must, by virtue of Kirchhof's current law, also be near zero. Observing the DC voltage on the return electrode, however will provide information about failures in the active compensation, and for this reason, the voltage should be measured as it is on the stimulating electrodes according to the present invention.

In order to measure the voltage across an electrode, to determine whether a compensation current needs to be applied to that electrode, the tissue potential must be known. To this end, it is required to use a reference electrode which is in contact with the tissue, but never has any current flowing in it; this electrode will give a reliable measurement of the tissue potential. All the electrodes should be formed from the same material such that the half-cell potential across each electrode is the same. This electrode should preferably be large in surface area (similar to the extra-cochlear return electrode), such that its impedance is relatively small.

A convenient location for a large reference electrode in cochlear implants is outside the cochlea; i.e. an exiting extra-cochlear electrode which does not participate in the stimulations could be used or, alternatively, a dedicated electrode could be added to the system. Another suitable choice for the reference electrode is a parallel combination of unused electrodes in the intra-cochlear electrode array. If the reference electrode is placed outside the cochlea, this may create an offset in the safe stimulation electrode DC voltage window (see below), due to typical natural DC voltage differences between the intra-cochlear electrodes and extra-cochlear electrodes of some tens of milli-volts.

The magnitude of the compensation currents should be chosen such that they can compensate for the largest expected error in the stimulation current. For example: if the maximum stimulation current is $I_M=2\text{mA}$, the stimulation phase

length is $T_P=50\mu s$, the minimum stimulation period on a particular electrode is $T_{SP}=500\mu s$, and the maximum stimulation current error is $E=2\%$ (typical numbers for cochlear implants), the worst-case DC current induced by the stimulation is

$$I_{DC} = I_M * T_P * E / T_{SP} = 4\mu A$$

- 5 Note that this magnitude is well below normal stimulation thresholds in cochlear implants. It is highly desirable that the compensation currents be kept below the level of user perception.

10 A simple way to control the delivery of the compensation current, I_{comp} is by means of the circuit shown in Figure 7. The output from the filter, V_{Filter} is compared with a safe window (say $V_{LO} = -200mV$ and $V_{HI} = 220mV$) and whenever the voltage is outside this range, a constant current is delivered to the electrode to move the voltage back. The magnitude of this current, I_{CC} should be chosen conservatively larger than the largest expected stimulation DC current; for the example above, maybe $I_{CC} = 5\mu A$.

- 15 It is noted that, because of the non-linearities of the electrode-tissue interface, ensuring zero DC voltage and zero DC current cannot be achieved at the same time. Figure 4 shows experimental data from a test where a pair of platinum electrodes was driven by a cochlear implant; the anodic phase length was varied, causing varying DC currents to be delivered during stimulation, while
- 20 DC voltages and DC currents are being measured. The current in nano-amperes appears on the y axis, while voltage in milli-volts is shown on the x axis. It can be seen that zero voltage and zero current does not appear at the same time. At high stimulation rates (towards 2kHz) and at high currents (towards 2mA), the DC voltage corresponding to zero DC current increases in magnitude.

- 25 The electrode tissue interface has a complex, non-linear impedance; further, this impedance may vary significantly over time, from implantee to implantee, and from electrode to electrode. For this reason, it is difficult to *a priori* determine the optimum parameter set for the control system. In any practical system, it is critical that in-vivo data is acquired for system verification. Another
- 30 consequence of these unknowns is that control loop may become unstable for particular electrode impedances. Thus, it is important that all key parameters are programmable. That is, it is recommended that

- Filter cut-off frequency (f_{LP}) should be programmable in the range 10Hz-1kHz.
- Filter order (N_{LP}) should be programmable as 1st, 2nd and 3rd order.
- 5 • Compensation current magnitudes (I_{CC}) should be programmable in the range 100nA-10uA.
- Compensation voltage thresholds (V_{LO} and V_{HI}) should be programmable in the range -400mV to 400mV.

Most implantable neural stimulators now have significant digital signal
10 processing capability. Thus, instead of using local hardware for each electrode to perform the compensation, data can be passed to a central DSP or microprocessor which acts on the data. This has the particular advantage that the compensation algorithm can be reprogrammed when more in-vivo data is available. Figure 8 shows an example of this, with the most central architecture,
15 where the raw, digitised (using a suitable analogue-to-digital converter (ADC)) electrode voltage measurements are passed to the DSP, which performs filtering ($H(z)$) and implements the control algorithm. In its simplest form, the control algorithm would be the same as described above: each electrode filter output is compared with compensation voltage thresholds and, if the filter output lies
20 outside the safe range, the need for and direction of a compensation current for that electrode is flagged. The compensation currents here are applied not via dedicated current sources but by perturbing the amplitude settings (received from the primary part of the system that calculates the stimulation data) on the stimulation current sources 801. In this embodiment the DC/LF voltage is
25 extracted using filter $H(z)$ and the charge error is determined in reference to the reference electrode 802 instead of synchronous sampling. Accordingly the amplitude of the asynchronous stimulation current may be adjusted without requiring a dead period. Other possible variations which could be implemented include:

- 30 • Performing electrode voltage sensing and filtering at the electrode sites, passing the DC voltage to the DSP. This variation relaxes the requirements to the ADC.

- Performing electrode voltage sensing, filtering and thresholding at the electrode sites, passing "out of range" information to the DSP. This variation eliminates the need for the ADC.

5 • Using dedicated compensation current sources for each electrode controlled by the DSP. This variation relaxes the requirements to the stimulating current sources.

10 • Using multiples (or a continuum) levels of compensation current depending on the electrode voltage error (using larger compensation current at higher voltage error). This variation enables a more advanced feed-back control system to be implemented.

- Using non-linear filters before a subsequent linear filtering such that the large electrode voltages present during stimulation are largely ignored by the control system. This variation would seek to keep the median, rather than the average, electrode voltage within the given bounds.

15 While the present invention has been described with respect to specific embodiments, it will be appreciated that various modifications and changes could be made without departing from the scope of the invention.

CLAIMS

1. A method of controlling DC/LF voltage in a body stimulating electrode system, including the steps of:

(a) measuring the residual DC/LF voltage associated with each stimulating
5 electrode;

(b) if a residual voltage is outside a predetermined range, then applying a compensating current of opposite polarity to reduce the residual voltage.

2. The method according to claim 1, wherein the residual voltage is measured relative to a reference electrode operatively in electrical contact with
10 body tissue, which does not carry a current.

3. The method according to claim 2, wherein the reference electrode has a relatively large surface area.

4. The method according to any one of the preceding claims, wherein the measured voltage is filtered so as to suppress stimulation artefacts.

15 5. The method according to any one of the preceding claims, wherein a feedback control system is provided associated with each stimulating electrode.

6. The method according to any one of claims 1 to 4, wherein a central control arrangement is provided to control the compensating currents.

7. The method according to claim 6, wherein the central control arrangement
20 also controls the stimulating currents.

8. A body stimulating device, operatively adapted to provide electrical stimuli within a body, the device including stimulating electrodes, stimulus generator, and electrode voltage sensors, said electrode voltage sensors operatively measuring the DC/LF voltage of the electrodes, wherein if the sensors determine that the
25 electrode voltage for an electrode is outside a predetermined range, then a compensating current is applied to that electrode, so as to reduce the voltage.

9. The device according to claim 8, wherein each voltage sensor operates between a respective stimulating electrode and a reference electrode, operatively in electrical contact with body tissue, which does not carry a current.
10. The device according to claim 9, wherein the reference electrode has a
5 relatively large surface area.
11. The device according to any one of claims 8 to 10, wherein the electrode voltage measured by each sensor is filtered so as to suppress stimulation artefacts.
12. The device according to any one of claims 8 to 11, wherein a feedback
10 control system is provided associated with each stimulating electrode.
13. The device according to any one of claims 8 to 11, wherein a central processor is provided to control the compensating currents.
14. The device according to claim 13, wherein the central processor is or controls the stimulus generator.

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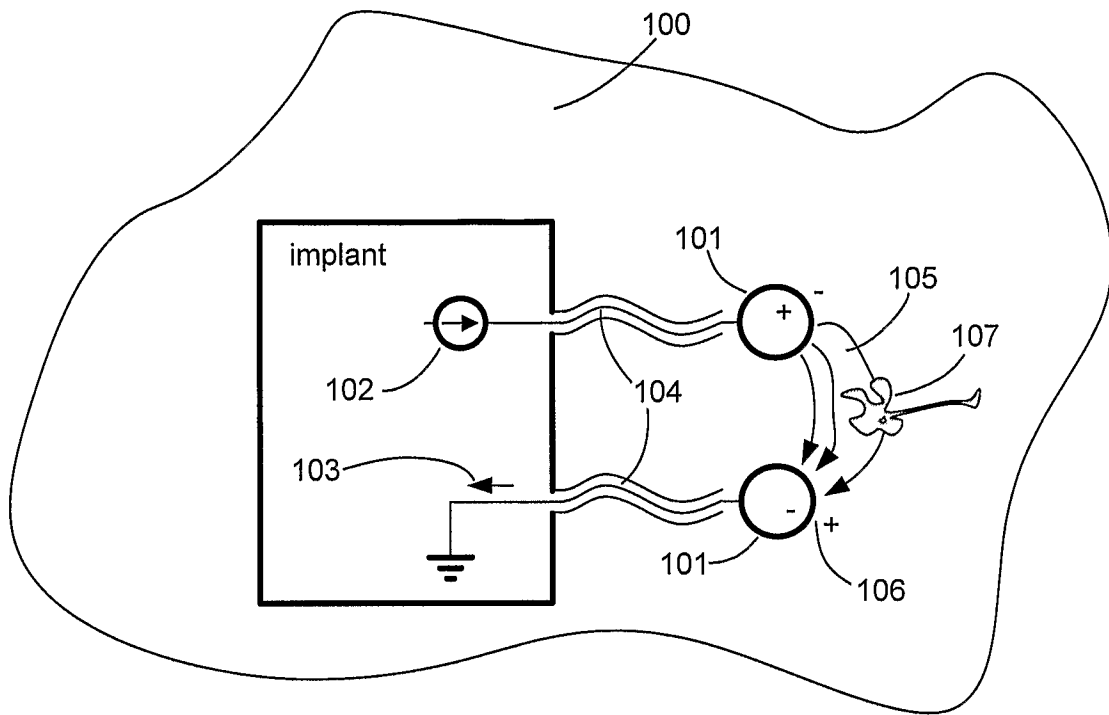


FIGURE 1

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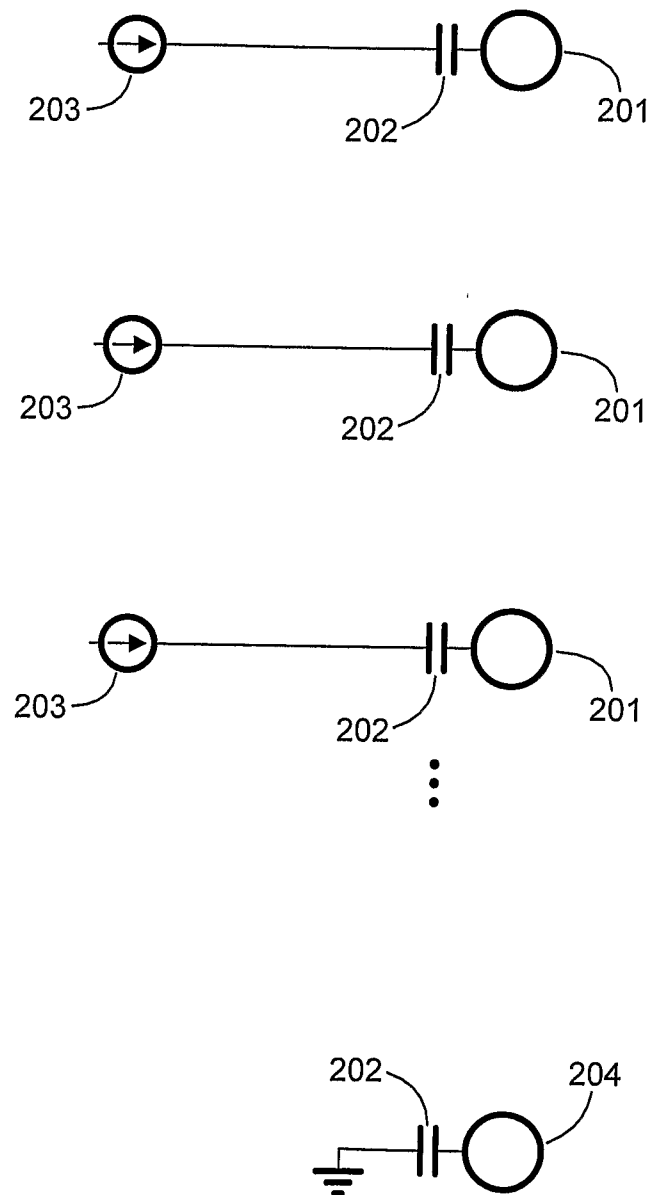


FIGURE 2

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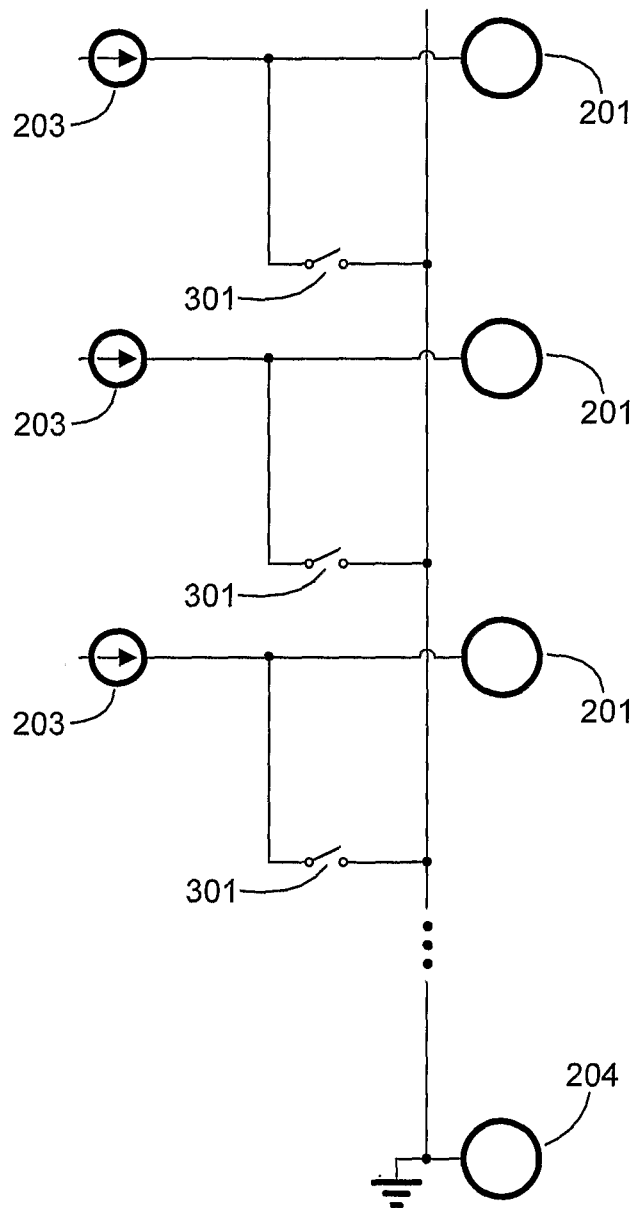


FIGURE 3

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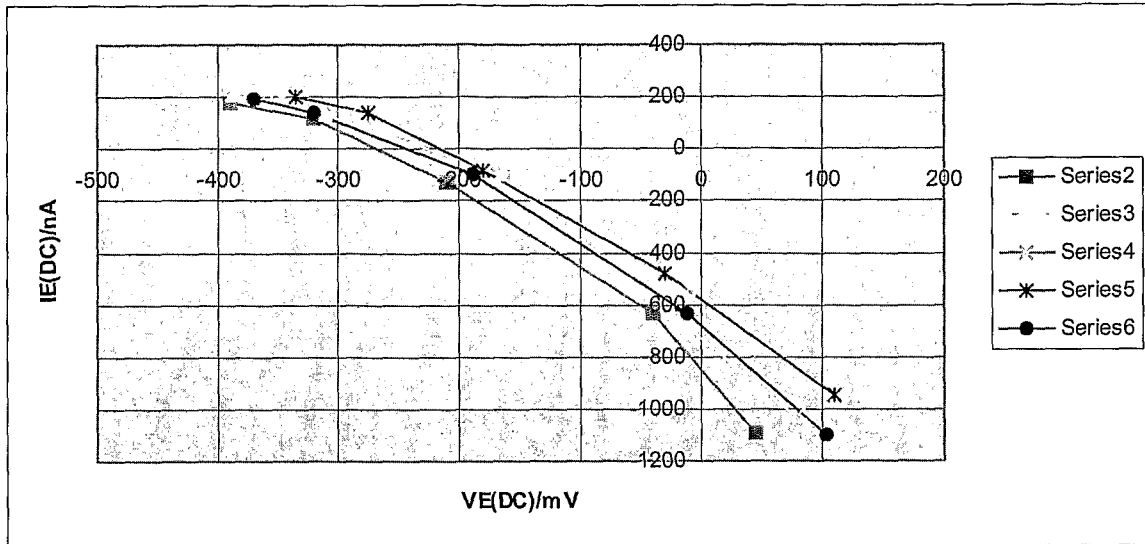


FIGURE 4

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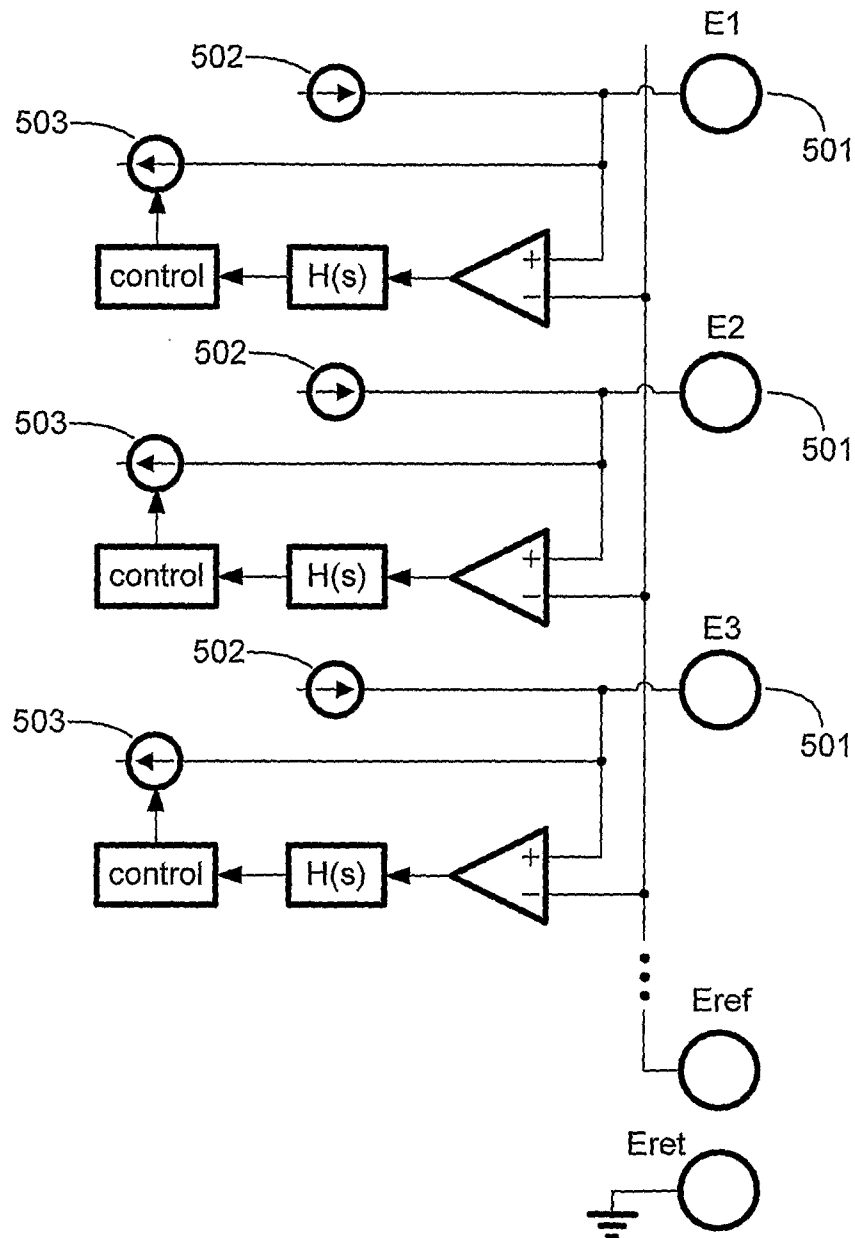


FIGURE 5

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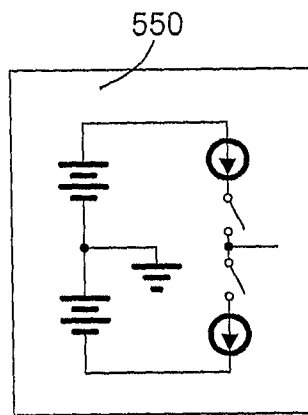


FIGURE 5a

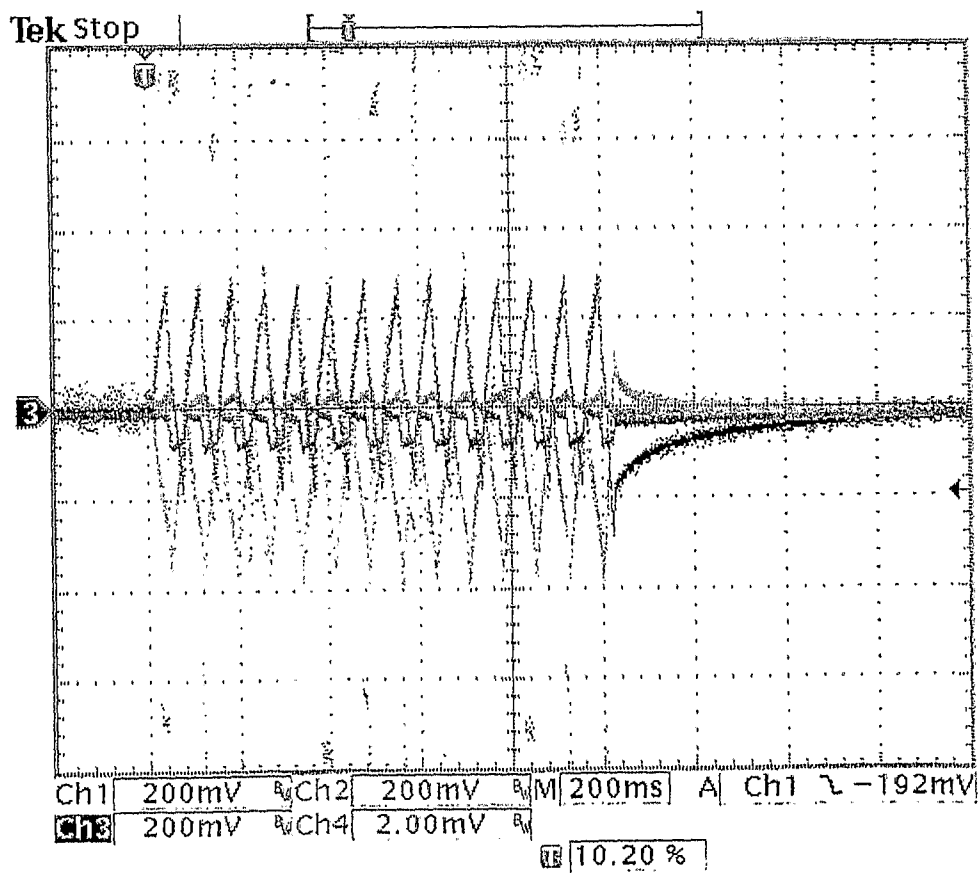


FIGURE 5b

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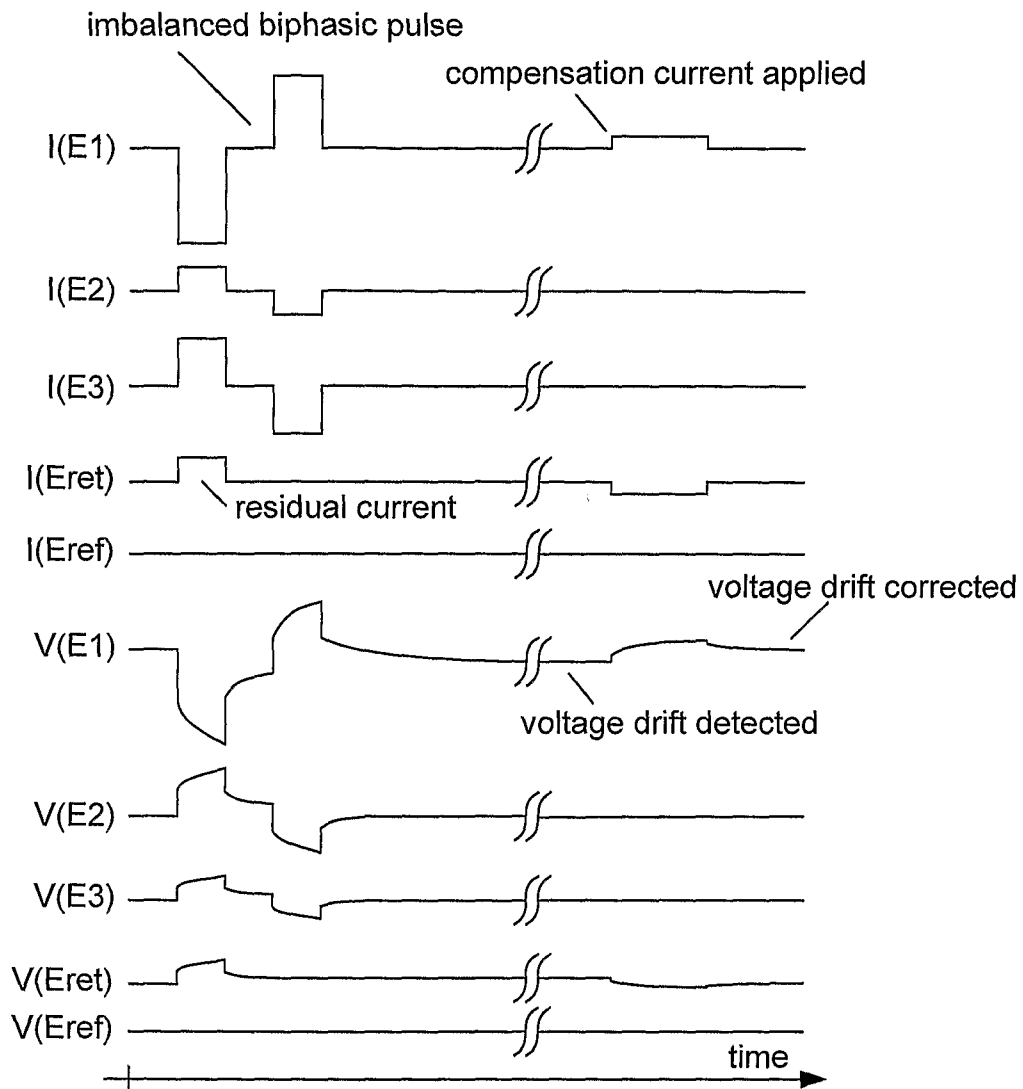


FIGURE 6

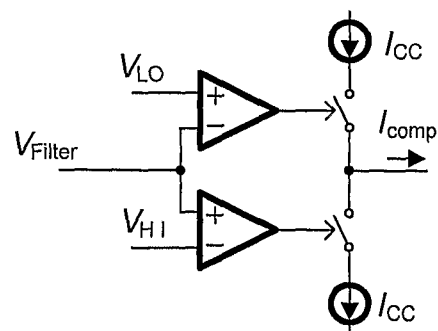


FIGURE 7

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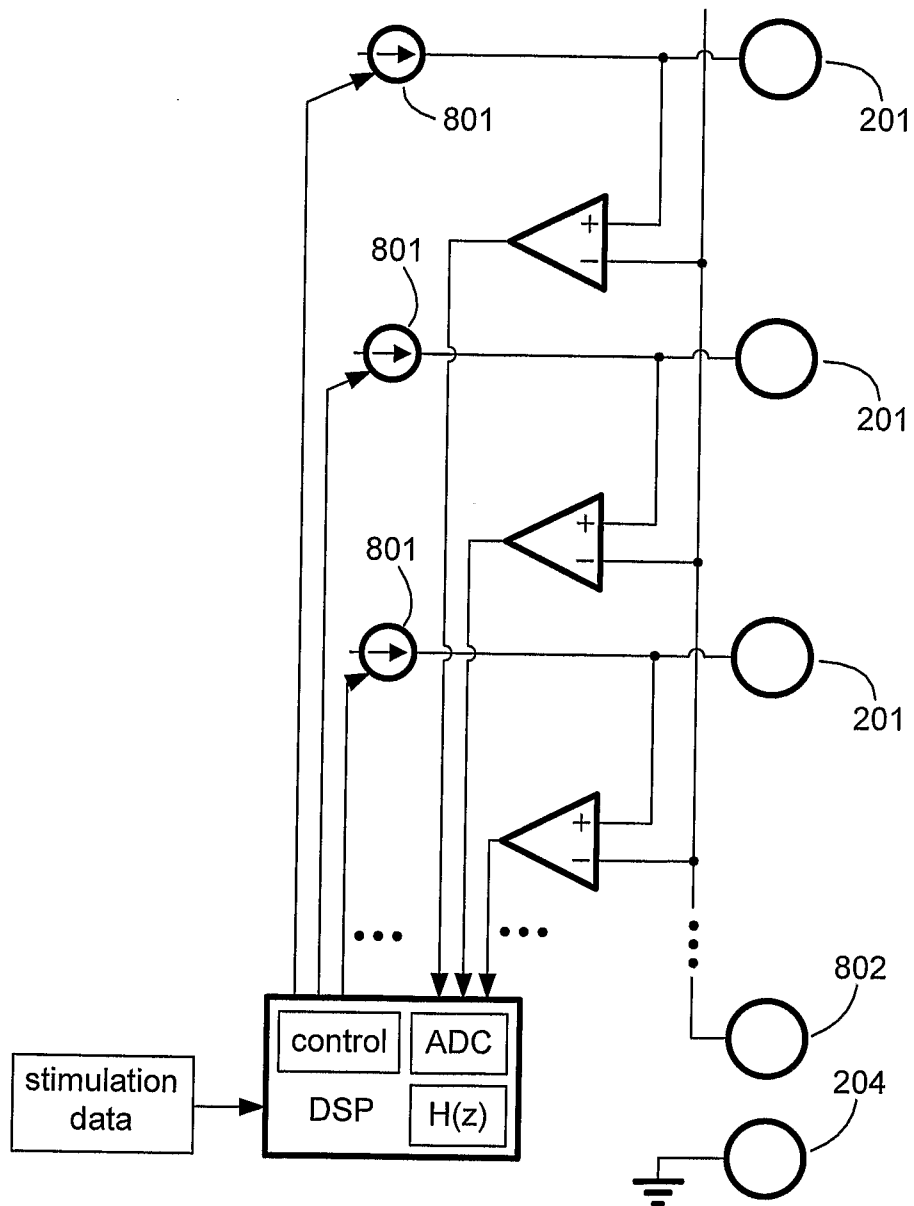


FIGURE 8

INTERNATIONAL SEARCH REPORT

International application No.
PCT/AU2008/001506

A. CLASSIFICATION OF SUBJECT MATTER Int. Cl. <i>A61F 11/04</i> (2006.01) <i>A61N 1/08</i> (2006.01) According to International Patent Classification (IPC) or to both national classification and IPC					
B. FIELDS SEARCHED Minimum documentation searched (classification system followed by classification symbols) Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched Electronic data base consulted during the international search (name of data base and, where practicable, search terms used) WPI and IPC marks A61F and A61N and keywords: electrode and residual and voltage and compensate and similar terms.					
C. DOCUMENTS CONSIDERED TO BE RELEVANT					
Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.			
P,X	US 2008/0015641 A1 (ARMSTRONG et al.) 17 January 2008 Paragraphs 25 to 32	1 – 14			
Y	DE 102004059973 A1 (IMI INTELLIGENT MEDICAL IMPLANTS) 22 June 2006 Paragraphs 17 to 22	1 – 14			
Y	WO 1997/021324 A1 (COCHLEAR LIMITED) 12 June 1997 Page 10 line 36 to page 11 line 30	1 – 14			
☒ Further documents are listed in the continuation of Box C ☒ See patent family annex					
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Date of the actual completion of the international search 11 November 2008		Date of mailing of the international search report 25 NOV 2008			
Name and mailing address of the ISA/AU AUSTRALIAN PATENT OFFICE PO BOX 200, WODEN ACT 2606, AUSTRALIA E-mail address: pct@ipaaustralia.gov.au Facsimile No. +61 2 6283 7999		Authorized officer DAVID MELHUSH AUSTRALIAN PATENT OFFICE (ISO 9001 Quality Certified Service) Telephone No : +61 2 6283 2426			

INTERNATIONAL SEARCH REPORT

International application No.

PCT/AU2008/001506

C (Continuation). DOCUMENTS CONSIDERED TO BE RELEVANT		
Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
Y	EP 1384434 A1 (HARBINGER MEDICAL, INC.) 28 January 2004 Paragraph 3	4, 11
A	US 5603725 A (SCHALDACH) 18 February 1997 Whole document	1, 8
A	US 2004/0267344 A1 (STETT et al.) 30 December 2004 Whole document	1, 8

INTERNATIONAL SEARCH REPORT

Information on patent family members

International application No.

PCT/AU2008/001506

This Annex lists the known "A" publication level patent family members relating to the patent documents cited in the above-mentioned international search report. The Australian Patent Office is in no way liable for these particulars which are merely given for the purpose of information.

Patent Document Cited in Search Report		Patent Family Member	
US	2008015641	NONE	
DE	102004059973	AU 2005315861 WO 2006063743	CA 2590208 EP 1827591
WO	9721324	AU 39755/95	EP 0867102 US 5674264
EP	1384434	JP 2004249080	US 7016731 US 2004054383
US	5603725	DE 4231602	EP 0660736 WO 9406512
US	2004267344	DE 10151650 WO 03033069	EP 1448264 US 7272447
Due to data integration issues this family listing may not include 10 digit Australian applications filed since May 2001.			
END OF ANNEX			