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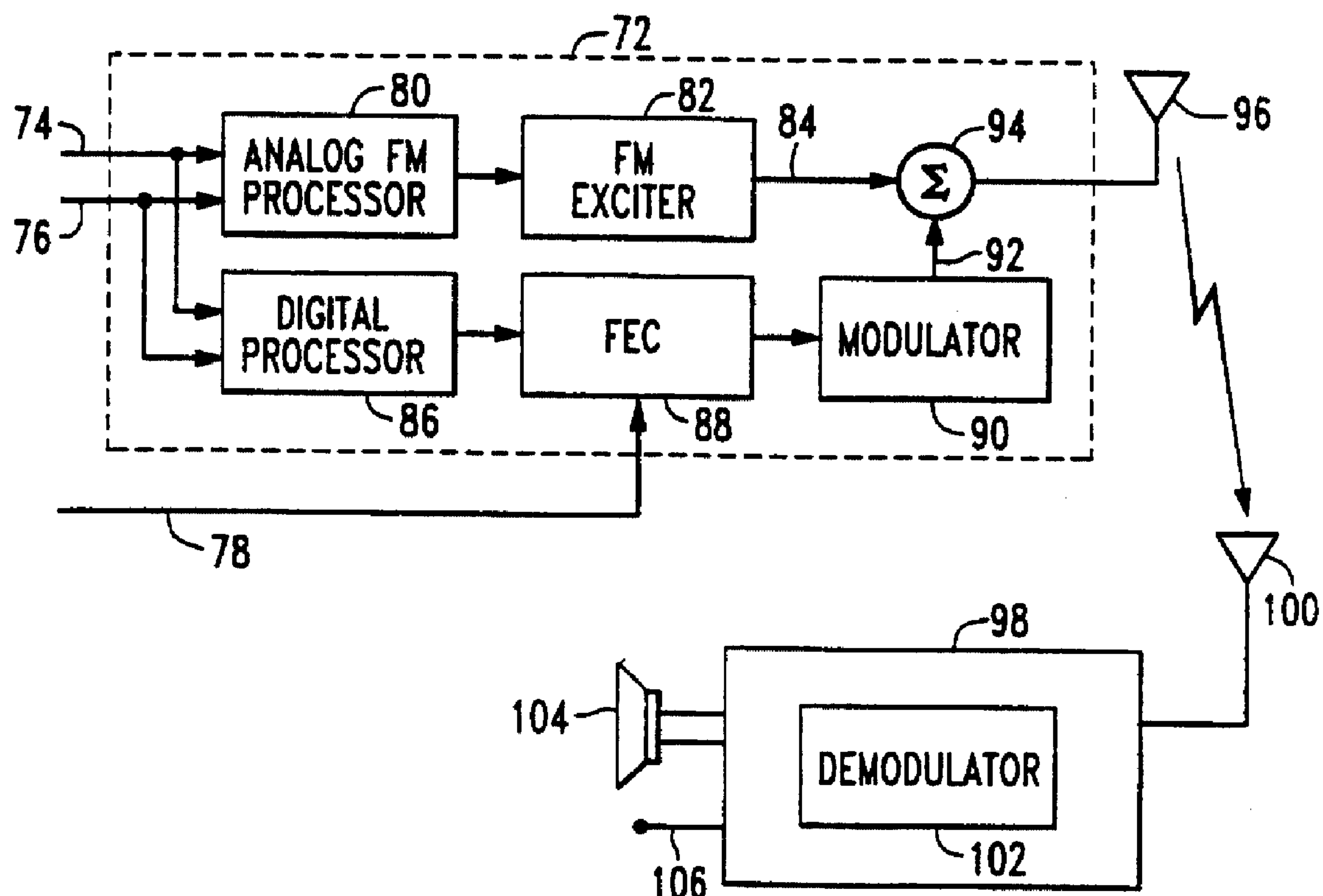
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(54) Titre : SYSTEME ET PROCEDE DE DIFFUSION AUDIONUMERIQUE DE MODULATION DE FREQUENCE DANS LE MEME CANAL, DANS LA MEME BANDE

(54) Title: HYBRID AND ALL DIGITAL FM IN-BAND-ON-CHANNEL DIGITAL AUDIO BROADCASTING METHOD AND SYSTEM



(57) Abrégé/Abstract:

This invention provides a method of broadcasting including the steps of providing a broadcast signal in a central frequency band of an FM radio channel; providing a first plurality of sub-carriers in an upper sideband of the FM radio channel; providing a second

(57) Abrégé(suite)/Abstract(continued):

plurality of sub-carriers in a lower sideband of the FM radio channel; modulating a first group of the first plurality of sub-carriers with a complementary punctured convolution coded encoded version of program material; modulating a first group of the second plurality of sub-carriers with the complementary punctured convolution coded version of the program material; and transmitting the broadcast signal, the first group of the first plurality of sub-carriers and the first group of the second plurality of sub-carriers. In an all-digital version of the invention, the central frequency band may include a plurality of sub-carriers transmitted at a lower power level than that of the sub-carriers in the upper and lower sidebands. These additional sub-carriers can be used to transmit additional data. Transmitters and receivers which utilized the method are also included.

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<table border="1" style="width: 100%; border-collapse: collapse;"> <tr> <td style="width: 50%; vertical-align: top; padding: 5px;"> <p>(21) International Application Number: PCT/US99/04328</p> <p>(22) International Filing Date: 24 February 1999 (24.02.99)</p> <p>(30) Priority Data: 09/049,210 27 March 1998 (27.03.98) US</p> <p>(71) Applicant: USA DIGITAL RADIO, INC. [US/US]; 8865 Stanford Boulevard, Columbia, MD 21045 (US).</p> <p>(72) Inventors: KROEGER, William; 12813 Amberwoods Way, Sykesville, MD 21784 (US). CAMMARATA, Denise, Maureen; 12087 Long Lake Drive, Owings Mill, MD 21117 (US). MARTINSON, Richard, Edward; 906 High Stepper Trail, Sykesville, MD 21784 (US). BAIRD, Jeffrey, Scott; 19882 Olde Woods Way, Columbia, MD 21044 (US). PEYLA, Paul, James; 6528 Meadowfield Court, Elkridge, MD 21075 (US). WALDEN, E., Glynn; 14 Landings Drive, Marlton, NJ 08053 (US).</p> <p>(74) Agent: LENART, Robert, L.; CBS Corporation, 11 Stanwix Street, Pittsburgh, PA 15222 (US).</p> </td> <td style="width: 50%; vertical-align: top; padding: 5px;"> <p>(81) Designated States: AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, CA, CH, CN, CU, CZ, DE, DK, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MD, MG, MK, MN, MW, MX, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, UA, UG, UZ, VN, YU, ZW, ARIPO patent (GH, GM, KE, LS, MW, SD, SZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GW, ML, MR, NE, SN, TD, TG).</p> <p>Published <i>With international search report.</i> <i>Before the expiration of the time limit for amending the claims and to be republished in the event of the receipt of amendments.</i></p> <p>(88) Date of publication of the international search report: 18 November 1999 (18.11.99)</p> </td> </tr> </table>			<p>(21) International Application Number: PCT/US99/04328</p> <p>(22) International Filing Date: 24 February 1999 (24.02.99)</p> <p>(30) Priority Data: 09/049,210 27 March 1998 (27.03.98) US</p> <p>(71) Applicant: USA DIGITAL RADIO, INC. [US/US]; 8865 Stanford Boulevard, Columbia, MD 21045 (US).</p> <p>(72) Inventors: KROEGER, William; 12813 Amberwoods Way, Sykesville, MD 21784 (US). CAMMARATA, Denise, Maureen; 12087 Long Lake Drive, Owings Mill, MD 21117 (US). MARTINSON, Richard, Edward; 906 High Stepper Trail, Sykesville, MD 21784 (US). BAIRD, Jeffrey, Scott; 19882 Olde Woods Way, Columbia, MD 21044 (US). PEYLA, Paul, James; 6528 Meadowfield Court, Elkridge, MD 21075 (US). WALDEN, E., Glynn; 14 Landings Drive, Marlton, NJ 08053 (US).</p> <p>(74) Agent: LENART, Robert, L.; CBS Corporation, 11 Stanwix Street, Pittsburgh, PA 15222 (US).</p>	<p>(81) Designated States: AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, CA, CH, CN, CU, CZ, DE, DK, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MD, MG, MK, MN, MW, MX, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, UA, UG, UZ, VN, YU, ZW, ARIPO patent (GH, GM, KE, LS, MW, SD, SZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GW, ML, MR, NE, SN, TD, TG).</p> <p>Published <i>With international search report.</i> <i>Before the expiration of the time limit for amending the claims and to be republished in the event of the receipt of amendments.</i></p> <p>(88) Date of publication of the international search report: 18 November 1999 (18.11.99)</p>
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<p>(54) Title: HYBRID AND ALL DIGITAL FM IN-BAND-ON-CHANNEL DIGITAL AUDIO BROADCASTING METHOD AND SYSTEM</p> <p>(57) Abstract</p> <p>This invention provides a method of broadcasting including the steps of providing a broadcast signal in a central frequency band of an FM radio channel; providing a first plurality of sub-carriers in an upper sideband of the FM radio channel; providing a second plurality of sub-carriers in a lower sideband of the FM radio channel; modulating a first group of the first plurality of sub-carriers with a complementary punctured convolution coded encoded version of program material; modulating a first group of the second plurality of sub-carriers with the complementary punctured convolution coded version of the program material; and transmitting the broadcast signal, the first group of the first plurality of sub-carriers and the first group of the second plurality of sub-carriers. In an all-digital version of the invention, the central frequency band may include a plurality of sub-carriers transmitted at a lower power level than that of the sub-carriers in the upper and lower sidebands. These additional sub-carriers can be used to transmit additional data. Transmitters and receivers which utilized the method are also included.</p>				

HYBRID AND ALL DIGITAL FM IN-BAND-ON-CHANNEL DIGITAL AUDIO BROADCASTING METHOD AND SYSTEM

BACKGROUND OF THE INVENTION

This invention relates to radio broadcasting, and more particularly, to modulation formats for FM In-Band-On-Channel (IBOC) Digital Audio Broadcasting (DAB) and broadcasting systems utilizing such modulation formats.

5 Digital Audio Broadcasting is a medium for providing digital-quality audio, superior to existing analog broadcasting formats. FM IBOC DAB can be transmitted in a hybrid format where the digitally modulated signal coexists with the currently broadcast analog FM signal. IBOC requires no new spectral allocations because each DAB signal is simultaneously transmitted within the spectral mask of an existing FM channel allocation.
10 IBOC promotes economy of spectrum while enabling broadcasters to supply digital quality audio to their present base of listeners. FM IBOC broadcasting systems using have been the subject of several United States patents including patents no. 5,465,396; 5,315,583; 5,278,844; and 5,278,826.

The advantages of digital transmission for audio include better signal
15 quality with less noise and wider dynamic range than with existing FM radio channels. Initially the hybrid format would be adopted allowing the existing receivers to continue to receive the analog FM signal while allowing new IBOC receivers to decode the digital signal. Hybrid format DAB systems are described in Kroeger et al., "Robust Modem And Coding Techniques For FM Hybrid IBOC DAB", IEEE Transactions on Broadcasting,
20 vol. 43, no. 4, December 1997; Kroeger et al., "Compatibility of FM Hybrid In-Band On-Channel (IBOC) System For Digital Audio Broadcasting", IEEE Transactions on Broadcasting, vol. 43, no. 4, December 1997; Kroeger et al., "Robust IBOC DAB AM And FM Technology For Digital Audio Broadcasting", 51st Annual Broadcast Engineering Conference (NAB), April 1997; and PCT published Patent Application WO 97 49207 A.
25 Some time in the future, when IBOC DAB receivers are abundant; broadcasters may elect to transmit an all-digital format. An all-digital format DAB system is described in Cupo et al., "An OFDM All Digital In-Band On-Channel (IBOC) AM And FM Radio Solution Using The PAC Encoder", IEEE Transactions on Broadcasting, vol. 44, no. 1, March

1998. The goal of FM hybrid IBOC DAB is to provide virtual-CD-quality stereo digital audio (plus data) while simultaneously transmitting the existing FM signal. The goal of FM all-digital IBOC DAB is to provide virtual-CD-quality stereo audio along with a data channel with capacity of up to about 200 kbps, depending upon a particular station's
5 interference environment.

Since there is expected to be a transition from hybrid to all-digital IBOC DAB formats, it would be desirable to devise a modulation format that can be used by both systems so that the transition can be made with minimal changes in transmitting equipment.

10 SUMMARY OF THE INVENTION

This invention provides a method of broadcasting comprising the steps of: providing a first plurality of sub-carriers in an upper sideband of an FM radio channel, the upper sideband ranging from about +100 kHz to +200 kHz from a center frequency of the radio channel; providing a second plurality of sub-carriers in a lower sideband of the FM
15 radio channel, the lower sideband ranging from about -100 kHz to -200 kHz from a center frequency of the radio channel; modulating a first group of the first plurality of sub-carriers with a digitally coded encoded version of program signal to be transmitted; modulating a first group of the second plurality of sub-carriers with the digitally coded version of the program signal to be transmitted; providing a
20 third plurality of sub-carriers in a central frequency band of the FM radio channel, wherein the power spectral density of the third plurality of sub-carriers is less than the power spectral density of the sub-carriers in the upper and lower sidebands; modulating the third plurality of sub-carriers with additional data; and transmitting the first group of the first plurality of sub-carriers, the first group of the second plurality of sub-carriers, and the third
25 plurality of sub-carriers.

When operating with a hybrid modulation format, the central frequency band signal comprises a carrier FM modulated by an analog program signal. When operating with an all-digital modulation format, the central frequency band signal comprises a third plurality of sub-carriers modulated by a digital signal and broadcast at an
30 average power spectral density less than that of the upper sideband and lower sideband sub-carriers.

The invention also encompasses transmitters and receivers that utilize the above method.

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Thus, according to one particular aspect of the invention, there is provided a transmitter for broadcasting in-band-on-channel digital audio signals, said transmitter comprising: means for producing a first plurality of sub-carriers in an upper sideband of an FM radio channel, said upper sideband lying within a frequency band extending from about +100 kHz to about +200 kHz from a center frequency of said radio channel, a second plurality of sub-carriers in a lower sideband of said FM radio channel, said lower sideband lying within a frequency band extending from about -100 kHz to about -200 kHz from the center frequency of said radio channel, and a third plurality of sub-carriers in a central frequency band of the FM radio channel, wherein the third plurality of sub-carriers have a lower power spectral density than that of the first and second plurality of sub-carriers; means for modulating a first group of said first plurality of sub-carriers with a coded version of program material; means for modulating a first group of said second plurality of sub-carriers with said coded version of said program material; means for modulating said third plurality of sub-carriers with additional data; and means for transmitting said first group of said first plurality of sub-carriers, said first group of said second plurality of sub-carriers, and said third plurality of sub-carriers.

There is also provided a receiver for receiving in-band-on-channel digital audio signals, said receiver comprising: means for receiving a first plurality of sub-carriers in an upper sideband of an FM radio channel, said upper sideband lying within a frequency band extending from about +100 kHz to about +200 kHz from a center frequency of said radio channel, said first plurality of sub-carriers being modulated with a complementary punctured convolution coded encoded version of the program material; a second

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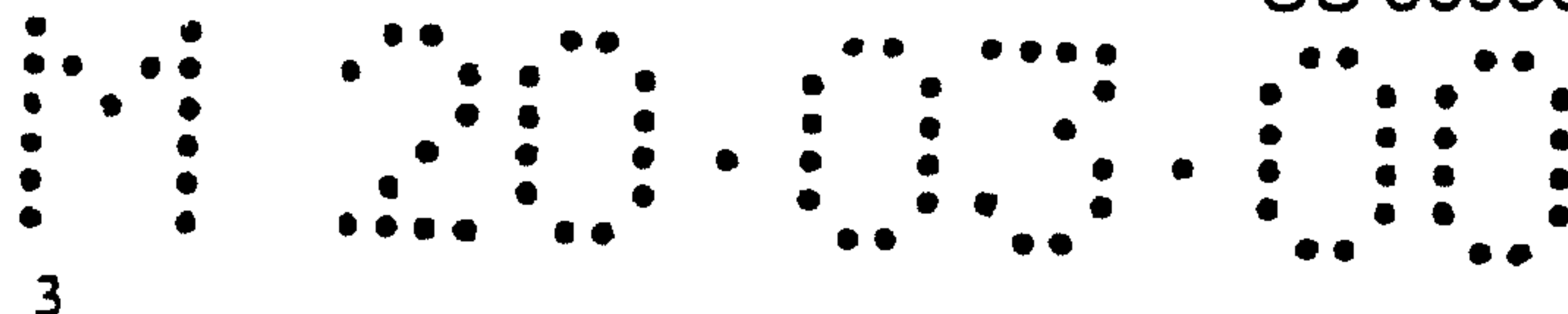
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plurality of sub-carriers in a lower sideband of said FM radio channel, said lower sideband lying within a frequency band extending from about -100 kHz to about -200 kHz from the center frequency of said radio channel,

5 said second plurality of sub-carriers being modulated with a complementary punctured convolution coded encoded version of the program material; and a third plurality of sub-carriers, said third plurality of sub-carriers being modulated with additional data having a lower power spectral density than

10 that of said first and second plurality of sub-carriers; means for demodulating a first group of said first plurality of sub-carriers, a first group of said second plurality of sub-carriers, and said third plurality of sub-carriers; and means for outputting said program material obtained from

15 demodulating said first group of said first plurality of sub-carriers and said first group of said second plurality of sub-carriers, and additional data obtained from demodulating said third plurality of sub-carriers.



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BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1 is a schematic representation of the frequency allocations and relative power spectral density of the signal components for a hybrid FM IBOC DAB signal;

5 Figure 2 is a schematic representation of the frequency allocations and relative power spectral density of the signal components for an all-digital FM IBOC DAB signal in accordance with the present invention;

Figure 3 is a schematic representation of the frequency allocations and relative power spectral density of the signal components for the upper sideband of the FM IBOC DAB signal in accordance with the present invention;

Figure 4 is a schematic representation of the frequency allocations and relative power spectral density of the signal components for the lower sideband of the FM IBOC DAB signal in accordance with the present invention;

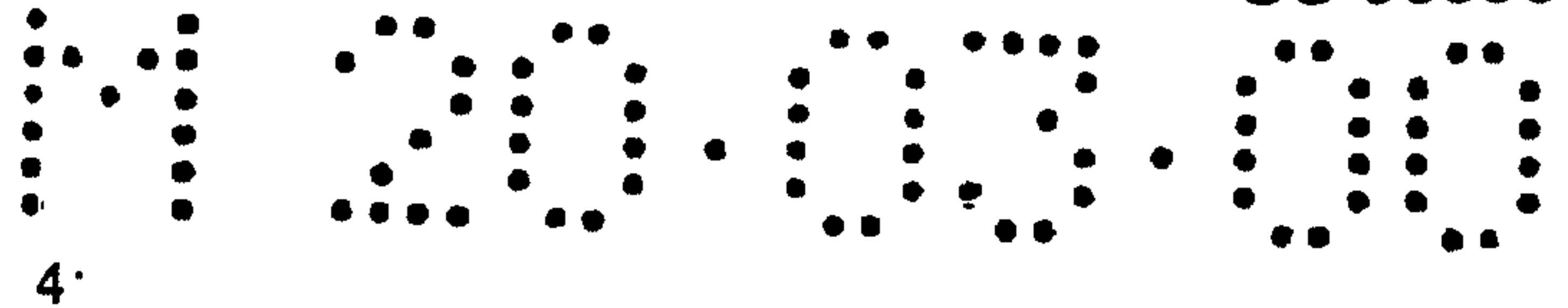
Figure 5 illustrates the potential interference between a channel broadcast in accordance with this invention and the left first adjacent analog FM channel in an IBOC DAB system;

Figure 6 illustrates the potential interference between a channel broadcast in accordance with this invention and the left first adjacent IBOC DAB channel in an IBOC DAB system; and

20 Figure 7 is a simplified block diagram of a broadcasting system, which may incorporate the modulation method of the present invention.

DESCRIPTION OF THE PREFERRED EMBODIMENTS

Referring to the drawings, Figure 1 is a schematic representation of the frequency allocations (spectral placement) and relative power spectral density of the signal components for a hybrid FM IBOC DAB signal 10 in accordance with the present invention. The hybrid format includes the conventional FM stereo analog signal 12 having a power spectral density represented by the triangular shape 14 positioned in a central, or central frequency band, 16 portion of the channel. The Power Spectral Density (PSD) of a typical analog FM broadcast signal is nearly triangular with a slope of about -0.35 dB/kHz from the center frequency. A plurality of digitally modulated evenly spaced sub-carriers are positioned on either side of the analog FM signal, in an upper sideband 18 and a lower sideband 20, and are transmitted concurrently with the analog FM signal. All of the carriers are transmitted at a power level that falls within the United States Federal

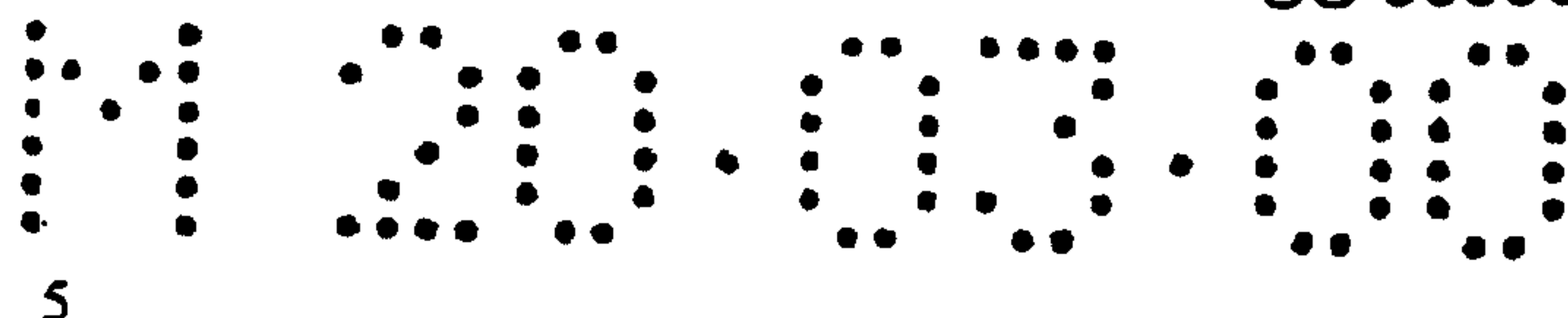


Communications Commission channel mask 22. The vertical axis in Figure 1 shows the peak power spectral density as opposed to a more conventional average power spectral density characterization. In this case, the total one-sided DAB signal power is 25 dB below the FM carrier power while the peak spectral power ratio appears significantly greater. The short-term FM spectrum is more "peaky" than the short-term DAB spectrum when both are observed in a 1 kHz bandwidth. As will be seen from the description below, the digitally modulated portion of the hybrid signal is a subset of the all-digital DAB signal that will be transmitted in the all-digital IBOC DAB format.

Signals from an adjacent FM channel (i.e. the first adjacent FM signals), if present, would be centered at a spacing of 200 kHz from the center of the channel of interest. In the hybrid FM IBOC modulation format, 95 evenly spaced orthogonal frequency division modulated (OFDM) sub-carriers are placed on each side of the host analog FM signal occupying the spectrum from about 129 kHz through 198 kHz away from the host FM center frequency as illustrated by the upper sideband 18 and the lower sideband 20 in Figure 1. In the hybrid system the total DAB power in the OFDM modulated sub-carriers in each sideband is set to about -25 dB relative to its host analog FM power.

The spectral placement and relative signal power density levels of the OFDM digital sub-carriers in an embodiment of the invention, referred to as the all-digital FM DAB format illustrated by item number 24, is shown in Figure 2. In this embodiment of the invention, the analog FM signal has been replaced by an optional additional group of OFDM sub-carriers, referred to as the extended all-digital signal 26, located in the central frequency band 28. Once again evenly spaced OFDM sub-carriers are positioned in an upper sideband 30 and a lower sideband 32. The sidebands of the all-digital format of Figure 2 are wider than the sidebands of Figure 1. In addition, the power spectral density level of the all-digital IBOC signal sidebands is set about 10 dB higher than that allowed in the hybrid IBOC sidebands. This provides the all-digital IBOC signal with a significant performance advantage. Furthermore the power spectral density of the extended all-digital signal is about 15 dB below that of the hybrid IBOC sidebands. This minimizes or eliminates any interference problems to adjacent hybrid or all-digital IBOC signals while providing additional capacity for other digital services.

It is recommended that the extended data sub-carriers be set at a relative level of roughly 15 dB below the level of the other main sub-carriers. This is a

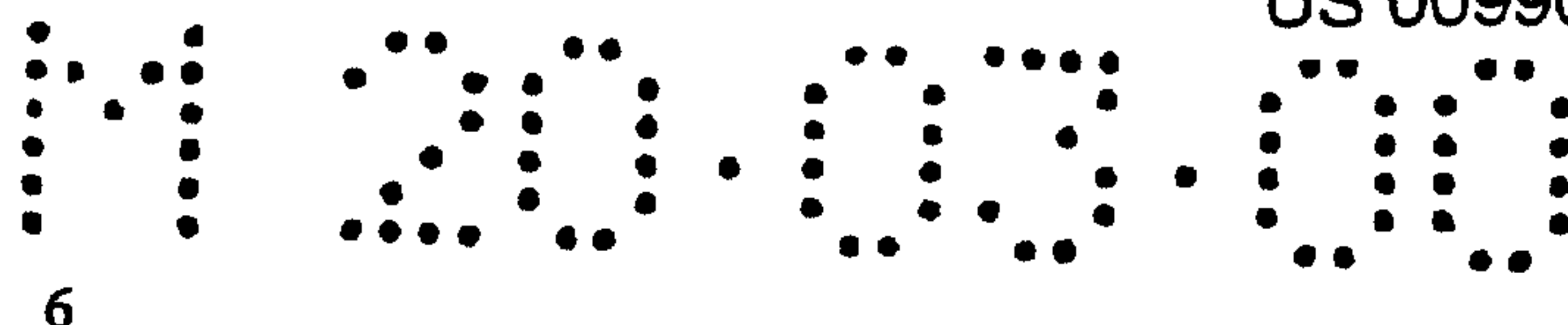


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compromise between the robustness of these extended sub-carriers and interference to the main sub-carriers of a first adjacent signal. To assess the potential interference situation, assume that the maximum relative level of the first adjacent all-digital station is -6 dB at the 54 dBu protected contour. This is the case for the pair of first-adjacent stations that meet the FCC guidelines, although there are exceptions that are close-spaced. The extended data sub-carriers would interfere with the first adjacent main sub-carriers at a relative level of -21 dB (-6dB-15dB). This level of interference includes some margin for fading and should not result in significant degradation of the main signal. However, the extended data sub-carriers would suffer when the first adjacent interferer is -6 dB lower since the interferer's main sub-carriers would be 9 dB higher than the extended data sub-carriers. FEC coding would be imposed upon the extended data sub-carriers such that one first adjacent interferer can be tolerated. If the extended data is perceived to be more valuable than indicative of the protection provided, then consider raising the level of the extended data sub-carriers to -10 dB instead of -15 dB.

Figure 3 is a schematic representation of the placement and relative power spectral density of the signal components for the upper sideband of FM IBOC DAB signal in accordance with the present invention. In Figures 3 and 4, the potential sub-carrier locations are indexed (assigned numbers) ranging from zero at the FM center frequency to plus or minus 273 at the edges of the 400 kHz bandwidth, with the positive assignments having carrier frequencies above the channel center frequency and the negative numbers having frequencies below the channel center frequency. The sub-carrier assignments shown in parentheses above the frequency scale Figure 3 include all optional sub-carriers in the upper sideband of both the hybrid and all-digital systems. In the preferred embodiments of the invention, the individual OFDM sub-carriers are QPSK modulated at 689.0625 Hz (44100/64) and are orthogonally spaced at about 726.7456055 Hz (44100*135/8192) after pulse shaping is applied (root raised cosine time pulse with 7/128 excess time functions as guard time). The frequency scale shows the frequency difference from the channel center frequency.

The upper sideband represented in Figure 3 is comprised of information-bearing sub-carriers 140 through 272 corresponding to subcarrier frequencies 101,744 Hz through 197,675 Hz. Sub-carrier 273 is an optional reference sub-carrier. The upper sideband is shown to be divided into several groups 34, 36, 38 and 40. Group 34 represents the main channel and contains sub-carriers 178 through 253. The main channel

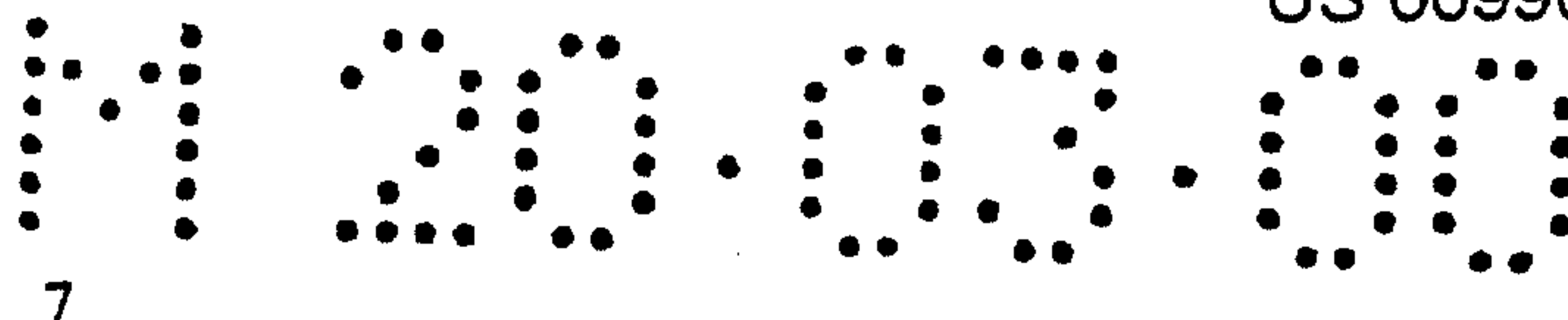


sub-carriers are used to transmit the program material to be broadcast in the form of data bits of the coding algorithm at a rate of at least 96 thousand bits per second (kbs). The main channel may include ancillary and auxiliary data. A second group of carriers 36 occupying sub-carrier positions 254 through 272 are used to transmit parity bits. A third group of carriers can be used to carry a 24 kbps delayed version of the program material for tuning and backup purposes. As will be discussed below, these sub-carriers are more likely to be corrupted by interferers than sub-carriers that are positioned closer to the center of the channel. The most expendable code bits are placed on the outer OFDM sub-carriers. The expendable bits contribute least to the free distance or coding gain of the combined code and they are least important to the error correction ability of the code. Therefore, the most vulnerable sub-carriers are used to carry these expendable bits.

Another group of sub-carriers 38 is used in the all-digital embodiment of the invention to carry parity bits or optional data. Sub-carrier group 40 includes sub-carrier positions 140 through 158 and is used in the all-digital embodiment to transmit a delayed backup version of the program material at a lower data rate, of for example 24 kbs. In the all-digital embodiment, the sub-carriers of group 40 provide data that can be used in the event of a loss of the signal transmitted in the main channel. The sub-carrier at location 273 represents an optional reference signal 42. This signal may be used for signal acquisition purposes if desired.

The lower sideband shown in Figure 4 has a mirror image of the upper sideband format with negative indexes and frequencies. Lower sideband main channel 44 contains the sub-carriers at locations -178 through -253 and is used to transmit the same program material as is transmitted in the upper sideband main channel. The sub-carriers in groups 46, 48 and 50 are utilized in the same manner as the sub-carriers of groups 36, 38 and 40 of the upper sideband. The sub-carrier in position -273 may be used to transmit an optional reference signal. The sub-carriers in both sidebands use orthogonal frequency division multiplexing and are FEC coded using Complementary Punctured Convolution (CPC) codes. CPC codes are known in the art, for example, see S. Kallel, "Complementary Punctured Convolution (CPC) Codes and Their Applications," IEEE Trans. Comm., Vol. 43, No. 6, pp. 2005-2009, June, 1995.

The reference sub-carriers, if used, are located at plus or minus 273 with center frequencies of plus or minus 198,402 Hz. The reference sub-carriers would be modulated with the same symbol phase used to modulate sub-carrier 272 for the previous



symbol time. This allows the receiver the option of performing differential detection in frequency starting with the reference sub-carrier, or differential detection in frequency starting with the time differential detection of sub-carrier 272. Ideally, in the absence of interference, but with fading, performance may be better using the reference sub-carrier.

5 However, it may be advantageous to eliminate the reference sub-carrier to minimize potential interference from a second adjacent DAB signal.

The 96 kbps PAC main channel occupies sub-carriers 178 through 253. This main channel is coded over both DAB sidebands using CPC codes, resulting in a rate $\frac{1}{2}$ CPC code. The reference (pilot) sub-carriers may be modulated with an alternating

10 sequence to permit assistance in frequency and symbol timing acquisition and tracking. The preferred embodiment of this invention uses a perceptual audio coding (PAC) algorithm. Perceptual audio coding algorithms have been the subject of numerous U. S. Patents such as 5,481,614; 5,285,498 and 5,040,217. However, it should be understood that this invention is not limited to the use of perceptual audio coding algorithms.

15 Sub-carriers 254 through 272 (upper and lower sidebands) carry either additional parity bits for the CPC code, or data. The transmission of parity bits here improves the FEC code rate over the main channel from $R=\frac{1}{2}$ to $R=\frac{2}{5}$, or $R=\frac{4}{5}$ on each sideband independently. In the presence of adjacent channel FM interference, these outer OFDM sub-carriers are most vulnerable to corruption, and the interference on the upper

20 and lower sidebands is independent. Since the PSD of an FM broadcast signal is nearly triangular, then the interference increases as the OFDM sub-carriers approach the frequency of a first adjacent signal. When parity bits are transmitted, the coding and interleaving may be specially tailored to deal with this nonuniform interference such that the communication of information is robust.

25 Sub-carriers 159 through 177 in group 38 of the upper sideband and sub-carriers -159 through -177 in group 48 of the lower sideband can carry either additional parity bits for the CPC code, or data. The transmission of parity bits here improves the FEC code rate over the main channel from $R=\frac{1}{2}$ to $R=\frac{2}{5}$, or $R=\frac{4}{5}$ on each independent DAB sideband. If parity bits are transmitted in both regions 159 through 177 and 254

30 through 272 (and corresponding sub-carriers in the lower sideband), then the overall code rate is $R=\frac{1}{3}$, or $R=\frac{2}{3}$ on each independent DAB sideband.

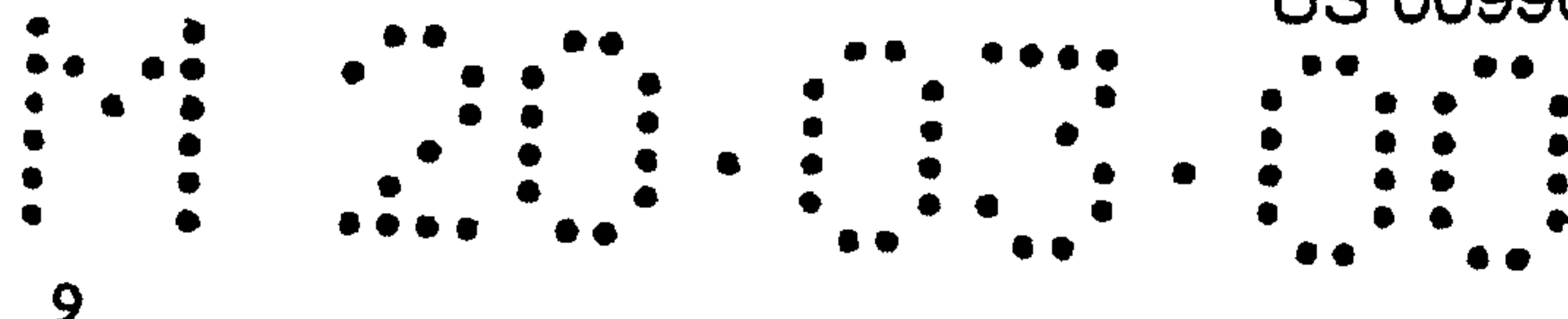
The IBOC DAB system will transmit all the digital audio information on each DAB sideband (upper or lower) of the FM channel. Although additional sub-carriers

beyond the baseline system can be activated to enable the transmission of all the code bits of the rate $1/3$ FEC code, the baseline system employs a code rate of $2/5$. Each sideband can be detected and decoded independently with an FEC coding gain achieved by a rate $4/5$ (optionally rate $2/3$) convolutional code. This redundancy permits operation on one sideband while the other is corrupted. However, usually both sides are combined to provide additional signal power and coding gain commensurate with a rate $2/5$ (optionally rate $1/3$) code. Furthermore special techniques can be employed to demodulate and separate strong first adjacent interferers such that a "recovered" DAB sideband can supplement the opposite sideband to improve coding gain and signal power over any one sideband.

The all-digital system will utilize sub-carriers 140 through 158 in group 40 of the upper sideband and sub-carriers -140 through -158 of the lower sideband to carry a lower data rate version of the data in the main channel, e.g. 24 kbps embedded PAC code. This lower rate backup data is delayed to enhance performance using time diversity. This backup data of the all-digital system replaces the analog FM blend of the hybrid system which is described in commonly owned U.S. Patent No. 6,178,317, entitled "A System And Method For Mitigating Intermittent Interruption In An Audio Radio Broadcast System", issued January 23, 2001. When the Main Channel data is corrupted, the backup data can fill-in the audio segment. Since the backup data is comprised of an embedded subset of the main channel data bits, the backup can enable additional error protection for the main channel.

In the all-digital embodiment, sub-carriers from index -139 to 139 which are located in the central frequency band 28 in Figure 2, can be used as an option to extend DAB capacity. The channel bit rate over this "extended" bandwidth without coding is about 384 kbps. Because half of this bandwidth can be corrupted by a first adjacent DAB signal, the CPC FEC coding technique should be applied to each half of the extended bandwidth, i.e. sub-carriers 1 through 139 should carry the same information as sub-carriers -1 through -139. Then, if either half becomes corrupted, there will still be a rate $2/3$ complementary code on the remaining half. In this case, the information capacity after rate $1/3$ coding is about 128 kbps.

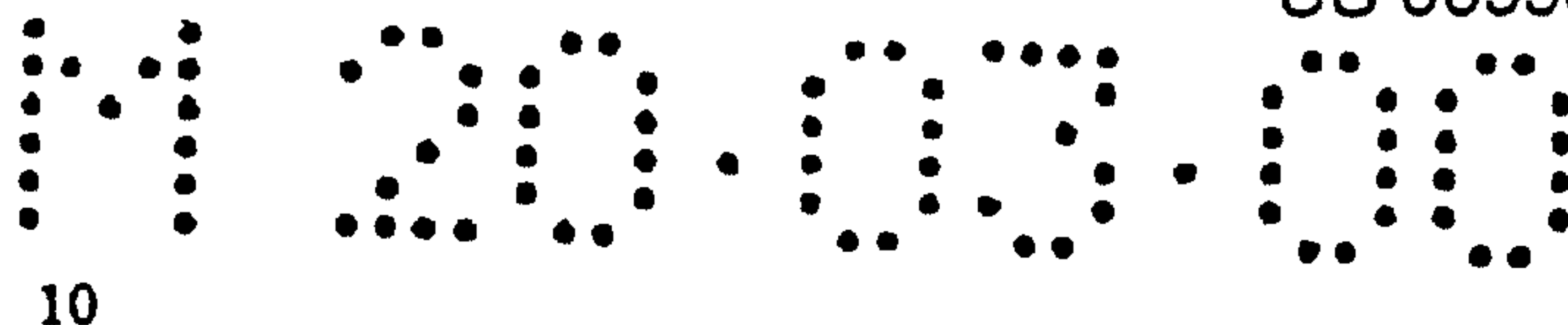
The extended all-digital band is exposed to interference only from a first-adjacent hybrid or all-digital interferer. Under present protected contour guidelines, the maximum level of the first adjacent interferer is -6dB relative to the host station. If this



first adjacent interferer is an all-digital IBOC, then the interferer can be up to 14 dB higher than the level of that half of the extended band. The extended band starts to positively contribute to the coding gain when the spectral density of the interferer is about the same level as the extended band signal. This implies that an all-digital first adjacent interferer must be at least 20 dB below the signal of interest (20 dB di/du) before that half of the extended band is useful. Reception of the extended data might be possible with both first adjacents present at -20 dB; however robust reception in fading probably requires at least one first adjacent at -30 dB or lower.

Consider the possibility of raising the level of the extended band as high as the hybrid DAB sideband levels. The extended band interference to a first adjacent hybrid is then only -6dB at the 54 dBu contour. Similarly the interference to an all-digital first adjacent signal is -16 dB. While the coverage area and robustness of the extended region are not as good as the all-digital sidebands, acceptable levels of performance should be achievable within the normal protected contour, except in areas where both first adjacent signals are significant. Possible uses of the extended all-digital bandwidth are surround-sound, slow-scan video, datacasting, etc. These extended services can be received where available.

The interference to and from the first adjacent channels placed ± 200 kHz away from the host signal can be derived from the relationship of the adjacent signals shown in the plot of Figure 5. Figure 5 shows a hybrid DAB signal 54 having a central frequency band signal 56 and upper and lower sidebands 58 and 60, and a conventional first adjacent left channel 62. FM stations are geographically placed such that the nominal received power of an undesired adjacent channel is at least 6 dB below the desired station's power at the edge of its coverage area. Then the D/U (desired to undesired power ratio in dB) is at least 6 dB. Knowledge of the ratio of each station's DAB signal power to its FM host permits assessment of first adjacent interference to DAB. Similarly the interference of the first adjacent DAB signal 64 (with central frequency band signal 66 and upper and lower sidebands 68 and 70) to the host FM signal can be assessed from the relationship shown in Figure 6. In this example the host signal is shown at 200 kHz offset from the interferer.



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Interference from a second adjacent DAB interference to the host DAB signal has also been addressed. This problem has been avoided by limiting the far edge of the DAB signal to within 200 kHz of its host carrier frequency to prevent spectral overlap.

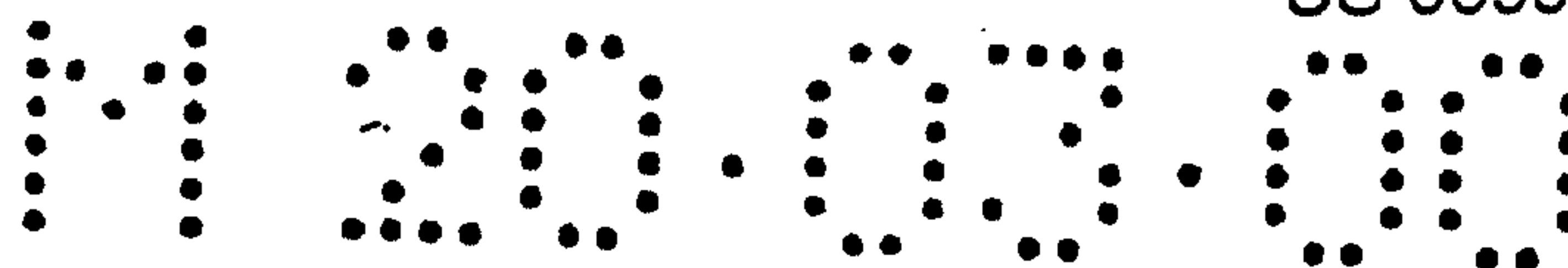
Analysis of the DAB to first adjacent interference at the edge of coverage
5 showed that the total DAB signal should be set at about -21 to -25 dB relative to its FM host power. This reduces the adjacent DAB interference ratio to the FM signal from about -24 dB to about -31 to -34 dB, assuming the D/U at the edge of coverage is 6 dB.

Although FM channel spacing in some countries is 100 kHz, these first
adjacents are geographically separated such that FM reception is not impaired within the
10 coverage area. Therefore this should pose no problem to the FM IBOC system. The DAB to DAB interference at 300 kHz spacing can impair performance on one sideband, but the CPC code is designed to tolerate this condition.

An OFDM technique has been described for IBOC DAB. An OFDM
signal consists of orthogonally spaced carriers all modulated at a common symbol rate.
15 The frequency spacing for rectangular pulse symbols (e.g., BPSK, QPSK, 8PSK or QAM) is equal to the symbol rate. For IBOC transmission of FM/DAB signals, a redundant set of OFDM sub-carriers is placed within about 100 kHz to 200 kHz on either side of a coexisting FM channel spectrum. The DAB power (upper or lower sideband) is set to about -25 dB relative to the FM signal. The level and spectral occupancy of the DAB
20 signal is set to limit interference to its FM host while providing adequate SNR for the DAB sub-carriers. First adjacent signals spaced at ± 200 kHz from the FM carrier can corrupt the DAB signal. However, at any particular location within a station's coverage area, it is unlikely that both first adjacents will significantly interfere with DAB. Therefore the upper and lower DAB sidebands carry the same redundant information such
25 that only one sideband is needed to communicate the information.

Inherent advantages of OFDM include robustness in the presence of multipath interference, and tolerance to non-gaussian short term noise or notches due to selective fading. The relatively long symbol integration times tend to "gaussianize" these short-term degradations.

30 Figure 7 is a greatly simplified block diagram of a digital audio broadcast system constructed in accordance with the invention. A transmitter 72 includes inputs 74 and 76 for receiving left and right channels of the program material. A separate data input 78 is included for an additional data signal, particularly for use with the all-digital



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modulation format of this invention. The transmitter includes an analog FM processor 80 and FM exciter 82, which operate in accordance with prior art processors, and excite to produce an analog FM broadcast signal on line 84. The inputs 74 and 76 are also fed to a coding processor 86 which converts the program material in complementary punctured convolution coded signals that are error corrected in block 88 and fed to a modulator 90 which applies the coded signals to the plurality of sub-carriers using orthogonal frequency division modulation. The output 92 of the modulator is summed with the signal on line 84 in summer 94 and sent to antenna 96. The receiver 98 receives the transmitted signal on antenna 100 and demodulates the signal in demodulator 102 to recover the program material and associated data, if included. The audio information is sent to a speaker 104 and additional data, if any, is provided to output 106, which may be fed to a display or other device that can further process the data.

The present invention provides a modulation format for an FM all-digital In-Band On-Channel (IBOC) Digital Audio Broadcast (DAB) systems. The all-digital IBOC modulation format is backward compatible with the FM hybrid IBOC system. The all-digital format embodiment of the present invention enables a substantially greater datacasting capacity. The modulation formats presented here allow a compatible transition for the broadcasters and listeners to migrate to digital signal virtual-CD audio quality while also providing a new datacasting medium.

The IBOC DAB modulation format of the present invention uses a complementary punctured convolution (CPC) coded version of the program material in two sidebands (upper sideband and lower sideband) which are potentially impaired by nearly independent interferers with independent fading. If one sideband is completely corrupted by a strong first adjacent FM signal in the vicinity of the receiver, the opposite sideband must be independently decodable at the receiver. Therefore each sideband must be coded with an independently decodable FEC code. However, when both sidebands contain useful information that is not completely corrupted by an interferer, then CPC codes provide additional coding gain above that achieved by power combining the two sides.

While the present invention has been illustrated in terms of its preferred embodiments, it should be understood that various changes may be made to the disclosed method and system without departing from the scope of the invention which is defined by the following claims. For example, while the above preferred embodiment shows the use

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of QPSK using CPC codes, various other modulation formats and code types may be used, such as 8PSK using 2/3 trellis code modulation followed by an optional Reed Soloman block code.

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CLAIMS:

1. A method of broadcasting comprising the steps of:

providing a first plurality of sub-carriers in an upper sideband of an FM radio channel, said upper sideband
5 ranging from about +100 kHz to +200 kHz from a center frequency of said radio channel;

providing a second plurality of sub-carriers in a lower sideband of said FM radio channel, said lower sideband ranging from about -100 kHz to -200 kHz from a center
10 frequency of said radio channel;

modulating a first group of said first plurality of sub-carriers with a digitally encoded version of a program signal to be transmitted;

modulating a first group of said second plurality
15 of sub-carriers with said digitally encoded version of said program signal to be transmitted;

providing a third plurality of sub-carriers in a central frequency band of said FM radio channel, wherein the power spectral density of the third plurality of sub-
20 carriers is less than the power spectral density of the sub-carriers in the upper and lower sidebands;

modulating said third plurality of sub-carriers with additional data; and

transmitting said first group of said first
25 plurality of sub-carriers, said first group of said second plurality of sub-carriers, and said third plurality of sub-carriers.

2. The method of claim 1, further comprising:

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modulating a second group of said first plurality of sub-carriers with a delayed digitally encoded version of program material; and

modulating a second group of said second plurality
5 of sub-carriers with said delayed digitally encoded version of program material.

3. The method of claim 2, wherein:

said second group of said first plurality of sub-carriers is positioned closer to the center of the FM radio
10 channel than said first group of said first plurality of sub-carriers; and

said second group of said second plurality of sub-carriers is positioned closer to the center of the FM radio channel than said first group of said second plurality of
15 sub-carriers.

4. The method of claim 2, wherein:

a sub-carrier in said upper sideband positioned farthest from the center of the FM radio channel is a reference sub-carrier; and

20 a sub-carrier in said lower sideband positioned farthest from the center of the FM radio channel is a reference sub-carrier.

5. The method of claim 2, further comprising:

modulating a third group of said first plurality
25 of sub-carriers with parity bits for said digitally encoded version of program material; and

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modulating a third group of said second plurality of sub-carriers with parity bits for said digitally encoded version of program material.

6. The method of claim 5, further comprising:

5 modulating a fourth group of said first plurality of sub-carriers with digitally encoded additional information; and

modulating a fourth group of said second plurality of sub-carriers with parity bits for said digitally encoded
10 additional information.

7. The method of claim 5, further comprising:

modulating a fourth group of said first plurality of sub-carriers with additional parity bits for said digitally encoded version of program material; and

15 modulating a fourth group of said second plurality of sub-carriers with additional parity bits for said digitally encoded version of program material.

8. The method of claim 5, wherein:

20 said third group of said first plurality of sub-carriers is positioned farther from the center of the FM radio channel than said first group of said first plurality of sub-carriers; and

said third group of said second plurality of sub-carriers is positioned farther from the center of the
25 FM radio channel than said first group of said second plurality of sub-carriers.

9. The method of claim 1, wherein:

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said digitally encoded version of said program material comprises a complementary convolutional code, with a first portion of said complementary convolutional code transmitted on said first group of said first plurality of sub-carriers and with a second portion of said complementary convolutional code transmitted on said first group of said second plurality of sub-carriers being independently decodable.

10. A transmitter for broadcasting in-band-on-channel digital audio signals, said transmitter comprising:

means for producing a first plurality of sub-carriers in an upper sideband of an FM radio channel, said upper sideband lying within a frequency band extending from about +100 kHz to about +200 kHz from a center frequency of said radio channel, a second plurality of sub-carriers in a lower sideband of said FM radio channel, said lower sideband lying within a frequency band extending from about -100 kHz to about -200 kHz from the center frequency of said radio channel, and a third plurality of sub-carriers in a central frequency band of the FM radio channel, wherein the third plurality of sub-carriers have a lower power spectral density than that of the first and second plurality of sub-carriers;

means for modulating a first group of said first plurality of sub-carriers with a coded version of program material;

means for modulating a first group of said second plurality of sub-carriers with said coded version of said program material;

means for modulating said third plurality of sub-carriers with additional data; and

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means for transmitting said first group of said first plurality of sub-carriers, said first group of said second plurality of sub-carriers, and said third plurality of sub-carriers.

5 11. The transmitter of claim 10, wherein:

a sub-carrier in said upper sideband positioned farthest from the center of the FM radio channel is a reference sub-carrier; and

10 a sub-carrier in said lower sideband positioned farthest from the center of the FM radio channel is a reference sub-carrier.

12. A receiver for receiving in-band-on-channel digital audio signals, said receiver comprising:

means for receiving a first plurality of sub-
15 carriers in an upper sideband of an FM radio channel, said upper sideband lying within a frequency band extending from about +100 kHz to about +200 kHz from a center frequency of said radio channel, said first plurality of sub-carriers being modulated with a complementary punctured convolution
20 coded encoded version of the program material; a second plurality of sub-carriers in a lower sideband of said FM radio channel, said lower sideband lying within a frequency band extending from about -100 kHz to about -200 kHz from the center frequency of said radio channel,
25 said second plurality of sub-carriers being modulated with a complementary punctured convolution coded encoded version of the program material; and a third plurality of sub-carriers, said third plurality of sub-carriers being modulated with additional data having a lower power spectral density than
30 that of said first and second plurality of sub-carriers;

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means for demodulating a first group of said first plurality of sub-carriers, a first group of said second plurality of sub-carriers, and said third plurality of sub-carriers; and

5 means for outputting said program material obtained from demodulating said first group of said first plurality of sub-carriers and said first group of said second plurality of sub-carriers, and additional data obtained from demodulating said third plurality of sub-
10 carriers.

13. The receiver of claim 12, further comprising:

means for differentially detecting a reference sub-carrier in said upper sideband; and

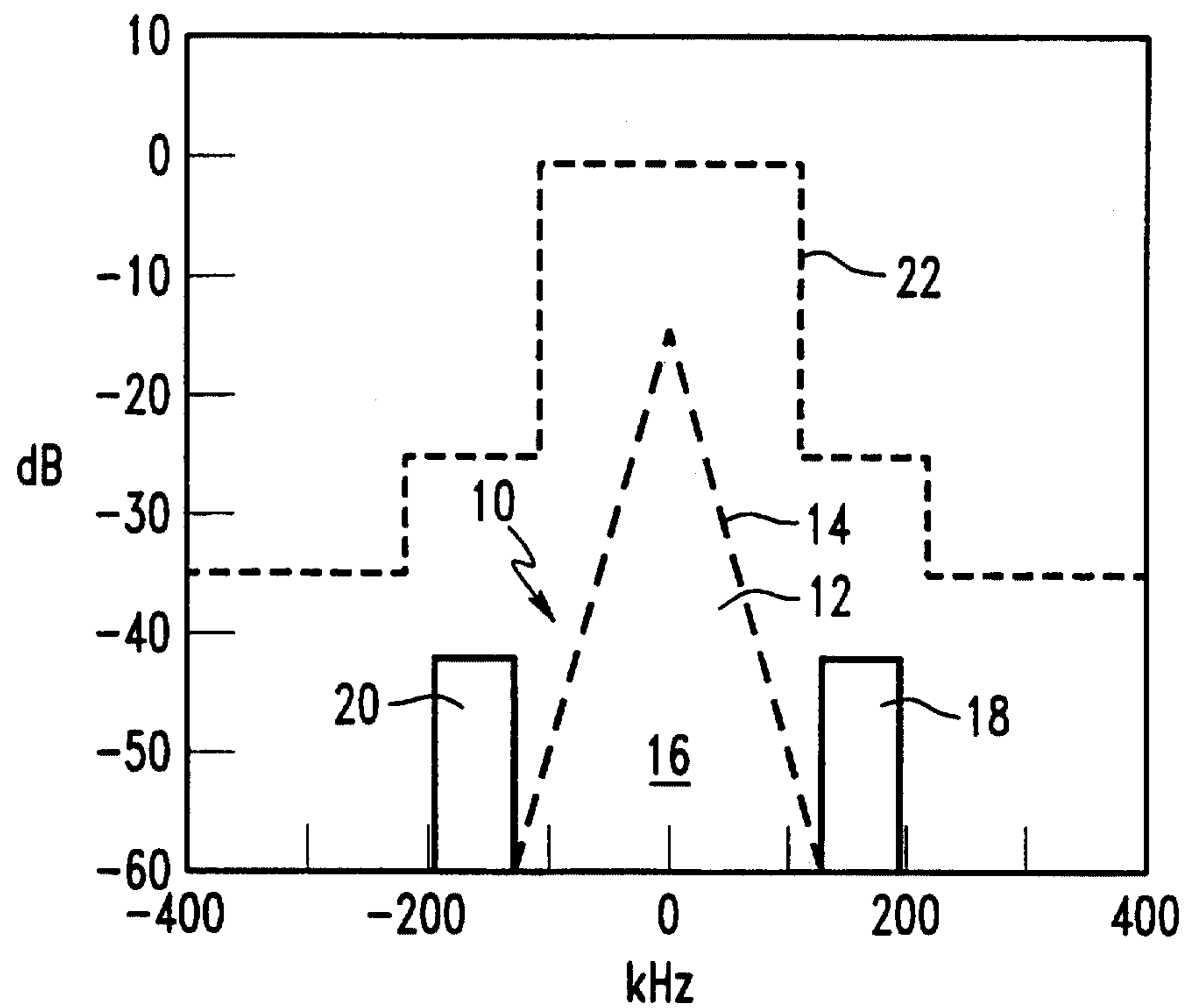
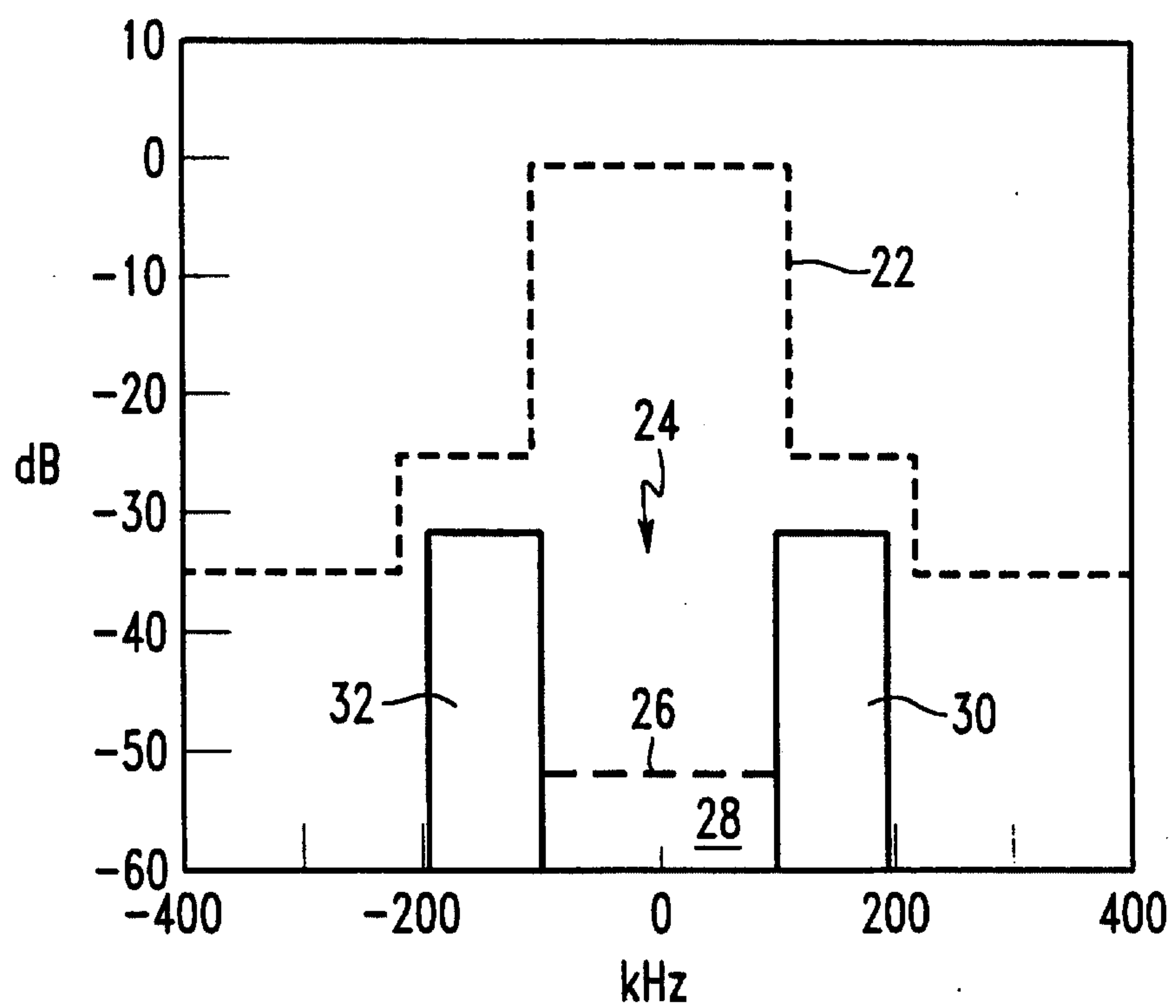
means for differentially detecting a reference
15 sub-carrier in said lower sideband.

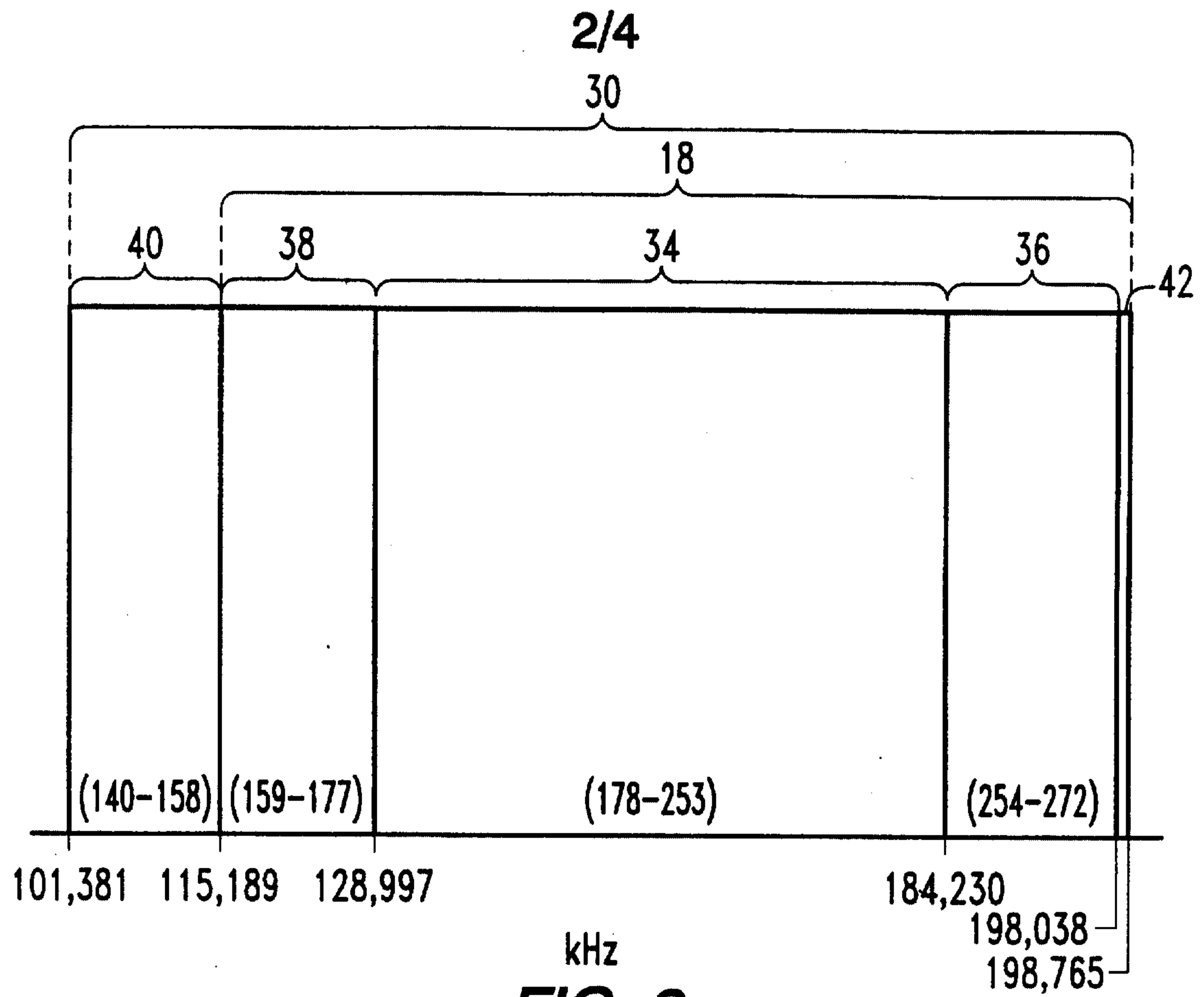
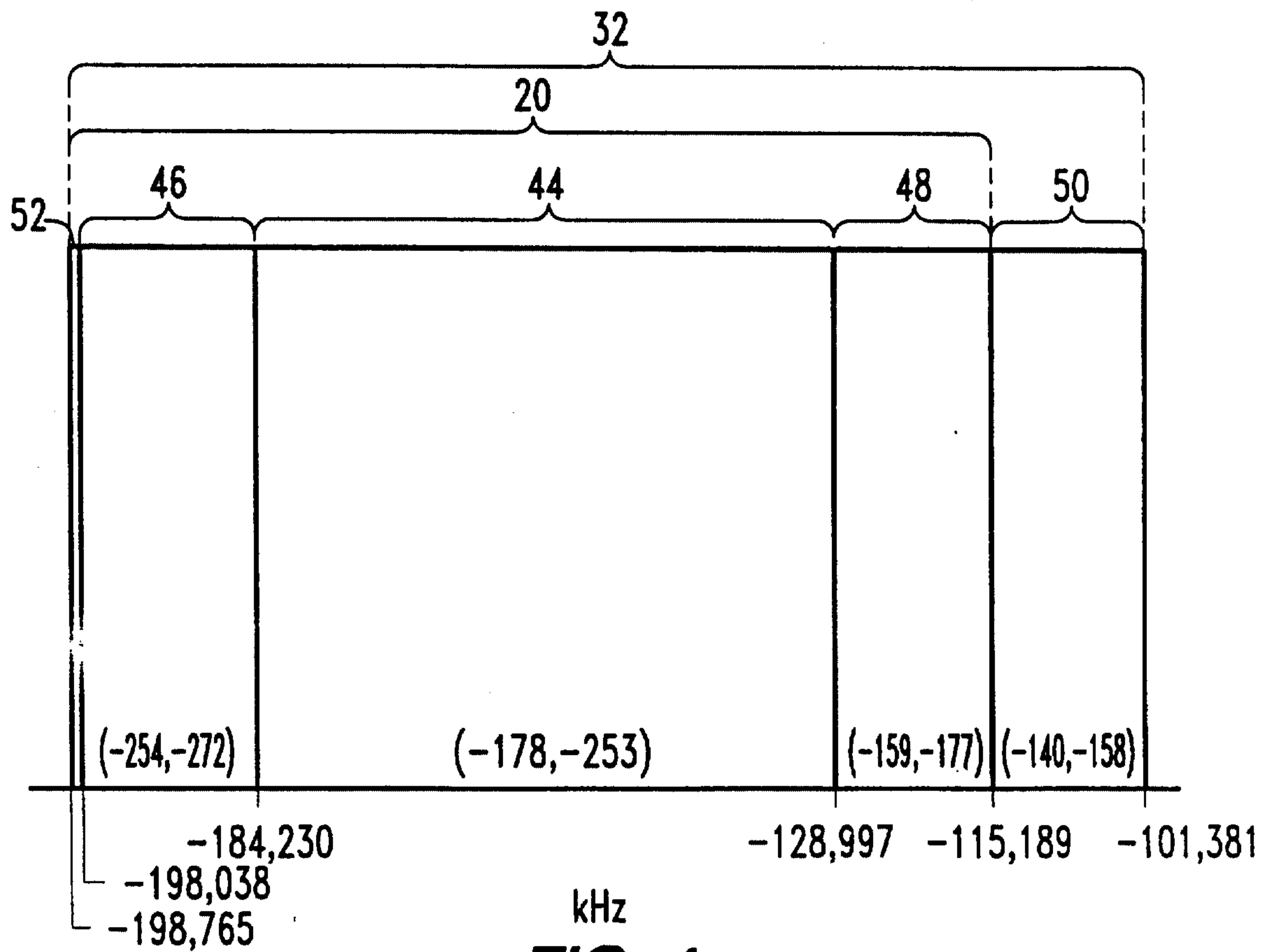
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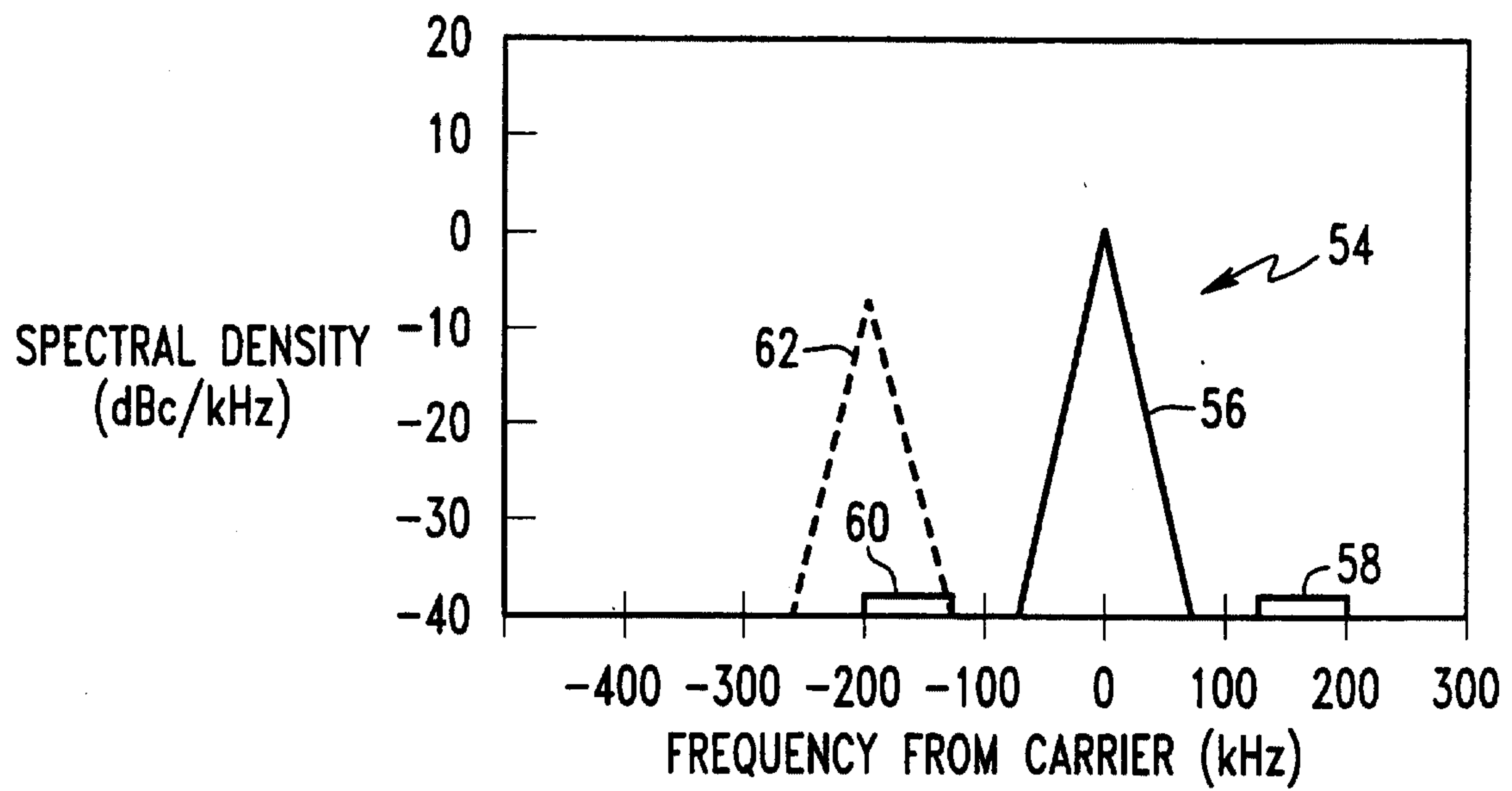
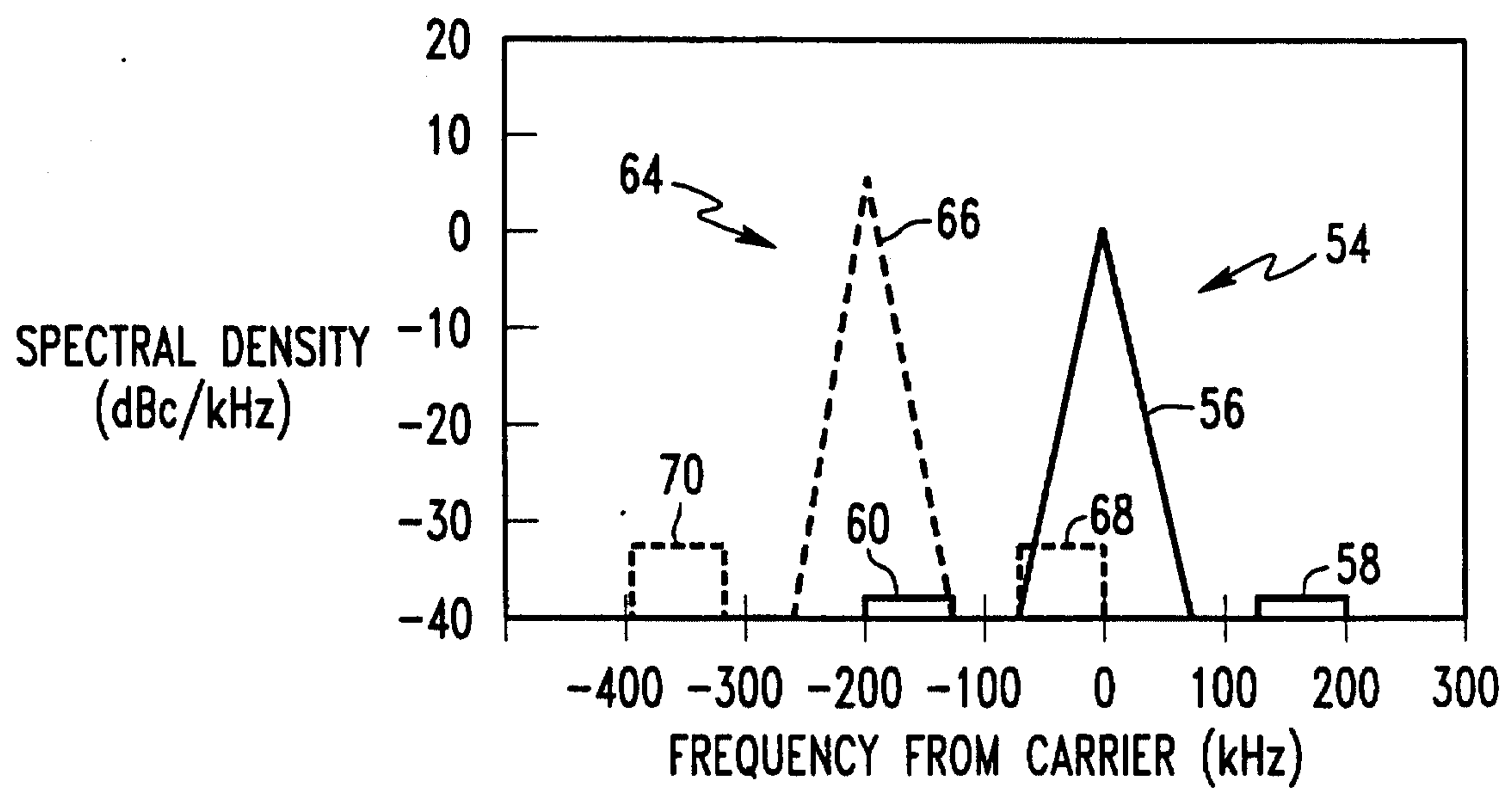
PATENT AGENTS

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**FIG. 1****FIG. 2**

**FIG. 3****FIG. 4**

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**FIG. 5****FIG. 6**

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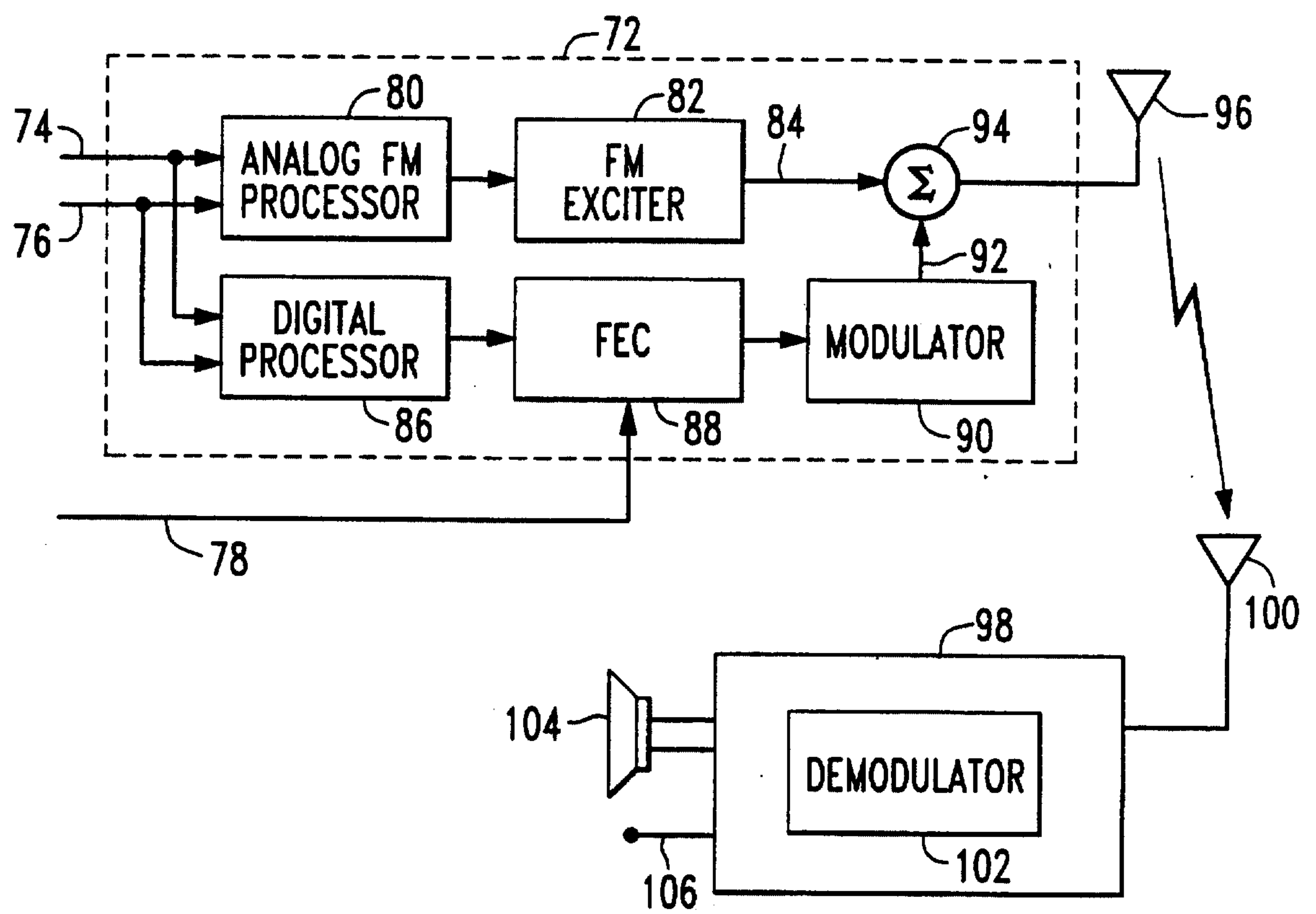


FIG. 7

