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Electronic musical instrument with waveform memory for storing waveform data based on external sound

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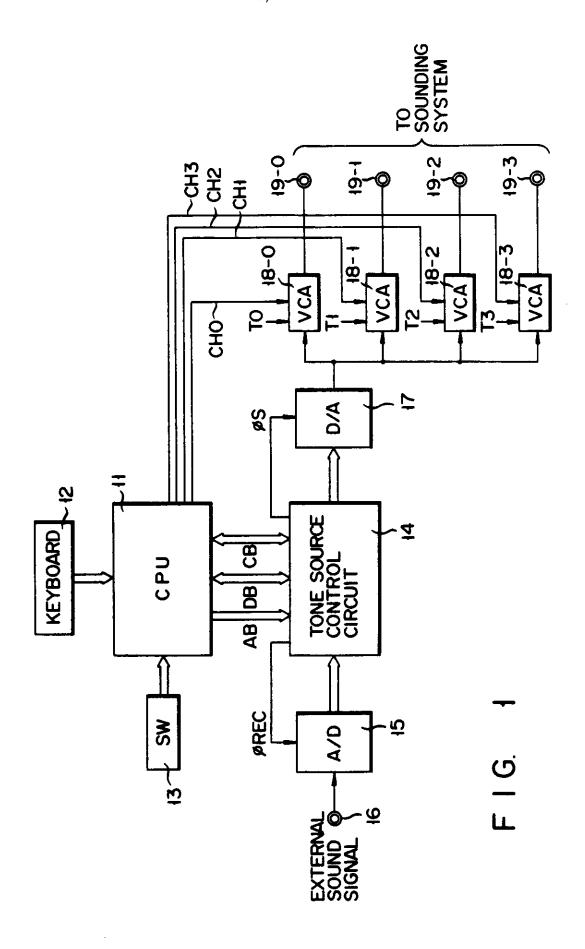
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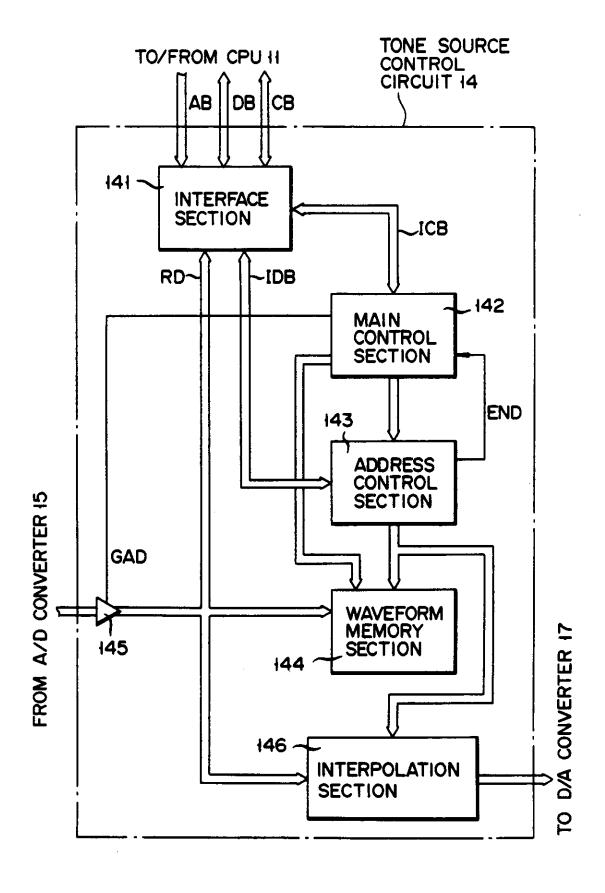
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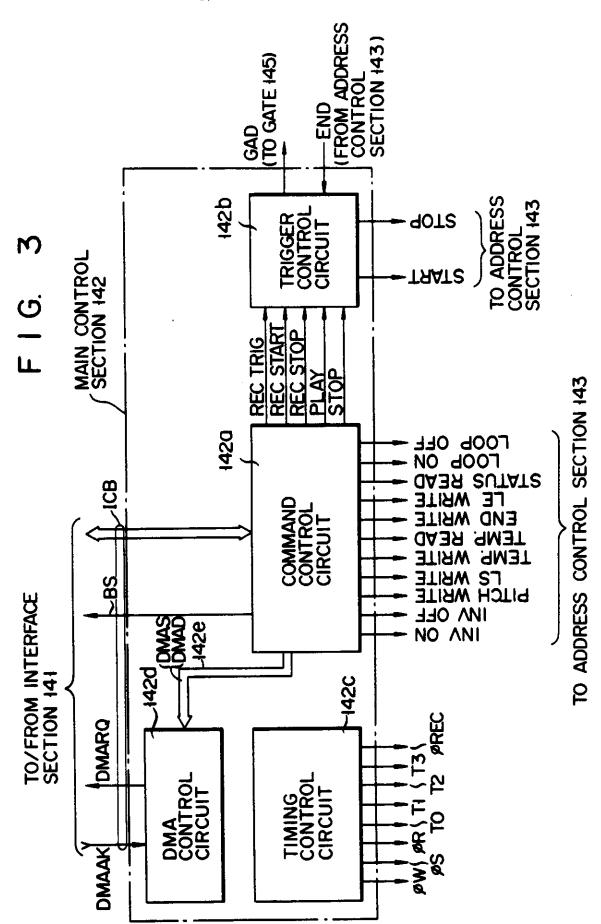
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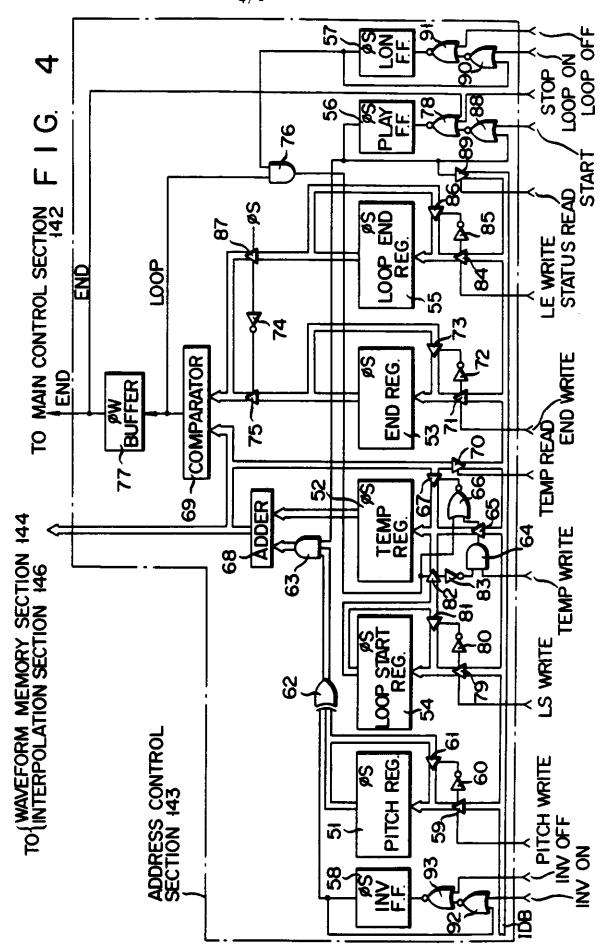
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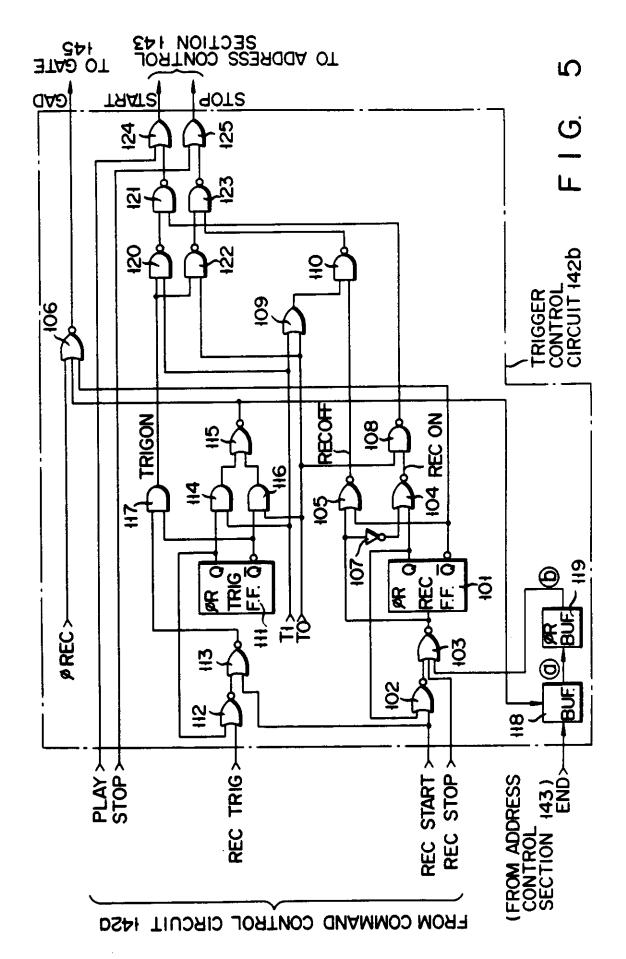


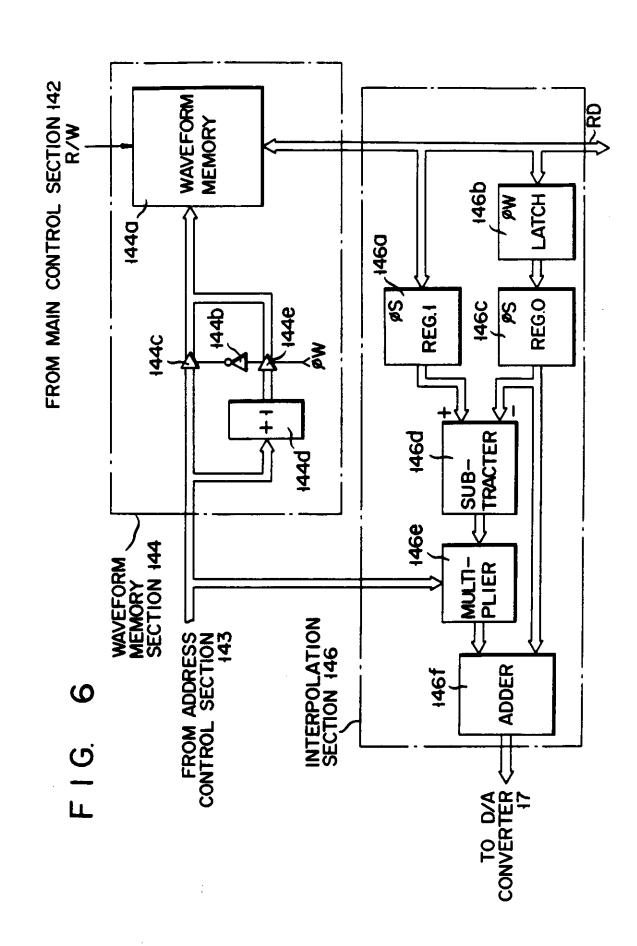
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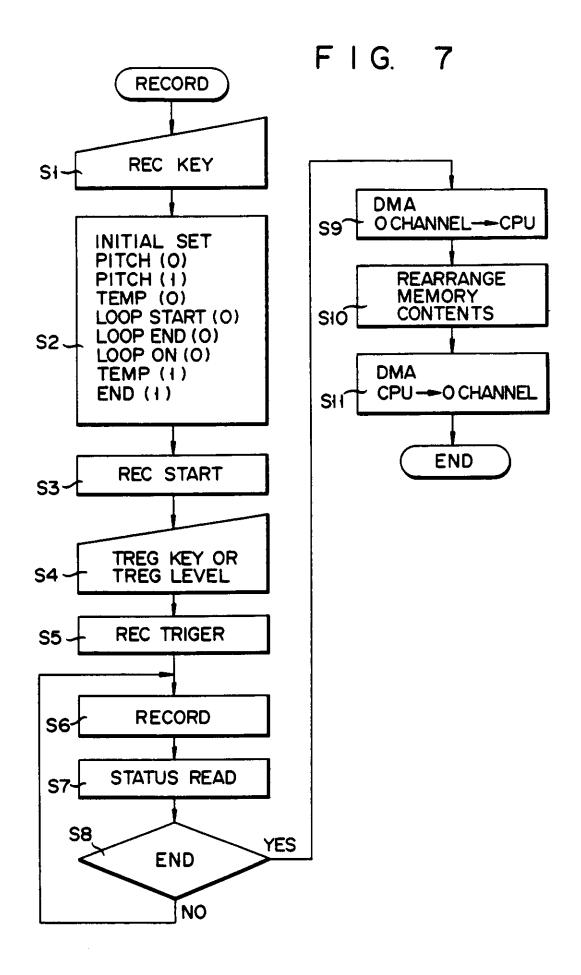


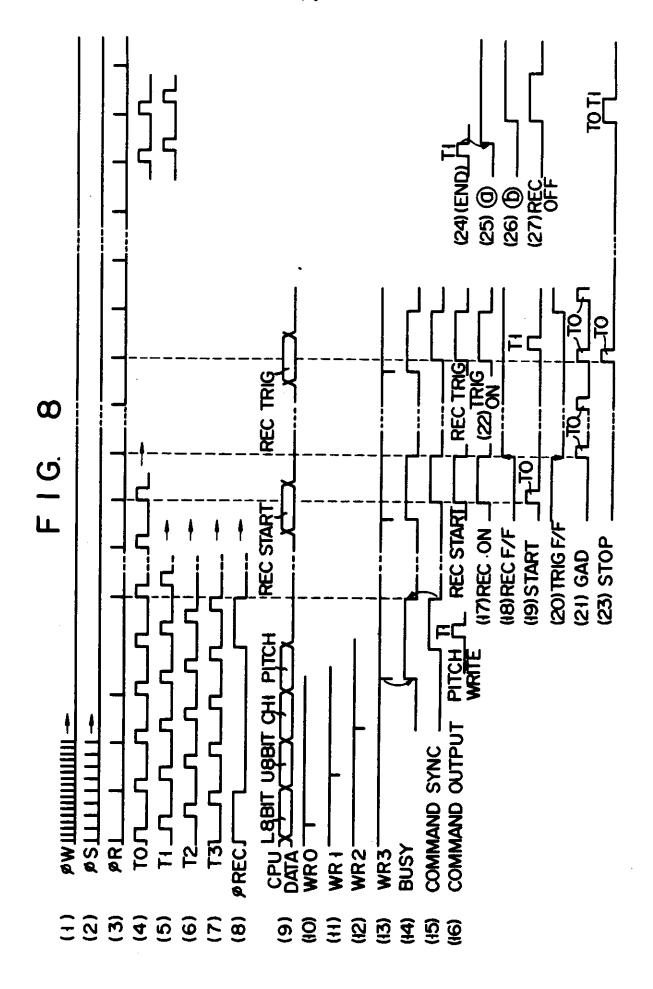












"ELECTRONIC MUSICAL INSTRUMENT WITH WAVEFORM MEMORY FOR STORING WAVEFORM DATA BASED ON EXTERNAL SOUND"

This invention relates to an electronic musical instrument of the type in which externally fed piano or like sounds are converted into a digital waveform signal which is stored in a memory, sounds having various waveforms being obtained thereby through control of the addressing of the memory.

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Heretofore, an electronic musical instrument has been developed in which a tone of, for instance, Cl, 10 obtained by depressing a Cl key of a piano, is converted through a microphone into a Cl tone signal which is A/D converted into a PCM digital signal to be stored in a waveform memory. In order to effectively use the memory area of the waveform memory, the operation of storing 15 the PCM digital signal in the waveform memory is started simultaneously with the depression of a note key such as the Cl note key. With this arrangement, however, it is likely that a leading, rising portion of the piano To overcome this drawback, the tone fails to be stored. 20 operation of storing data in the waveform memory may be started before the key depression. In this case, however, it is necessary to use a waveform memory having a large capacity, leading to a cost increase.

Meanwhile, in the prior art electronic musical instrument noted above, the waveform data stored in the waveform memory is read out under control by a

DMAC (direct memory access controller), and waveform data is DMA transferred from the waveform memory to a FIFO (first in first out) buffer every time a key on the keyboard is operated. The waveform data stored in the FIFO buffer is fed to a D/A converter according to the output of a VCO, which provides a frequency signal corresponding to the note of the depressed key, whereby a tone having a predetermined The DMA transfer of the waveform pitch is obtained. data from the waveform memory to the FIFO buffer is done under control of a CPU. In a polyphonic electronic musical instrument where signals are processed through time division basis processing, therefore, the CPU is used exclusively for the DMA processing, such that it can devote insufficient time to other processing.

Further, the polyphonic electronic musical instrument noted above requires pluralities of VCOs, FIFO buffers, D/A converters, etc., so that its price is inevitably high.

The object of the invention is to provide an electronic musical instrument in which the processing time required by the CPU for the writing and reading of externally fed tones can be reduced, with external sound signals being efficiently and accurately stored in a waveform memory of a simple circuit construction.

This application has been divided from GB 8810919, which in turn was divided from GB 8520019, and contains subject matter in common with both of these applications.

According to the invention, there is provided a polyphonic electronic musical instrument comprising: waveform memory means for writing therein a sound waveform signal which is a digital signal obtained by A/D conversion of an external sound signal;

address signal generator means for time-

divisionally generating address signals, which correspond to a plurality of channels, for writing and reading the sound waveform signal in and from said waveform memory means;

control means for providing commands to the address signal generator means to write and read the sound waveform signal in and from said waveform memory means;

write control means for writing the sound waveform signal such that an address signal is supplied to said waveform memory means by at least one of said plurality of channels when said control means commands said waveform memory means to write the sound waveform signal therein;

performance means operable to initiate reading out of the sound waveform signal written in said waveform memory means for each of a plurality of sounds at a respective desired pitch; and

read control means for time-divisionally supplying read address signals, which correspond to said plurality of sounds, from said plurality of channels to said waveform memory means and outputting sound waveform signals corresponding to said plurality of sounds, in accordance with the operation of said performance means.

This invention can be more fully understood from the following detailed description when taken in conjunction with the accompanying drawings, in which:

Fig. 1 is a block diagram showing an embodiment of the invention;

Fig. 2 is a block diagram showing a tone source control circuit shown in Fig. 1;

Fig. 3 is a block diagram showing the main control section shown in Fig. 2;

Fig. 4 is a schematic representation of an address control section shown in Fig. 2;

Fig. 5 is a schematic representation of a trigger control circuit shown in Fig. 3;

Fig. 6 is a block diagram showing a waveform memory section and an interpolation section shown in Fig. 2;

Fig. 7 is a flow chart explaining a recording operation of the embodiment shown in Figs. 1 to 6; and

Fig. 8 is a time chart illustrating the operation

of the embodiment in relation to Fig. 7.

Now, an embodiment of the invention will be described in detail with reference to the drawings.

Fig. 1 shows the overall construction of the embodiment.

A CPU 11 produces key input signals by detecting the operation of performance keys on a keyboard 12, and it also receives switch input signals provided through operation of switches in a control switch section 13.

The control switch section 13 includes a record switch and a reproduce switch, the outputs of these switches being fed as control signals to the CPU 11. An address bus AB, a data bus DB and a control bus CB are connected at one end to the CPU 11 and at the other end to a tone source control circuit 14 which effects recording and reproduction of tones.

An external sound signal is fed through an input terminal 16 to an analog-to-digital (A/D) converter 15 to obtain a digital waveform signal which is fed to the tone source control circuit 14. The input terminal 16 may be connected to, for instance, a microphone so that a note signal of, for instance, the note Cl, obtained when a corresponding performance key of a piano is depressed, can be fed as an external sound signal to the A/D converter 15. The tone source control circuit 14 feeds a sampling clock signal \$\phi_{REC}\$ to the A/D converter 15. The external sound signal is sampled by the A/D converter 15, and the amplitude level of the sampled

external sound signal is PCM converted to a digital waveform signal.

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The output signal of the A/D converter 15 is stored in a waveform memory provided in the tone source control circuit 14 under the control of an address control section operable under the control of commands from the CPU The waveform data stored in the waveform memory is read out according to the output from an address control section and converted, in a D/A converter 17, into an analog signal according to a waveform read clock ϕ_S from the tone source control circuit 14, the analog signal being fed to four VCAs 18-0 to 18-3 having a four-tone polyphonic structure. Four-channel time division timing signals To to T3 are fed as enable signals to gates (not shown) provided in an input stage of the VCAs 18-0 to 18-3. The output of the D/A converter 17 is fed to the VCAs 18-0 to 18-3 through their respective gates during corresponding channel periods. Further, the CPU 11 feeds channel switching signals CHO to CH3 to the VCAs 18-0 to 18-3, a tone signal being provided from only a 20 designated channel. The tone signal thus provided is fed from one of the corresponding channel output terminals 19-0 to 19-3 to a sound system having an amplifier, a loudspeaker, etc. (not shown), from which the corresponding tone is produced. 25

Now, the detailed construction of the tone source control circuit 14 will be described with reference to

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Fig. 2. Referring to the Figure, the tone source control circuit 14 includes an interface section 141, a main control section 142, an address control section 143 and a waveform memory section 144 including a RAM and an interpolating section 146. The interface section 141 is coupled with the CPU 11 via the address bus AB, data bus DB and control bus CB, and serves as an interface for data between the CPU 11 and tone source control circuit 14. From the interface section 141 various control signals are fed, through an internal control bus ICB, to the main control section 142. Also, initialization data, such as the first address of a memory in the section 144, and pitch data of the tone to be generated are fed from the interface section 141 to the address control section 143 through an internal data bus IDB. Further, waveform data is transferred between the interface section 141 and waveform memory section 144 via a RAM data bus RD. The RAM data bus RD also serves as a path for both waveform data fed from the A/D converter 15 through a gate 145 and waveform data being fed to the interpolation section 146. The main control section 142 controls the entire tone source control circuit 14.

Fig. 3 shows the main control section 142 in detail. As shown, a control signal fed through the internal control bus ICB is decoded in a command control circuit 142a having a decoder construction, and its various commands are then fed to the address control

section 143 and an internal trigger control circuit 142b. The trigger control circuit 142b realizes a delay trigger function of effecting a preliminary recording and a regular recording at the start of recording so as to eliminate the loss of the starting portion of the 5 recorded sound. START and STOP commands are fed from the trigger control circuit 142b to the address control section 143, and a gate enable signal GAD is fed to the gate 145. The section 142 further includes a timing control circuit 142c which provides various timing 10 signals, including clock signals ϕ_{REC} and ϕ_{S} as well as timing signals T0 to T3. Figs. 8(1) to 8(8) show various timing signals ϕ_W , ϕ_S , ϕ_R , T0 to T3 and ϕ_{REC} . The section 142 further includes a DMA control circuit 142d which provides a DMA request signal DMARQ via an 15 internal control bus ICB and effects DMA control according to a DMA acknowledge signal. A signal DMAD for determining the direction of DMA and a DMA start command DMAS are provided from a command control circuit 142a via a line 142e. The command control circuit 142a 20 also provides a signal BS, which is a switching signal for switching data transfer directions through the data bus DB when data is read into the CPU 11 through the internal control bus ICB.

25 The address control section 143 designates addresses of a waveform memory 144a (to be described later) in the waveform memory section 144. It renews

memory addresses under control of the main control section 142 and, when the renewing is done, it feeds an END signal to the trigger control circuit 142b of the main control section 142. Address data provided from the address control section 143 includes an integer portion and a decimal fraction portion, data representing the integer portion being fed to the waveform memory section 144, and data representing the decimal fraction portion being fed to the interpolation section 146.

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The waveform memory section 144 records waveform data fed from the A/D converter 15 and feeds the stored waveform data to the CPU 11 or to the interpolation section 146 through the RAM data bus RD.

The interpolation section 146 effects linear interpolation of waveform data read out from the waveform memory section 144 and provides the interpolated data to the D/A converter 17.

Fig. 4 shows the details of the address control section 143. Referring to the Figure, reference numeral 51 is a pitch register for storing pitch data for determining the pitch of the tone to be generated, numeral 52 a temporary storage register for storing address data of the waveform memory 144a, numeral 53 an end register for storing the end value of a renewal of the content of the temporary storage register 52, numeral 54 a loop start register for storing start address data of loop designation addresses, numeral 55 a loop end register

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for storing end address data of the loop designation addresses, numeral 56 a play flip-flop for controlling the start/stop of the renewal of the content of the temporary storage register 52, numeral 57 a loop on/off flip-flop for on/off controlling loop address designation, and numeral 58 an inverter flip-flop for inverting the polarity of each bit of pitch data read out from the pitch register 51. The elements 51 to 58, noted above, all consist of four-stage shift registers for shifting data in synchronism with the timing signal ϕ_S . In other words, these elements constitute a four-channel fourtone polyphonic structure which is driven on a time division basis at timings TO to T3. Pitch data is fed to the pitch register 51 via the internal data bus IDB and set therein, as a gate 59 is enabled by a command PITCH WRITE provided from the main control section 142, another gate, gate 61, being disabled through an inverter 60. When the pitch data is set, the output of the inverter 60 is inverted, and the set pitch data is circulated through the gate 61 and fed to an AND gate 63 through an exclusive OR gate 62. Address data is fed to the temporary storage register 52 via the internal data bus IDB, and a command TEMP WRITE, provided from the main control section 142, is fed to a gate 65 through an AND gate 64, as well as to a gate 67 through a NOR gate 66, whereby the address data is set in the temporary storage register 52. The set address data is fed to an

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adder 68 to be added to pitch data fed through the AND gate 63. The data is also fed to a comparator 69 and, subsequently, returned to the temporary storage register 52 through the gate 67. Seventeen bits constituting the integer portion data of the set address data are provided as address designation data to the waveform memory 144a. Meanwhile, thirteen bits constituting the decimal fraction portion of the set address data are provided as interpolation data to the interpolation section 146. Further, when a command TEMP READ is provided from the main control section 142, a gate 70 is enabled so that the content of the temporary storage register 52 is provided to the internal data bus IDB. End address data is fed to the end register 53 through the internal data bus IDB and set therein, as a command END WRITE, provided from the main control section 142, enables a gate 71 and disables a gate 73 through an inverter 72. The set end address data is fed to the comparator 69 through a gate 75, to which the timing signal ϕ_S is fed through an inverter 74. The comparator 69 compares end address data from the end register 53 with address data provided from the temporary storage register 52 through the adder 68, and, if the address data from the adder 68 is greater, it provides a signal LOOP. The signal LOOP is fed to an AND gate 76 as well as into a buffer 77 in synchronism with the timing signal $\phi_{W^{\bullet}}$. The output signal of the buffer 77 is fed as an end signal END to

both the main control section 142 and a NOR gate 78. Loop start address data is fed to the loop start register 54 via the internal data bus IDB and set therein, as a command LS WRITE, provided from the main control section 142, enables a gate 79 and disables a 5 gate 81 through an inverter 80. The set loop start address data is circulated through the gate 81 and set in the temporary storage register 52 through a gate 82, as the signal LOOP enables a gate 82 through the AND gate 76, disables the AND gate 64 10 through an inverter 83 and disables the gate 67 through the NOR gate 66. The loop end address data is fed to the loop end register 55 via the internal data bus IDB and set therein, as a command LE WRITE, provided from the main control section 142, enables the gate 84 and 15 disables the gate 86 through the inverter 85. The set loop end address data is circulated through a gate 86 and fed to a comparator 69 through a gate 87, which is enabled by the timing signal ϕ_S . The data fed to the comparator 69 for comparison with the data 20 of the temporary storage register 52 is the content of the loop end register 55 when the timing signal ϕ_S prevails, and the content of the end register 53 in the absence of the timing signal ϕ_S . The play flip-flop 56 is set when a command START is fed from the main control 25 section 142 to a NOR gate 88, and reset when a command STOP from either the section 142 or the end signal

from the buffer 77 is fed to the NOR gate 78. The output of the play flip-flop 56 is returned to the NOR gate 88 and fed to enable the AND gate 63. Further, when a command STATUS READ is provided from the main control section 142, a gate 89 is enabled to pass the output of the play flip-flop 56 to the internal data bus The loop-on flip-flop 57 is set when a command LOOP ON is fed from the main control section 142 to a NOR gate 90. The output of the loop-on flip-flop 57 is returned to the NOR gate 90 and fed to the AND gate The inverse flip-flop 58 is set when a command INV ON is fed from the main control section 142 to a NOR gate 92, and reset when a command INV OFF is fed to a NOR gate 93. The output of the inverse flip-flop 58 is returned to the NOR gate 92 and fed to the exclusive NOR gate 62 to cause inversion of the pitch data from the pitch register 51.

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Fig. 5 shows, in detail, the trigger control circuit 142b in the main control section 142. Referring to the Figure, reference numeral 101 designates a record flip-flop which is set when a command REC START from the command control circuit 142a is fed to a NOR gate 102 in synchronism with the timing signal ϕ_R , and reset when a command REC STOP is fed to a NOR gate 103. The \overline{Q} output of the record flip-flop 101 is returned to the NOR gate 102 and fed through a NOR gate 104 as a signal REC ON. The Q output, conversely, is

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fed through a NOR gate 105 as a signal REC OFF. output of the NOR gate 103 is fed to both the NOR gate 105 and the NOR gate 104 through an inverter 107. The signal REC ON is fed to a NAND gate 108 to which the timing signal TO is fed as one input. signal REC OFF is fed to a NAND gate 110 to which the timing signal TO or Tl is fed through an OR gate 109. Reference numeral 111 designates a trigger flip-flop which is set when a command REC TRIG is fed from the command control circuit 142a to a NOR gate 112 in synchronism with the timing signal ϕ_R , and reset when the command REC START, noted above, is fed to a NOR gate 113. The Q output of the trigger flip-flop 111 is returned to the NOR gate 112 and fed to a NOR gate 115 through an AND gate 114 to which the timing signal The is fed as one input. The \overline{Q} output, in contrast, is fed to a NOR gate 115 through an AND gate 116 to which the timing signal TO is fed as one input, and is also fed through an AND gate 117 to which the output of the NOR gate 113 is fed as one input, thereby appearing as a signal TRIG ON. The output of the NOR gate 115 is fed to the NOR gate 106, and also fed, as a read signal, to a buffer 118. The buffer 118 receives the END signal from the buffer 77 in the address control section 143, and feeds its output to a buffer 119 under the control of the timing signal ϕ_R . The output of the buffer 119 is, in turn, fed to the NOR gate 103. The signal TRIG

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ON, provided from the AND gate 117, is fed to a NAND gate 121 through a NAND gate 120 to which the timing signal Tl is fed as one input, and is also fed to a NAND gate 123 through a NAND gate 122 to which the timing signal TO is fed as one input. The output of the NAND gate 108 is fed to another input terminal of the NAND gate 121, while the output of the NAND gate 110 is fed to another input terminal of the NAND gate 123. output of the NAND gate 121 is fed as a START signal to the address control section 143 through an OR gate 124 to which a command PLAY is fed as one input from the command control section 142a. The output of the NAND gate 123 is fed as a STOP signal to the address control section 143 through an OR gate 125 to which a command STOP is fed as one input from the command control circuit 142a. The timing signal $\phi_{\mbox{\scriptsize REC}}$ is fed to a NOR gate 106 to which the output of the NOR gate 115 and the $\overline{\mathbb{Q}}$ output of the record flip-flop 101 are also fed, and the output of the NOR gate 106 is fed as a GAD signal to the gate 145 to enable this gate 145 to lead the waveform data from the A/D converter 15 to the RAM data bus RD.

Fig. 6 shows the waveform memory section 144 and interpolation section 146 in detail. Of the address data provided from the address control section 143, seventeen bits in the integer number portion are fed to a waveform memory 144a through a gate 144c to which the timing signal ϕ_W is fed through an inverter 144b, these

bits being incremented by +1 by a +1 incrementing circuit 144d and fed to the waveform memory 144a through a gate 144e enabled by the timing signal ϕ_W . To the waveform memory 144a is fed a read/write signal R/W from the main control section 142.

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The waveform data provided from the designated address of the waveform memory 144a is fed into a register 146a via the RAM data bus RD and in synchronism with the timing signal ϕ_S . Similarly, it is fed into a latch 146b in synchronism with the timing signal ϕ_W , and then led into a register 146c in synchronism with the timing signal ϕ_S . The data led into the register 146a is fed to a subtracter 146d for subtraction from it of the data having been fed into the register 146c, the difference data being fed to a multiplier 146e. To the multiplier 146e is also fed the decimal fraction portion of the address data from the address control section The multiplier 146e multiplies the decimal fraction data and data from the subtracter 146d and feeds the product data to an adder 146f. To the adder 146f is also fed the output of the register 146c. The adder 146f adds the two inputs and feeds the sum data to the D/A converter 17 shown in Fig. 1.

The operation of the embodiment of the above

construction will now be described with reference to

Figs. 7 and 8. Fig. 7 is a flow chart illustrating a

record routine. When recording external sound, a record

key in the key switch group 13 is first turned on (step Then, given initial data are set, this initial data having been set in advance in the registers 51 to More specifically, the pitch data, start address data, loop start address data, loop end address data, 5 end address data, loop-on data, etc., are fed from the key switch group 13 (step S2). At this time, the CPU 11 feeds the 16-bit data as separate lower (L) and upper (U) 8-bit data, as shown in (9) in Fig. 8. The operation timings of the CPU ll are asynchronous to the 10 timings in the tone source control circuit 14, as shown in (1) through (8) in Fig. 8. When the pitch data is of channel CH1, for instance, channel CH1 designation data and pitch designation data are provided subsequent to the pitch data, as shown in (9) in Fig. 8. The writing 15 of these data in the tone source control circuit 14 is done under the control of write signals WRO to WR3, generated from the command control circuit 142a, as shown in (10) to (13) in Fig. 8. The upper and lower bit data is fed, through the interface section M1, to 20 the internal data bus IDB in response to the signals WR1, and WR0, respectively, and a BUSY signal, as shown in (8) in Fig. 8, is provided, in response to the signal WR3, from the command control circuit 142a to the CPU 11 to inhibit execution of the 25 next instruction. A COMMAND SYNC signal, as shown in (15) in Fig. 8, which is a timing signal for

synchronizing the CPU 11 and tone source control circuit 14, rises in the main control section 142 in response to the timing signal op provided while the BUSY signal prevails. Command output is provided from the main control section 142. The command control circuit 142a 5 in the main control section 142 provides a command PITCH WRITE in response to the timing signal Tl (see (16) in Fig. 8). Meanwhile, the COMMAND SYNC signal falls in response to the next timing signal ϕ_R , and, with this fall, causes the BUSY signal to fall. When the command 10 PITCH WRITE is provided from the command control circuit 142a, the gate 59 in the address control section 143 is enabled so that the pitch data provided to the internal data bus IDB is set, in response to the timing signal ϕ_S , in the pitch register 51 for channel CH1. 15 The same pitch data setting operation takes place for the other channel registers as well.

It is now assumed that the following initial data are set:

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PITCH(0) = 0.25

PITCH(1) = 0.25

TEMP(0) = 00000

LOOP START(0) = 00000

LOOP END(0) = 01000

LOOP ON(0) = set

TEMP(1) = 01000

END(1) = 08000

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where (0) and (1) represent respective channel numbers, and TEMP represents the temporary storage register 52.

When the setting of the initial data is completed, the CPU 11 generates a record start command (step S3). This record start command is written under the control of the write signal WR3, and the command control circuit 142a generates a command REC START at the time of appearance of the COMMAND SYNC signal during the presence of the BUSY signal. The command REC START is fed to the record flip-flop 101 through the NOR gates 102 and 103 in the trigger control circuit 142b. The record flip-flop 101 is set in response to the next timing signal ϕ_R . The Q output of the record flip-flop 101 is thus inverted from "0" to "1" so that a REC ON signal, as shown in (17) in Fig. 8, is generated. The record flip-flop 101 provides the Q output as shown in (18) in Fig. 8. The REC ON signal is fed to the NAND gate 108, the output of which is "0" during the presence of the timing signal TO. The NAND gate output is also fed as a command START (i.e., START signal as shown in (19) in Fig. 8) to the address control section 143 through the NAND gate 121 and OR gate 124. Meanwhile, the command REC START is fed to the trigger flip-flop 111 through the NOR gate 113 to reset the trigger flip-flop 111 in response to the timing signal ϕ_R . The trigger flip-flop lll thus provides its Q output as shown in (20) in Fig. 8.

The timing signal T0 is fed through the AND gate 116, NOR gate 115 and NOR gate 106 to be fed as gate open signal GAD to the gate 145. Thus, waveform data sampled in the A/D converter 15 is fed through the gate 145 to the RAM data bus RD in response to every T0 timing signal.

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The starting command (i.e., START signal) provided from the trigger control circuit 142b is fed to the play flip-flop 56 through the NOR gates 88 and 78 in the address control section 143, and set in the play flip-flop 56 in response to the timing signal ϕ_S . In this state, preliminary recording is started. More specifically, when the play flip-flop 56 is set, its output enables the AND gate 63 so that the pitch data is passed from the pitch register 51 to the adder 68. In the temporary storage register 52 the data "00000" (representing address 0) has been set, while in the pitch register 51 pitch data "0.25" has been set. Thus, the adder 68 progressively adds 0.25 to the content of the temporary storage register 52. The output data of the adder 68 is fed to the waveform memory section 144 to designate successive addresses of the waveform memory 144a from the address 0, whereby the waveform data sampled by the A/D converter 15 is successively stored in the designated addresses of the waveform memory 144a from the address 0. Meanwhile, data "01000" (representing address 1,000) has been set in the loop

end register 55 so that, when the address data output of the adder 68 coincides with 1,000, the comparator 69 provides the LOOP signal. Further, since the loop-on flip-flop 57 has been set, the AND gate 76 is enabled to enable the gate 82 so that the address data "00000" set in the loop start register 81 is transferred to the temporary storage register 52. Subsequently, the address addition operation proceeds again according to the pitch data. In this way, a recording operation is performed by repeatedly designating addresses from the loop start address set in the loop start register 54 until the loop end address set in the loop end register 55 is reached. This recording state is the preliminary recording state.

Subsequently, when a trigger key in the key switch group 13 is operated, or when the recording level exceeds a predetermined level, the CPU 11 provides a command REC TRIGGER (steps S4 and S5 and also see (9) in Fig. 8). This command is written under the control of the write signal WR3, as provided by the command control circuit 142a. This command is fed to the trigger flipflop 111, through the NOR gates 112 and 113 in the trigger control circuit 142b, to set the trigger flipflop 111 in response to the timing signal \$\phi_R\$. Further, a TRIG ON signal as shown in (22) in Fig. 8 is provided from the AND gate 117 and fed to the NAND gate 122. Thus, the timing signal T0 is fed, as a STOP signal,

through the NAND gates 122 and 123, and OR gate 125 to the NOR gate 78 of the address control section 143 to reset the play flip-flop 56 for channel CHO, as shown in (23) in Fig. 8. As a result, the AND gate 63 is disabled to interrupt the address renewal. The TRIG ON signal is also fed to the NAND gate 120, while the timing signal Tl is fed through the NAND gates 120 and 121, and OR gate 124, and provided as the START signal, as shown in (19) in Fig. 8. This START signal is fed to the NOR gate 88 of the address control section 143 to set the play flip-flop for channel CH1.

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Since the data "01000" "08000" and "0.25" have been set in the temporary storage register 52, end register 53 and pitch register 51 for the channel CH1, respectively, the address renewing operation is started from address 1,000. That is, waveform data is written from address 1,000 of the waveform memory 144a (step S6). This operation is the regular recording operation. The CPU 11 reads out the set status of the play flip-flop 56 by periodically providing a command STATUS READ (step S7). If it detects that the play flip-flop 56 has been set, it judges that recording is in force, so that it refrains from proceeding to the following process (step S8). When the address data from the adder 68 coincides with the address data "08000" from the end register 53, the comparator 69 provides the LOOP signal. At this time, the AND gate 76 is not enabled since the loop-on

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flip-flop 57 for channel CH1 has not been set. Meanwhile, the LOOP signal is written in the buffer 77 and provided as an END signal which is fed to both the NOR gate 78 to reset the play flip-flop 56, and the buffer 118 of the trigger control circuit 142b of the main control section 142. The END signal fed to the buffer 118 is written therein in response to the rising of the timing signal Tl provided through the AND gate 114 to which the Q output of "1" is fed from the trigger flip-flop 111, i.e., in response to the falling of the signal Tl said END signal being then written in the buffer 119 in response to the next timing signal ϕ_R fed to the NOR gate 103. As a result, the record flip-flop 101 is reset to provide a \overline{Q} output of "1" which is fed through the NOR gate 105 as the REC ON signal (see (25) through (27) in Fig. 8). Thus, the timing signals TO and Tl are fed from the NAND gate 110 and OR gate 109, and the outputs $\overline{\textbf{T}}\textbf{0}$ and $\overline{\textbf{T}}\textbf{1}$ are thus fed through the NAND gate 123 and OR gate 125 so that the STOP signal will be provided in response to the timing signals TO and Tl (see (23) in Fig. 8).

The STOP signal is fed to NOR gate 78 in the address control section 143, thus resetting the play flip-flops for both the channels CHO and CHl. The AND gate 63 is thus disabled to interrupt the address renewing. The CPU 11 reads out the content of the play flip-flop 56 under the control of the command STATUS

READ and, if it detects that the flip-flop 56 is reset, it proceeds to the next process.

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In the process up to step S8, repeated recording from address 0 to address 1,000 of the waveform memory is done for the channel CHO (preliminary recording), while recording is done for channel CH1 (regular recording) from address 1,000 to address 8,000 of the waveform memory 144a. In the next process, the preliminary record portion and regular record portion are joined together. More specifically, in step S9 the waveform data recorded in the waveform memory 144a from address 0 to address 1,000 for channel CHO is DMA transferred to a memory (not shown) in the CPU 11. More specifically, the DMA start signal and signal representing the direction of DMA are provided from the command control circuit 142a to the DMA control circuit 142d (here from the waveform memory 144a to the CPU 11), and the DMA control circuit 142d provides a DMA request signal RQ to the CPU 11. When the DMA processing is ready to perform completion of the preceding process, the CPU 11 provides the DMA acknowledgement signal AK to cause start of the DMA transfer. In the memory in the CPU 11, the waveform data from address 0 to address 1,000 in the waveform memory 144a are stored and rearranged into a correct sequence of data (step S10). If the preliminary recording has terminated at address 600 with the appearance of the loop end signal,

recorded data one loop before address 601 to address 1,000 remains. In this case, the data is rearranged into a sequence of data from addresses 601 to 1,000 and data in addresses 0 to 600 in the mentioned order.

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At this time, the data in the temporary storage register 52 in the address control section 143 is "00600". By setting data "00000" and "01000" in the loop start register 54 and loop end register 55, respectively, and by setting the loop-on register 57, the content of the temporary storage register 52 is updated 601 + 1,000, 0 + 600 so that data is read out from the waveform memory 144a in the correct sequence. It is also possible to readout data in addresses 0 to 1,000 of the waveform memory 144a without condition, as well as to rearrange the readout data in the memory in the CPU 11. To this end, the CPU 11 provides a

in the CPU 11. To this end, the CPU 11 provides a command TEMP READ to enable the gate 70 to readout the content in the temporary storage register 52. If address 600 is detected, the process noted above is possible. In a subsequent step S11, the content in the memory in the CPU 11 is DMA transferred to addresses 0 to 1,000 of the waveform memory 144a.

The recording process is completed in the above way. The reproducing process will now be described.

Reproduction may be done in one of two ways, i.e., one in which the reproduction is done at notes corresponding to keys operated on the keyboard 12, or

one in which the recorded sound is reproduced as such by operating a monitor switch in the key switch group 13. Here, the former method will be described. First, a reproduce mode is set by depressing a reproduce key in the key switch group 13 such that one of the channels CHO to CH3 is designated. In the above recording example, the tone waveform data is recorded in addresses 0 to 8,000 for channel CHl. Therefore, channel CHl is designated, and initial data "00000" and "08000" are set in the temporary storage register 52 and end register 10 53, respectively, this data setting operation being the same as in the case of recording. Then, pitch data is set in the pitch register 51 by depressing a key on the keyboard 12. When the CPU 11 provides a reproduce command, the command control circuit 142a of the main 15 control section 142 provides a command PLAY. command PLAY is fed to the address control section 143 through the OR gate 124 of the trigger control circuit 142b to set the play flip-flop 56 through the NOR gates 88 and 78 in synchronism with the timing signal ϕ_S . 20 With the setting of the play flip-flop 56, the AND gate 63 is enabled so that address data in the temporary storage register 52 is renewed according to the pitch data set in the pitch register 51, as in the case of recording. If the inverse flip-flop 58 has been set by 25 the command INV ON at this time, a "l" signal is fed from the inverse flip-flop 58 to the exclusive OR gate

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62. Thus, the data provided from the pitch register 51 through the exclusive OR gate 62 is inverted. Thus, a complement to the pitch data is added to the data of the temporary storage register 52 in the adder 68, i.e., subtraction is done. It is thus possible to produce—inverse reproduction from the waveform memory 144a. The resetting of the inverse flip-flop 58 is done under the control of the command INV OFF.

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Of the address data provided from the adder 68, the integer number portion of 17 bits is fed to the waveform memory 144, and the decimal fraction portion of 13 bits is fed to the interpolation section 146. address data fed to the waveform memory section 144 is incremented by +1 in the +1 incrementing circuit 144d in response to the timing signal ϕ_W the incremented data designating the corresponding address of the waveform memory 144a, while it also designates as such the corresponding address of the waveform memory 144a in response to the timing signal $\overline{\phi}_{W^{\bullet}}$ That is, an address in question and the next address are designated in a time division fashion. The waveform data read out from the waveform memory 144a is fed through the RAM data bus RD, and, in synchronism with the timing signal ϕ_W , the waveform data in the +1-incremented address is set in the latch 146b. In synchronism with the next timing signal ϕ_S , the data in the latch 146b is written in the register 146c, and the waveform data read out under the

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control of the address data which is not incremented by +1 is written in the register 146a. In the subtracter 146d, the content of the register 146c is substracted from the content of the register 146a. The difference data is multiplied in the multiplier 146e by the decimal fraction portion of the address data from the address control section 143, whereby the ratio of the decimal fraction portion to the waveform data is designated by the integer number portion of the address data. output of the subtracter 146d is added in the adder 146f to the data stored in the register 146c to effect linear interpolation. The output of the adder 146f is fed to the D/A converter 17 which converts the input digital waveform data into a corresponding analog value that is fed through the VCA 18-1, which has been rendered operative with the designation of channel CH1, to the tone output terminal 19-1. When the key is depressed prolongedly, the sound is stopped upon completion of a reading of all the stored waveform data, if the loop-on flip-flop 56 is not set. If the loop-on flip-flop 56 is set and suitable data are set in the loop start register 54 and loop end register 55, the sound is continued for the duration during which the key is depressed. By resetting the loop-on register 57 when the key is released, all the waveform data is read out before the stop.

As has been described in the foregoing, according

to the invention, an address control logic is provided which effects waveform memory address control under the sole control of commands such as start and end commands from the CPU. The address control logic thus relieves the burden on the CPU in the recording and reproduction of external sound. In addition, a obtained by constructing polyphonic arrangement is shift registers, each having a plurality of stages as a temporary storage register for storing addresses of the waveform memory: an end register for storing the end address, a loop start register for storing the loop start address, a loop end register for storing the loop end address, a pitch register for storing the pitch data, etc., these registers being driven on a time division basis. A polyphonic system is realized without increase of components or cost.

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Further, since the preliminary recording is done repeatedly before regular recording and the regular recording is initiated by the start of trigger recording, there is no possibility of an interruption at the start of recording, thereby eliminating a wasteful non-recorded portion.

Furthermore, it is possible to permit reproduction in one channel while recording is made in a different channel, just as it is also possible to obtain mixing, i.e., recording the reproduced sound.

The digital waveform signal used in the described

embodiment is formed by the PCM processing. This digital waveform signal, however, may also be formed by any known pulse modulation method.

CLAIMS:

1. A polyphonic electronic musical instrument comprising;

waveform memory means for writing therein a sound waveform signal which is a digital signal obtained by A/D conversion of an external sound signal;

address signal generator means for timedivisionally generating address signals, which correspond to a plurality of channels, for writing and reading the sound waveform signal in and from said waveform memory means;

control means for providing commands to the address signal generator means to write and read the sound waveform signal in and from said waveform memory means;

write control means for writing the sound waveform signal such that an address signal is supplied to said waveform memory means by at least one of said plurality of channels when said control means commands said waveform memory means to write the sound waveform signal therein;

performance means operable to initiate reading out of the sound waveform signal written in said waveform memory means for each of a plurality of sounds at a respective desired pitch; and

read control means for time-divisionally supplying read address signals, which correspond to said plurality of sounds, from said plurality of channels to said waveform memory means and outputting sound waveform signals corresponding to said plurality of sounds, in accordance with the operation of said performance means.

2. The electronic musical instrument according to claim 1, wherein said address signal generator means comprises:

temporary storage register means for storing address signals of said waveform memory means;

renewal means for renewing the content of said temporary storage register means;

pitch register means for storing pitch data for determining the renewal speed of said renewal means;

end register means for storing the end address of the renewed content; and

end signal output means for comparing the content of said temporary storage register means with the content of said end register means and supplying an end signal to said renewal means when the content of said temporary storage register means coincides with or exceeds the content of said end register means, and

said temporary storage register means, said pitch register means, and said end register means each have a plurality of channels operable when designated on a time division basis.

- 3. The electronic musical instrument according to claim 2, wherein said renewal means includes means for providing a complement of the pitch data output from said pitch register means to complement update the content of said temporary storage register means and step the address signal in reverse order.
- 4. The electronic musical instrument according to claim 1, wherein said address signal generator means comprises:

temporary storage register means for storing address signals of said waveform memory means;

renewal means for renewing the content of said temporary storage register means;

pitch register means for storing pitch data for determining the renewal speed of said renewal means;

loop start register means for storing a leading address for designating repetition of addresses of said waveform memory means;

loop end register means for storing the end address for designating repetition of addresses of said waveform memory means; and

transfer means for outputting a loop end signal when the content of said temporary storage register means coincides with or exceeds the content of said loop end register means and transferring the content of said loop start register means to said temporary storage register means, and

said temporary storage register means, said pitch register means, said loop start register means and said loop end register means each have a plurality of channels operable when designated on a time division basis.

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**** END OF REGISTER ENTRY ****