Redundancy circuitry for a semiconductor memory device

Circuit de redondance pour un dispositif de mémoire à semi-conducteur

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Description

The present invention relates to a redundancy circuitry for a semiconductor memory device. Semiconductor memory devices are generally organized in a bidimensional array (memory matrix) wherein the single memory elements are located at the intersection of rows ("word lines") and columns ("bit lines") of the matrix; to access a given memory element, it is necessary to select the word line and the bit line at the intersection of which said memory element is located; to this purpose, the memory address bus is divided into row and column address signals, which are decoded independently.

It is known that in the manufacture of semiconductor memory devices defects are frequently encountered that affect a limited number of memory elements in the memory matrix. The reason for the high probability of defects of this type resides in that in a semiconductor memory device the greatest part of the chip area is occupied by the memory matrix; moreover, it is in the memory matrix, and not in the peripheral circuitry, that the manufacturing process characteristics are usually pushed to limits.

In order to avoid that the presence of a limited number of defective memory elements on many millions forces the rejection of the entire chip, and therefore to increase the manufacturing process yield, the technique is known of providing for the manufacture of a certain number of additional memory elements, commonly called "redundancy memory elements", to be used as a replacement of those elements that, during testing of the memory device, prove defective; the selection circuits, with which the integrated component must necessarily be provided, and which allow the abovementioned functional replacement of a defective memory element with a redundancy memory element, are indicated as a whole with the name of "redundancy circuitry", while the set of redundancy memory elements and circuitry is defined for short as "redundancy".

The redundancy circuitry comprises programmable non-volatile memory registers (redundancy registers) suitable to store those address configurations corresponding to the defective memory elements; such registers are programmed once and for all during the memory device testing, and must retain the information stored therein even in absence of the power supply.

In practical implementations of redundancy in memory devices, both word lines and bit lines of redundancy memory elements are generally provided in the memory matrix; each redundancy word line or bit line is associated to a respective row or column redundancy register, wherein the address of a defective word line or bit line is stored so that, whenever the defective word line or bit line is addressed, the corresponding redundancy word line or bit line is selected.

The adoption of redundancy in semiconductor memory devices is profitable from the point of view of the increase in the manufacturing process yield only if the increase in the overall chip size is not very high, so that, once the statistical defectivity of the process is taken into account, the number of "good" memory device chips per wafer is in the average higher than that obtainable without implementing redundancy. To limit the chip area necessary for implementing redundancy, a careful evaluation of how many redundancy word lines and bit lines are to be provided in the memory matrix is essential, as well as an optimized design of the physical layout of the redundancy circuitry.


In view of the state of the art described, the object of the present invention is to provide a redundancy circuitry which minimizes the chip size overhead due to the implementation of redundancy.

According to the present invention as claimed in claim 1, such object is attained by means of a redundancy circuitry for a semiconductor memory device, the redundancy circuitry comprising a first plurality of programmable non-volatile memory registers for the selection of redundancy bit lines of redundancy memory elements and a second plurality of programmable non-volatile memory registers for the selection of redundancy word lines of redundancy memory elements, characterized in that it comprises an array of programmable non-volatile memory elements for storing the addresses of defective bit lines and word lines which must be functionally replaced respectively by redundancy bit lines and word lines, and in that it is divided in identical layout strips which are perpendicular to said array of memory elements and which comprise each a first and a second strip sides located at opposite sides of the array of memory elements, the first strip side containing at least one memory register of the first plurality and being crossed by a column address signal bus running parallel to the array of memory elements, the second strip side containing one memory register of the second plurality and being crossed by a row address signal bus running parallel to the array of memory elements.

Further enhancements are provided by the sub-claims.

Thanks to the present invention, and particularly to the fact that circuit blocks interacting both with each other and with the same signals have been physically grouped in a same chip region, the redundancy circuitry is very compact, and the increase in the overall chip size is therefore limited.

These and other features of the present invention will be now made more evident by the following detailed description of a particular embodiment, described as a non-limiting example in the annexed drawings, wherein:

Figure 1 is a simplified view of a redundancy circuitry layout according to the present invention;
Figure 2 is a schematic view of one strip of the layout of Figure 1;
In the following, a redundancy circuitry for a semiconductor memory device will be described initially from the circuitual point of view, and successively from the physical layout point of view. The redundancy circuitry will, by way of example, be thought to be integrated in a word-organized Flash EEPROM device (i.e., with sixteen bits in the external input/output data bus), wherein the memory elements (represented by stacked-gate MOS transistors) are located at the intersection of rows (word lines WL) and columns (bit lines BL) of either one or another of two bidimensional arrays or half-matrices HM1, HM2 (Fig. 6), with respect to an architecture providing for only one bidimensional array of memory elements, the shown architecture allows to limit the word line and bit line length, thus reducing as known to anyone skilled in the art the memory device access time. Furthermore, as usual in Flash EEPROM devices, individually-addressable memory sectors are provided (let's say four sectors), each ideally divided in sector portions representing the sector memory space dedicated to respective bits in the memory device external input/output data bus; a sector portion can be imagined as being made up by a respective group of bit lines; in the present case of a word-organized memory device, each one of the four memory sectors comprises therefore sixteen sector portions or bit line groups.

The memory device is provided with both word lines and bit lines of redundancy memory elements, which in the following will be referred to as "redundancy word lines" and "redundancy bit lines", respectively. Accordingly, the redundancy circuitry comprises a row redundancy circuitry and a column redundancy circuitry. Furthermore, each memory sector is provided with dedicated redundancy bit lines, let's say four redundancy bit lines per sector: a defective bit line in a given memory sector can be only reduned by one of the four redundancy bit lines associated to such sector.

A column redundancy circuitry for a word-organized sectored memory device is described in the copending European Patent Application EP-A-0 668 562 and comprises a plurality of programmable non-volatile memory registers (column redundancy registers), each one associated to a respective redundancy bit line and each one suitable to store an address of a defective bit line which must be replaced by the associated redundancy bit line. In the present example, four column redundancy registers per memory sector are necessary, for a total of sixteen column redundancy registers.

With reference to Figure 7, each column redundancy register CRR of the circuitry according to the present invention is substantially made up of a first part, represented in Figure 7 by a block 1, wherein the address of a defective bit line can be programmed, and of a second part, represented in Figure 7 by a block 2, suitable to store in a coded form an information suitable to determine in which sector portion of a given memory sector (i.e., in which of the sixteen groups of bit lines constituting the memory sector) the defective bit line whose address is stored in block 1 has been found.

The block 1 comprises a number of programmable non-volatile memory cells CMC0-CMC5 equal to the number of column address signals C0-C5 which, together with their respective logic-complemented signals C0N-C5N, constitutes a column address signal bus C- BUS also supplying in a per-se known way a column decoding circuitry (not shown) for addressing (i.e., selecting) single bit lines of a given memory sector. Each memory cell CMC0-CMC5 is supplied, for programming purposes, with a respective column address signal C0-C5 and with the respective complemented signal C0N-C5N; each memory cell CMC0-CMC5 has an output signal CMCSO-CMCSS, representing the memory cell status, which is supplied, together with the respective column address signal C0-C5, to a respective comparator CCMP whose output signal CCMP0-CCMP5 is activated only when the memory cell status, represented by the signal CMCSO-CMCSS, coincides with the current state of the respective column address signal C0-C5. All the signals CCMP0-CCMP5 supply a redundancy bit line selection circuit RBLSC which is also supplied with a memory sector selection signal SS taken from a sector selection signal bus SBUS; the bus SBUS is, in the present example, made up of four sector selection signals, each one allowing the selection, for reading or programming, of one respective memory sector. The four column redundancy registers CRR associated to each memory sector are commonly supplied with one of the four sector selection signals of the bus SBUS.

Coming back to Figure 7, when all the signals CCMP0-CCMP5 and the sector selection signal SS are activated (i.e., when the current state of the column address signals C0-C5 coincides with the logic state stored
in the memory cells CMCO-CMC5, which means that a
defective column address is supplied to the memory de-
vice, and the currently addressed memory sector is that
to which the redundancy bit line associated to the col-
umn redundancy register CRR belongs) the redundancy
bit line selection circuit RBLSC activates a respective
redundancy bit line selection signal RBLSS. Said signal
RBLSS is connected, through a respective switch SW
controlled by the sector selection signal SS, to one of
four signals of a bus RBSBUS which runs from the re-
dundancy circuitry to redundancy bit lines selection means (not shown). Each signal of the bus RBSBUS is
associated to four redundancy bit lines, each of which
is associated to a different memory sector, and to four
column redundancy registers CRR, each one associat-
ed to a different memory sector. The switches SW in
each one of said four column redundancy registers per-
form a multiplexing action, so that when a given memory
sector is addressed, the four signals of the bus RBSBUS
are electrically connected to the signals RBLSS of the
four column redundancy registers CRR associated to
said addressed sector. In this way, with a bus RBSBUS
of just four signals common to all the sixteen column
redundancy registers, it is possible to individually select
sixteen redundancy bit lines.

As regards the block 2 in each column redundancy
register CRR, it comprises a number of memory cells
CMC6-CMC9 sufficient to store in coded form an iden-
tifying code for identifying the memory sector portion
wherein the defective bit line whose address is stored
in the memory cells CMCO-CMC5 of the block 1 has
been found; in the example shown, referring to a word-
organized memory device with sixteen bits in the exter-
nal input/output data bus, a four-bit code is sufficient
to identify the sixteen different memory sector portions
of each memory sector. Each memory cell CMC6-CMC9
is supplied, for programming purposes, with a respecti-

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tive row address signal R0-R3 and with the respective
complemented signal R0N-R3N, taken from a row ad-

dress signal bus R Abbas also supplying in a per-se

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known way a row decoding circuitry (not shown) for ad-
dressing (i.e. selecting) single word lines inside each
one of the two memory half-matrices HM1, HM2 of Fig-
ure 6. The memory cell status output signals
CMCS6-CMCs9 is grouped together in a local identify-
ing code bus ICBUS. The local identifying code bus ICB-
US is connected to a common identifying code bus
ICBUS through a respective multiple switch MSW (a
switch with four input channels and four output chan-
nels) controlled by the redundancy bit line selection sig-
nal RBLSS. The common identifying code bus ICBUS
is made up of four signals and is common to all the six-
teen column redundancy registers CRR. The multiple
switches MSW in each column redundancy register
CRR perform a multiplexing action, so that at any time
the signals of the bus ICBUS are electrically connected
to the signals of the local bus ICBUS' of the column
redundancy register CRR which is associated to the cur-
rently addressed memory sector and which stores the
address of the currently addressed defective bit line. If
the currently addressed bit line in the currently ad-

dressed memory sector is not defective, all the switches
MSW are open, and the signals of the bus ICBUS are
left floating.

Each memory cell CMCO-CMC9 is further supplied
with a memory cell program enable signal PGEN sup-
plied by a respective ground/high-voltage switch CH-
VSW. Said switch CHVSW is controlled by an output sig-
nal of a respective three-input AND gate 6, which has a
first input supplied with a column redundancy register
program enable signal CRRPGE activated by a control

circuitry (not shown) generally provided in the memory
device, a second input supplied with a respective row
address signal of the set of row address signals R5-R8
also taken from the row address bus R Abbas, and a third
input supplied with the sector selection signal SS. The
AND gate 6 and the switch CHVSW represent a column
redundancy register programming selection circuitry.
The four column redundancy registers CRR associated
to a given memory sector are each supplied by one dif-

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ferent row address signal of the set R5-R8, to select one
of them for programming.

Supposing that a defective bit line in a given mem-
ory sector portion of a given memory sector is found, to
program one of the four column redundancy register
CRR associated to such memory sector, the address of
the defective bit line is supplied to the memory device
and is carried by the column address signal bus CABUS;
the memory device is further supplied with a row ad-

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dress such that the row address signals R0-R3 carry
the identifying code of the memory sector portion to
which the defective bit line belongs; the sector selection
signal SS of the memory sector to which the defective
bit line belongs is further activated, and one of the four
row address signals R5-R8 is also activated, to select
one among the four column redundancy registers asso-
ciated to the addressed sector. Finally, the internal con-
trol circuitry activates the signal CRRPGE: in the se-
lected column redundancy register CRR, the signal
PGEN is driven to a high-voltage value, and the memory
cells CMCO-CMC9 are therefore programmed accord-
ing to the logic state of the respective signals C0, C1C5,
C5N and R0, R0N, R3N, R3N supplying them.

A row redundancy circuitry for a semiconductor
memory device with an architecture similar to that of Fig-
ure 6 is described in the copending European Patent

Each half-matrix HM1, HM2 comprises an equal
number, let's say eight, of redundancy word lines.

Similarly to the column redundancy circuitry just de-
scribed, the row redundancy circuitry according to the
present invention comprises a plurality of programmable
non-volatile memory registers (row redundancy reg-
isters), but each row redundancy register is associated
to a respective pair of redundancy word lines, for a total
of eight row redundancy registers, four per half-matrix.
The reason for such a difference resides in that it has been recognized that the most frequent defect affecting word lines consists in short-circuits between adjacent word lines (which causes that when the selection of one of two short-circuited word lines is attempted, the potential of such word line cannot rise to the designed value, being linked by the short-circuit to the potential of the adjacent non-selected word line). When therefore during testing a defective word line is found, it is assumed that such word line is short-circuited with the adjacent word line (the one which, in the testing scanning sequence, follows the defective word line), and both said word lines must be replaced by two respective redundancy word lines; from then on, the two defective word lines will never be selected.

Since defective word lines always come in pairs, it is known to design the row redundancy registers in such a way as to store each one a pair of row addresses, and to be each one associated to a respective pair of redundancy word lines: into each row redundancy register can therefore be programmed the addresses of two adjacent short-circuited word lines. In memory devices the association between the word lines in the memory matrix and the respective selection signals generated by the row address signal decoding circuitry is generally such that adjacent word lines have addresses which only differ in one bit; since however such bit can be anyone of the bits constituting the row address signal set, it follows that in order to be always able to replace two adjacent short-circuited word lines whatsoever, each row redundancy register should store two full row addresses: this means that each row redundancy register must comprise a number of memory cells equal to twice the number of row address bits. Since the memory cells occupy each a significant chip area, this would lead to an excessive increase in the overall chip size, so that the overall process yield is decreased instead of increased. This is why a compromise is generally reached between the repairability rate for defective word line pairs and the increase in chip size: in practice, designers give up the possibility to replace two adjacent short-circuited word lines whatsoever, limiting such replacement to adjacent word lines whose addresses only differ in one (or more) bit belonging to a given subset of the whole row address signal set: defining by m the number of bits in the row address signal set, it can be thought of as being the sum of two subsets n and q such that n contains the most significant row address bits, while q contains the least significant row address bits; limiting the replacement of defective word line pairs to take place for adjacent word lines having addresses differing in one (or more) bits of the subset q, it is sufficient to store in a given row redundancy register the full row address m for only one of the pair of adjacent word lines, and the subset q for the other word line of the pair. In this way each row redundancy register must be made up of (m+q) memory cells, instead of 2m; the impact on the repairability rate can be appreciated by considering that the probability of having a short-circuit defect between two adjacent word lines with addresses differing in one (or more) bit in the subset n (a defect which cannot be repaired) is 1/2^n.

Referring to the present example, as shown in Figure 8, each row redundancy register RRR comprises a first group G1 of programmable non-volatile memory cells RMC4-RMC9 (six in the shown example) and two second groups G2 and G2' of programmable non-volatile memory cells RMC0-RMC3 (four in the example).

The memory cells RMC4-RMC9 of the first group G1 of each redundancy register RRR are each one supplied with a respective row address signal R4-R9 and with the respective logic-complemented signal R4N-R9N, the twelve signals R4, R4N-R9, R9N are taken from the row address signal bus RABUS and correspond to the most significant row address bits. Each memory cell RMC4-RMC9 has an output signal RMCS4-RMCS9, representing the memory cells status, which is supplied, together with the respective row address signal R4-R9, to a respective comparator RCMP whose output signal RCMP4-RCMP9 is activated only when the memory cell status (represented by the signal RMCS4-RMCS9) coincides with the current state of the respective row address signal R4-R9. All the signals RCMP4-RCMP9 supply a first-level redundancy word line selection circuit FRWSC which is further supplied with a half-matrix selection signal HMS1 or HMS2, such signal allows the selection, for reading or programming, of one of the two half-matrices HM1, HM2. The four row redundancy registers RRR associated to the same half-matrix are commonly supplied with the signals HMS1, while the remaining four row redundancy registers RRR, associated to the other half-matrix, will be supplied with a signal HMS2 which is the logic complement of the signal HMSS1, and which allows the selection of said other half-matrix. Coming back to Figure 8, when all the signals RCMP4-RCMP9 and the half-matrix selection signal HMS1 or HMS2 are activated (i.e. when the current state of the most significant row address signals R4-R9 coincides with the logic state stored in the memory cells RMC4-RMC9 of the first group G1) the first-level redundancy word line selection circuit FRWSC activates a respective first-level redundancy word line selection signal FRWSS.

Each memory cell RMC0-RMC3 in the two second groups G2 and G2' is supplied with a respective row address signal R0-R3 and with the respective logic-complemented signal R0N-R3N; the eight signals R0, R0N, R3, R3N are taken from the row address signal bus RABUS and correspond to the least significant row address bits; similarly to the memory cells RMC4-RMC9 in the first group G1, each memory cell RMC0-RMC3 in the groups G2 and G2' has an output signal RMCS0-RMCS3, representing the memory cell status, which is supplied, together with the respective row address signal R0-R3, to a respective comparator RCMP whose output signal RCMP0-RCMP3 is activated only when the memory cell status, represented by the signal
RCMP0-RCMP3 supplies a respective second-level redundancy word line selection circuit SRWSC and SRWSC' which, when all the signals RCMP0-RCMP3 of the respective group are activated (i.e. when the current state of the least significant row address signals R0-R3 coincides with the logic state stored in the memory cells RMC0-RMC3 of the respective second group G2 and G2') activates a respective second-level redundancy word line selection signal SRWSC and SRWSC'. Each one of the two second-level selection signals SRWSC and SRWSC' is supplied, together with the first-level selection signal FRWSC and RWSC' whose output RWSS and RWSS' is activated when both the first-level selection signal FRWSC and the respective second-level selection signal SRWSC and SRWSC' are activated.

The signals RWSS and RWSS' are connected to two respective signals of a redundancy word line selection signal bus RWSBUS1 or RWSBUS2 through a respective switch RSW and RSW'; the switches RSW and RSW' are controlled by a respective control circuitry SWCNT and SWCNT' which can selectively commutate the switch RSW and RSW' from the respective redundancy word line selection signal RWSS,RWSC' to a fixed voltage supply line VDD. This is useful for performing particular in-factory device tests, e.g. for testing the absence of defects in the redundancy word lines. In this case, since such test is performed before the row redundancy registers are programmed, it is necessary to address the various redundancy word lines. Two bus RWSBUS1 and RWSBUS2 are provided, one supplying selection means (not shown) for the redundancy word lines of the first half-matrix HM1, and the other supplying selection means (also not shown) for the redundancy word lines of the second half-matrix HM2. Each bus RWSBUS1 and RWSBUS5 contains eight signals.

The activation of the signal RWSS determines the selection of one of the pair of redundancy word lines associated to the row redundancy register RRR, while the activation of the signal RWSS' determines the selection of the other redundancy word line of said pair. Further, the activation of the signal RWSS or RWSS' prevents the defective word line whose address is stored in the memory cells of the first group G1 and of the second group G2 or G2' of memory cells of the row redundancy register from being selected.

Each memory cell RMC4-RMC9 of the first group G1 and each memory cell RMC0-RMC3 of the second group G2 is further supplied with a first memory cell program enable signal PRGEN supplied by a first ground/high-voltage switch RHVSW, while each memory cell RMC0-RMC3 of the second group G2' is further supplied with a second memory cell program enable signal PRGEN' supplied by a second ground/high-voltage switch RHVSW'. The switch RHVSW is supplied with a first control signal CNTS and with a column address signal C0-C3 taken from the column address signal bus CABUS; the switch RHVSW' is also supplied with the column address signal C0-C3 and with a second control signal CNTS'.

In Figure 9 there is shown a circuit for the generation of the two control signals CNTS and CNTS'; two of such circuits are provided, one for the row redundancy registers RRR associated to the half-matrix HM1 and one for the row redundancy registers RRR associated to the half-matrix HM2. The activation of the two control signal CNTS and CNTS' is mutually exclusive, i.e. they are never simultaneously activated; their activation is submitted to the activation of a row redundancy register program enable signal RRRPRGEN by the internal control circuitry, and to the activation of the respective half-matrix selection signal HMSS1 or HMSS2. A column address signal C4 is used to selectively activate either the signal CNTS or CNTS'.

According to what previously noted, only those pairs of adjacent short-circuited word lines whose addresses differ in one or more bits belonging to their least significant part, represented by the row address signals R0-R3, can be replaced by redundancy word line pairs; in fact, each row redundancy register RRR comprises a unique group (the first group G1) of memory cells RMC4-RMC9, to store the most significant bits of the row address of a pair of word lines. Pairs of adjacent defective word lines with addresses differing in one or more bits belonging to their most significant part, represented by the row address signals R4-R9, cannot therefore be replaced by redundancy word line pairs. The impact on the repairability rate can be appreciated by considering that the probability of having a defect between two adjacent word lines with addresses differing in one or more bits in their most significant part is 1/16. Such a compromise allows to limit the number of memory cells required in each redundancy register to 6+4+4. It is obviously possible to increase the repairability rate, by decreasing the number of memory cells in the common first group G1, and increasing the number of memory cells in the two second groups G2 and G2', at the expense however of an increase in the overall number of memory cells, and thus in the chip size: if for example the first group G1 of memory cells is made to comprise five memory cells, and each one of the two second groups G2 and G2' five memory cells, the probability of having a non-repairable defect lower to 1/32, but the number of memory cells required for each redundancy register increases of these memory cells is to 5+5+5).

Supposing that a pair of adjacent defective word line is found in a given half-matrix, the addresses of the two word lines must be programmed into one of the four row redundancy registers RRR associated to said half-matrix. Firstly, the address of one of the two defective word lines is supplied to the memory device, so that the row address signals R0-R3 carry the least significant row address bits, and the row address signals R4-R9...
carry the most significant row address bits; one of the four column address signals C0-C3 is activated, to select for programming one of the four row redundancy registers RRR associated to said half-matrix; also, the column address signal C4 is kept low, to enable the activation of the control signal CNTS; when the internal control circuitry drives the signal RRRPGEN low, the signal CNTS goes low, and the signal PRGEN is driven to the high-voltage value, so that the memory cells RMC0-RMC3 of the group G2 and the memory cells RMC4-RMC9 of the first group G1 can be programmed. Secondly, the row address supplied to the memory device is changed so that the row address signals R0-R3 carry the least significant row address bits of the other word line pair; also, the column address signal C4 is driven high, to enable the activation of the control signal CNTS; when the internal control circuitry drives the signal RRRPGEN low, the signal CNTS' is driven low, and the signal PRGEN' is driven to the high-voltage value; the memory cells RMC0-RMC3 of the group G2' are therefore programmed to store the least significant row address bits of the second word line of the pair.

As described in the European Patent Application EP-A-0 655 743, and as shown in Figure 10, each programmable non-volatile memory cell CMCO-CMC9 and RMC0-RMC9 comprises a pair of programmable non-volatile memory elements, for example two stacked-gate MOSFETs T0 and T1 with source connected to a ground voltage GND and control gate connected to a supply voltage VG which can be switched by the internal control circuitry from a reading voltage value (typically of 5 V) to a programming high-voltage value (of about 12 V); the drains of T0 and T1 are connected to a reading load circuit LC whose output represents a memory cell status signal MCS; the drains of T0 and T1 are also connected to a programming load circuit substantially represented by two MOSFETs T0 and T1 whose drains are respectively connected to one of a pair of logic-complemented programming data lines PDL and PDLN; the gates of the MOSFETs T0 and T1 are commonly connected to a cell program enable signal CPGEN. In each column redundancy register CRR, the programming data lines PDL and PDLN of each one of the memory cells CMCO-CMC5 of the block 1 are respectively connected to a respective pair of logic-complemented column address signals C0,CON-C5,C5N, while the programming data lines PDL and PDLN of each one of the memory cells CMCO-CMC9 of the block 2 are respectively connected to a respective pair of logic-complemented row address signals R0,RON-R3,R3N; the cell program enable signal CPGEN of all the memory cells CMCO-CMC9 is connected to the signal PGEN. In each row redundancy register RR, the programming data lines PDL and PDLN of the memory cells RMC0-RMC9 are respectively connected to a respective pair of logic-complemented row address signals R0,RON-R9,R9N; the cell program enable signal CPGEN of the memory cells RMC4-RMC9 of the first group G1 and of the memory cells RMC0-RMC3 of the second group G2 is connected to the first memory cell program enable signal PRGEN, while the cell program enable signal CPGEN of the memory cells RMC0-RMC3 of the second group G2' is connected to the second memory cell program enable signal PRGEN'.

The redundancy circuitry previously described from the circuitual point of view will be now described from the physical layout point of view.

As shown in Figure 1, the redundancy circuitry layout is split in an upper portion UP and in a lower portion LO; the chip area amidst said upper and lower portions UP and LO is dedicated to the integration of other circuit blocks of the memory device; in the present example, referring to a Flash EEPROM device, said circuit blocks are represented by a counter COUNT, which is supplied by an external address signal bus EABUS containing signals generated by the address input buffer circuits (per-se known and therefore not shown), and by a predecoding circuitry PREDEC, performing a preliminary decoding of the row address signals RABUS and of the column address signals CABUS supplied by the counter COUNT. In a Flash EEPROM device the counter COUNT is necessary to automatize the preprogramming operation, which is the preliminary programming step to which all the memory elements of a given memory sector are submitted before they are erased; in this phase, the counter COUNT generates internally to the chip the address signals necessary to sequentially select the memory elements which are to be preprogrammed. During normal reading or programming operations, the counter COUNT is instead transparent, which means that the row address signal bus RABUS and the column address signal bus CABUS are directly connected to the external address signal bus EABUS.

The physical disposition of circuit blocks shown in Figure 1 is particularly effective in reducing the overall chip size, since all the circuit blocks which must be supplied with the address signals are placed in the same chip region, thus eliminating the necessity of providing long interconnection lines running along the chip.

The upper and lower portions UP and LO of the redundancy circuitry layout comprises each a central array of memory elements MAR wherein the programmable non-volatile memory elements (i.e. the stacked gate MOSFETs T0 and T1) of the memory cells CMCO-CMC9 and RMC0-RMC9 of the column and row redundancy registers CRR and RR are described and are located, and four identical layout strips LS1-LS4, each strip being parted in two parts by the central array of memory elements MAR. The column address signal bus CABUS and the row address signals bus RABUS run parallel to each other at opposite sides of the central array of memory elements MAR.

The structure of each layout strip LS1-LS4 is shown in expanded view in Figure 2; each layout strip LS1-LS4 is splitted in two sides, located at opposite sides with respect to the central array of memory elements MAR.
Each layout strip LS1-LS4 represents the chip area wherein two column redundancy registers CRRA and one row redundancy register RRR are obtained: more precisely, the side of each layout strip on the right of the central array of memory elements MAR represents the chip area dedicated to the integration of two column redundancy registers CRRA and CRRB, while the side of each layout strip LS1-LS4 on the left of the central array of memory elements MAR represents the chip area dedicated to the integration of one row redundancy register RRR. The four layout strips LS1-LS4 in the upper portion UP of the redundancy layout contain the four row redundancy registers RRR associated to the half-matrix HM1, while the four layout strips LS1-LS4 in the lower portion LO of the redundancy layout contain the four row redundancy registers RRR associated to the half-matrix HM2.

The right side of each layout strip LS1-LS4 is designed so that the two column redundancy registers CRRA and CRRB are partially interlaced one to another: in other words, a given memory cell CMC0-CMC5, associated to a given column address signal C0-C5, of one of the two column redundancy registers CRRA is not physically adjacent to another memory cell of the same column redundancy register, associated to another column address signal, but to that memory cell of the other column redundancy register CRRB which is associated to the same column address signal. In Figure 4, schematically showing two interlaced memory cells of the two column redundancy registers CRRA and CRRB, CMC0A is the memory cell of the column redundancy register CRRA associated to the column address signal C0, and CMC0B is a memory cell of the other column redundancy register CRRB also associated to the column address signal C0. Such a technique allows a reduction in area, since it reduces the number of address signal lines that must be traced from the column address signal bus CABUS to the layout blocks that will be described in the following.

Figure 3 is an expanded view of the right side of a given layout strip LS1-LS4. Just beside the central array of memory elements MAR, amidst this last and the column address signal bus CABUS, three layout blocks are provided: an upper block LC1 contains the load circuits LC of the interlaced memory cells CMC0-CMC2, a lower block LC2 contains the load circuits LC of the interlaced memory cells CMC3-CMC5, and a central block LC3 contains the load circuits LC of the memory cells CMC6-CMC9 of the two column redundancy registers CRRA and CRRB. The blocks LC1, LC2 and LC3 are respectively connected to the blocks PL1, PL2 and PL3 by means of respective interconnection lines IL4, IL5 and IL6. In LC1 and LC2 the load circuits LC are interlaced (Fig. 4), which means that the load circuit LC of the memory cell CMC0A of CRRA is adjacent to the load circuit LC of the memory cell CMC0B of CRRB, and so on.

On the right of the blocks LC1 and LC2, two blocks CMP1 and CMP2 contain respectively the comparators CCMP for the interlaced memory cells CMC0-CMC2 and CMC3-CMC5; again, the comparators CCMP are interlaced, which means that the comparator CCMP associated to the memory cell CMC0A of CRRA is adjacent to the comparator CCMP associated to the memory cell CMC0B of CRRA, and so on (Fig. 4). Amidst the blocks CMP1 and CMP2, a central block MSWB contains instead the two multiple switches MSW of the two column redundancy registers CRRA and CRRB. Interconnection lines IL7 and IL8 representing respectively the memory cell status signals CMC0-CMC2 and CMC3-CMC5 are provided to interconnect the blocks LC1 and LC2 to the blocks CMP1 and CMP2, respectively; column address signal lines C0-C2 and C3-C5 are also provided to respectively connect respective signal lines in the column address signal bus CABUS to the comparators CCMP in the blocks CMP1 and CMP2. Interconnection lines IL9 representing the two local buses ICBUS' of the two column redundancy registers CRRA and CRRB are wired from the block LC3 to the block MSWB. Further, interconnection lines IL15 are provided to connect the output channel of the multiple switches MSW in the block MSWB to respective signal lines in the bus ICBUS, that runs on the right side of the upper and lower portions UP and LO.
of the redundancy layout.

Finally, on the right of the blocks CMP1, MSWB and CMP2, a central block RBSC contains the two redundancy bit line selection circuits RBLS and the two switches SW of the two column redundancy registers CRRA and CRRB. An upper block PS1 contains the programming selection circuitry (i.e., the AND gate 6 and the ground/high-voltage switch CHVSW in Figure 7) for one of the two column redundancy registers. Let's say CRRA, while a lower block PS2 contains the programming selection circuitry for the other column redundancy register, CRRB. Interconnection lines IL10 and IL11, representing respectively the output signals CCMP0-CCMP2 and CCMP3-CCMP5 of the comparators CCMP in the blocks CMP1 and CMP2 are provided to respectively interconnect the blocks CMP1 and CMP2 and the block RBSC. A sector selection signal line SS is further provided to supply a respective signal in the sector selection signal bus SBUS to the block RBSC and to the block MSWB. A signal line R5-R8 is provided to supply a row address signal of the set R5-R8 in the row address signal bus RABUS to the block PS1; another signal line R5-R8 is provided to supply another row address signal of the set R5-R8 in the row address signal bus RABUS to the block PS2. Two interconnection lines, representing the program enable signals PGENA and PGENB for the column redundancy registers CRRA and CRRB are respectively provided to connect the blocks PS1 and PS2 to both the blocks PL1, PL2 and PL3, wherein they are connected to the gates of the MOSFETs T0, T1 of the programming load circuits.

Two interconnection lines IL16 and IL17 are provided to connect the two switches SW in RBSC to two respective signal lines of the bus RBSBUS, which runs parallel to the bus ICBUS.

The side of each layout strip LS1-LS4 on the left of the central array of memory elements MAR represents the chip area wherein one row redundancy register RRR is obtained.

Figure 5 is an expanded view of the left side of a given layout strip LS1-LS4. Just beside the central array of memory elements MAR there are provided three layout blocks PL4, PS3 and PS4. The block PL4 is located in central position, and contains the programming load circuits for the memory elements TF0, TF1 of the memory cells RMC0-RMC9 of the row redundancy register RRR; as already told, the programming load circuits are connected to the drain of respective stacked-gate MOSFETs TF0, TF1 in MAR and with drain connected to a respective row address signal line R0, R0N-R9, R9N; interconnection lines IL12 are therefore provided to connect the MOSFETs T0, T1 in the block PL4 to the stacked-gate MOSFETs TF0, TF1 in the central array of memory elements MAR, and row address signal lines R0, R0N-R9, R9N are provided to connect respective signal lines in the row address signal bus RABUS to the MOSFETs T0, T1 in the block PL4. The blocks PS3 and PS4 respectively represent the high-voltage switches RHVSW and RHVSW' of Figure 8; two interconnection lines C0-C3 are also provided to supply the switches RHVSW and RHVSW' in the blocks PS3 and PS4 with a same column address signal of the set C0-C3 in the column address signal bus CABUS. Furthermore, two interconnection lines, respectively representing the signals PRGEN and PRGEN' of Figure 8, are provided to connect the blocks PS3 and PS4 to the block PL4.

On the right of the row address bus RABUS, there is provided a block LC4 which contains the load circuits LC of the memory cells RMC0-RMC9 of the row redundancy register RRR, interconnection lines IL13 are provided to connect the block PL4 to the block LC4. On the right of the block LC4, there is provided a layout block CMP3 containing the comparators RCMP of the row redundancy register RRR, and, on the right of the block CMP3, a layout block RWSCB containing the first- and second-level redundancy word-line selection circuits FRWSC, RWSC and RWSC' of Figure 8. Interconnection lines IL14, representing the memory cell status signals RMCG0-RMCG9 of Figure 8, are provided to connect the block LC4 to the block CMP3; interconnection lines IL19, representing the signals RCMP0-RCMP9 of Figure 8, are provided to connect the block CMP3 to the block RWSCB.

The comparators RCMP in CMP3 are also connected to respective row address signal lines R0-R9 of the row address signal bus RABUS by means of interconnection lines IL18.

Above and below the blocks LC4, CMP3 and RWSCB there are provided two layout blocks RWSB and RWSB', respectively containing the switch RSW and its control circuit SWCNT, and the switch RSW' and its control circuit SWCNT' (Fig. 8). Two interconnection lines IL20 and IL21, representing the redundancy word line selection signals RWSS and RWSS' in Figure 8, are provided between the block RWSCB and the blocks RWSB and RWSB', respectively.

The switches RSW and RSW' in the blocks RWSB and RWSB' are respectively connected to a respective signal line of the bus RWSBUS1 (for the upper portion UP) or RWSBUS2 (for the lower portion LO).

The blocks RWSB and RWSB' are respectively connected to the row address signal bus RABUS by interconnection lines IL22 and IL23, and to a half-matrix selection signal bus HMMBSBUS by two respective interconnection lines IL24 and IL25 representing either the half-matrix selection signal HMMSS1 or HMMSS2.

The layout according to the present invention, as shown in Figures 1 and 2, makes the exchange of signals between the row and column address signal buses and the circuit blocks constituting the redundancy circuitry easy, in particular between the row address signal bus RABUS and the column redundancy registers on the right side of each layout strip, and between the column address signal bus CABUS and the row redundancy registers on the left side of each layout strip.
Claims

1. Redundancy circuitry for a semiconductor memory device, the redundancy circuitry comprising a first plurality of programmable non-volatile memory registers (CCR) for the selection of redundancy bit lines of redundancy memory elements and a second plurality of programmable non-volatile memory registers (RRR) for the selection of redundancy word lines of redundancy memory elements, characterized in that it comprises an array (MAR) of programmable non-volatile memory elements (TF0,TF1), a load circuit (LC) of said first memory cells (CMC0-CMC5) for storing the addresses of defective bit lines and word lines which must be functionally replaced respectively by redundancy bit lines and word lines, and in that it is divided in identical layout strips (LS1-LS4) which are perpendicular to said array (MAR) of memory elements (TF0,TF1) and which comprise each a first and a second strip sides located at opposite sides of the array (MAR) of memory elements (TF0,TF1), the first strip side containing at least one memory register (CRR) of the first plurality and being crossed by a column address signal bus (CABUS) running parallel to the array (MAR) of memory elements (TF0,TF1), the second strip side containing one memory register (RRR) of the second plurality and being crossed by a row address signal bus (RABUS) running parallel to the array (MAR) of memory elements (TF0,TF1).

2. Redundancy circuitry according to claim 1, each memory register (CCR) of said first plurality of programmable non-volatile memory registers comprising first memory cells (CMC0-CMC5) supplied with respective column address signals (C0,C0N,C5,C5N), first comparator means (CCMP) for comparing said column address signals (C0-C5) with output signals (CMC0-CMC5) of said first memory cells (CMC0-CMC5), first selection means (RLS) supplied with output signals (CCMP0-CCMP5) of the first comparator means (CCMP) for selecting one respective redundancy bit line, each memory cell of said first memory cells (CMC0-CMC5) comprising at least one programmable non-volatile memory element (TF0,TF1), a load circuit (LC) for reading the information stored in said memory element (TF0,TF1), and a programming load circuit (T0,T1) for electrically connecting the non-volatile memory element (TF0,TF1) to the respective column address signal (C0,C0N,C5,C5N), further characterized in that each first strip side comprises a first region (PL1,PL2) adjacent to the array (MAR) of memory elements (TF0,TF1) wherein the programming load circuits (T0,T1) of said first memory cells (CMC0-CMC5) of said at least one memory register (CRRA,CRRB) are obtained, a second region (LC1,LC2) wherein the load circuits (LC) of said first memory cells (CMC0-CMC5) of said at least one memory register (CRRA,CRRB) are obtained, the column address signal bus (CABUS) being interposed between said first region (PL1,PL2) and said second region (LC1,LC2), a third region (CMP1,CMP2) adjacent to the second region (LC1,LC2) wherein said first comparator means (CCMP) of said at least one memory register (CRRA,CRRB) are obtained, and a fourth region (RWSCB) adjacent to the third region (CMP1,CMP2) wherein said selection means (RBLSB) of said at least one memory register (CRRA,CRRB) are obtained.

3. Redundancy circuitry according to claim 1, each memory memory register (RRR) of said second plurality of programmable non-volatile memory registers comprising second memory cells (RMC0-RMC9) supplied with respective row address signals (R0,R0N,R9,R9N), second comparator means (RCMP) for comparing said row address signals (R0-R9) with output signals (RMC0-RMC9) of said second memory cells (RMC0-RMC9), second selection means (FRWSC,SRWSC,SRWSC') supplied with output signals (RCMP0-RCMP9) of said second comparator means (RCMP) for selecting at least one respective redundancy word line, each memory cell of said second memory cells (RMC0-RMC9) comprising at least one programmable non-volatile memory element (TF0,TF1), a load circuit (LC) for reading the information stored in said memory element (TF0,TF1), and a programming load circuit (T0,T1) for electrically connecting the non-volatile memory element (TF0,TF1) to the respective row address signal (R0,R0N,R9,R9N), further characterized in that each second strip side comprises a first region (PL4) adjacent to the array (MAR) of memory elements (TF0,TF1) wherein the programming load circuits (T0,T1) of said second memory cells (RMC0-RMC9) of said one memory register (RRR) are obtained, a second region (LC4) wherein the load circuits (LC) of said second memory cells (RMC0-RMC9) of said one memory register (RRR) are obtained, the row address signal bus (RABUS) being interposed between said first region (PL4) and said second region (LC4), a third region (CMP3) adjacent to the second region (LC4) wherein the comparator means (RCMP) of said one memory register (RRR) are obtained, and a fourth region (RWSCB) adjacent to the third region (CMP3) wherein said second selection means (FRWSC,SRWSC,SRWSC') of said one memory register (RRR) are obtained.

4. Redundancy circuitry according to claim 2, further characterized in that the first strip side contains two memory registers (CRRA,CRRB) of said first plurality of programmable non-volatile memory registers, each one associated to a respective redundancy bit
line, the first region (PL1, PL2) of the first strip side containing the programming load circuits (T0, T1) of said first memory cells (CMC0-CMC5) of both the two memory registers (CRRRA, CRRRB), the second region (LC1, LC2) of the first strip side containing the load circuits (LC) of said first memory cells (CMC0-CMC5) of both the two memory registers (CRRRA, CRRRB), the third region (CMP1, CMP2) and the fourth region (RBSC) of the first strip side containing respectively the comparator means (CCMP) and the selection means (RBLSC) of both the memory registers (CRRRA, CRRRB).

5. Redundancy circuitry according to claim 4, further characterized in that said first memory cells (CMC0-CMC5) of one (CRRRA) of the two memory registers (CRRRA, CRRRB) are physically interlaced with said first memory cells (CMC0-CMC5) of the other (CRRRB) of the two memory registers (CRRRA, CRRRB) in such a way that pairs of memory cells of the two memory registers (CRRRA, CRRRB) supplied by identical column address signals (C0, C0N-C5, C5N) are physically adjacent one to another.

6. Redundancy circuitry according to claim 1, further characterized in that it is further divided in two sides (UP, L0), each of said side (UP, L0) comprising a central array (MAR) of memory elements (TF0, TF1) and being divided in layout strips (LS1-LS4) perpendicular to the respective central array (MAR) of memory elements, the two sides (UP, L0) being separated by predetermined circuit blocks (COUNT, PREDEC) of the memory device which are also supplied by the column address signals (CABUS) and by the row address signals (RABUS).

Patentansprüche

1. Redundanzschaltung für eine Halbleiterspeichervorrichtung, mit
   einer ersten Mehrzahl programmierbarer nicht-flüchtiger Speicherregister (CRR) für die Auswahl von Redundanzbitleitungen von Redundanzspeicherzellen und
   einer zweiten Mehrzahl programmierbarer nicht-flüchtiger Speicherregister (RRR) für die Auswahl von Redundanzwortleitungen von Redundanzspeicherzellen,

   gekennzeichnet durch
   ein Feld (MAR) programmierbarer nicht-flüchtiger Speicherelemente (TF0, TF1) zum Lesen der in dem Speicherelement (TF0, TF1) gespeicherten Information und
   eine Programmierlastschaltung (T0, T1) zum elektrischen Verbinden des nicht-flüchtigen Speicherelementes (TF0, TF1) mit dem entsprechenden Spaltenadreßsignal (C0, C0N-C5, C5N),

   weiter dadurch gekennzeichnet,
   daß sie in identische Layout-Streifen (LS1-LS4) unterteilt sind, die senkrecht zu dem Feld (MAR) von Speicherelementen (TF0, TF1) angeordnete Seitenstreifen aufweisen, wobei der erste Seitenstreifen mindestens ein Streifenregister (CRRRA, CRRRB) der ersten Mehrzahl enthält und von einem Spaltenadreßsignalbus (CABUS) gequert wird, der parallel zu dem Feld (MAR) von Speicherelementen (TF0, TF1) verläuft, wobei der zweite Seitenstreifen ein Speicherregister (RRR) der zweiten Mehrzahl enthält und von einem Zeilenadreßsignalbus (RABUS) gequert wird, der parallel zu dem Feld (MAR) von Speicherelementen (TF0, TF1) verläuft.

2. Redundanzschaltung nach Anspruch 1, bei dem jedes Speicherregister (CRR) der ersten Mehrzahl von programmierbaren nicht-flüchtigen Speicherregistern aufweist
   erste Speicherzellen (CMC0-CMC5), die mit entsprechenden Spaltenadreßsignalen (C0, C0N-C5, C5N) beliefert werden, ein erstes Vergleichsmittel (CCMP) zum Vergleichen der Spaltenadreßsignale (C0-C5) mit Ausgangssignalen (CMC0-CMC5) der ersten Speicherzellen (CMC0-CMC5), ein erstes Auswahlmittel (RBLSC), das mit Ausgangssignalen (CMC0-CMP5) des ersten Komparatormittels (CCMP) zum Auswahl einer entsprechenden redundanten Bitleitung beliefert wird, wobei jede Speicherzelle der ersten Speicherzellen (CMC0-CMC5) mindestens ein programmierbares nicht-flüchtiges Speicherelement (TF0, TF1) aufweist, eine Lastschaltung (LC) zum Lesen der in dem Speicherelement (TF0, TF1) gespeicherten Information und
   eine Programmierlastschaltung (T0, T1) zum elektrischen Verbinden des nicht-flüchtigen Speicherelementes (TF0, TF1) mit dem entsprechenden Spaltenadreßsignal (C0, C0N-C5, C5N),

   weiter dadurch gekennzeichnet,
   daß jeder erste Seitenstreifen aufweist
   einen ersten Bereich (PL1, PL2) benachbart zu dem Feld (MAR) von Speicherelementen (TF0, TF1), in dem die Programmierlastschaltungen (T0, T1) der ersten Speicherzellen (CMC0-CMC5) mindestens einen Speicherregister (CRRRA, CRRRB) erhalten sind,
einen zweiten Bereich (LC1, LC2), in dem die Lastschaltungen (LC) der ersten Speicherzellen (CMC0-CMC5) des mindestens einen Speicherregisters (CRRA, CRRB) erhalten sind, wobei der Spaltenadressebus (CABUS) zwischen den ersten Bereich (PL1, PL2) und den zweiten Bereich (LC1, LC2) eingefügt ist, einen dritten Bereich (CMP1, CMP2) benachbart zu dem zweiten Bereich (LC1, LC2), in dem das erste Komparatormittel (CCMP) des mindestens einen Speicherregisters (CRRA, CRRB) erhalten ist, und einen vierten Bereich (RBSC) benachbart zu dem dritten Bereich (CMP1, CMP2), in dem das Auswahlmittel (RBLSC) des mindestens einen Speicherregisters (CRRA, CRRB) erhalten ist.

3. Redundanzschaltung nach Anspruch 1, wobei jedes Speicherregister (RRR) der zweiten Mehrzahl von programmierbaren nicht-flüchtigen Speicherregistern aufweist

zweite Speicherzellen (RMC0-RMC9), die mit entsprechenden Zeilenadresseignalen (R0, R0N-R9, R9N) beliefert werden, ein zweites Komparatormittel (RCMP) zum Vergleichen der Zeilenadresseignale (R0-R9) mit Ausgangssignalen (RMC0-RMC9) der zweiten Speicherzellen (RMC0-RMC9), ein zweites Auswahlmittel (FRWSC, SRWSC, SRWSC'), das mit Ausgangssignalen (RCMP0-RCMP9) des zweiten Komparatormittels (RCMP) beliefert wird, zum Auswählen von mindestens einer entsprechenden Redundanzworteitung, wobei jede Speicherzelle der zweiten Speicherzellen (RMC0-RMC5) aufweist

mindestens ein programmierbares nicht-flüchtiges Speicherelement (TFO, TF1), eine Lastschaltung (LC) zum Lesen der in dem Speicherelement (TFO, TF1) gespeicherten Informationen und eine Programmierlastschaltung (T0, T1) zum elektrischen Verbinden des nicht-flüchtigen Speicherelementes (TFO, TF1) mit dem entsprechenden Zeilenadresseignal (R0, R01-R9, R9N), dadurch gekennzeichnet, daß jeder zweite Seitenstreifen aufweist

einen ersten Bereich (PL4) benachbart zu dem Feld (MAR) von Speicherelementen (TFO, TF1), in dem die Programmierlastschaltungen (T0, T1) der zweiten Speicherzellen (RMC0-RMC9) des einen Speicherregisters (RRR) erhalten sind, einen zweiten Bereich (LC4), in dem die Lastschaltungen (LC) der zweiten Speicherzellen (RMC0-RMC9) des einen Speicherregisters (RRR) erhalten sind, wobei der Zeilenadressebus (RABUS) zwischen den ersten Bereich (PL4) und den zweiten Bereich (LC4) eingefügt ist, einen dritten Bereich (CMP3) benachbart zu dem zweiten Bereich (LC4), in dem das Komparatormittel (RCMP) des einen Speicherregisters (RRR) erhalten ist, und einen vierten Bereich (RWSCB) benachbart zu dem dritten Bereich (CMP3) in dem das zweite Auswahlmittel (FRWCS, SRWSC, SRWSC') des einen Speicherregisters (RRR) erhalten ist.

4. Redundanzschaltung nach Anspruch 2, weiter dadurch gekennzeichnet, daß der erste Seitenstreifen zwei Speicherregister (CRRA, CRRB) der ersten Mehrzahl von programmierbaren nicht-flüchtigen Speicherregistern enthält, von denen jedes mit einer entsprechenden Redundanzbitleitung verbunden ist, wobei der erste Bereich (PL1, PL2) des ersten Seitenstreifens die Programmierlastschaltungen (T0, T1) der ersten Speicherzellen (CMC0-CMC5) beider Speicherregister (CRRA, CRRB) enthält, wobei der zweite Bereich (LC1, LC2) des ersten Seitenstreifens die Lastschaltung (LC) der ersten Speicherzellen (CMC0-CMC5) beider Speicherregister (CRRA, CRRB) enthält, wobei der dritte Bereich (CMP1, CMP2) und der vierte Bereich (RBSC) der ersten Seitenstreifens das Komparatormittel (CMP) bzw. das Auswahlmittel (RBLSC) beider Speicherregister (CRRA, CRRB) enthält.

5. Redundanzschaltung nach Anspruch 4, weiter dadurch gekennzeichnet, daß die ersten Speicherzellen (CMC0-CMC5) eines Speicherregisters (CRRA, CRRB) physikalisch mit den ersten Speicherzellen (CMC0-CMC5) des anderen Speicherregisters (CRRA, CRRB) auf solche Weise verflochten sind, daß Paare von Speicherzellen der zwei Speicherregister (CRRA, CRRB), an die identische Spaltenadresseignale (C0, CON-C5, C5N) geliefert werden, physikalisch benachbart sind.

6. Redundanzschaltung nach Anspruch 1, weiter dadurch gekennzeichnet, daß sie weiter in zwei Seiten (UP, LO) unterteilt ist, wobei jede Seite (UP, LO) ein Mittelfeld
2. Circuit à redondance selon la revendication 1, cha-

Revendications

1. Circuit à redondance pour un dispositif mémoire à

semiconducteur, le circuit à redondance compren-
ant une première pluralité de registres mémoire
non volatils programmables (CRR) pour la sélection
de lignes de bit redondantes d'éléments mémoire
redondants et une seconde pluralité de registres mémoire
non volatils programmables (RRR) pour
la sélection de lignes de mot redondantes d'élé-
ments mémoire redondants, caractérisé en ce qu'il
comprend un réseau (MAR) d'éléments mémoire
non volatils programmables (TF0, TF1) pour mémor-
eriser les adresses de lignes de bit et de lignes de
mot défectueuses qui doivent être fonctionnelle-
ment remplacées respectivement par des lignes de
bit et des lignes de mot redondantes, et en ce qu'il
est divisé en bandes de topologie identiques
(LS1-LS4) qui sont perpendiculaires au réseau
(MAR) d'éléments mémoire (TF0, TF1) et qui com-
priment chacune un premier et un second côté de
bande disposés des côtés opposés du réseau
(MAR) d'éléments mémoire (TF0, TF1), le premier
côté de bande contenant au moins un registre mé-
oire (CRR, CRBB) de la première pluralité et
étant croisé par un bus de signal d'adresse de co-
lonne (CABUS) s'étendant parallèlement au réseau
(MAR) d'éléments mémoire (TF0, TF1), le second
côté de bande contenant un registre mémoire
(RRR) de la seconde pluralité et étant croisé par un
bus de signal d'adresse de rangée (RABUS) s'éten-
dant parallèlement au réseau (MAR) d'éléments
mémoire (TF0, TF1).

2. Circuit à redondance selon la revendication 1, cha-
que registre mémoire (CRR) de la première pluralité
de registres mémoire programmables non volatils
comprenant des premières cellules mémoire
(CMC0, CMC5) alimentées par des signaux
d'adresse de colonne respectifs (C0, CON-C5,
C5N), des premiers moyens comparateurs (CCMP)
pour comparer les signaux d'adresse de colonne
(C0-C5) à des signaux de sortie (CMCS0-CMCS5)
des premières cellules mémoire (CMC0-CMC5),
des premiers moyens de sélection (RBLSC) ali-
mentés par des signaux de sortie

(CCMP0-CCMP5) du premier moyen comparateur
(CCMP) pour sélectionner une ligne de bit redon-
dante respective, chaque cellule mémoire des premières
cellules mémoire (CMC0-CMC5) compren-
ant au moins un élément mémoire non volatil (TF0, TF1), un circuit de charge (LC)
pour lire les informations mémorisées dans l'élé-
ment mémoire (TF0, TF1) et un circuit de charge de
programmation (T0, T1) pour connecter électriquem-
ment l'élément mémoire non volatil (TF0, TF1) au
signal d'adresse de colonne respectif (C0, CON-C5,
C5N), caractérisé en outre en ce que chaque premi-
côté de bande comprend une première région
(PL1, PL2) voisine du réseau (MAR) d'éléments
mémoire (TF0, TF1) dans laquelle les circuits de
charge de programmation (T0, T1) des premières
cellules mémoire (CMC0-CMC5) dudit au moins un
registre mémoire (CRR, CRBB) sont obtenus, une
seconde région (LC1, LC2) dans laquelle les cir-
cuits de charge (LC) des premières cellules mémoire
(CMC0-CMC5) dudit au moins un registre mé-
oire (CRR, CRBB) sont obtenus, le bus de signal
d'adresse de colonne (CABUS) étant interposé en-
tre la première région (PL1, PL2) et la seconde re-
gion (LC1, LC2), une troisième région (CMP1,
CMP2) voisine de la seconde région (LC1, LC2),
dans laquelle les premiers moyens comparateurs
(CCMP) dudit au moins un registre mémoire (CRR,
CRBB) sont obtenus, et une quatrième région
(RBSC) voisine de la troisième région (CMP1,
CMP2) dans laquelle les moyens de sélection
(RBLS) dudit au moins un registre mémoire (CRR,
CRBB) sont obtenus.

3. Circuit à redondance selon la revendication 1, cha-
que registre mémoire (RRR) de la seconde pluralité
de registres mémoire programmables non volatils
comprenant des secondes cellules mémoire
(RMC0-RMC9) alimentées par les signaux d'adresse
de rangée respectifs (R0, R0N-R9, R9N), des
secondes moyens comparateurs (RCMP) pour com-
parer des signaux d'adresse de rangée (R0-R9) aux
signaux de sortie (RMC0-RMC9) des secondes
cellules mémoire (RMC0-RMC9), des secondes
moyens de sélection (FRW, SRW, SRW')
alimentés par les signaux de sortie
(RCMP0-RCPM9) des secondes moyens compara-
teurs (RCMP) pour sélectionner au moins une ligne
de mot redondante respective, chaque cellule mé-
oire des secondes cellules mémoire
(RMC0-RMC5) comprenant au moins un élément
mémoire non volatil programmable (TF0, TF1), un
circuit de charge (LC) pour lire les informations mémor-
isées dans l'élément mémoire (TF0, TF1), et un
circuit de charge de programmation (T0, T1) pour
connecter électriquement l'élément mémoire non
volatil (TF0, TF1) au signal d'adresse de rangée
respectif (R0, R0N-R9, R9N), caractérisé en outre
en ce que chaque second côté de bande comprend une première région (PL4) voisine du réseau (MAR) d'éléments mémoire (T0, T1) dans laquelle les circuits de charge de programmation (T0, T1) des secondes cellules mémoire (RMC0-RMC9) dudit registre mémoire (RRR) sont obtenus, une seconde région (LC4) dans laquelle les circuits de charge (LC) des secondes cellules mémoire (RMC0-RMC9) dudit registre mémoire (RRR) sont obtenus, le bus de signal d’adresse de rangée (RABUS) étant interposé entre la première région (PL4) et la seconde région (LC4), une troisième région (CMP3) voisine de la seconde région (LC4) dans laquelle les moyens comparateurs (RCMP) dudit registre mémoire (RRR) sont obtenus, et une quatrième région (RWSCB) voisine de la troisième région (CMP3) dans laquelle les seconds moyens de sélection (FRWSC, SRWSC, SRWSC’) dudit registre mémoire (RRR) sont obtenus.

4. Circuit à redondance selon la revendication 2, caractérisé en outre en ce que le premier côté de bande contient deux registres mémoire (CRRA, CRRB) de la première pluralité de registres mémoire programmables non volatils, chacun étant associé à une ligne de bit redondante respective, la première région (PL1, PL2) du premier côté de bande contenant les circuits de charge de programmation (T0-T1) des premières cellules mémoire (CMC0-CMC5) des deux registres mémoire (CRRA, CRRB), la seconde région (LC1, LC2) du premier côté de bande contenant les circuits de charge (LC) des premières cellules mémoire (CMC0-CMC5) des deux registres mémoire (CRRA, CRRB), la troisième région (CMP1, CMP2) et la quatrième région (RBSC) du premier côté de bande contenant respectivement les moyens comparateurs (CCMP) et les moyens de sélection (RBLSC) des deux registres mémoire (CRRA, CRRB).

5. Circuit à redondance selon la revendication 4, caractérisé en outre en ce que les premières cellules mémoire (CMC0-CMC5) de l’un (CRRA) des deux registres mémoire (CRRA, CRRB) sont physiquement entrelacées avec les premières cellules mémoire (CMC0-CMC5) de l’autre (CRRB) des deux registres mémoire (CRRA, CRRB) de telle façon que des paires de cellules mémoire des deux registres mémoire (CRRA, CRRB) alimentées par des signaux d’adresse de colonne identiques (C0, C0, C5, C5N) sont physiquement voisines l’une de l’autre.

6. Circuit à redondance selon la revendication 1, caractérisé en outre en ce qu’il est en outre divisé en deux côtés (UP, L0), chaque côté (UP, L0) comprenant un réseau central (MAR) d’éléments mémoire (TF0, TF1) et étant divisé en bandes de configura-
FIG. 4

FIG. 9