

[54] IMPEDANCE PLETHYSMOGRAPH 3,566,233 2/1971 Kahn 128/2.1 Z
 3,593,718 7/1971 Krasner et al. 128/419 P
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 [22] Filed: Aug. 15, 1973
 [21] Appl. No.: 388,541

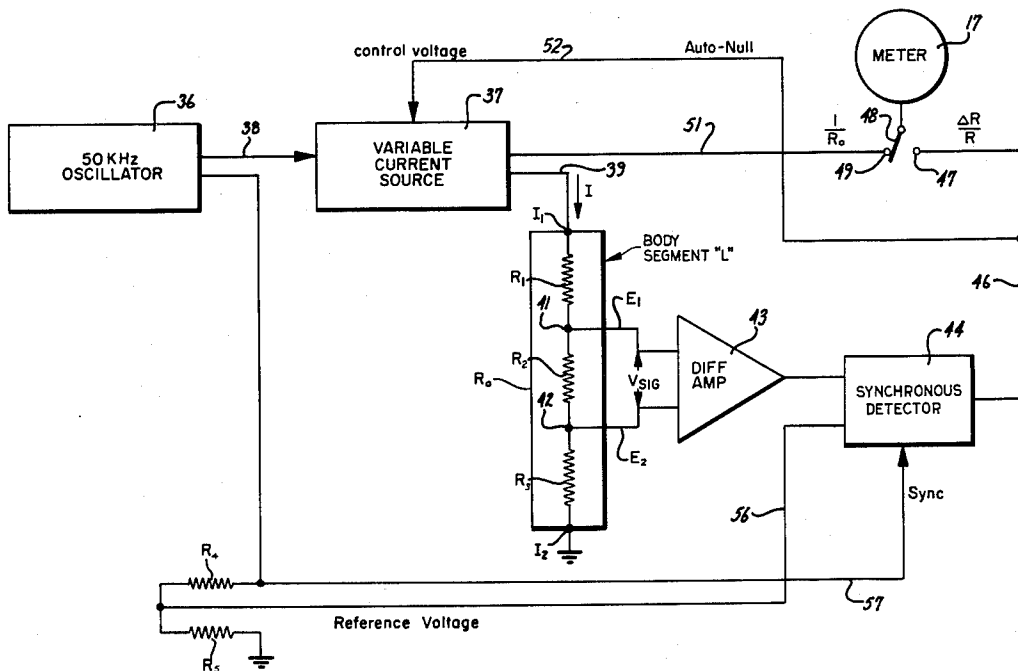
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Related U.S. Application Data
 [63] Continuation of Ser. No. 190,900, Oct. 20, 1971, abandoned.
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 [51] Int. Cl. A61b 5/04
 [58] Field of Search 128/2.05 P, 2.05 R, 2.05 V, 128/2.08, 2.1 M, 2.1 P, 2.1 R, 2.1 Z, 145.5-148, 419 P, 419 R, 421, 422, 2.05 F, 2 L

[57] **ABSTRACT**
 Plethysmograph having current and voltage electrodes with a variable current source connected to the current electrodes to supply varying current to a biological segment to provide a voltage which is utilized to provide a signal which represents a percent of change of resistance of the biological segment.

[56] **References Cited**
 UNITED STATES PATENTS
 3,149,627 9/1964 Bagno 128/2.1 Z

5 Claims, 4 Drawing Figures



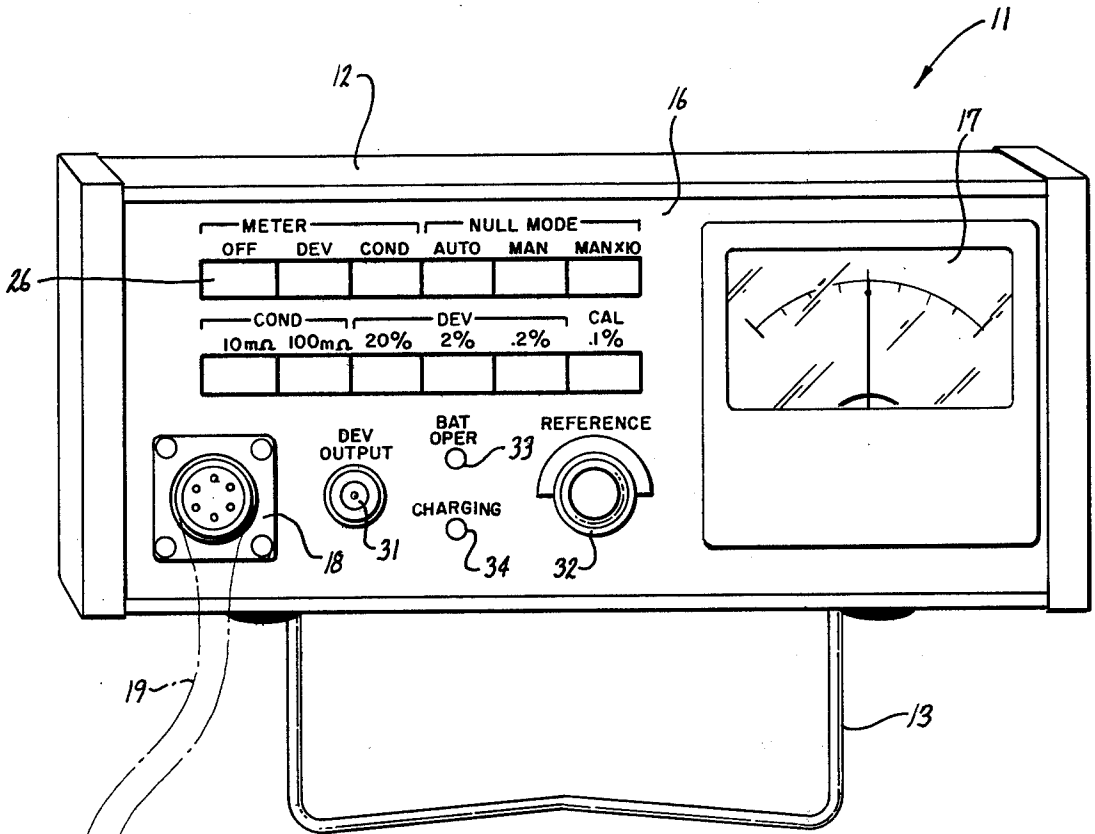


Fig. 1

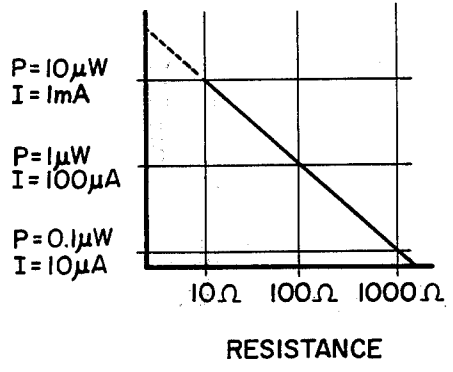
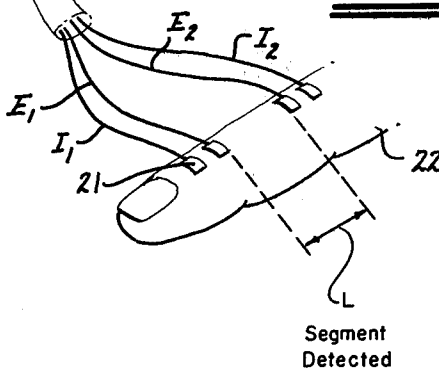


Fig. 3

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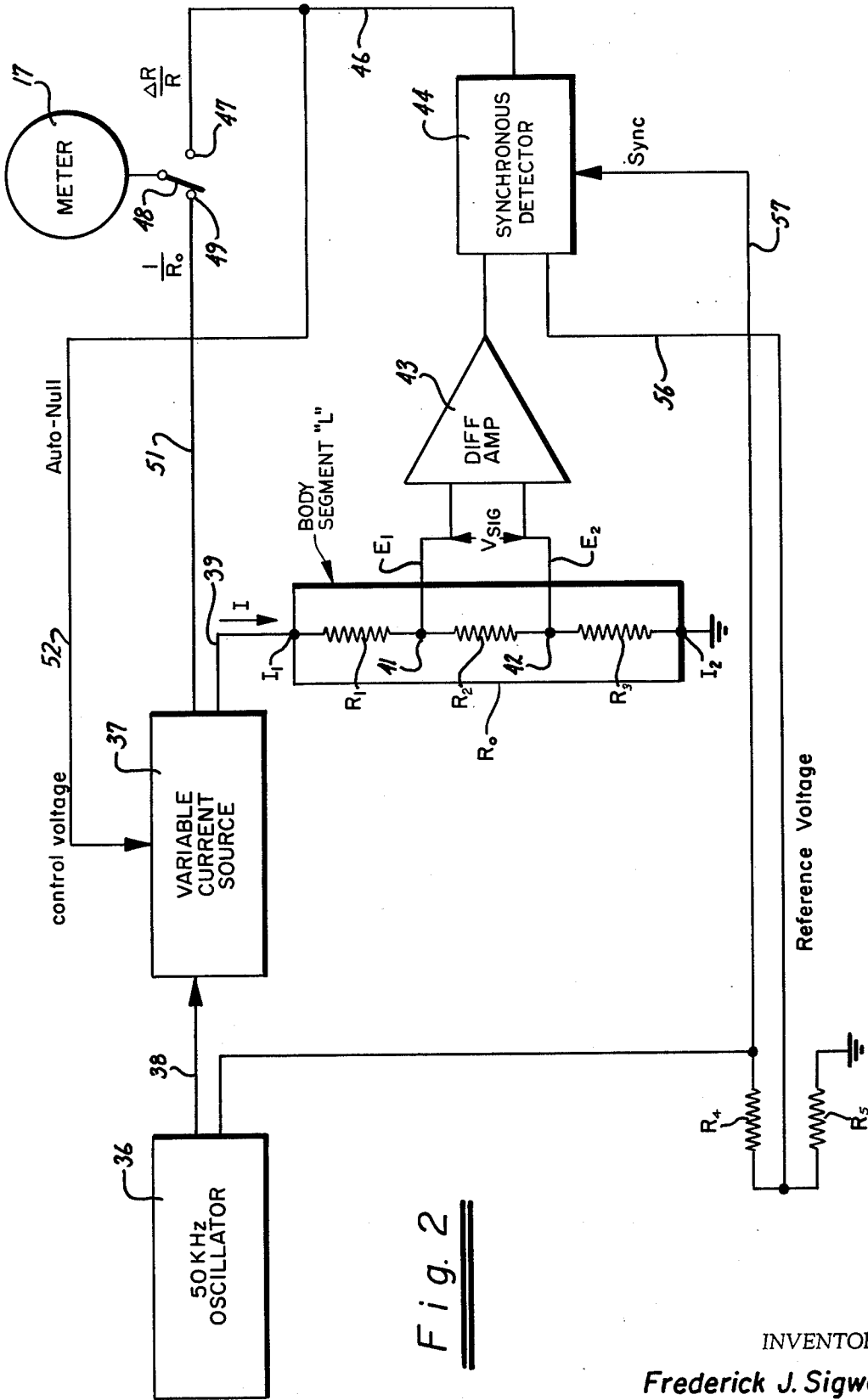


Fig. 2

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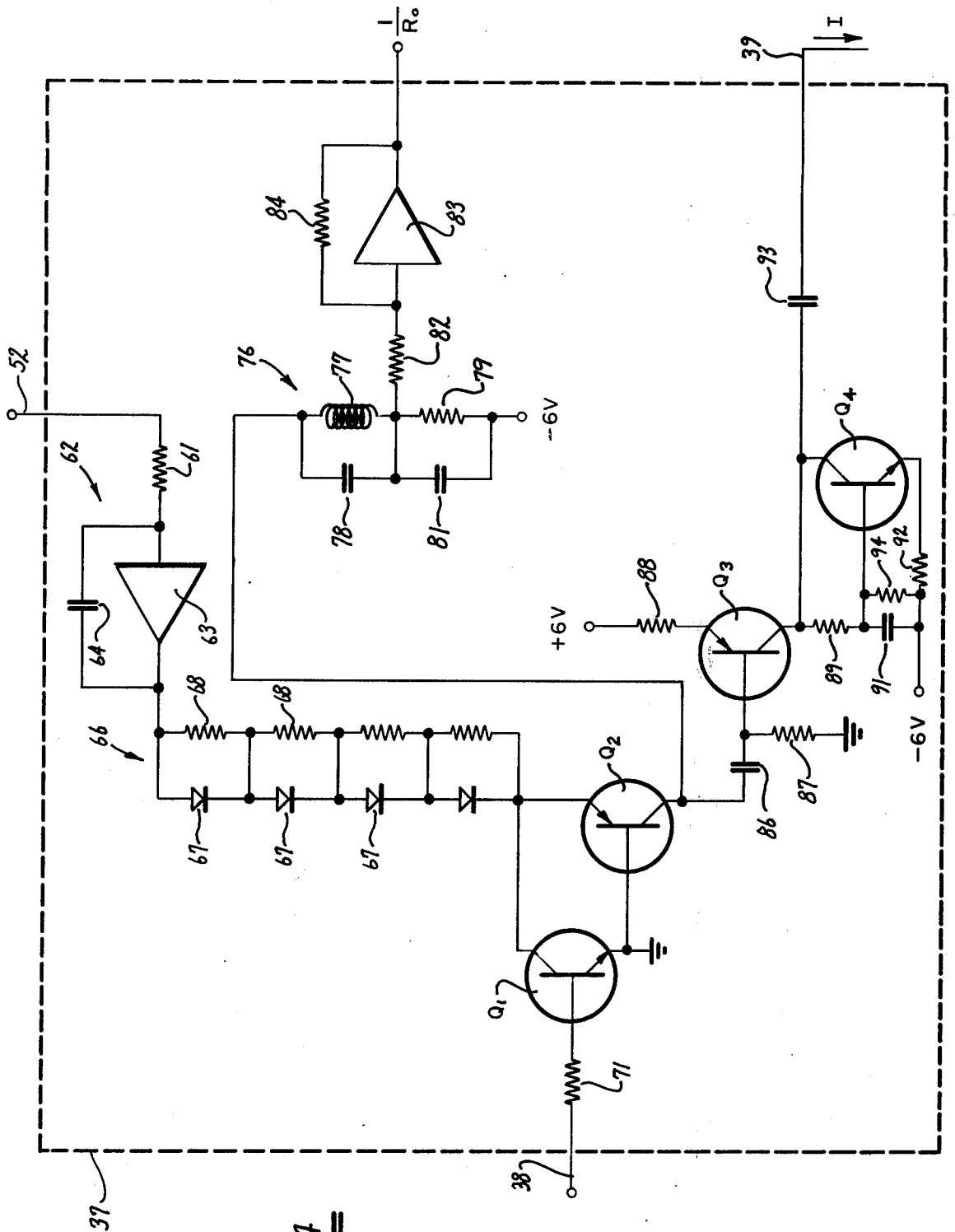


Fig. 4

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IMPEDANCE PLETHYSMOGRAPH

CROSS REFERENCES

This is a continuation of application Ser. No. 190,900, filed Oct. 20, 1971, now abandoned.

BACKGROUND OF THE INVENTION

In U.S. Pat. No. 3,149,627, there is disclosed a plethysmograph which can be utilized for making impedance measurements on biological segments. A bridge arrangement is provided which must be manually nulled. Because the variations in resistance in the biological segments are very small as, for example, on the order of 0.1% or less, it is very difficult to null the meter because of drift. There is, therefore, a need for a new and improved plethysmograph.

SUMMARY OF THE INVENTION AND OBJECTS

The plethysmograph consists of first and second current electrodes and first and second voltage electrodes adapted to be applied in pairs across the biological segment with one current electrode adjacent one voltage electrode and the other current electrode adjacent the other voltage electrode. A high frequency oscillator is provided. A variable current source is connected to the output of the oscillator and to the current electrodes so that an oscillatory current is supplied to the current electrodes. Amplifier means is coupled to the voltage electrodes for measuring the difference voltage developed across the voltage electrodes. Means is coupled to the amplifier and to the high frequency oscillator for detecting the output of the amplifier using the output of the oscillator as a reference. Means is provided for supplying the output of the detector means to provide a control voltage for the variable current source to automatically null the same. The output of the detector produces a signal which represents a percent of change of the resistance of the biological segment.

In general, it is an object of the present invention to provide an impedance plethysmograph which is quite accurate and reliable.

Another object of the invention is to provide a plethysmograph of the above character which is relatively easy to operate.

Another object of the invention is to provide a plethysmograph which can provide a readout in either percent of deviation or in conductance.

Another object of the invention is to provide a plethysmograph of the above character in which an automatic nulling function is provided.

Another object of the invention is to provide a plethysmograph of the above character which can be battery operated.

Another object of the invention is to provide a plethysmograph of the above character in which the readout is in percent variation rather than an absolute resistance indication.

Additional objects and features of the invention will appear from the following description in which the preferred embodiment is set forth in detail in conjunction with the accompanying drawings.

BRIEF DESCRIPTION OF THE DRAWING

FIG. 1 is a perspective view of an impedance plethysmograph incorporating the present invention connected to a body segment.

FIG. 2 is a block diagram of the impedance plethysmograph.

FIG. 3 is a graph showing current flow for various values of resistance.

FIG. 4 is a detailed circuit diagram of the variable current source in FIG. 2.

DESCRIPTION OF PREFERRED EMBODIMENT

The impedance plethysmograph incorporating the present invention is in the form of an electronic instrument 11 as shown in FIG. 1 which includes a case 12. The case 12 is provided with a stand 13 so that the front panel 16 of the instrument can be elevated above the top surface of a table or other support provided for the instrument. A meter 17 is mounted in the front panel and is capable of measuring percent deviation according to the deviation scaling selected as hereinafter described. It is also capable of reading millimhos according to the scaling mode selected as hereinafter described. The front panel is also provided with a connector 18 which is connected to a cable 19. The cable 19 is provided with four conductors which have been identified as first and second current electrodes I_1 , I_2 and first and second voltage electrodes E_1 , E_2 shown in FIG. 1 and which are preferably color coded in a suitable manner. For example, the conductors I_1 and I_2 can be color coded red, whereas the conductors E_1 and E_2 can be color coded black. Each of the conductors which are provided, I_1 , I_2 and E_1 , E_2 is connected to relatively thin metallic strips 21 formed of a suitable material such as aluminum. These strips serve as electrodes which are secured to the biological segment L of which the measurements are to be made such as a human body segment in the form of a finger 22 by suitable means such as an adhesive tape (not shown). The electrodes are arranged in pairs across the body segment L with one current electrode adjacent one voltage electrode and the other current electrode adjacent the other voltage electrode. Thus, electrodes I_1 and E_1 are adjacent to each other and the electrodes E_2 and I_2 are adjacent to each other.

A total of 12 pushbuttons 26 are arranged in two spaced parallel rows, the function of which will hereinafter be described. Three output jacks are provided as a part of the instrument in which one of the jacks 31 is provided on the front panel. These jacks are simultaneously active. The front jack 31 provides deviation, that is, recording resistance change related to blood or gas volume variation. The two rear jacks (not shown) provide a derivative signal which is the rate of change of the deviation signal and a conductance signal which monitors the relatively slow changes in the conductance of the examined segment during change in posture or in response to vasodilation or vasoconstriction. A reference potentiometer 32 is mounted on the front panel. In addition, two lights 33 and 34 are provided. Light 33 indicates that the instrument is on battery operation and light 34 indicates that the battery is being charged.

A block diagram of the electronic circuitry within the case 12 is shown in FIG. 2. As shown therein, it consists of an oscillator 36 operating at a suitable frequency as, for example, 50 kilohertz (KHz). The oscillator 36 supplies a timing signal through a conductor 38 to a variable current source 37. The variable current source supplies on output line 39 an output current which is related in amplitude to the control voltage which is sup-

plied to the current source. This output current is identified by the letter I and is fed into the body segment L which is represented by the resistance R_0 consisting of three resistors R_1 , R_2 and R_3 in series. R_1 and R_3 represent the contact resistance between the electrodes I_1 and I_2 and the skin of the body member to which the electrodes are secured. Two taps 41 and 42 are provided on opposite sides of the resistor R_2 and correspond to the two electrodes connected to the conductors E_1 and E_2 . The conductors E_1 and E_2 are connected to a differential amplifier 43 which measures the difference voltage developed across the body segment resistance R_2 and is indicated as V_{sto} . The output of the differential amplifier 43 is supplied to a synchronous detector 44. The detector 44 has an output 46 which is connected to a terminal 47 of a switch 48 that has an additional terminal 49. The switch 48 is connected to the meter 17. When the switch 48 is connected to the terminal 47, it will be reading the resistance changes in the body segment itself which is supplied by the synchronous detector. When the switch is connected to the terminal 49 which is connected to the variable current source 37 by conductor 51, the meter will be reading the conductance of the body segment itself by measuring the amount of current that the current source is putting into the body segment. The output of the synchronous detector is also supplied to the variable current source by a conductor 52 to provide a control voltage as hereinafter described.

A reference voltage is supplied to the synchronous detector 44 through the conductor 56 which is connected between two serially connected resistors R_4 and R_5 that form a voltage divider network connected between ground and the oscillator 36. A synchronizing voltage is supplied to the synchronous detector 44 from the oscillator 36 by a conductor 57.

By way of example, the reference voltage supplied to the synchronous detector can be on the order of 100 millivolts. If the voltage supplied by the differential amplifier is greater than 100 millivolts, the output of the synchronous detector is positive and if the voltage from the differential amplifier is less than 100 millivolts, then the output from the detector 44 is negative. Thus, the detector 44 provides a d.c. voltage which gives information as to whether or not the output of the differential amplifier is above or below the null point. Thus, there is supplied a control voltage to the variable current source which performs an automatic nulling function. Thus, when the output from the synchronous detector 46 is negative, the variable current source 47 will be instructed to supply an additional voltage across the body segment to increase the current flow through the body segment. Conversely, if the output of the synchronous detector 44 is positive, the variable current source will be instructed to supply a lesser voltage to the body segment to thereby decrease the current flow through the body segment. It can be seen that this is similar to nulling a bridge where the differential signal voltage developed across the body segment is compared with an internal reference voltage. In the present nulling system, the current supplied to the body segment itself is varied which means that the voltage level developed across the body segment is a long term constant, although short-term small changes in the body segment resistance will show up as variations superimposed on the 100 millivolt output from the differential amplifier

and, therefore, will show up as instantaneous variations in the output of the synchronous detector 44.

The control of the current passing through the body segment L to provide a nominally constant voltage makes it possible for the output read on the meter 17 to represent the percent of value change rather than absolute resistance. Thus, for example, if the body segment resistance is 100 ohms, it would develop a 10 millivolt signal across the body segment which requires 0.1 of a milliamperere to pass through the body segment resistance. The 10 millivolts across the body segment is picked off by the conductors E_1 and E_2 and amplified as, for example, ten times in the differential amplifier 43 to provide the 100 millivolt output to the synchronous detector 44. The relatively slow response time of the control voltage ensures that the current level will remain constant for short-term variations. If there is an instantaneous change of resistance in the body segment, as for example, by 0.1 ohm, this would mean a change in the signal voltage measured across the body segment of the 0.1 ohm times a 0.1 of a milliamperere or 10 microvolts of change at the output of the synchronous detector 44. The synchronous detector 44 would then amplify by a suitable gain such as 1,000 to provide 100 millivolts of signal which is supplied to the meter 17 which would show a variation of 0.1%, i.e., 0.1 of an ohm out of 100 ohm.

If the segment resistance is 50 ohms instead of 100 ohms, then in order to obtain 100 millivolts at the input to the detector or 10 millivolts at the input to the differential amplifier, the feedback loop would cause the current source to supply 0.2 of a milliamperere of current to the body segment. Then if there is 0.1% change in the resistance of 50 ohms which would be 0.05 ohms, the current level would be double which would provide the same 10 microvolts of signal being supplied to the differential amplifier and thus eventually would result in the same 100 millivolt signal to the meter 17.

In FIG. 3, there is a curve which shows the current and power involved in the measurement process with current and power variations. The X-axis shows body segment resistances from 10 ohms to 1 k. The Y-axis shows both the power and current involved in the body segment at each resistance. A logarithmic scale is used in both axes. As shown in the graph, there were 100 microamperes of current flowing through the body segment when the body segment had a resistance of 100 ohms. If the body segment resistance instead of being 100 ohms is 10 ohms, the feedback loop would change the variable current source 37 until the current level would be 1 milliamperere and the power would be 10 microwatts.

The chart shows that the current level is inversely proportional to the body segment resistance. It also shows that only very small current and power levels are required for making the desired measurements, so small that no sensation is felt or health hazard is encountered by use of the device.

Of the blocks shown in FIG. 2, the oscillator 36, the differential amplifier 43 and the synchronous detector 44 are of a conventional type. By way of example, the synchronous detector 44 can be like the phase sensitive detector 15 disclosed in U.S. Pat. No. 3,149,627.

A detailed circuit diagram of the variable current source 37 is shown in FIG. 4. As shown therein, the control voltage from the detector 44 is supplied through the lead 52 to an integrator 62 consisting of an

amplifier 63, a resistor 61 and a capacitor 64 which is connected between the input and the output of the amplifier 63. The output of the integrator 62 is supplied to a log function generator 66 of a type well known to those skilled in the art which consists of a plurality of serially connected diodes 67 and a plurality of serially connected resistors 68 with one of the resistors 68 being connected across each of the diodes 67. The log function generator 66 produces a voltage across the same which is the logarithm of the current flowing through the log function generator. The use of the log function generator makes it possible for the system to have a constant response time regardless of the body segment resistance and the current level that is fed into the segment L. Transistors Q1, Q2, Q3 and Q4 are provided in the circuit shown in FIG. 4. They are all of a conventional type and each includes base, collector and emitter elements as shown.

The 50 KHz signal from the oscillator 36 is fed onto the line 38 through a resistor 71 to the base of a transistor Q1 which serves as a switching transistor for switching the signal supplied by the log function generator away from the transistor Q2 for nominally half of the time. The other half of the time Q1 is turned off under the control of the signal supplied on the line 38 and all of the current through the function generator is supplied to the emitter of the transistor Q2. The result is that a square wave current output is supplied by the collector of Q2.

The square wave output from the collector of the transistor Q2 is supplied to a tuned circuit 76 which is resonant at the same frequency as the oscillator 36 and consists of an inductance 77 and a capacitor 78 connected across the inductance. The tuned circuit 76 converts the square wave into a uniform sine wave. The current exciting the tuned circuit 76 is also supplied through a sampling resistor 79 for the conductance readout by the meter 17. A capacitor 81 connected across the resistor 79 serves as a bypass capacitor. The resistor 79 is connected to a -6 volt source as indicated in FIG. 4. The resistor 79 measures the average value of current flowing through the log function generator 66 and through the transistor Q2 and through the inductance 77. This current is proportional to the output current from the entire current source 37. This current is supplied through a resistor 82 to an operational amplifier 83 which has a resistor 84 connected between the input and the output of the amplifier 83. The output of the amplifier is connected to a terminal which is identified as $1/R_0$ which is adapted to be connected to the meter 17 to provide a conductance readout on the meter 17.

The 50 KHz sine wave signal developed across the tuned circuit 76 is also supplied through a capacitor 86 to the base of a current source transistor Q3. The base is also connected to ground through a resistor 87. The emitter of the transistor is connected to a +6 volt supply voltage through a resistor 88.

The collector of the transistor Q3 is connected through a resistor 89 to the base of a transistor Q4. The base of Q4 is also connected through a capacitor 91 and a resistor 94 to a -6 volt source as indicated in FIG. 4. The emitter of the transistor Q4 is connected through a resistor 92 to the -6 volt source. The collector is coupled through a capacitor 93 to the current output line 39 which is connected to the resistor R_0 as shown in FIG. 2.

The transistor Q4 and its associated components, capacitors 91 and 93, serve as a d.c. current source to provide operating bias for the transistor Q3. The output impedance of the current source is high, limited only by the collector impedances of Q3 and Q4 and the value of resistor 89.

The instrument is provided with a battery which is not shown. Means is also provided in the instrument for charging the battery when the instrument is not in use. Circuitry is provided which is associated with the pushbuttons 26 on the front panel so that switching from the off button to either the conductance or the deviation button disconnects the instrument from the a.c. power supply and switches to battery power. The battery operating light 33 will turn on if the battery is operational. When the instrument is not in use, it can be connected to an a.c. power outlet. Pushing the off button will activate the battery charging circuit and the light 34 will be lit.

Additional circuitry is provided which is connected to the other pushbuttons which are utilized in connection with the operation of the instrument as hereinafter described.

Let it be assumed that it is desired to make an impedance measurement of a body segment as, for example, of one finger as shown in FIG. 1. The electrodes are applied by adhesive tape as hereinbefore described and positioned in the manner shown in FIG. 1.

Now let it be assumed that it is desired to operate the instrument in the manual balancing mode. If the segmental resistance ($E_1 - E_2$) is below 100 ohms, the deviation button, the manual button and the 20% deviation mode button should be pushed. The reference potentiometer 32 is adjusted for a zero center reading on the meter 17. Alternatively, the 2% and 0.2% buttons can be pushed to obtain finer tuning by the use of the reference potentiometer 32.

When the segmental resistance is above 100 ohms, the "manual $\times 10$ " button is pushed and the reference potentiometer is switched from zero to a 1,000 ohm scaling. The deviation can be read directly from the meter 17 according to the scaling (20%, 2%, 0.2%), or the deviation may be displayed on an oscilloscope or recorded on a laboratory recorder by connecting the same to the deviation jack 31 provided on the front panel. When the conductance button is pushed, the conductance in millimhos can be read on the meter. The reference potentiometer 32 reads directly in ohms. Thus, in a manual mode, the potentiometer 32 reads from zero to 100 ohms. In the manual $\times 10$ mode, the potentiometer 32 reads zero to 1,000 ohms. The reading on the reference potentiometer 32 corresponds with the conductance reading on the meter 17.

Let it be assumed that automatic null operation is desired. When this is the case, the deviation button, the automatic null button and one of the deviation scaling buttons (20%, 2%, 0.2%) should be pushed. The deviation can be read directly from the meter 17 or displayed on a laboratory recorder from the output jack 31 as hereinbefore described. When the conductance button is pushed, the conductance can be read directly off the meter or displayed on a laboratory recorder from the output jack provided on the back side of the instrument. Nulling is performed automatically in the automatic mode and the reference potentiometer is not used in the automatic null configuration.

In practice, it may be desirable to determine the resistance from the reference potentiometer 32 by using the instrument either in manual or manual $\times 10$ mode and then switching to the automatic null mode to obtain conductance.

During these measurements, it should be appreciated that since the instrument is being operated from a battery, there is no possibility of supplying voltage to the patient. In addition, it should be appreciated that the voltage and currents which are being applied to the biological segments of the patient are so low that they will not cause any harm.

In the diagram shown in FIG. 2, the meter 17 is connected to $1/R_o$ terminal which corresponds to the condition when the conductance button is pushed. The meter 17 registers the body segment conductance which is reciprocal of the resistance. Thus, when automatic nulling is taking place, the current passing through the body segment is inversely proportional to the segment resistance. Therefore, the current supplied into the body segment is proportional to the segment conductance. The monitoring voltage which is supplied at the $1/R_o$ terminal shows the amount of current which is being supplied to the body segment which can be read out directly on the meter 17. The meter 17 is provided with two scales. The meter normally rests in the center of the two scales. One scale is in deviation and provides deviation in percent in both directions. The second scale is calibrated in millimhos of conductance from zero to 10 or zero to 100 millimhos depending upon which pushbutton is pressed.

The measurements hereinbefore described by the use of the four electrode impedance plethysmograph makes it possible to measure changes in electrical resistance related to blood and gas volumes in biological segments. During each cardiac cycle as blood is distributed from the heart to various organs and tissues, the new blood volume results in a change in resistance which increases current conduction from the plethysmograph and results in a pulsatile voltage output which can be detected and recorded for qualitative and quantitative evaluation.

In a similar manner, increases and decreases in gas volume during respiratory activity result in variations in electrical resistance of the thoracic segment which can be detected and recorded as an index of pulmonary function.

The high frequency current from the plethysmograph penetrates through superficial and deep body structures, and each segment defined by the placement of electrodes may be interpreted as a homogeneous volume conductor having a measure of resistance referred to as the base line resistance.

This measured segmental resistance is influenced by the pulsatile, gas or blood volume variations as well as increases or decreases in total blood or body tissue fluid within the segment which may occur during vasoconstriction, vasodilation or postural variation. The impedance plethysmograph is capable of recording both the pulsatile changes (deviation) and variations in base line resistance which is registered as the reciprocal conductance.

In summary, the basic principle utilized by the impedance plethysmograph depends upon changes in the electrical characteristics of tissue during each cardiac or respiratory cycle as a new volume of blood or gas enters the segment. In terms of pulsatile blood volume

changes, the conductive properties of the segment and the new volume of blood parallel each other during systole and create a change in resistance which is proportional to the pulse volume. Variations in resistance occur in the thoracic cavity during respiration as a result of the difference between blood and gas volumes in the chest during respiratory activity.

The impedance plethysmograph offers the clinician, medical specialist, educator or biologist and researcher a reliable, easy to use and highly versatile electronic instrument. The plethysmograph output can be connected to oscillographs and pen recorders for simultaneous recording of ECG and changes in blood pulse volume and respiratory volume of biological segments. The use of the impedance plethysmograph does not cause trauma to the patient. The system does not require venous occlusion or methods for detecting small and large changes in electrical resistance related to blood pulse volume variations and is useful in defining pulmonary gas volume when electrodes are positioned over the thorax area.

The impedance plethysmograph can be useful in medical schools for teaching principles of cardiovascular and pulmonary physiology, in hospitals for the management of patients pre- and post-operatively, in animal and clinical diagnostic laboratories for drug evaluation, and in offices of the general practitioner for early detection of cardio-vascular disorders. The deviation output (ΔR) is located on the front and rear panels. The base line (conductance) and the derivative $d(\Delta R/R)/dt$ outputs are available at the rear panel for simultaneous recordings. The selector pushbuttons activate the meter for additional monitoring of all three simultaneous recorder outputs.

It is apparent from the foregoing that there has been provided an impedance plethysmograph which has many desirable features. It is portable and it is simple to operate. It can be operated in either manual balancing or automatic balancing modes.

The incorporation of a constant current source of variable output as a nulling element in the impedance plethysmograph is particularly important because it makes it possible to provide a readout which is a percent of the percent variation rather than an absolute resistance indication on a meter.

I claim:

1. In a plethysmograph first and second current input electrodes, first and second voltage drop measuring electrodes adapted to be applied in pairs across a biological segment with one current electrode adjacent one voltage electrode and the other current electrode adjacent the other voltage electrode, an oscillator having a high frequency output, a variable current source connected to the oscillator and to the first and second current input electrodes for receiving a timing signal from the oscillator so that an oscillatory output current is provided to the current input electrodes, amplifier means coupled to said first and second voltage drop electrodes for receiving the voltage developed across said voltage drop electrodes and producing an output responsive thereto, means for providing a predetermined reference voltage, means coupled to the amplifier means for detecting the output of the amplifier means and providing an output responsive thereto, said means for detecting receiving said predetermined reference voltage, means for supplying the output of the means for detecting to the variable current source to

provide a control voltage for the variable current source so that the output from the variable current source provides an output from the amplifier means matching said predetermined reference voltage, the output of said means for detecting providing a percent deviation signal which represents a percent of change of resistance value of the biological segment, indicating means for receiving the percent deviation signal from the means for detecting for indicating the percent of change of resistance, said indicating means also being for indicating conductance of the biological segment, and means for alternately connecting said indicating means for receiving the output of the variable current source.

2. In a plethysmograph first and second current input electrodes, first and second voltage drop measuring electrodes adapted to be applied in pairs across a biological segment with one current electrode adjacent one voltage electrode and the other current electrode being adjacent the other voltage electrode, an oscillator having a high frequency output, a variable current source connected to the oscillator and to the first and second current input electrodes for receiving a timing signal from the oscillator so that an output oscillatory current is provided to the current input electrodes, amplifier means coupled to said first and second voltage drop electrodes for receiving the voltage developed across said voltage drop electrodes and producing an output responsive thereto, means for providing a predetermined reference voltage, means coupled to the amplifier means for detecting the output of the amplifier means and providing an output responsive thereto, said means for detecting receiving said predetermined reference voltage, means for supplying the output of the means for detecting to the variable current source to provide a control voltage for the variable current source so that the output from the variable current source provides an output from the amplifier means matching said predetermined reference voltage, the output of the means for detecting providing a percent deviation signal which represents a percent of change of resistance value of the biological segment, and indicating means calibrated in terms of two quantities, one of said quantities being conductance and the other of said quantities being percent of resistance change, together with means for connecting said indicating means alternately to said variable current source and said means for detecting respectively.

3. In a plethysmograph first and second current input electrodes, first and second voltage drop measuring electrodes adapted to be applied in pairs across a biological segment with one current electrode adjacent one voltage electrode and the other current electrode being adjacent the other voltage electrode, an oscillator having a high frequency output, a variable current source connected to the oscillator and to the first and second current input electrodes for receiving a timing signal from the oscillator so that an output oscillatory current is provided to the current input electrodes, amplifier means coupled to said first and second voltage drop electrodes for receiving the voltage developed across said voltage drop electrodes and producing an output responsive thereto, means for providing a predetermined reference voltage, means coupled to the amplifier means for detecting the output of the amplifier means and providing an output responsive thereto, said means for detecting receiving said predetermined

reference voltage, means for supplying the output of the means for detecting to the variable current source to provide a control voltage for the variable current source so that the output from the variable current source provides an output from the amplifier means matching said predetermined reference voltage, wherein said current source includes a logarithmic function generator electrically coupled between the control voltage input and the oscillatory current output so that the plethysmograph will have a constant response time regardless of the biological segment resistance and the current level supplied to the biological segment, the output of the means for detecting providing a percent deviation signal which represents a percent of change of resistance value of the biological segment, and indicating means for receiving the percent deviation signal from the means for detecting and for indicating the percent change of resistance.

4. In a plethysmograph, first and second current input electrodes, first and second voltage drop measuring electrodes adapted to be applied in pairs across a biological segment with one current electrode adjacent one voltage electrode and the other current electrode being adjacent the other voltage electrode, an oscillator having a high frequency output, a variable current source connected to the oscillator and to the first and second current input electrodes for receiving a timing signal from the current input electrodes, said variable current source including at least two active electronic amplification means connected in series and providing an output impedance which is high with respect to the impedance of the biological segment so that plethysmograph calibration will not be substantially affected by changes in the impedance of the biological segment, amplifier means coupled to said first and second voltage drop electrodes for receiving voltage developed across said voltage drop electrodes and producing an output responsive thereto, means for providing a predetermined reference voltage, detecting means coupled to the amplifier means and to the means for providing a predetermined reference voltage for detecting the output of the amplifier means and providing an output responsive thereto, means for supplying the output of the detecting means to the variable current source to provide a control voltage for the variable current source so that the output from the variable current source develops a voltage across said first and second voltage drop electrodes which provides output from the amplifier means automatically nulling said predetermined reference voltage, the output of the detecting means providing a percent deviation signal which represents a percent of change of the value of resistance of the biological segment, a plurality of serially connected diodes and a plurality of serially connected resistors with one of said resistors connected across each of said diodes included in said variable current source for providing a voltage thereacross which is the logarithm of the current therethrough, whereby a constant system response time is obtained regardless of biological segment resistance or current source output level, and indicating means for receiving the percent deviation signal from the means for detecting and for indicating the percent of change of resistance.

5. In a plethysmograph, first and second current input electrodes, first and second voltage drop measuring electrodes adapted to be applied in pairs across a biological segment with one current electrode adjacent

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one voltage electrode and the other current electrode being adjacent the other voltage electrode, an oscillator having a high frequency output, a variable current source connected to the oscillator and to the first and second current input electrodes for receiving a timing signal from the oscillator so that an oscillator output current is provided to the current input electrodes, amplifier means coupled to the voltage drop electrodes for receiving the voltage developed across the voltage electrodes and producing an output responsive thereto, means coupled to the amplifier means for detecting the output of the amplifier means and providing an output responsive thereto, means for supplying the output of the means for detecting to the variable current source to provide a control voltage for the variable current

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source so that the variable current source provides an output producing a long term constant voltage across the first and second voltage drop electrodes, said variable current source including a logarithmic function generator for providing a constant output response time over a range of biological segment resistance values covering substantially two orders of magnitude, the output of the means for detecting providing a percent deviation signal which represents a percent of change of resistance value of the biological segment, and means connected to receive the percent deviation signal for indicating the biological segment percent change of resistance.

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