



(11) **EP 2 779 507 A1**

(12) **EUROPEAN PATENT APPLICATION**

(43) Date of publication:
17.09.2014 Bulletin 2014/38

(51) Int Cl.:
H04L 1106 (2006.01) **H04L 25/02 (2006.01)**
H04B 11707 (2011.01) **H04B 7/26 (2006.01)**

(21) Application number: **14171776.9**

(22) Date of filing: **05.12.2001**

(84) Designated Contracting States:
AT BE CH CY DE DK ES FI FR GB GR IE IT LI LU MC NL PT SE TR

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(30) Priority: **07.12.2000 US 254013 P**

(62) Document number(s) of the earlier application(s) in accordance with Art. 76 EPC:
09006756.2 / 2 086 147
04015676.2 / 1 463 227
01996123.4 / 1 340 334

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Remarks:

This application was filed on 10-06-2014 as a divisional application to the application mentioned under INID code 62.

(54) **Simple block space time transmit diversity using multiple spreading codes**

(57) A base station, comprising a first channelization device to receive data and spread the data with a first channelization code to produce a first spread data field; a second channelization device to receive the same data as the first channelization device and spread the received data with a second channelization code that is different from the first channelization code to produce a second spread data field; a first training sequence device in communication with the first channelization device, the first training sequence device to combine the first spread data field with a first midamble to produce a first communication burst, the first midamble for estimating a channel impulse response; a second training sequence device in communication with the second channelization device, the second training sequence device to combine the second spread data field with a second midamble to produce a second communication burst, the second midamble for estimating a channel impulse response; a first scrambling device to scramble the first spread data field by a scrambling code associated with the base station; a second scrambling device to scramble the second spread data field by the scrambling code associated with the base station; a first diversity antenna to transmit the first communication burst; and a second diversity antenna to transmit the second communication burst.

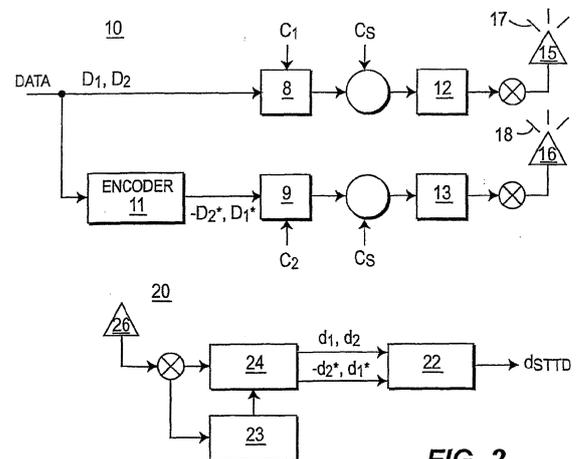


FIG. 2

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Description

BACKGROUND

[0001] The present invention relates to communications systems imploring code division multiple access (CDMA) techniques. More particularly, the present invention relates to a transmission diversity scheme which can be applied to a CDMA communication system

[0002] Spacial diversity has been proposed for support of very high data rate users within third generation wide band code division multiple access systems such as CDMA. Using multiple antennas, the systems achieve better gains and link quality, which results in increased system capacity. Classically, diversity has been exploited through the use of either beam steering or through diversity combining.

[0003] More recently, it has been realized that coordinated use of diversity can be achieved through the use of space-time codes. Such systems can theoretically increase capacity by up to a factor equaling the number of transmit and receive antennas in the array. Space-time block codes operate on a block of input symbols producing a matrix output over antennas and time.

[0004] In the past, space-time transmit diversity systems have transmitted consecutive symbols simultaneously with their complex conjugates. This type of system, though may result in symbol overlap at the receiving end, with the amount of overlap being dependent on the length of the impulse response of the propagation channel. In time division duplex (TDD) mode, this symbol overlap will have to be accounted for in the joint detection receiver. The joint detector will have to estimate the transmitted symbols and their conjugates, resulting in an increase in complexity of the joint detection.

[0005] In order to alleviate this increase in joint detection, systems have been created which transmit two similar but different data fields. The first data field, having a first portion, D_1 , and a second portion, D_2 , is transmitted by the first antenna. A second data field is produced by modifying the first data field. The negation of the conjugate of D_2 , D_2^* , is the first portion of the second data field and the conjugate of D_1 , D_1^* , is the second portion. The second data field is simultaneously transmitted by the second antenna. This type of system results in the joint detection implemented at the receiver needing only to estimate the same amount of symbols as in the case of a single transmit antenna. A block diagram of this system is illustrated in Figure 1.

[0006] Although the above system reduces the complexity of joint detection for a single data block, joint detection requires the use of two joint detectors at the receiver in a system employing two transmit diversity antennas. Each joint detection device estimates the data from one of the antennas. The estimated data is combined to produce the original data. Therefore, the receiver in such a system has a high complexity resulting in higher receiver expense.

[0007] Accordingly, there exists a need for a transmit diversity system requiring less complexity and receiver expense.

5 SUMMARY

[0008] The present invention is a system and method for use in a CDMA communication system including a plurality of base stations and a user equipment (UE), each for communicating with each other. The base station has a transmitter which includes a first and second antenna for transmitting a data field of symbols. The first spreading device spreads the first data field using a first channelization code and the second spreading device spreads the second data field using a second channelization code, each channelization code being uniquely associated with one of the first and second antennas. The UE has a receiver for receiving a signal including the first and second spread data fields. The UE includes a joint detection device for detecting the symbols of the first and second data fields using the first and second channelization codes and a decoder for decoding the detected data fields to generate a single data field of symbols.

BRIEF DESCRIPTION OF THE DRAWINGS

[0009]

30 Figure 1 is a block diagram of a prior art communication system employing space-time transmit diversity.

35 Figure 2 is a block diagram of a transmitter and receiver in a communication system in accordance with the preferred embodiment of the present invention.

Figure 3 is a flow diagram of the transmit diversity system of the present invention.

40 Figure 4 is a graph of the performance of the transmit diversity system of the present invention.

45 Figure 5 is a block diagram of a transmitter and receiver in a communication system in accordance with an alternative embodiment of the present invention.

Figure 6 is a flow diagram of an alternative transmit diversity system of the present invention.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

50 **[0010]** Figure 2 is a block diagram of a transmitter 10, preferably located at a base station, and a receiver 20, preferably located at a user equipment (UE), in a CDMA communication system in accordance with the preferred embodiment of the present invention. Although it is preferable to have the transmitter located at a base station and the receiver located at the UE, the receiver and transmitter may switch locations and the present invention op-

erate on an uplink communication. The transmitter 10 comprises a block encoder 11, a plurality of channelization devices 8, 9, a plurality of spreading sequence insertion devices 12, 13, and a plurality of antennas 15, 16. Although Figure 1 illustrates a transmitter comprising two (2) antennas, it should be apparent to those having skill in the art that more than two (2) antennas may be used, such as N antennas.

[0011] A typical communication burst has two data fields separated by a midamble sequence. Preferably, the same encoding procedure, as discussed in the following, for one data field is also performed on the other data field. Data to be transmitted by the transmitter 10 is produced by a data generator (not shown). The resulting data symbols ($S_1, S_2, \dots, S_{N/2}, (S_{N/2+1}, S_{N/2+2}, \dots, S_N)$) of the first data field, which can be represented by sub-data fields D_1 and D_2 , are input into the block encoder 11, preferably a block space-time transmit diversity (BSTTD) encoder. The block encoder 11 encodes the input symbols and generates the complex conjugate of D_1 and the negation of the conjugate of D_2 : D_1^* , $-D_2^*$. The encoder 11 also changes the order of the symbols so that $-D_2^*$ is ahead of D_1^* . Preferably, an analogous encoding of the second data field is also performed.

[0012] In accordance with the preferred embodiment of the present invention, the data fields, D_1, D_2 and $-D_2^*, D_1^*$ are forwarded to a first and second channelization device 8, 9, respectively. The first channelization device 8 spreads the data blocks D_1, D_2 by a first channelization code, and $-D_2^*, D_1^*$ by the second channelization device 9 using a second different channelization code. Each of the spread data blocks from the first and second channelization devices 8, 9 are then scrambled by the scrambling code associated with the transmitter 10.

[0013] Once the symbols $D_1, D_2, -D_2^*, D_1^*$ have been scrambled, they are mixed with a first and second midamble through training sequence insertion devices 12, 13, producing two communication bursts 17, 18. The two bursts 17, 18 are modulated and simultaneously transmitted to the receiver 20 over antenna 15 and diversity antenna 16, respectively.

[0014] The receiver 20 comprises a joint detection device (JD) 24, a BSTTD decoder 22, a channel estimation device 23 and an antenna 26. The antenna 26 of the UE receives various RF signals including the communication bursts 17, 18 from the transmitter 10. The RF signals are then demodulated to produce a baseband signal.

[0015] The baseband signal is then forwarded to the joint detection device 24 and the channel estimation device 23. As those skilled in the art know, the channel estimation device 23 provides channel information, such as channel impulse responses, to the joint detection device 24.

[0016] The joint detection device 24, coupled to the channel estimation device 23 and BSTTD decoder 22, utilizes the channel information and the channelization codes to detect the soft data symbols $d_1, d_2, -d_2^*, d_1^*$ in the received signal. The channel impulse response for

each burst is determined using that burst's midamble sequence. Since each burst was transmitted using a different spreading code, the joint detection device 24 treats each burst as being transmitted by a different user. As a result, any joint detection device which can recover data from different transmitter sites may be used. Such joint detection devices include zero forcing block linear equalizers, detection devices using Cholesky or approximate Cholesky decomposition, as well as many others. The joint detection device 24 estimates the data symbols of each of the bursts 17, 18 output by the transmitter antennas 15, 16 and forwards the estimates to the BSTTD decoder 22.

[0017] The BSTTD decoder 22, coupled to the joint detection device 24, receives the estimated soft data symbols d_1, d_2 and $-d_2^*, d_1^*$ corresponding to the antennas 15, 16 and decodes the symbols to yield a single data field's soft symbols, d_{STTD} .

[0018] The flow diagram of the present invention is illustrated in Figure 3. A data generator generates data to be transmitted to the receiver 20 (step 301). Each data field is separated into two sub-data fields D_1, D_2 (step 302). The sub-data fields D_1, D_2 are forwarded to the block encoder 11 and the first channelization device 8 (step 303). The sub-data fields forwarded to the block encoder 11 are encoded (step 304) and forwarded to the second channelization device 9 (step 305). Each channelization device 8, 9 spreads their respective data input using a separate channelization code associated with a respective antenna 15, 16 (step 306). The two spread signals are then scrambled, using the scrambling code associated with the base station (step 307) and transmitted to the receiver 20 over diversity antennas 15, 16 (step 308).

[0019] The receiver 20 receives a RF communication signal including the two spread signals from the diversity antennas 15, 16 (step 309), demodulates the signal and forwards the demodulated signal to the channel estimation device 23 and joint detection device 24 (step 310). The received signal is processed by the channel estimation device 23 (step 311) and the channel information applied by the joint detection device 24 along with the channelization codes, to estimate the transmit symbols from the diversity antennas 15, 16 (step 312). The detected sub-data fields, corresponding to each diversity antenna 15, 16, are forwarded to the BSTTD decoder (step 313), which decodes the soft symbol sub-fields to yield a single data field's soft symbols, d_{STTD} (step 314).

[0020] Similar to the preferred embodiment disclosed above, Figure 5 is a block diagram of an alternative transmitter 40, preferably located at a base station, and a receiver 50, preferably located at a user equipment (UE) in a communication system. The transmitter 40 comprises a plurality of channelization devices 48, 49, a plurality of spreading sequence insertion devices 42, 43, and a plurality of antennas 45, 46.

[0021] Data to be transmitted by the transmitter 40 is produced by a data generator (not shown). The resulting

data symbols ($S_1, S_2, \dots, S_{N/2}$), ($S_{N/2+1}, S_{N/2+2}, \dots, S_N$) of the first data field, which can be represented by sub-data fields D_1 and D_2 , are input to a first and second channelization device 48,49, respectively. The first channelization device 8 spreads the data blocks D_1, D_2 by a first channelization code, and the second channelization device 49 spreads the data blocks D_1, D_2 by a second different channelization code. Each of the spread data blocks from the first and second channelization devices 48, 49 are scrambled by the scrambling code associated with the transmitter 40.

[0022] Once the symbols have been scrambled, they are mixed with a first and second midamble through training sequence insertion devices 42, 43, producing two communication bursts 44,45. The two bursts 44, 45 are modulated and simultaneously transmitted to the receiver 50 over antenna 46 and diversity antenna 47, respectively.

[0023] The receiver 50 comprises a joint detection device (JD) 54, a decoder 22, a channel estimation device 53 and an antenna 51. The antenna 51 of the UE receives various RF signals including the communication bursts 44,45 from the transmitter 40. The RF signals are then demodulated to produce a baseband signal.

[0024] The baseband signal is then forwarded to the joint detection device 54 and the channel estimation device 53. The joint detection device 54, coupled to the channel estimation device 53 and decoder 52, utilizes the channel information and the channelization codes to detect the soft data symbols d_1, d_2 , in the received signal. The channel impulse response for each burst is determined using that burst's midamble sequence. Since each burst was transmitted using a different spreading code, the joint detection device 54 treats each burst as being transmitted by a different user. The joint detection device 54 estimates the data symbols of each of the signals 44,45 output by the transmitter antennas 46, 47 and forwards the estimates to the decoder 52.

[0025] The decoder 52, coupled to the joint detection device 54, receives the estimated soft data symbols d_1, d_2 corresponding to the antennas 46, 47 and decodes the symbols to yield a single data field's soft symbols, d .

[0026] The flow diagram of the alternative embodiment is illustrated in Figure 6. A data generator generates data to be transmitted to the receiver 40 (step 601). Each data field is separated into two sub-data fields D_1, D_2 (step 602). The sub-data fields D_1, D_2 are forwarded to the first channelization device 48 and to the second channelization device 49 (step 603). Each channelization device 48, 49 spreads their respective data input using a separate channelization code associated with each antenna 46, 47 (step 604). The two spread signals are then scrambled, using the scrambling code associated with the base station (step 605) and transmitted to the receiver 50 over diversity antennas 46, 47 (step 606).

[0027] The receiver 50 receives a RF communication signal including the two spread signals from the diversity antennas 46, 47 (step 607), demodulates the signal and

forwards the demodulated signal to the channel estimation device 53 and joint detection device 54 (step 608). The received signal is processed by the channel estimation device 53 (step 609) and the channel information applied by the joint detection device 54 along with the channelization codes, to estimate the transmit symbols from the diversity antennas 46,47 (step 610). The detected sub-data fields, corresponding to each diversity antenna 46, 47, are forwarded to the decoder 52 (step 611), which decodes the soft symbol sub-fields to yield a single data field's soft symbols, d_{STTD} (step 612).

[0028] By using additional channelization codes, the above approaches can be applied to an antenna array having any number of antennas. Each antenna has its own associated channelization code and midamble. If a block encoder is used, the data field transmitted by each of the antennas has a unique encoding, allowing the use of a single joint detector at the receiver.

[0029] The BSTTD transmitter with two channelization codes of the present invention allows for the use of a cheaper and simpler method of transmit diversity. The use of different channelization codes per transmit antenna requires only one joint detection device at the receiver resulting in a less complex receiver system than those of the prior art. Figure 4 is a graph showing the raw BER of various block STTD decoders. The model is based on all the receivers using a block linear equalizer (BLE) based approach to JD. NTD means the single antenna case, i.e., no transmit diversity. STTD with 1 code is the traditional block STTD JD. STTD with 2 code is the disclosed block STTD transmitter. Simple STTD with 2 code is the transmission system disclosed in the alternative embodiment As illustrated, the benefit of 2 codes for STTD can be summarized as follows: 1) there is up to a 0.5 dB gain at 0.01 raw Bit error rate over 1 code STTD; and 2) by eliminating the encoding block in simple STTD with 2 code, the performance degradation is only 0.2 dB at 0.1 raw BER and no degradation at 0.01 raw BER. The performance improvement over NTD is still 1.0 dB and 2.7 dB at 0.1 and 0.01 raw BER.

ASPECTS OF THE INVENTION

[0030]

1. A user equipment using a diversity scheme in a code division multiple access, CDMA, telecommunications system comprising:

receiving means (26) for receiving a signal including a first communication burst transmitted by a first transmitting means and a second communication burst transmitted by a second transmitting means, the first and the second communication burst being produced using a first and a second channelization code on data fields of symbols, respectively, each channelization code being uniquely associated with one of said

first and second transmitting means;
 detecting means (24) for detecting the symbols
 of said first and second communication burst by
 using the first and second channelization code;
 and
 decoding means (22) for decoding said detected
 symbols of the first and second communication
 burst to generate a single data field of symbols.

2. The user equipment of aspect 1, wherein the received first and second communication burst include a first and a second training sequence inserted as midamble between two data fields, respectively.

3. The user equipment of aspect 2, further comprising estimating means (23) for estimating channel information of the first communication burst as first channel information by using the midamble of the first communication burst and for estimating channel information of the second communication burst as a second channel information by using the midamble of the second communication burst.

4. The user equipment of aspect 3, wherein the detecting means (24) is adapted for detecting the symbols of the first and second communication burst by using the first and the second channel information.

5. The user equipment of one of the preceding aspects, wherein the detecting means (24) comprises a joint detector.

6. The user equipment of one of the preceding aspects, wherein detecting means (24) comprises zero forcing block linear equalizers, BLE.

7. The user equipment of one of the preceding aspects, wherein the decoding means (22) comprises a block space-time transmit diversity, BSTTD, decoder.

8. The user equipment of one of the preceding aspects, wherein the estimating means (23) provides channel impulse responses of the first and the second communication burst.

9. The user equipment of one of the preceding aspects, further comprising demodulating means for producing a baseband signal from the received signal.

Claims

1. A base station, comprising:

a first channelization device to receive data and spread the data with a first channelization code

to produce a first spread data field;
 a second channelization device to receive the same data as the first channelization device and spread the received data with a second channelization code that is different from the first channelization code to produce a second spread data field;
 a first training sequence device in communication with the first channelization device, the first training sequence device to combine the first spread data field with a first midamble to produce a first communication burst, the first midamble for estimating a channel impulse response;
 a second training sequence device in communication with the second channelization device, the second training sequence device to combine the second spread data field with a second midamble to produce a second communication burst, the second midamble for estimating a channel impulse response;
 a first scrambling device to scramble the first spread data field by a scrambling code associated with the base station;
 a second scrambling device to scramble the second spread data field by the scrambling code associated with the base station;
 a first diversity antenna to transmit the first communication burst; and
 a second diversity antenna to transmit the second communication burst.

2. A system for diversity transmission, the system comprising:

a base station, comprising:

a first channelization device to receive data and spread the data with a first channelization code to produce a first spread data field, and wherein the first channelization code is utilized exclusively for transmission on the first antenna;
 a second channelization device to receive the same data as the first channelization device and spread the received data with a second channelization code that is different from the first channelization code to produce a second spread data field, and wherein the second channelization code is utilized exclusively for transmission on the second antenna;
 a first training sequence device in communication with the first channelization device, the first training sequence device to combine the first spread data field with a first midamble to produce a first communication burst, the first midamble for estimating a

channel impulse response;
 a second training sequence device in communication with the second channelization device, the second training sequence device to combine the second spread data field with a second midamble to produce a second communication burst, the second midamble for estimating a channel impulse response;
 a first scrambling device to scramble the first spread data field by a scrambling code associated with the base station;
 a second scrambling device to scramble the second spread data field by the scrambling code associated with the base station;
 a first diversity antenna to transmit the first communication burst; and
 a second diversity antenna to transmit the second communication burst; and

a user equipment, comprising:

a data detection device having an input configured to receive the signal being diversity transmitted, the received signal having the first communication burst and the second communication burst, the first and second communication bursts being transmitted from the same base station in a same time slot over different antennas and the first and second communication bursts having the same data in a same order, each of the first and second communication bursts having the inserted midamble training sequence used for estimating a channel impulse response, and a scrambling code associated with a transmitting base station, the data detection device configured to process the received signal with the first channelization code associated with the first transmission antenna to produce first data and the data detection device configured to process the received signal with the second channelization associated with the second transmission antenna to produce second data, wherein the first and second data are the same data.

- 3. The system of claim 2 further comprising a decoder configured to process the first and second data to produce the same data.
- 4. The system of claim 2 wherein the data detection device is a joint detection device.
- 5. The system equipment of claim 2 further comprising a channel estimation device configured to provide the channel impulse response estimates for the first

- and second communication bursts to the data detection device using the inserted midamble training sequences.
- 6. The system of claim 2 wherein the data detection device is further configured to descramble the first and second communication bursts using the scrambling code associated with the base station.

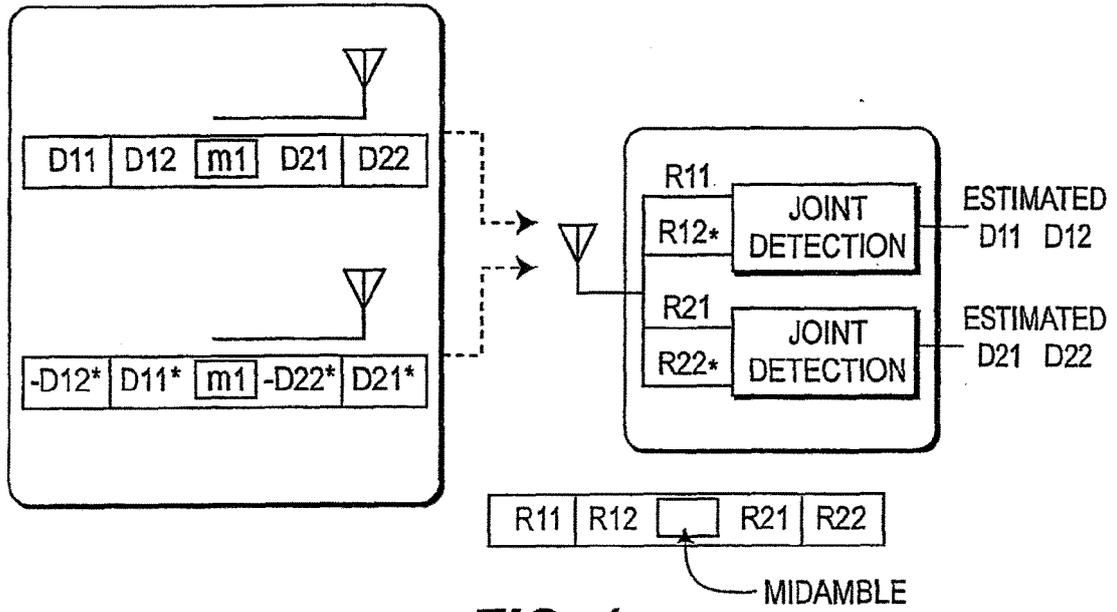


FIG. 1

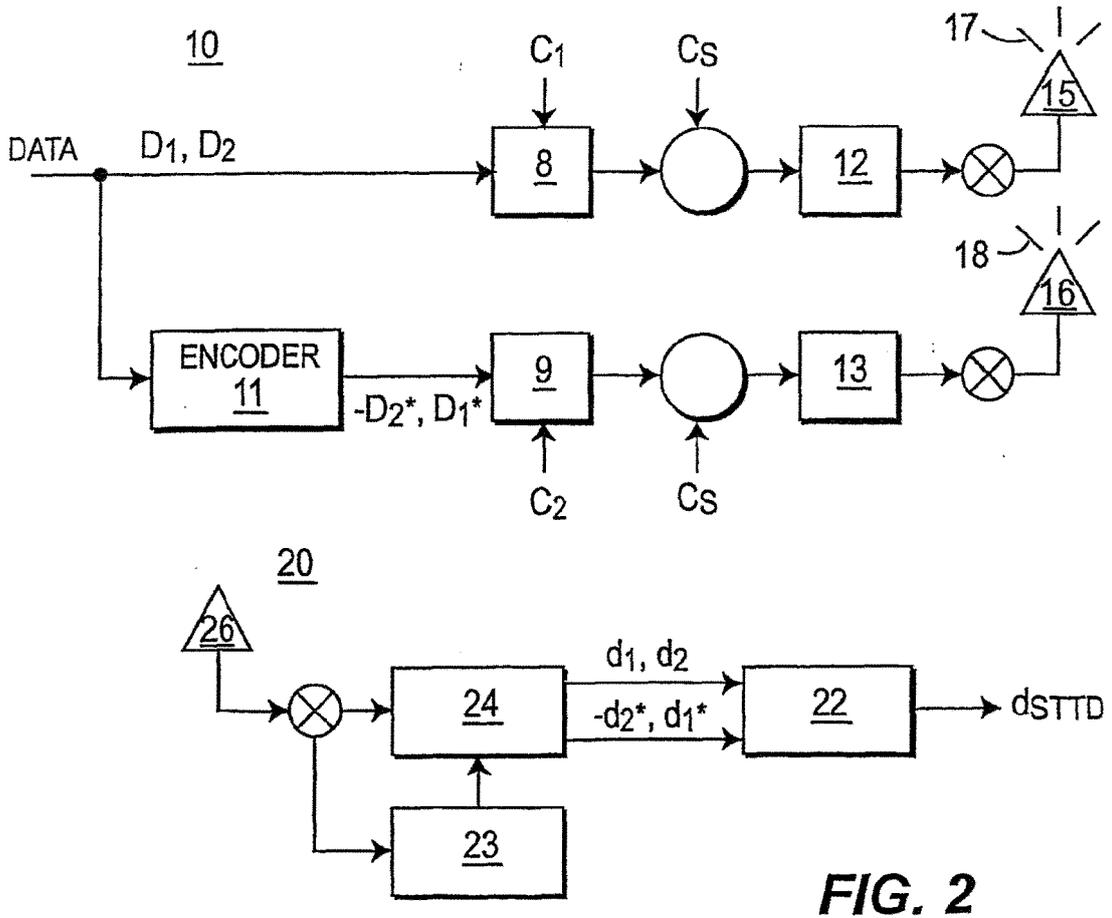


FIG. 2

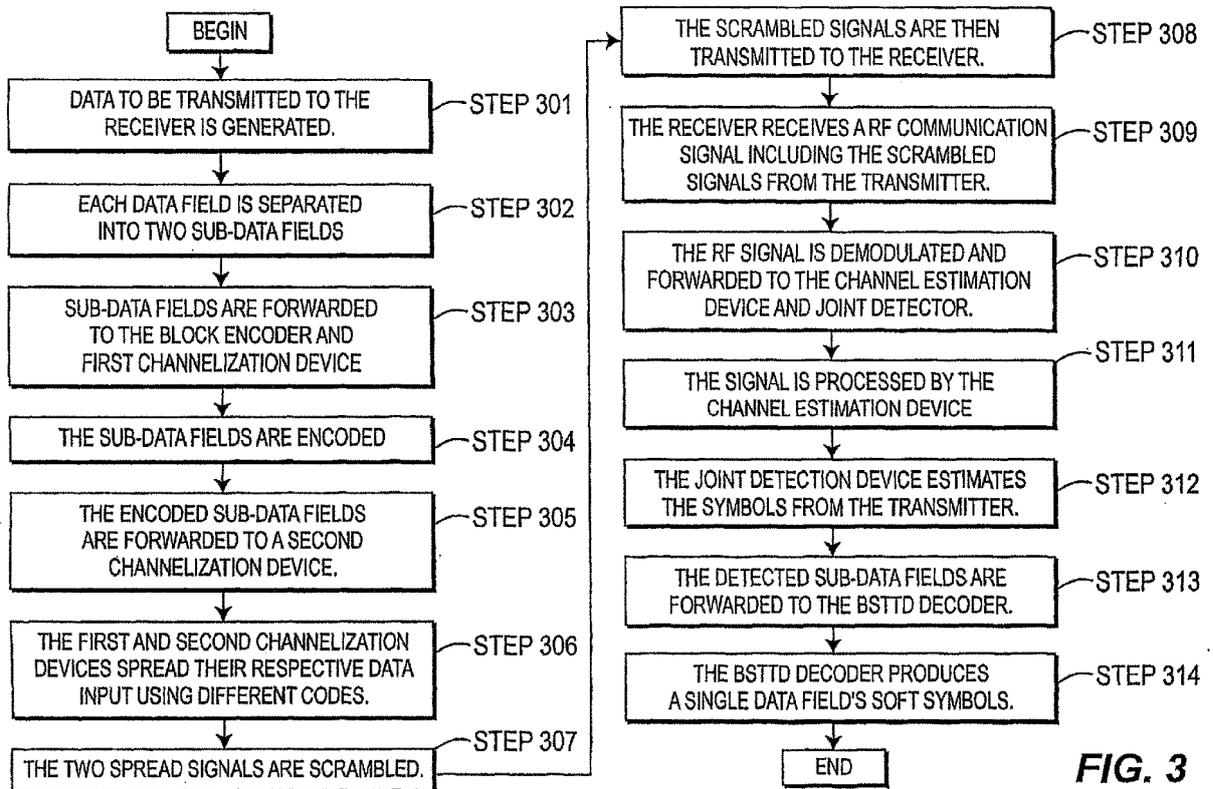


FIG. 3

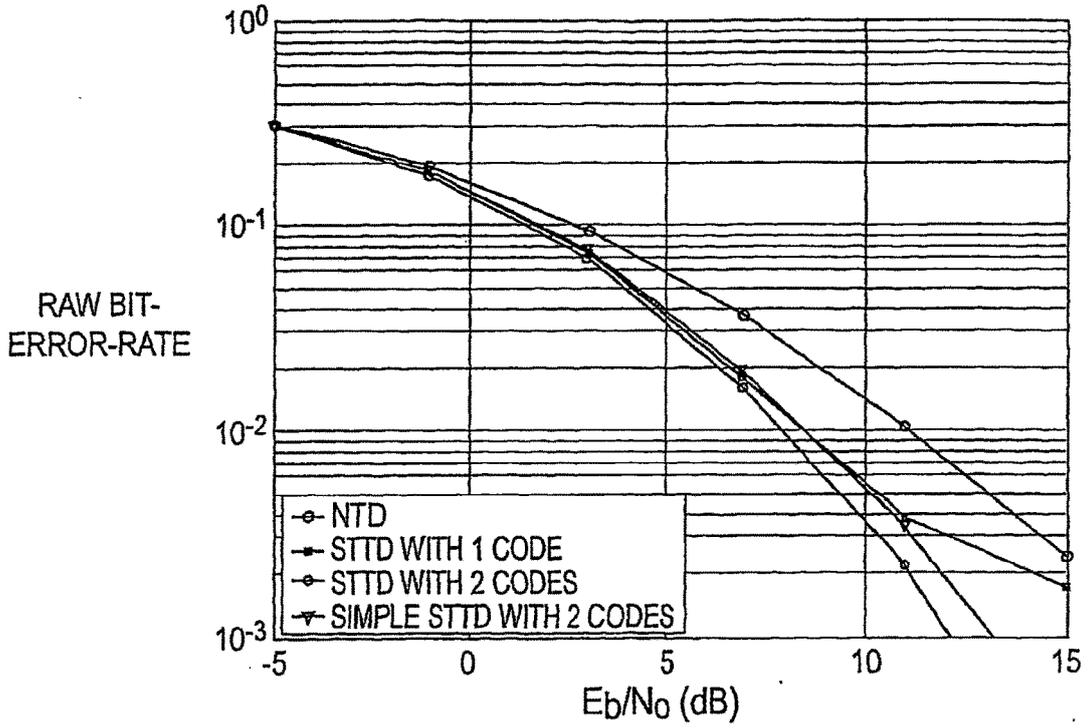


FIG. 4

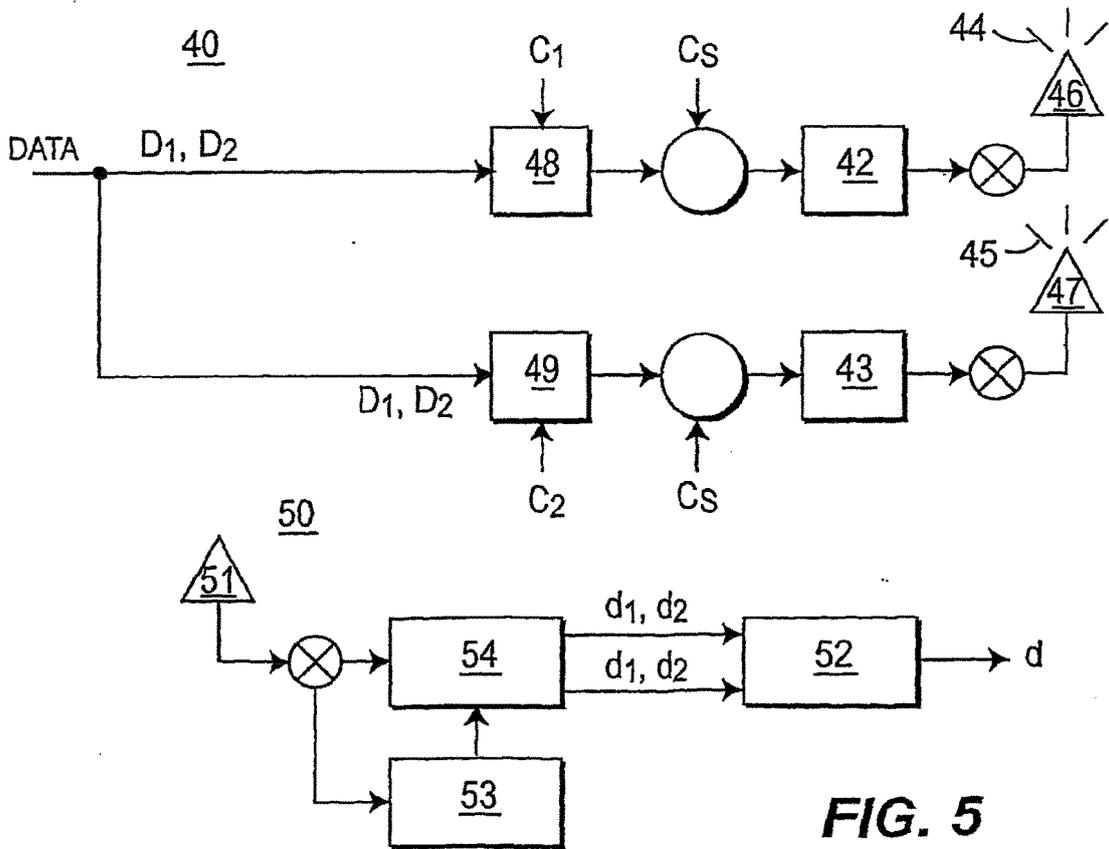


FIG. 5

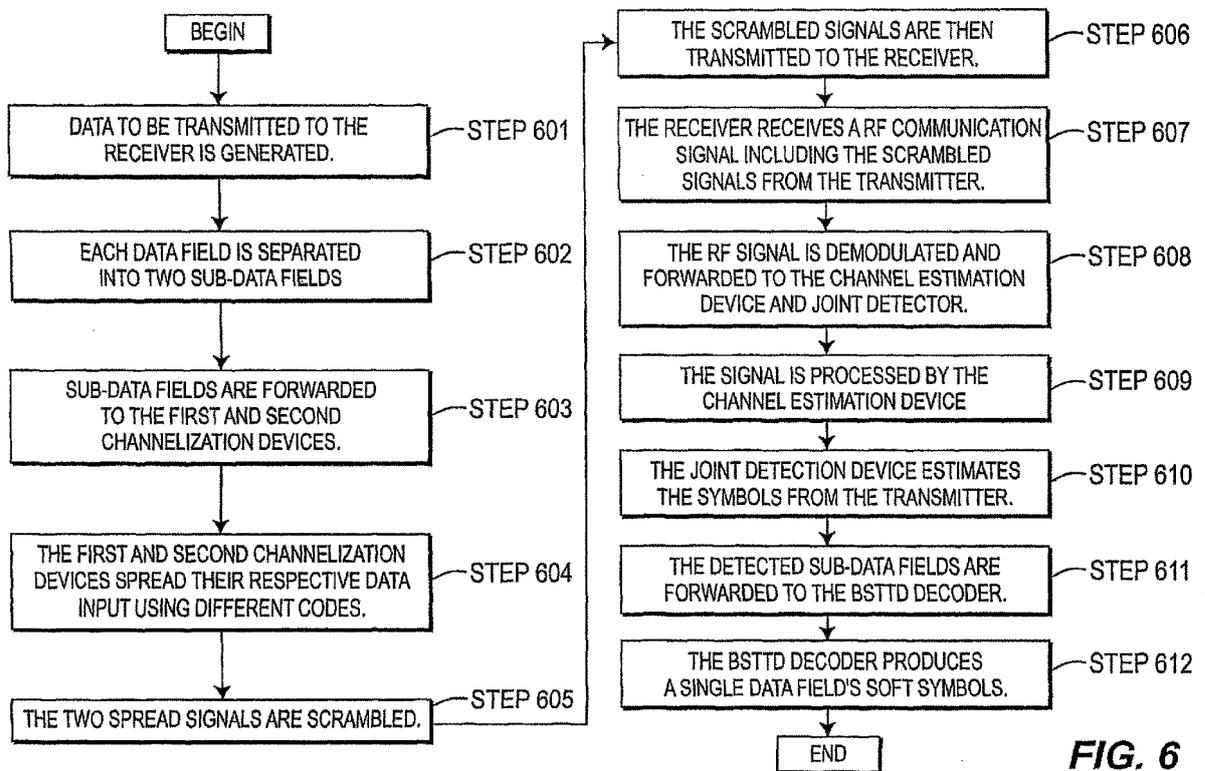


FIG. 6



EUROPEAN SEARCH REPORT

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摘要

一个基站包括：第一信道化装置，该第一信道化装置以接收数据和使用第一信道化编码扩展数据，以便产生第一扩展数据字段；第二信道化装置，该第二信道化装置接收与所述第一信道化装置相同的数
据，并采用第二信道化编码扩展所接收的数据以便产生第一扩展数据字段，所述第二信道化编码不同于
所述第一信道化编码；第一训练序列装置，该第一训练序列装置与所述第一信道化装置通信，该第一训
练序列装置组合所述第一扩展数据字段与第一中间码以产生第一通信突发，所述第一中间码用于估计一个
信道脉冲响应；第二训练序列装置，该第二训练序列装置与所述第二信道化装置通信，该第二训练序
列装置组合所述第二扩展数据字段与第二中间码以产生第二通信突发，所述第二中间码用于估计一个信
道脉冲响应；第一扰码器，所述第一扰码器通过与该基站相连的一个干扰码扰乱第一扩展数据字段；第
二扰码器，所述第二扰码器通过与该基站相连的一个干扰码扰乱第二扩展数据字段；第一分集天线，所
述第一分集天线传输所述第一通信突发；以及第二分集天线，所述第二分集天线传输所述第二通信突
发。