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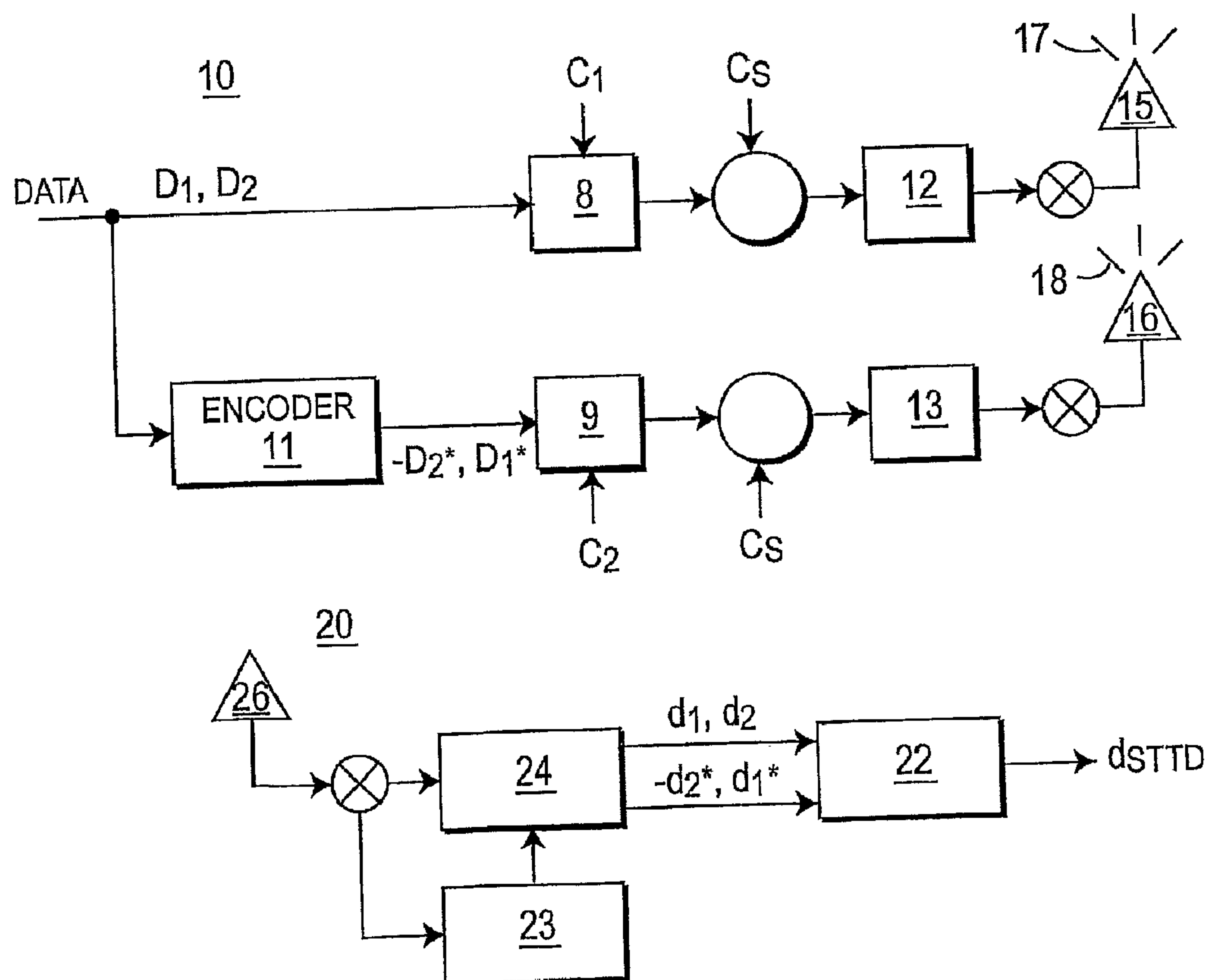
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(54) Titre : DIVERSITE D'EMISSION PAR BLOCS SPATIO-TEMPORELS SIMPLES UTILISANT DES CODES
D'ETALEMENT MULTIPLES

(54) Title: SIMPLE BLOCK SPACE TIME TRANSMIT DIVERSITY USING MULTIPLE SPREADING CODES



(57) Abrégé/Abstract:

The present invention is a method for a base station to transmit a data field of symbols. The method consists of generating a first data field of symbols, encoding the first data field to produce a second data field, spreading the first and second data fields using

(57) **Abrégé(suite)/Abstract(continued):**

respective first and second channelization codes uniquely associated with respective first and second antennas and transmitting an RF signal including the first and second spread data fields over the first and second antennas. Preferably, the second data field has symbols that are the complex conjugates of the symbols of the first field.

ABSTRACT

The present invention is a method for a base station to transmit a data field of symbols. The method consists of generating a first data field of symbols, encoding the first data field to produce a second data field, spreading the first and second data fields using respective first and second channelization codes uniquely associated with respective first and second antennas and transmitting an RF signal including the first and second spread data fields over the first and second antennas. Preferably, the second data field has symbols that are the complex conjugates of the symbols of the first field.

SIMPLE BLOCK SPACE TIME TRANSMIT DIVERSITY USING MULTIPLE SPREADING CODES

This application is a divisional of Canadian Patent Application Serial No.
5 2,635,909, which in turn is a divisional of Canadian Patent Application No. 2,430,720
filed internationally on December 5, 2001 and entered nationally on June 3, 2003.

BACKGROUND

The present invention relates to communications systems imploring code
10 division multiple access (CDMA) techniques. More particularly, the present invention
relates to a transmission diversity scheme which can be applied to a CDMA
communication system

Spatial diversity has been proposed for support of very high data rate users
within third generation wide band code division multiple access systems such as
15 CDMA. Using multiple antennas, the systems achieve better gains and link quality,
which results in increased system capacity. Classically, diversity has been exploited
through the use of either beam steering or through diversity combining.

More recently, it has been realized that coordinated use of diversity can be
achieved through the use of space-time codes. Such systems can theoretically increase
20 capacity by up to a factor equaling the number of transmit and receive antennas in the
array. Space-time block codes operate on a block of input symbols producing a matrix
output over antennas and time.

In the past, space-time transmit diversity systems have transmitted consecutive
symbols simultaneously with their complex conjugates. This type of system, though
25 may result in symbol overlap at the receiving end, with the amount of overlap being
dependent on the length of the impulse response of the propagation channel. In time
division duplex (TDD) mode, this symbol overlap will have to be accounted for in the
joint detection receiver. The joint detector will have to estimate the transmitted

- 2 -

symbols and their conjugates, resulting in an increase in complexity of the joint detection.

In order to alleviate this increase in joint detection, systems have been created which transmit two similar but different data fields. The first data field, having a first portion, D_1 , and a second portion, D_2 , is transmitted by the first antenna. A conjugate of D_2 , $-D_2^*$, is the first portion of the second data field and the conjugate of D_1 , D_1^* , is the second portion. The second data field is simultaneously transmitted by the second antenna. This type of system results in the joint detection implemented at the receiver needing only to estimate the same amount of symbols as in the case of a single transmit antenna. A block diagram of this system is illustrated in Figure 1.

Although the above system reduces the complexity of joint detection for a single data block, joint detection requires the use of two joint detectors at the receiver in a system employing two transmit diversity antennas. Each joint detection device estimates the data from one of the antennas. The estimated data is combined to produce the original data. Therefore, the receiver in such a system has a high complexity resulting in higher receiver expense.

U.S. Patent No. 5,652,764 discloses a radio communication system using two transmission antennas. Data to be transmitted is mixed with a first and second orthogonal code. The first mixed data is transmitted by a first antenna and the second mixed data is transmitted by a second antenna. A receiver receives the data transmitted by each antenna. A first matched filter matched to the first orthogonal code filters the received data and a second matched filter matched to the second orthogonal code filters the received data. A result of both matched filters is combined to recover the original data.

Accordingly, there exists a need for a transmit diversity system requiring less complexity and receiver expense.

SUMMARY

The invention relates to a method for a base station to transmit a data field of symbols. The method consists of generating a first data field of symbols, encoding the first data field to produce a second data field, spreading the first and second data fields
5 using respective first and second channelization codes uniquely associated with respective first and second antennas and transmitting an RF signal including the first and second spread data fields over the first and second antennas. Preferably, the second data field has symbols that are the complex conjugates of the symbols of the first field.

10 According to a first broad aspect of the present invention, there is disclosed a method for a base station to transmit a data field of symbols comprising: generating a first data field of symbols; encoding said first data field producing a second data field having complex conjugates of the symbols of said first data field; spreading said first and second data fields, wherein said first data field is spread using a first
15 channelization code that is uniquely associated with a first antenna and said second data field is spread using a second channelization code that is uniquely associated with a second antenna; and transmitting an RF signal including said first and second spread data fields over the first and second antennas.

20 According to a second broad aspect of the present invention, there is disclosed a method for a base station to transmit a data field of symbols including a transmitter, the method comprising the steps of: generating a first data field of symbols; spreading said first data field using a first channelization code producing a first spread data field, wherein the first channelization code is uniquely associated with a first antenna; spreading said first data field using a second channelization code producing a second
25 spread data field, wherein the second channelization code is uniquely associated with a second antenna; and transmitting an RF signal including said first and second spread data fields over the first and second antennas.

According to a third broad aspect of the present invention, there is provided a method for a base station to transmit a data field of symbols comprising:

- 4 -

generating a first data field of symbols; spreading said first data field using a first channelization code that is exclusively associated with a first antenna; spreading a second data field using a second channelization code that is exclusively associated with a second antenna, the second data field including the same data as the first data field; combining the first spread data field with a first midamble to produce a first communication burst, the first midamble for estimating a channel impulse response; combining the second spread data field with a second midamble to produce a first communication burst, the second midamble for estimating a channel impulse response; scrambling the first spread data field by a scrambling code associated with the base station; scrambling the second spread data field by a scrambling code associated with the base station; transmitting the first communication burst over a first diversity antenna; and transmitting the second communication burst over a second diversity antenna.

According to a fourth broad aspect of the present invention, there is provided a base station, comprising: a first channelization device to receive data and spread the data with a first channelization code to produce a first spread data field; a second channelization device to receive the same data as the first channelization device and spread the received data with a second channelization code that is different from the first channelization code to produce a second spread data field; a first training sequence device in communication with the first channelization device, the first training sequence device to combine the first spread data field with a first midamble to produce a first communication burst, the first midamble for estimating a channel impulse response; a second training sequence device in communication with the second channelization device, the second training sequence device to combine the second spread data field with a second midamble to produce a second communication burst, the second midamble for estimating a channel impulse response; a first scrambling device to scramble the first spread data field by a scrambling code associated with the base station; a second scrambling device to scramble the second spread data field by the scrambling code associated with the

- 5 -

base station; a first diversity antenna to transmit the first communication burst; and a second diversity antenna to transmit the second communication burst.

According to a fifth broad aspect of the present invention, there is provided a user equipment comprising: a data detection device having an input configured to
5 receive a signal being diversity transmitted, the received signal having a first communication burst and a second communication burst, the first and second communication bursts being transmitted from a same base station in a same time slot over different antennas and the first and second communication bursts having the same data in a same order, each of the first and second communication bursts
10 having an inserted midamble training sequence used for estimating a channel impulse response, and a scrambling code associated with a transmitting base station, the data detection device configured to process the received signal with a first channelization code associated with a first transmission antenna to produce first data and the data detection device configured to process the received signal with a
15 second channelization associated with a second transmission antenna to produce second data, wherein the first and second data are the same data; wherein the first and second channelization codes are different from one another and the first channelization code is utilized exclusively for transmission on the first antenna and the second channelization code is utilized exclusively for transmission on the
20 second antenna.

According to a sixth broad aspect of the present invention, there is provided a system for diversity transmission, the system comprising: a base station, comprising: a first channelization device to receive data and spread the data with a first channelization code to produce a first spread data field, and wherein the first
25 channelization code is utilized exclusively for transmission on the first antenna; a second channelization device to receive the same data as the first channelization device and spread the received data with a second channelization code that is different from the first channelization code to produce a second spread data field, and wherein the second channelization code is utilized exclusively for transmission

- 6 -

on the second antenna; a first training sequence device in communication with the first channelization device, the first training sequence device to combine the first spread data field with a first midamble to produce a first communication burst, the first midamble for estimating a channel impulse response; a second training

5 sequence device in communication with the second channelization device, the second training sequence device to combine the second spread data field with a second midamble to produce a second communication burst, the second midamble for estimating a channel impulse response; a first scrambling device to scramble the first spread data field by a scrambling code associated with the base station; a

10 second scrambling device to scramble the second spread data field by the scrambling code associated with the base station; a first diversity antenna to transmit the first communication burst; and a second diversity antenna to transmit the second communication burst; and a user equipment, comprising: a data detection device having an input configured to receive the signal being diversity

15 transmitted, the received signal having the first communication burst and the second communication burst, the first and second communication bursts being transmitted from the same base station in a same time slot over different antennas and the first and second communication bursts having the same data in a same order, each of the first and second communication bursts having the inserted midamble training

20 sequence used for estimating a channel impulse response, and a scrambling code associated with a transmitting base station, the data detection device configured to process the received signal with the first channelization code associated with the first transmission antenna to produce first data and the data detection device configured to process the received signal with the second channelization associated

25 with the second transmission antenna to produce second data, wherein the first and second data are the same data.

BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1 is a block diagram of a prior art communication system employing space-time transmit diversity.

Figure 2 is a block diagram of a transmitter and receiver in a communication system in accordance with the preferred embodiment of the present invention.

Figure 3 is a flow diagram of the transmit diversity system of the present invention.

Figure 4 is a graph of the performance of the transmit diversity system of the present invention.

Figure 5 is a block diagram of a transmitter and receiver in a communication system in accordance with an alternative embodiment of the present invention.

Figure 6 is a flow diagram of an alternative transmit diversity system of the present invention.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

Figure 2 is a block diagram of a transmitter 10, preferably located at a base station, and a receiver 20, preferably located at a user equipment (UE), in a CDMA communication system in accordance with the preferred embodiment of the present invention. Although it is preferable to have the transmitter located at a base station and the receiver located at the UE, the receiver and transmitter may switch locations and the present invention operate on an uplink communication. The transmitter 10 comprises a block encoder 11, a plurality of channelization devices 8, 9, a plurality of spreading sequence insertion devices 12, 13, and a plurality of antennas 15, 16. Although Figure 1 illustrates a transmitter comprising two (2) antennas, it should be apparent to those having skill in the art that more than two (2) antennas may be used, such as N antennas.

A typical communication burst has two data fields separated by a midamble sequence. Preferably, the same encoding procedure, as discussed in the following, for

- 8 -

one data field is also performed on the other data field. Data to be transmitted by the transmitter 10 is produced by a data generator (not shown). The resulting data symbols ($S_1, S_2, \dots, S_{N/2}$), ($S_{N/2+1}, S_{N/2+2}, \dots, S_N$) of the first data field, which can be represented by sub-data fields D_1 and D_2 , are input into the block encoder 11, preferably a block
 5 space-time transmit diversity (BSTTD) encoder. The block encoder 11 encodes the input symbols and generates the complex conjugate of D_1 and the negation of the conjugate of D_2 : $D_1^*, -D_2^*$. The encoder 11 also changes the order of the symbols so that $-D_2^*$ is ahead of D_1^* . Preferably, an analogous encoding of the second data field is also performed.

10 In accordance with the preferred embodiment of the present invention, the data fields, D_1, D_2 and $-D_2^*, D_1^*$ are forwarded to a first and second channelization device 8, 9, respectively. The first channelization device 8 spreads the data blocks D_1, D_2 by a first channelization code, and $-D_2^*, D_1^*$ by the second channelization device 9 using a
 15 second different channelization code. Each of the spread data blocks from the first and second channelization devices 8, 9 are then scrambled by the scrambling code associated with the transmitter 10.

Once the symbols $D_1, D_2, -D_2^*, D_1^*$ have been scrambled, they are mixed with a first and second midamble through training sequence insertion devices 12, 13, producing two communication bursts 17, 18. The two bursts 17, 18 are modulated and
 20 simultaneously transmitted to the receiver 20 over antenna 15 and diversity antenna 16, respectively.

The receiver 20 comprises a joint detection device (JD) 24, a BSTTD decoder 22, a channel estimation device 23 and an antenna 26. The antenna 26 of the UE receives various RF signals including the communication bursts 17, 18 from the
 25 transmitter 10. The RF signals are then demodulated to produce a baseband signal.

The baseband signal is then forwarded to the joint detection device 24 and the channel estimation device 23. As those skilled in the art know, the channel estimation

- 9 -

device 23 provides channel information, such as channel impulse responses, to the joint detection device 24.

The joint detection device 24, coupled to the channel estimation device 23 and BSTTD decoder 22, utilizes the channel information and the channelization codes to
5 detect the soft data symbols d_1 , d_2 , $-d_2^*$, d_1^* in the received signal. The channel impulse response for each burst is determined using that burst's midamble sequence. Since each burst was transmitted using a different spreading code, the joint detection device 24 treats each burst as being transmitted by a different user. As a result, any joint detection device which can recover data from different transmitter sites may be
10 used. Such joint detection devices include zero forcing block linear equalizers, detection devices using Cholesky or approximate Cholesky decomposition, as well as many others. The joint detection device 24 estimates the data symbols of each of the bursts 17, 18 output by the transmitter antennas 15, 16 and forwards the estimates to the BSTTD decoder 22.

15 The BSTTD decoder 22, coupled to the joint detection device 24, receives the estimated soft data symbols d_1 , d_2 and $-d_2^*$, d_1^* corresponding to the antennas 15, 16 and decodes the symbols to yield a single data field's soft symbols, d_{STTD} .

The flow diagram of the present invention is illustrated in Figure 3. A data generator generates data to be transmitted to the receiver 20 (step 301). Each data field
20 is separated into two sub-data fields D_1 , D_2 (step 302). The sub-data fields D_1 , D_2 are forwarded to the block encoder 11 and the first channelization device 8 (step 303). The sub-data fields forwarded to the block encoder 11 are encoded (step 304) and forwarded to the second channelization device 9 (step 305). Each channelization device 8, 9 spreads their respective data input using a separate channelization code
25 associated with a respective antenna 15, 16 (step 306). The two spread signals are then scrambled, using the scrambling code associated with the base station (step 307) and transmitted to the receiver 20 over diversity antennas 15, 16 (step 308).

- 10 -

The receiver 20 receives a RF communication signal including the two spread signals from the diversity antennas 15, 16 (step 309), demodulates the signal and forwards the demodulated signal to the channel estimation device 23 and joint detection device 24 (step 310). The received signal is processed by the channel estimation device 23 (step 311) and the channel information applied by the joint detection device 24 along with the channelization codes, to estimate the transmit symbols from the diversity antennas 15, 16 (step 312). The detected sub-data fields, corresponding to each diversity antenna 15, 16, are forwarded to the BSTTD decoder (step 313), which decodes the soft symbol sub-fields to yield a single data field's soft symbols, d_{STD} (step 314).

Similar to the preferred embodiment disclosed above, Figure 5 is a block diagram of an alternative transmitter 40, preferably located at a base station, and a receiver 50, preferably located a user equipment (UE) in a communication system. The transmitter 40 comprises a plurality of channelization devices 48, 49, a plurality of spreading sequence insertion devices 42, 43, and a plurality of antennas 45, 46.

Data to be transmitted by the transmitter 40 is produced by a data generator (not shown). The resulting data symbols $(S_1, S_2, \dots, S_{N/2})$, $(S_{N/2+1}, S_{N/2+2}, \dots, S_N)$ of the first data field, which can be represented by sub-data fields D_1 and D_2 , are input to a first and second channelization device 48, 49, respectively. The first channelization device 8 spreads the data blocks D_1, D_2 by a first channelization code, and the second channelization device 49 spreads the data blocks D_1, D_2 by a second different channelization code. Each of the spread data blocks from the first and second channelization devices 48, 49 are scrambled by the scrambling code associated with the transmitter 40.

Once the symbols have been scrambled, they are mixed with a first and second midamble through training sequence insertion devices 42, 43, producing two communication bursts 44, 45. The two bursts 44, 45 are modulated and simultaneously transmitted to the receiver 50 over antenna 46 and diversity antenna 47, respectively.

- 11 -

The receiver 50 comprises a joint detection device (JD) 54, a decoder 22, a channel estimation device 53 and an antenna 51. The antenna 51 of the UE receives various RF signals including the communication bursts 44, 45 from the transmitter 40. The RF signals are then demodulated to produce a baseband signal.

5 The baseband signal is then forwarded to the joint detection device 54 and the channel estimation device 53. The joint detection device 54, coupled to the channel estimation device 53 and decoder 52, utilizes the channel information and the channelization codes to detect the soft data symbols d_1 , d_2 , in the received signal. The channel impulse response for each burst is determined using that burst's midamble
10 sequence. Since each burst was transmitted using a different spreading code, the joint detection device 54 treats each burst as being transmitted by a different user. The joint detection device 54 estimates the data symbols of each of the signals 44, 45 output by the transmitter antennas 46, 47 and forwards the estimates to the decoder 52.

15 The decoder 52, coupled to the joint detection device 54, receives the estimated soft data symbols d_1 , d_2 corresponding to the antennas 46, 47 and decodes the symbols to yield a single data field's soft symbols, d .

20 The flow diagram of the alternative embodiment is illustrated in Figure 6. A data generator generates data to be transmitted to the receiver 40 (step 601). Each data field is separated into two sub-data fields D_1 , D_2 (step 602). The sub-data fields D_1 , D_2 are forwarded to the first channelization device 48 and to the second channelization
25 device 49 (step 603). Each channelization device 48, 49 spreads their respective data input using a separate channelization code associated with each antenna 46, 47 (step 604). The two spread signals are then scrambled, using the scrambling code associated with the base station (step 605) and transmitted to the receiver 50 over diversity
antennas 46, 47 (step 606).

 The receiver 50 receives a RF communication signal including the two spread signals from the diversity antennas 46, 47 (step 607), demodulates the signal and forwards the demodulated signal to the channel estimation device 53 and joint

- 12 -

detection device 54 (step 608). The received signal is processed by the channel estimation device 53 (step 609) and the channel information applied by the joint detection device 54 along with the channelization codes, to estimate the transmit symbols from the diversity antennas 46, 47 (step 610). The detected sub-data fields, corresponding to each diversity antenna 46, 47, are forwarded to the decoder 52 (step 611), which decodes the soft symbol sub-fields to yield a single data field's soft symbols, d_{STTD} (step 612).

By using additional channelization codes, the above approaches can be applied to an antenna array having any number of antennas. Each antenna has its own associated channelization code and midamble. If a block encoder is used, the data field transmitted by each of the antennas has a unique encoding, allowing the use of a single joint detector at the receiver.

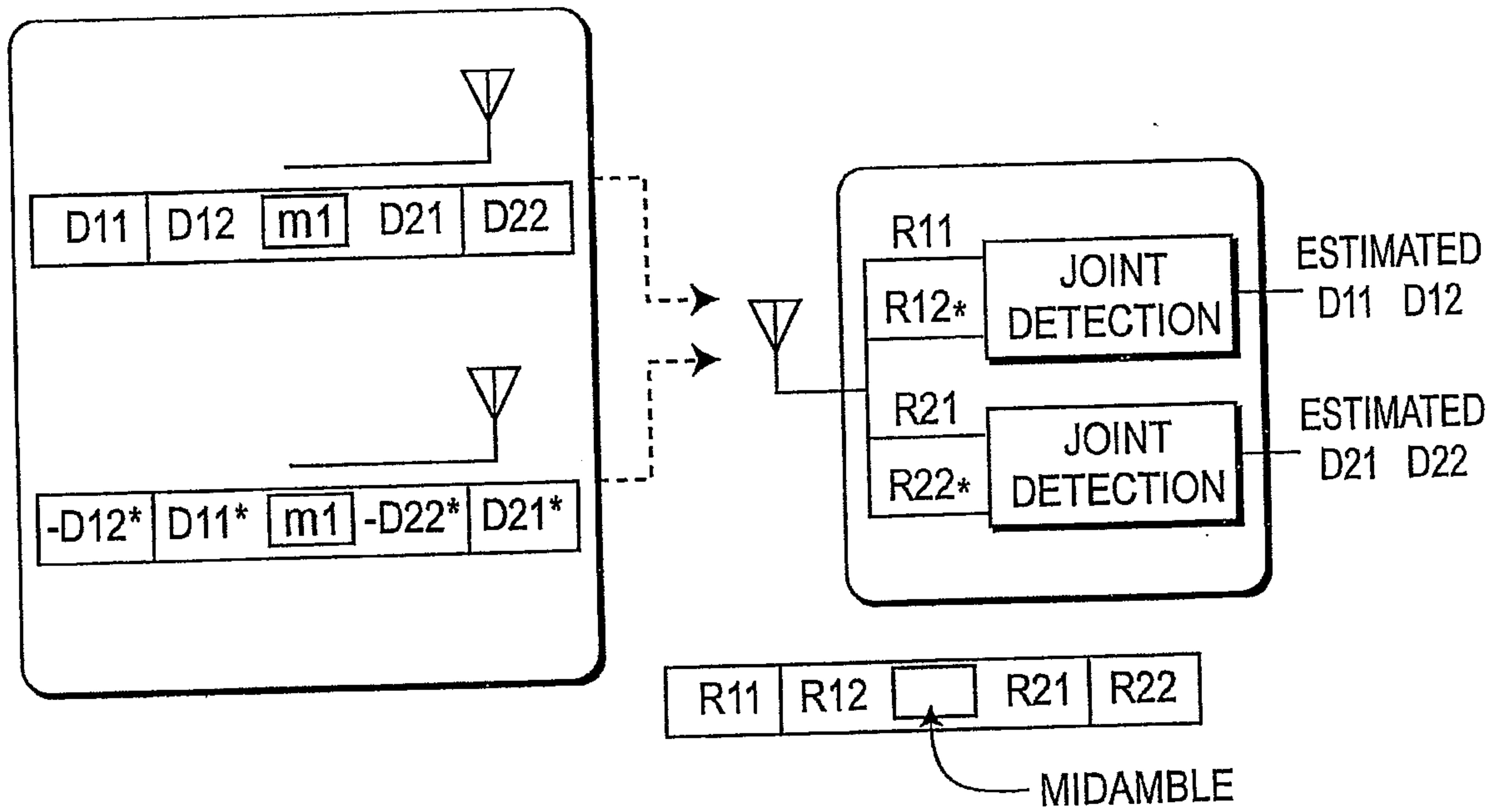
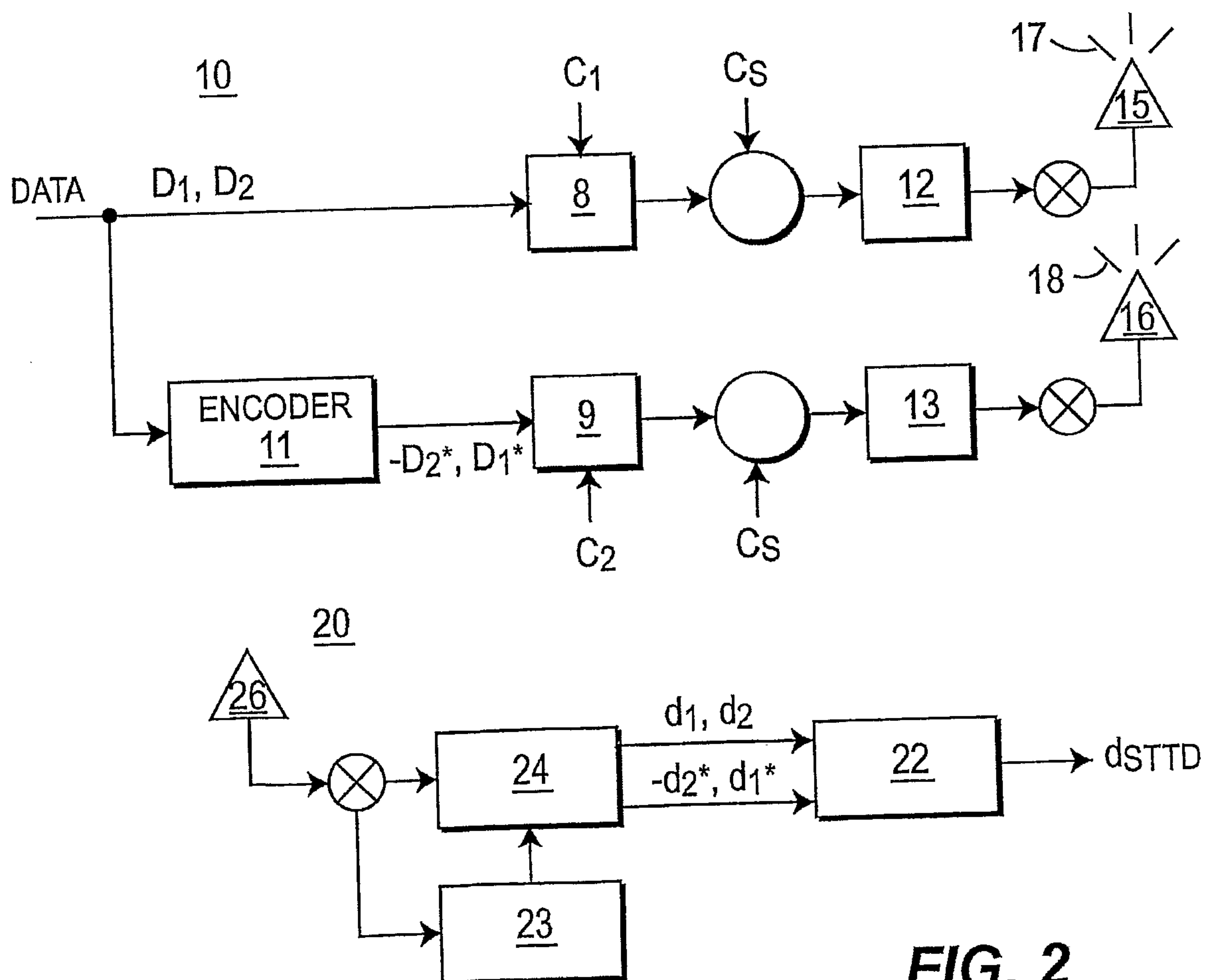
The BSTTD transmitter with two channelization codes of the present invention allows for the use of a cheaper and simpler method of transmit diversity. The use of different channelization codes per transmit antenna requires only one joint detection device at the receiver resulting in a less complex receiver system than those of the prior art. Figure 4 is a graph showing the raw BER of various block STTD decoders. The model is based on all the receivers using a block linear equalizer (BLE) based approach to JD. NTD means the single antenna case, i.e., no transmit diversity. STTD with 1 code is the traditional block STTD JD. STTD with 2 code is the disclosed block STTD transmitter. Simple STTD with 2 code is the transmission system disclosed in the alternative embodiment. As illustrated, the benefit of 2 codes for STTD can be summarized as follows: 1) there is up to a 0.5 dB gain at 0.01 raw Bit error rate over 1 code STTD; and 2) by eliminating the encoding block in simple STTD with 2 code, the performance degradation is only 0.2 dB at 0.1 raw BER and no degradation at 0.01 raw BER. The performance improvement over NTD is still 1.0 dB and 2.7 dB at 0.1 and 0.01 raw BER.

- 13 -

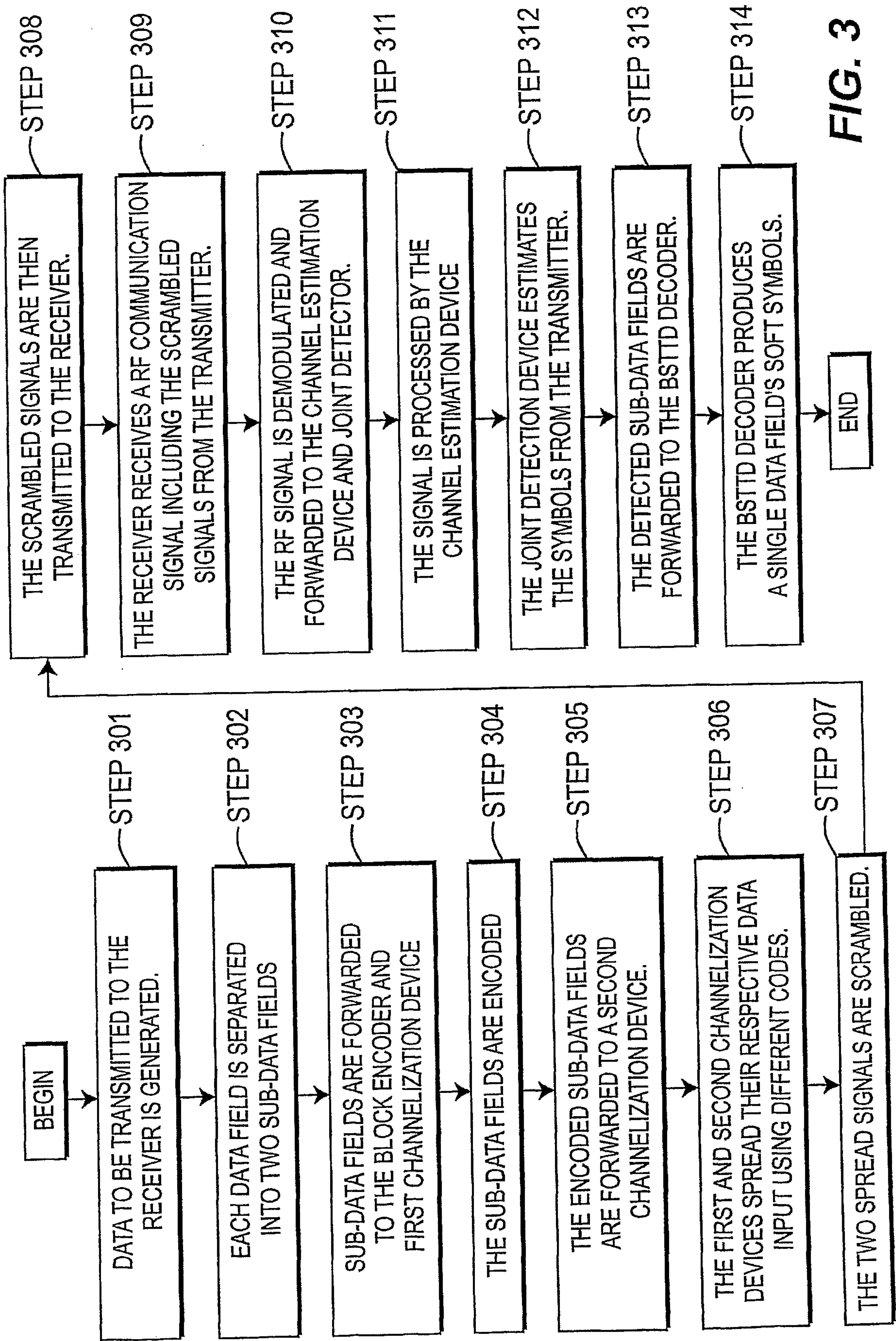
CLAIM:**1. A user equipment comprising:**

a data detection device having an input configured to receive a signal being diversity transmitted, the received signal having a first communication burst and a second communication burst, the first and second communication bursts being transmitted from a same base station in a same time slot over different antennas and the first and second communication bursts having the same data in a same order, each of the first and second communication bursts having an inserted midamble training sequence used for estimating a channel impulse response, and a scrambling code associated with a transmitting base station, the data detection device configured to process the received signal with a first channelization code associated with a first transmission antenna to produce first data and the data detection device configured to process the received signal with a second channelization associated with a second transmission antenna to produce second data, wherein the first and second data are the same data; wherein the first and second channelization codes are different from one another and the first channelization code is utilized exclusively for transmission on the first antenna and the second channelization code is utilized exclusively for transmission on the second antenna.

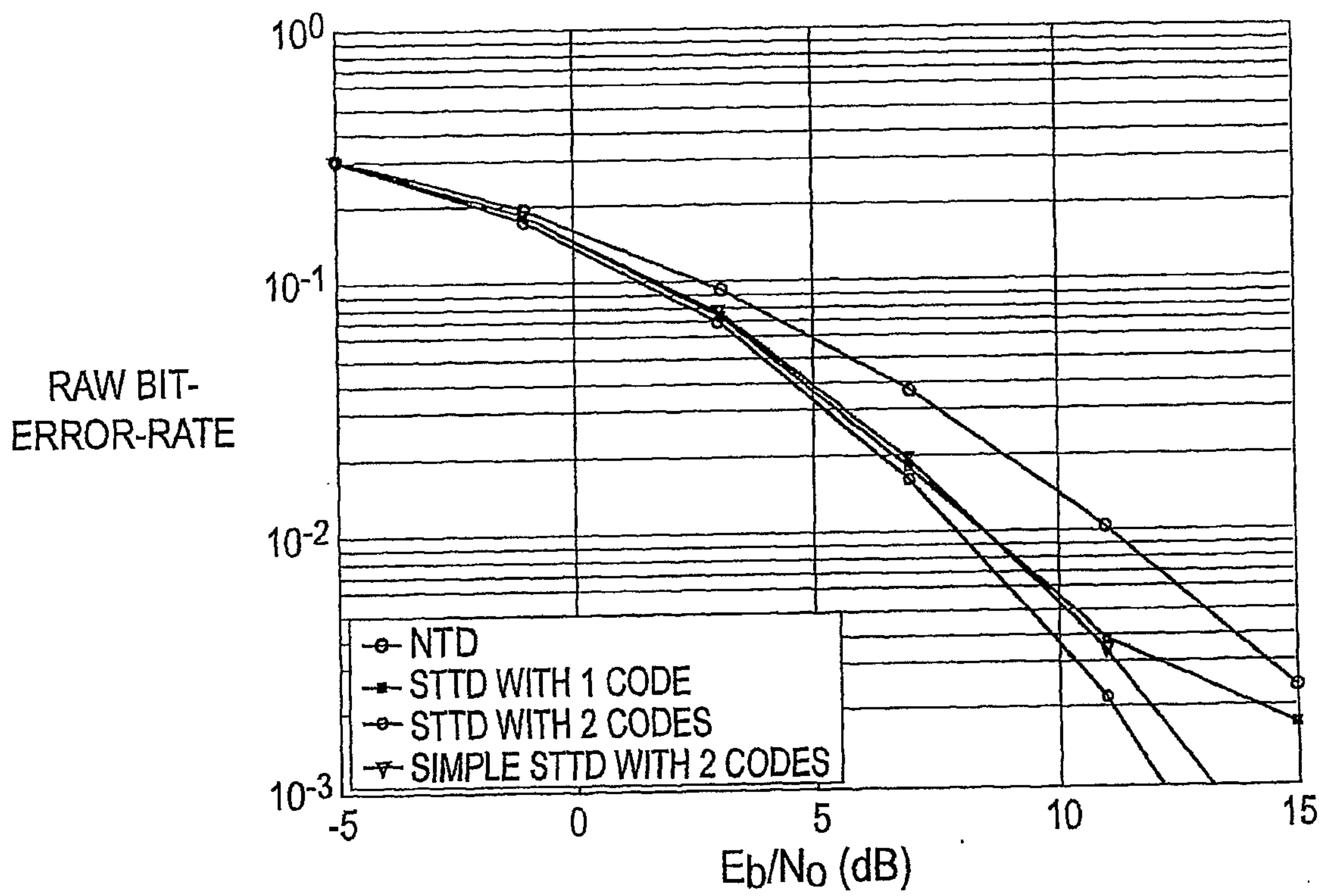
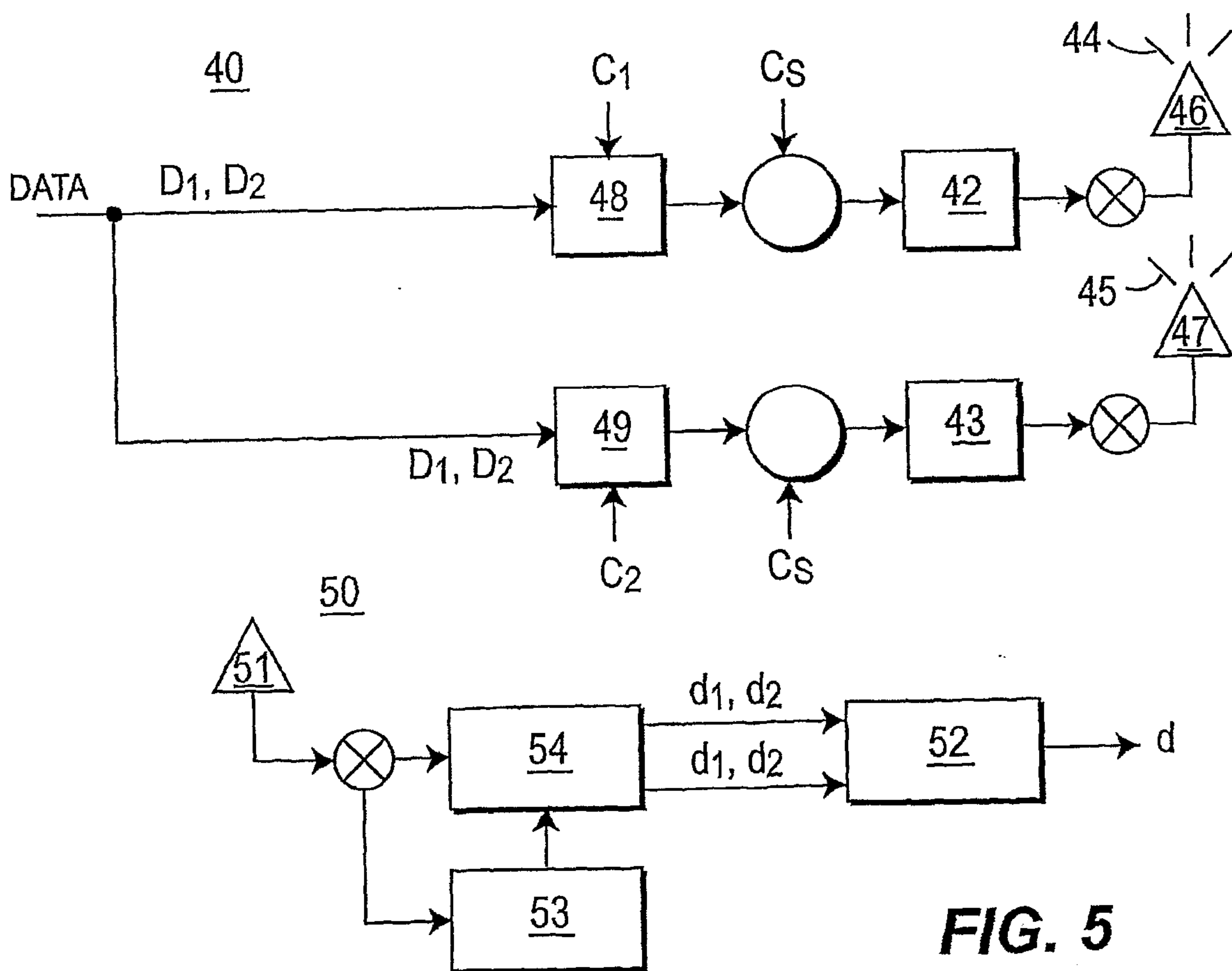
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**FIG. 1****FIG. 2**

2/4



3/4

**FIG. 4**

4/4

