



US007990219B2

(12) **United States Patent**
Gleason et al.

(10) **Patent No.:** **US 7,990,219 B2**
(45) **Date of Patent:** **Aug. 2, 2011**

(54) **OUTPUT COMPENSATED VOLTAGE REGULATOR, AN IC INCLUDING THE SAME AND A METHOD OF PROVIDING A REGULATED VOLTAGE**

(75) Inventors: **Jeffrey A. Gleason**, Burnsville, MN (US); **David W. Kelly**, Eagan, MN (US); **Paul Mazur**, Cottage Grove, MN (US)

(73) Assignee: **Agere Systems Inc.**, Allentown, PA (US)

(*) Notice: Subject to any disclaimer, the term of this patent is extended or adjusted under 35 U.S.C. 154(b) by 498 days.

(21) Appl. No.: **12/250,427**

(22) Filed: **Oct. 13, 2008**

(65) **Prior Publication Data**

US 2010/0090667 A1 Apr. 15, 2010

(51) **Int. Cl.**
H92M 3/04 (2006.01)

(52) **U.S. Cl.** **330/258; 330/253**

(58) **Field of Classification Search** **330/149, 330/145, 250-260, 277; 327/112-113, 359, 327/361, 346; 323/235, 282**

See application file for complete search history.

(56) **References Cited**

U.S. PATENT DOCUMENTS

3,872,390 A	3/1975	Nash	
4,048,573 A *	9/1977	Evans et al.	330/2
5,115,205 A *	5/1992	Holmes, Jr.	330/10
6,163,218 A *	12/2000	Hashimoto et al.	330/257
6,281,752 B1 *	8/2001	Stochino	330/255
6,724,257 B2 *	4/2004	Wrathall	330/253
7,161,431 B2 *	1/2007	Kronmueller	330/255

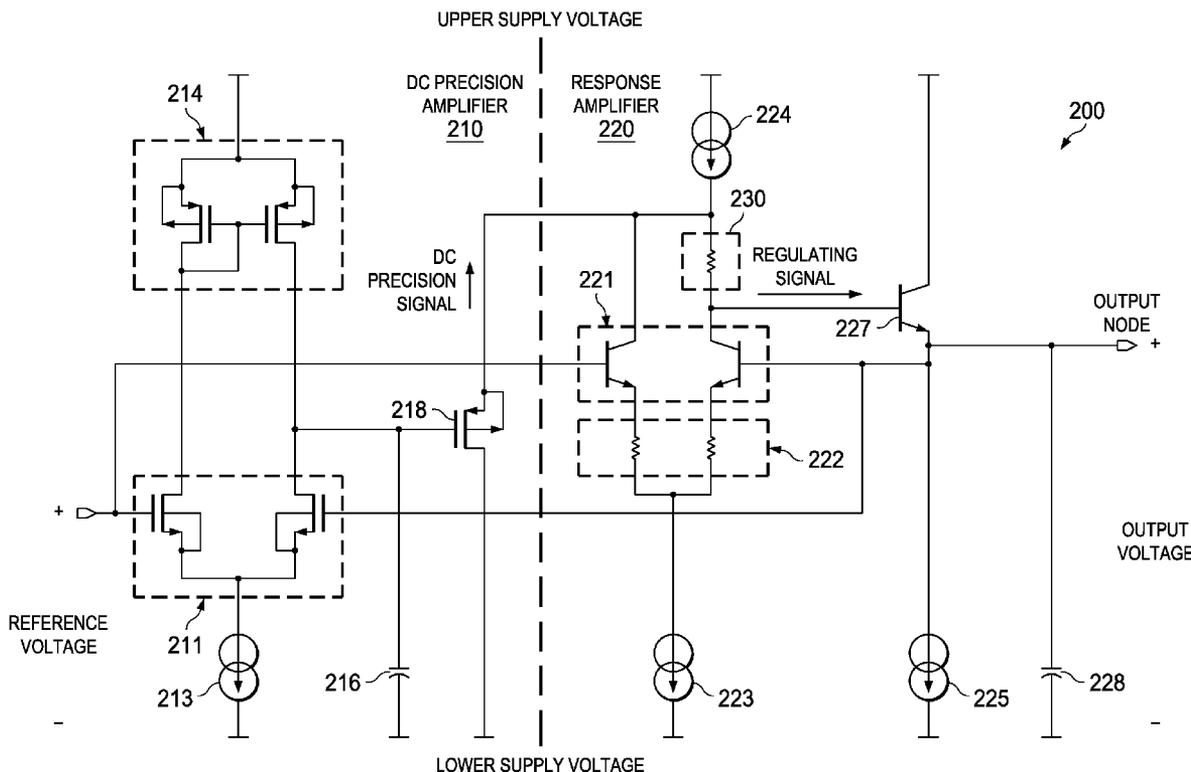
* cited by examiner

Primary Examiner — Rajnikant B Patel

(57) **ABSTRACT**

A voltage regulator, a method of regulating voltage and an IC including a voltage regulator. In one embodiment, the voltage regulator includes: (1) a DC precision amplifier configured to generate a DC precision signal based on a reference voltage and a regulated output of the voltage regulator and (2) a response amplifier, coupled in parallel with the DC precision amplifier, configured to generate an error signal based on the reference voltage and the regulated output, the response amplifier further configured to generate the regulated output based on a regulating signal comprised of the error signal and the DC precision signal. The DC precision amplifier may be a CMOS amplifier and the response amplifier may be a NPN amplifier.

24 Claims, 2 Drawing Sheets



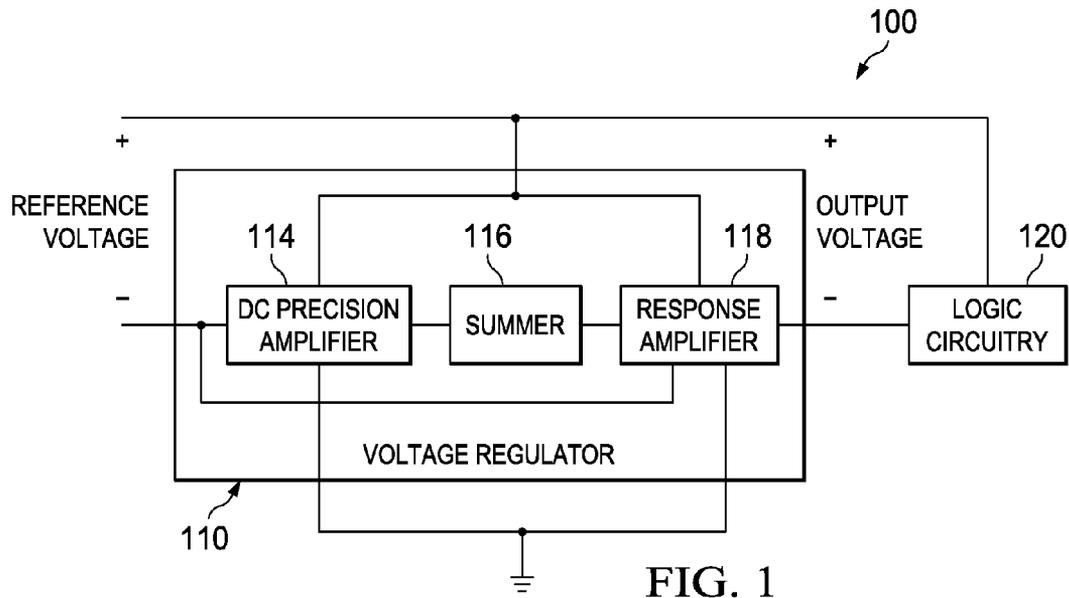


FIG. 1

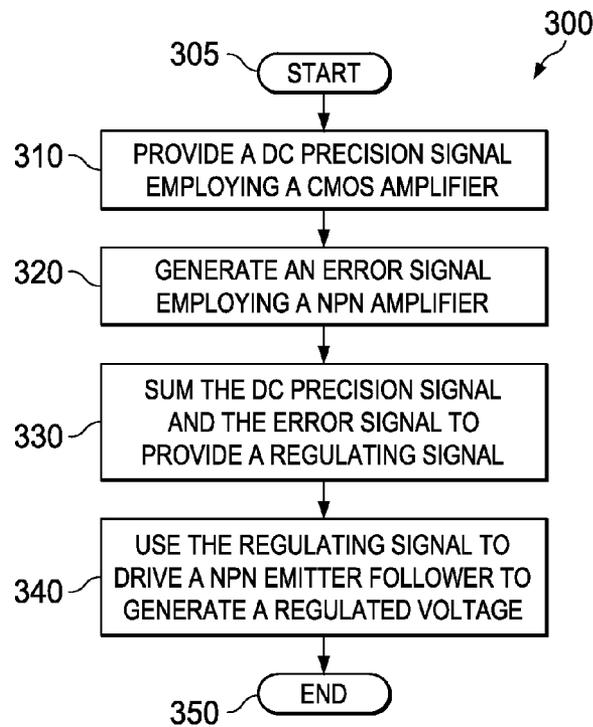
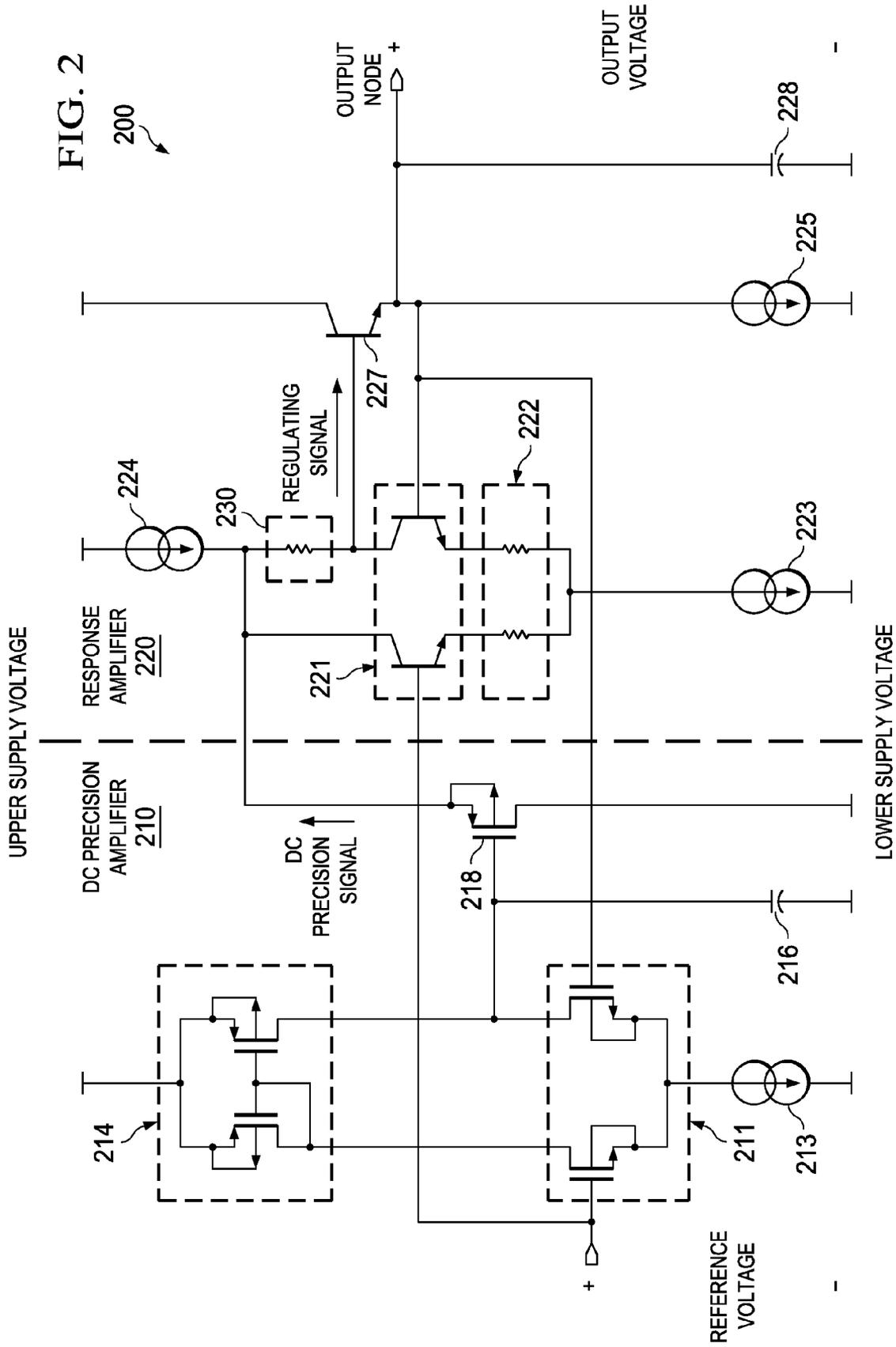


FIG. 3



1

**OUTPUT COMPENSATED VOLTAGE
REGULATOR, AN IC INCLUDING THE SAME
AND A METHOD OF PROVIDING A
REGULATED VOLTAGE**

TECHNICAL FIELD

The invention is directed, in general, to voltage regulators and, more specifically, to a voltage regulator that provides DC precision and a quick response to load changes.

BACKGROUND

Voltage regulators are often used in electrical circuits to provide a constant voltage for logic circuitry. This can prove to be a difficult task due to the influence of the load on the voltage regulators. For example, voltage regulators may be used in a pre-amplifier to provide a regulated voltage for write circuitry of a hard disk drive. As the data of the write circuitry transitions from one state to another, the output of the voltage regulator may dip. The output voltage of the regulator, therefore, can become dependent on the load current. As such, a voltage regulator that does not suffer from pattern dependency would be desirable in the art.

SUMMARY

To address the above-discussed deficiencies of the prior art, one aspect of the invention provides a voltage regulator. In one embodiment, the voltage regulator includes (1) a DC precision amplifier configured to generate a DC precision signal based on a reference voltage and a regulated output of the voltage regulator and (2) a response amplifier, coupled in parallel with the DC precision amplifier, configured to generate an error signal based on the reference voltage and the regulated output, the response amplifier further configured to generate the regulated output based on a regulating signal comprised of the error signal and the DC precision signal.

In yet another embodiment, the voltage regulator includes: (1) a CMOS amplifier configured to receive a reference voltage and employ a source-follower transistor to generate a DC precision signal based on the reference voltage and a regulated output of the voltage regulator, (2) a NPN amplifier configured to receive the DC precision signal, the reference voltage and the regulated output, the NPN amplifier employing a NPN emitter-follower transistor to generate the regulated output based on a regulating signal and (3) a summer configured to generate the regulating signal based on the DC precision signal and an error signal generated by the NPN amplifier.

In still another aspect, the invention provides an IC. In one embodiment, the IC includes (1) logic circuitry and (2) a voltage regulator configured to provide a regulated output for the logic circuitry. The voltage regulator includes: (2A) a DC precision amplifier configured to generate a DC precision signal based on a reference voltage and the regulated output of the voltage regulator and (2B) a response amplifier, coupled in parallel with the DC precision amplifier, configured to generate an error signal based on the reference voltage and the regulated output, the response amplifier further configured to generate the regulated output based on a regulating signal comprised of the error signal and the DC precision signal.

In yet a different aspect, the invention provides a method of providing a regulated voltage. In one embodiment, the method includes: (1) providing a DC precision signal employing a CMOS amplifier, (2) generating an error signal

2

employing a NPN amplifier, (3) summing the DC precision signal and the error signal to provide a regulating signal and (4) using the regulating signal to drive a NPN emitter-follower to generate a regulated voltage.

The foregoing has outlined certain aspects and embodiments of the invention so that those skilled in the pertinent art may better understand the detailed description of the invention that follows. Additional aspects and embodiments will be described hereinafter that form the subject of the claims of the invention. Those skilled in the pertinent art should appreciate that they can readily use the disclosed aspects and embodiments as a basis for designing or modifying other structures for carrying out the same purposes of the invention. Those skilled in the pertinent art should also realize that such equivalent constructions do not depart from the scope of the invention.

BRIEF DESCRIPTION OF THE DRAWINGS

For a more complete understanding of the invention, reference is now made to the following descriptions taken in conjunction with the accompanying drawings, in which:

FIG. 1 is a block diagram of an embodiment of an integrated circuit having a voltage regulator constructed according to the principles of the present invention;

FIG. 2 is schematic diagram of an embodiment of a voltage regulator constructed according to the principles of the present invention; and

FIG. 3 is a flow diagram of an embodiment of a method of regulating a voltage carried out according to the principles of the present invention.

DETAILED DESCRIPTION OF CERTAIN
ASPECTS AND EMBODIMENTS

To reduce pattern dependency on a voltage regulator from load current, a high speed response regulator is disclosed. In addition to a high speed response, the disclosed voltage regulator also provides DC precision needed for the output. Thus, a DC accurate voltage regulator with a quick response to load changes is disclosed herein.

The voltage regulator includes two amplifiers coupled together in parallel to provide DC precision and rapid load response. One of the amplifiers is a DC precision amplifier configured to provide DC precision centered around the DC operating point designated for the voltage regulator. The other amplifier is a response amplifier configured to provide adjustments to the output of the voltage regulator in response to changes in the load. As discussed below, the DC precision amplifier may be a low speed/high gain CMOS amplifier that maintains DC accuracy of the voltage regulator's output voltage. The fast response amplifier may be a high speed/low gain NPN amplifier that drives a bypass capacitor located at the output node of the voltage regulator.

FIG. 1 is a block diagram of an embodiment of an integrated circuit (IC) **100** having a voltage regulator **110** constructed according to the principles of the present invention. In addition to the voltage regulator **110**, the IC **100** also includes logic circuitry **120**. Unlike an off-chip voltage regulator with a large output capacitor, the voltage regulator **110** provides an on-chip alternative with a low impedance connection to the load, the logic circuitry **120**. The voltage regulator **110** allows high DC precision, quick response, voltage regulation that is proximate the circuit, the logic circuitry **120**, requiring the regulated voltage.

The IC **100** receives an external voltage (i.e., supplied to the IC **100**) that is used by the voltage regulator **110** and the

logic circuitry **120**. From the external voltage, a reference voltage is provided to the voltage regulator **110**. The value of the reference voltage may vary according to factors such as the voltage rating of the logic components of the logic circuitry **120**. For example, the reference voltage may be 1.2 volts, 1.5 volts or 2 volts with respect to the upper or lower supplied voltage. In addition to the voltage regulator **110** and the logic circuitry **120**, the IC **100** may include additional components that are not illustrated but are typically included on a conventional IC.

The logic circuitry **120** is configured to perform logic functions in accordance with designated applications. In one embodiment, the logic circuitry **120** is Complementary metal-oxide-semiconductor (CMOS) logic. The CMOS logic may include complementary MOS transistors configured to perform the designated logic functions. For example, the IC **100** may be a chip of a pre-amplifier for a HDD that is used to write data on a platter of the HDD. As such, the logic circuitry **120** may be the CMOS data path used to write the data. Of course, the logic circuitry **120** may be configured to perform other functions in different embodiments. Additionally, the logic circuitry **120** may be or have other types of logic besides CMOS logic including, for example, n-type MOS (NMOS) logic, p-type MOS (PMOS) logic, Transistor-Transistor Logic (TTL) and Emitter Coupled Logic (ECL)s.

In contrast to traditional, internally compensated voltage regulators, the voltage regulator **110** contains a high speed amplifier loop which is output compensated making the regulator capable of reacting quickly to rapid load changes. The quick response of the voltage regulator **110** may result in less supply bounce on the supply rails of the logic circuitry **120**. As such, jitter along the supply rails can be reduced. For example, considering the logic circuitry **120** as the CMOS data path, data path jitter may be reduced.

The voltage regulator **110** is configured to provide a substantially constant output voltage, a regulated voltage, to the logic circuitry **120**. The voltage regulator **110** includes a DC precision amplifier **114**, a summer **116** and a response amplifier **118**. The DC precision amplifier **114** and the response amplifier **118** are coupled in parallel to provide the output voltage. The DC precision amplifier **114** is configured to generate a DC precision signal based on a reference voltage and the regulated output of the voltage regulator **110**. The DC precision signal provides DC precision for the regulated output.

The response amplifier **118** is configured to generate an error signal based on the reference voltage and the regulated output. The error signal is generated in response to load changes associated with the voltage regulator **110**. The summer **116** is configured to add the DC precision signal and the error signal from the response amplifier **118** to provide a regulating signal. Based on the regulating signal, the response amplifier **118** generates the regulated output.

The summer **116** may be a resistor coupled between the DC precision amplifier **114** and the response amplifier **118**. The response amplifier **118** may be a high speed, low gain NPN amplifier and the DC precision amplifier **114** may be a high gain, low bandwidth CMOS amplifier. Embodiments of a high gain, low bandwidth CMOS amplifier and a high speed, low gain NPN amplifier are illustrated and discussed below with respect to FIG. 2.

Turning now to FIG. 2 is schematic diagram of an embodiment of a voltage regulator **200** constructed according to the principles of the present invention. The voltage regulator **200** is an on-chip regulator configured to provide a regulated voltage to a proximate load. The load may be CMOS circuitry that is located on the same chip as the voltage regulator **200**.

The voltage regulator **200** includes a DC precision amplifier **210** and a response amplifier **220** coupled together in parallel. An external voltage source provides an upper and a lower supply voltage for the voltage regulator **200** that is used to obtain a reference voltage. The supply voltage and reference voltage may be provided through conventional means typically employed with ICs. The value of the reference voltage may vary according to the desired regulated voltage, output voltage, needed for the load. In some embodiments, the reference voltage may be 1.5 volts or about 1.5 volts with respect to the upper or lower supplied voltage.

The DC precision amplifier **210** is a low speed/high gain CMOS amplifier that maintains DC accuracy of the output of the voltage regulator **200** at the output node. The high gain of the DC precision amplifier **210** facilitates the DC accuracy of the voltage regulator **200**. The DC precision amplifier **210** can be considered a Gm stage with a large output capacitor, the output capacitor **216**.

In addition to being a high gain amplifier, the DC precision amplifier **210** is a low bandwidth amplifier. The bandwidth of the DC precision amplifier **210** is sufficiently low in order to allow the response amplifier **220** and output capacitance at the output node to dominate the high speed frequency response of the voltage regulator **200**. The DC precision amplifier **210** dominates the low frequency behavior of the voltage regulator **200**. Thus, the DC precision amplifier **210** may be a slow amplifier having a bandwidth that is lower than the response amplifier **220**.

The DC precision amplifier **210** includes a CMOS differential pair **211**, a differential pair current source **213**, a CMOS current source **214**, an output capacitor **216** and a DC driver **218**. The CMOS differential pair **211** compares the reference voltage to the output voltage of the voltage regulator **200**. The resulting comparison signal is an error signal that is used to drive the DC driver **218** and the output capacitor **216**.

The differential pair current source **213** is a conventional current source that provides the bias current for the CMOS differential pair **211**. The CMOS current mirror **214** is a conventional component that operates as an active load for the DC precision amplifier **210**. The output capacitor **216** is coupled to the gate of the DC driver **218** and is selected to provide a pole that is decades slower (e.g., at least two decades slower) than the dominant pole for the response amplifier loop, set by output driver **227** and bypass capacitor **228**.

The DC driver **218** provides the output of the DC precision amplifier **210** to the response amplifier **220** and is used to set the DC output level for the voltage regulator **200**. The DC driver **218** receives operating current from a current source (i.e., response amplifier current source **224**) of the response amplifier **220**, and presents a low impedance input to the summing resistor **230**. The DC driver **218** may be a PMOS source-follower as illustrated.

The response amplifier **220** is a high speed, low gain amplifier. The high speed/low gain, response amplifier **220** is a NPN amplifier that drives a bypass capacitor **228** located at the output node of the voltage regulator **200**. The response amplifier **220** may have a low gain less than 20 dB to ensure an adequate phase margin. In one embodiment discussed below having a reference voltage of 1.5 volts, the gain of the response amplifier **220** may be 15 dB or about 5. In this embodiment, the DC precision amplifier **210** may have a gain of 60 dB or about 60 dB. The overall gain of the combined amplifiers, the response amplifier **220** and the DC precision amplifier **210**, may also be 60 dB (or about 60 dB) at low frequency.

In addition to the bypass capacitor **228**, the response amplifier **220** includes an output differential pair **221**, differential pair resistors **222**, an output differential pair current source **223**, a response amplifier current source **224**, an output current source **225** and an output transistor **227**. The output differential pair **221** is a NPN differential pair that receives the reference voltage and the output voltage and generates an error signal therefrom. The differential pair resistors **222** provide a load on the emitter of the NPN transistors of the output differential pair **221**. The output differential pair current source **223**, the response amplifier current source **224** and the output current source **225** are conventional devices that provide current for the response amplifier **220**.

The output transistor **227** drives the output node of the voltage regulator **200**. The output transistor **227** may be a NPN emitter-follower transistor. The output transistor **227** provides a low impedance output for the base current, the regulating signal, and is used to drive the output capacitor **228**. The output transistor **227** can then provide the needed current at the output node while the bypass capacitor **228** provides the needed voltage.

The output node needs a sufficient amount of bypass capacitance to provide the dominant pole in the loop stability of the response amplifier **220**. As such, the bypass capacitor **228** is selected to provide enough capacitance at the output node to keep the pole at the output node from approaching the secondary pole of the response amplifier **220**. The secondary pole of the response amplifier **220** is found at the output of the output differential pair **221** which produces the error signal that is used to drive the output transistor **227**. All internal nodes of the response amplifier **220** are designed to have a higher bandwidth than the output node.

A large amount of bypass capacitance at the output node also aids the voltage regulator **200** in maintaining tight output tolerance and is used as compensation for the high speed amplifier loop of the response amplifier **220**. Due to the high speed of the response amplifier **220**, the response amplifier **220** is quick to react to fast load transients, but because of the low gain, the response amplifier **220** suffers from lack of DC accuracy. The response amplifier **220** is then used to dominate the high frequency behavior of the voltage regulator **200** while the DC precision amplifier **210** is used to provide an accurate DC output level.

The voltage regulator also includes a summer **230** that combines the output of the DC precision amplifier **210**, the DC precision signal, with the error signal from the output differential pair **221**, to generate a regulating signal that is used to drive the output transistor **227**. As illustrated in FIG. **2**, the summer **230** may be a resistor coupled to differential pair **221** and the base of the output transistor **227**.

One skilled in the art will understand that the ratings and values of the various components of both the DC precision amplifier **210** and the response amplifier **220** are based on the desired output voltage. In one embodiment, where the reference voltage is 1.5 volts to provide a regulated voltage of 1.5 volts, the differential pair current source **213**, the output differential pair current source **223**, the response amplifier current source **224** and the output current source **225** may be rated at 12 μ A, 1.8 mA, 2.1 mA and 1.8 mA, respectively. Additionally, the output capacitor **216** may be a 10 pF capacitor and the bypass capacitor **228** may be a 400 pF capacitor. Furthermore, each of the differential pair resistors **222** may be 15 ohm resistors and the summer **230** may be a 500 ohm resistor.

FIG. **3** is a flow diagram of an embodiment of a method of regulating a voltage carried out according to the principles of the present invention. The method may be realized by

employing a voltage regulator as illustrated in FIGS. **1** and **2**. The method **300** begins in a step **305** with an intent to provide a regulated voltage.

The method **300** continues by providing a DC precision signal employing a CMOS amplifier in a step **310**. In one embodiment, the DC precision signal can be generated by a source-follower transistor of the CMOS amplifier. The source-follower transistor may be driven by a CMOS error signal from a CMOS differential pair of the CMOS amplifier.

Additionally, an error signal is generated employing a NPN amplifier in a step **320**. The NPN error signal is generated in one embodiment by a NPN differential pair of the NPN amplifier. The error signal is generated based on a comparison of a reference voltage and an output voltage.

The DC precision signal and the error signal are summed in a step **330** to provide a regulating signal. A summer may be used to combine the NPN error signal and the regulating signal. In one embodiment, the summer may be a resistor coupled between the CMOS amplifier and the NPN amplifier.

In a step **340**, the regulating signal is used to drive a NPN emitter-follower to generate a regulated voltage. The base of the NPN emitter-follower, which is an output transistor for the NPN amplifier receives the regulating signal to generate the regulated voltage. The NPN emitter-follower provides a low impedance output that is beneficial in providing proximate voltage regulation for logic loads such as CMOS logic. In a step **350**, the method **300** ends.

Those skilled in the art to which the invention relates will appreciate that other and further additions, deletions, substitutions and modifications may be made to the described embodiments without departing from the scope of the invention.

What is claimed is:

1. A voltage regulator comprising:

a DC precision amplifier configured to generate a DC precision signal based on a reference voltage and a regulated output of said voltage regulator;

a response amplifier, coupled in parallel with said DC precision amplifier, configured to generate an error signal based on said reference voltage and said regulated output, said response amplifier further configured to generate said regulated output based on a regulating signal comprised of said error signal and said DC precision signal.

2. The voltage regulator as recited in claim **1** wherein said DC precision signal provides DC precision for said voltage regulator.

3. The voltage regulator as recited in claim **1** wherein said error signal is generated in response to load changes associated with said voltage regulator.

4. The voltage regulator as recited in claim **1** wherein said DC precision amplifier includes an output capacitor selected to set the dominant pole for said voltage regulator.

5. The voltage regulator as recited in claim **1** wherein said DC precision amplifier includes an output capacitor selected to provide a pole response that is at least two decades slower than the dominant pole of said response amplifier.

6. The voltage regulator as recited in claim **1** wherein said DC precision amplifier is a high gain, low bandwidth amplifier.

7. The voltage regulator as recited in claim **1** wherein said response amplifier is a high speed, low gain amplifier.

8. The voltage regulator as recited in claim **1** further comprising a summer configured to provide said regulating signal based on said DC precision signal and said error signal.

7

9. A voltage regulator comprising:
 a first amplifier configured to receive a reference voltage
 and employ a first transistor to generate a DC precision
 signal based on said reference voltage and a regulated
 output of said voltage regulator;

a second amplifier configured to receive said DC precision
 signal, said reference voltage and said regulated output,
 said second amplifier employing a second transistor to
 generate said regulated output based on a regulating
 signal;

a summer configured to generate said regulating signal
 based on said DC precision signal and an error signal
 generated by said second amplifier.

10. The voltage regulator as recited in claim 9 wherein said
 first amplifier is a CMOS amplifier, said first transistor is a
 source-follower, said second amplifier is a NPN amplifier and
 said second transistor is a NPN emitter-follower.

11. The voltage regulator as recited in claim 9 wherein said
 summer couples said first amplifier to said second amplifier.

12. The voltage regulator as recited in claim 9 wherein said
 second amplifier includes a bypass capacitor coupled to an
 output node of said voltage regulator, said bypass capacitor
 selected to provide a secondary pole for said voltage regula-
 tor.

13. The voltage regulator as recited in claim 12 wherein
 said first amplifier includes an output capacitor coupled to
 said first transistor, said output capacitor selected to provide a
 pole response that is at least two decades slower than said
 secondary pole.

14. The voltage regulator as recited in claim 9 wherein said
 second amplifier has a gain of about five.

15. An integrated circuit, comprising:
 logic circuitry; and

a voltage regulator configured to provide a regulated output
 for said logic circuitry and including:

a DC precision amplifier configured to generate a DC
 precision signal based on a reference voltage and said
 regulated output of said voltage regulator; and

8

a response amplifier, coupled in parallel with said DC
 precision amplifier, configured to generate an error
 signal based on said reference voltage and said regu-
 lated output, said response amplifier further config-
 ured to generate said regulated output based on a
 regulating signal comprised of said error signal and
 said DC precision signal.

16. The integrated circuit as recited in claim 15 wherein
 said DC precision signal provides DC precision for said regu-
 lated output.

17. The integrated circuit as recited in claim 15 wherein
 said logic circuitry is CMOS logic circuitry.

18. The integrated circuit as recited in claim 17 wherein
 said integrated circuit is at least part of a preamplifier for a
 data writer.

19. The integrated circuit as recited in claim 15 wherein
 said DC precision amplifier is a high gain, low bandwidth
 CMOS amplifier.

20. The integrated circuit as recited in claim 19 wherein
 said response amplifier is a high speed, low gain NPN ampli-
 fier.

21. The integrated circuit as recited in claim 15 further
 comprising a summer configured to provide said regulating
 signal based on said DC precision signal and said error signal.

22. A method of providing a regulated voltage, comprising:
 providing a DC precision signal employing a first ampli-
 fier;

generating an error signal employing a second amplifier
 coupled in parallel with said first amplifier;

summing said DC precision signal and said error signal to
 provide a regulating signal; and

using said regulating signal to drive a transistor to generate
 a regulated voltage.

23. The method as recited in claim 22 wherein said first
 amplifier is a CMOS amplifier and said second amplifier is a
 NPN amplifier.

24. The method as recited in claim 23 wherein said tran-
 sistor is a NPN emitter-follower transistor.

* * * * *