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(54) **RF SLIPRING RECEIVER FOR A
COMPUTERIZED TOMOGRAPHY SYSTEM**

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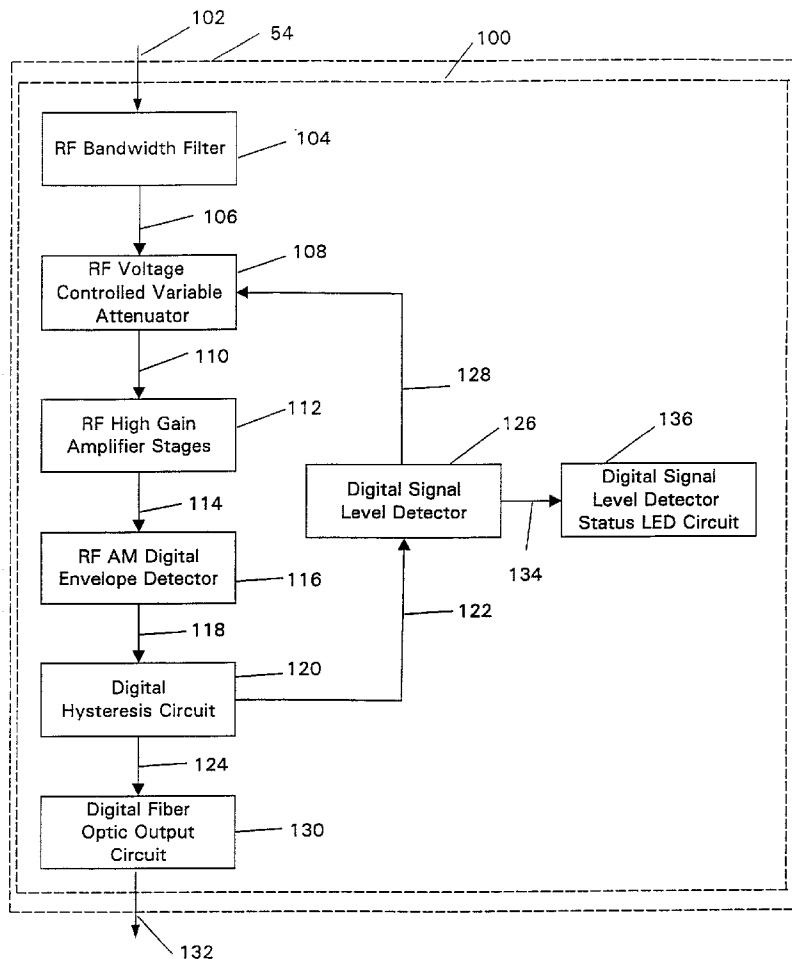
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(57) **ABSTRACT**

A computerized tomography (CT) system having an apparatus and method for receiving high data rate communication is disclosed herein. The apparatus includes a controlled variable attenuator for attenuating an encoded digital data signal having a wide range of power levels, a digital envelope detector for de-encoding the encoded digital data signal, a noise filtering circuit for filtering out the residual undesirable encoded signal components, and a feedback loop for maintaining a desired signal level such that the encoded digital data signal is converted to its pre-encoded state suitable for processing to generate a CT image.



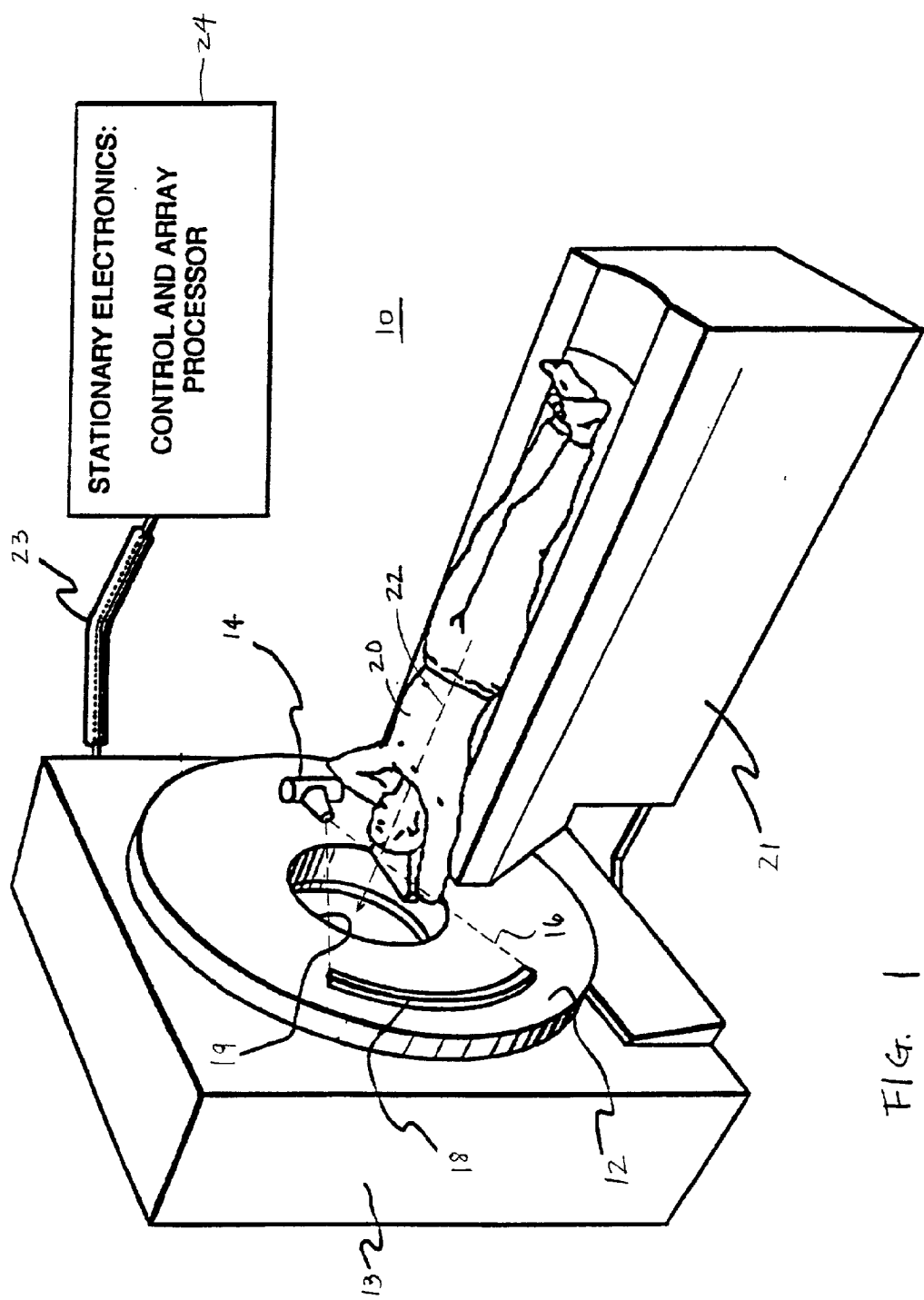


Fig. 1

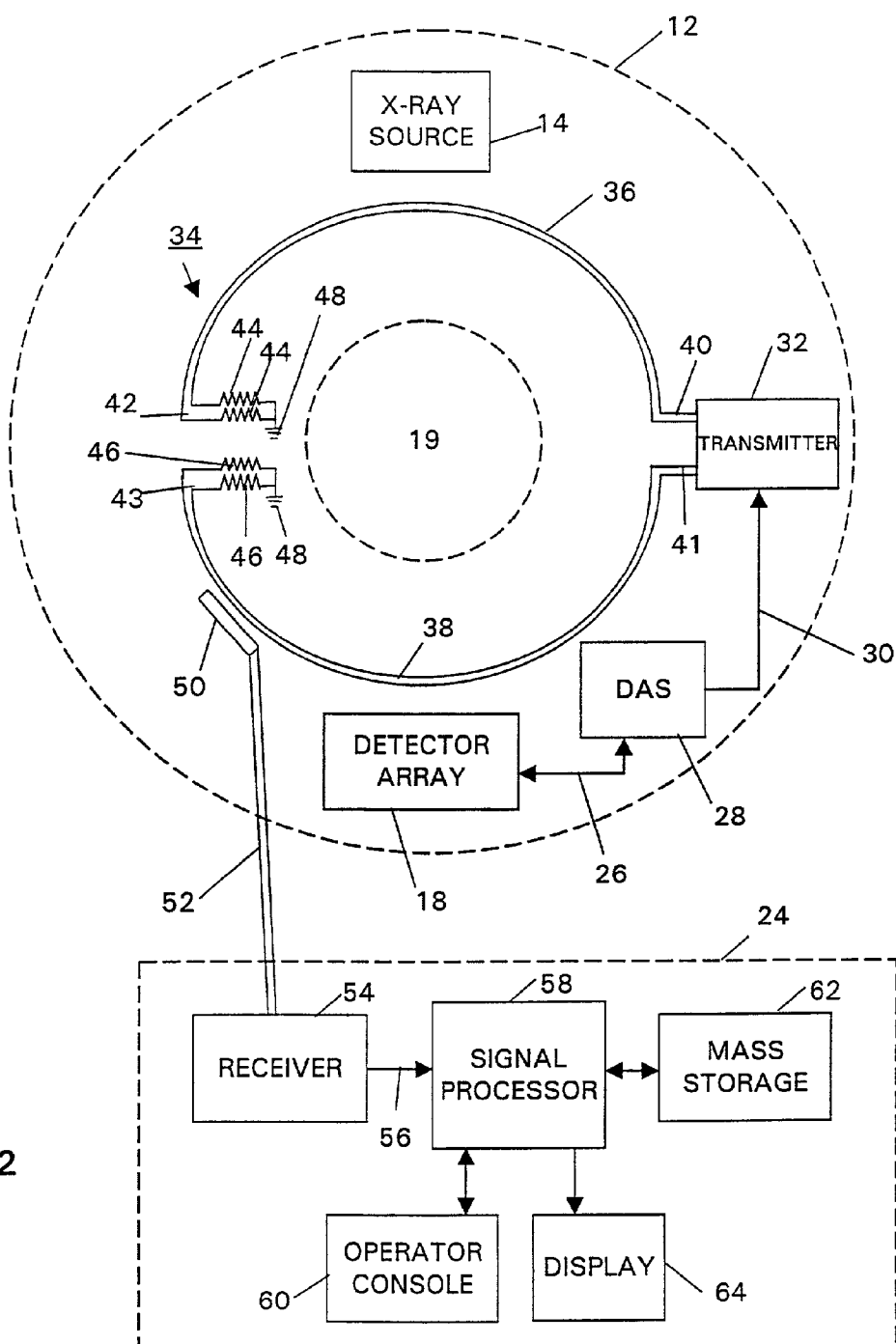


FIG. 2

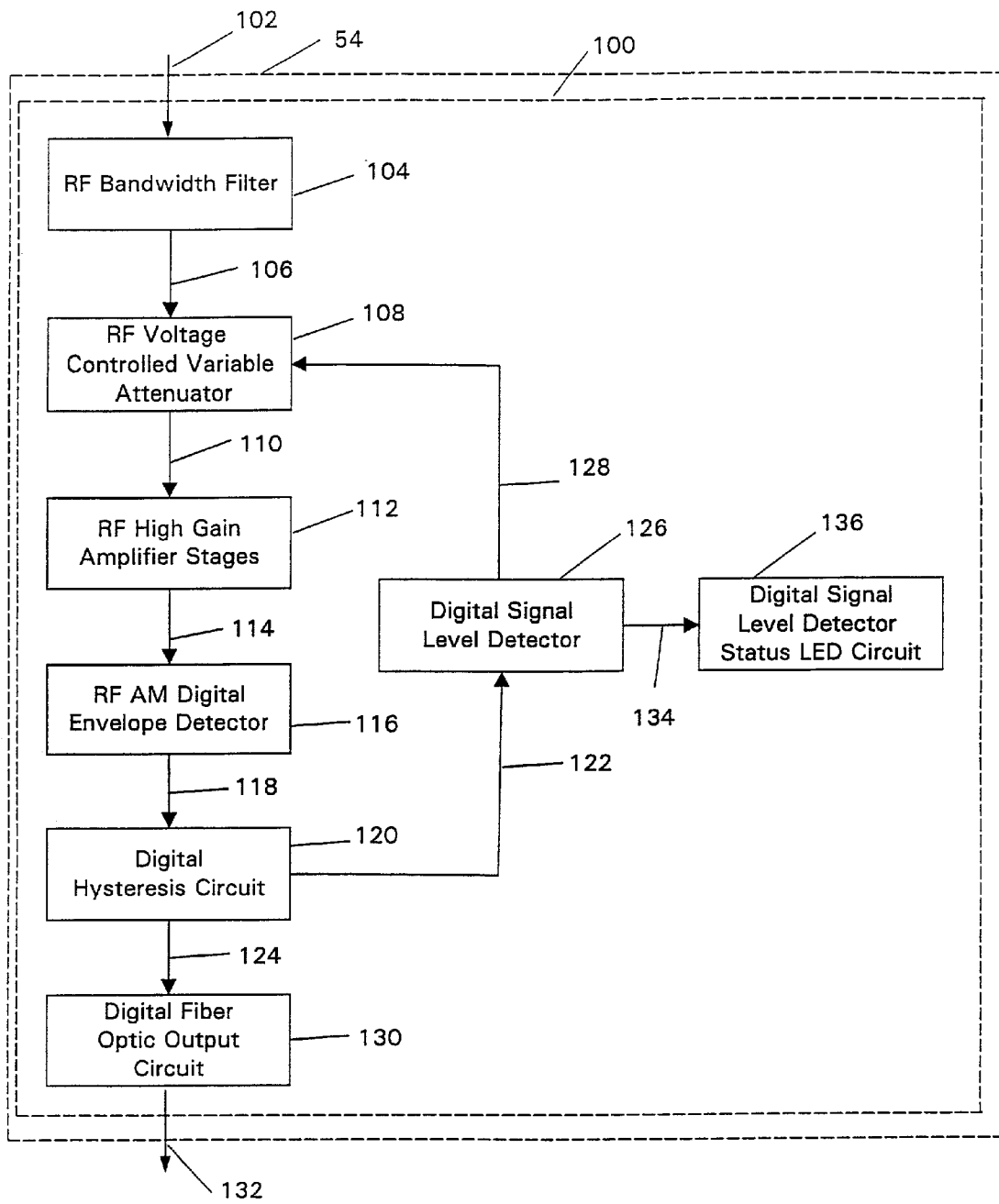


FIG. 3

RF SLIPRING RECEIVER FOR A COMPUTERIZED TOMOGRAPHY SYSTEM

BACKGROUND OF THE INVENTION

[0001] The present invention relates generally to computerized tomography (CT) systems. More particularly, the present invention relates to an apparatus and method for receiving high data rate communication in a CT system.

[0002] CT systems are used to obtain non-invasive sectional images of test objects, particularly internal images of human tissue for medical analysis and treatment. Current CT systems position the test object, such as a patient, on a table within a central aperture of a rotating frame, or gantry, which is supported by a stationary frame. The gantry includes an x-ray source and a detector array positioned on opposite sides of the aperture, within an x-y plane of a Cartesian coordinate system (generally referred to as the "imaging plane"), such that both rotate with the gantry around the test object being imaged. At each of several angular positions along the rotational path of the gantry (also referred to as "projections"), the x-ray source emits a fan-shaped collimated beam which passes through the imaging slice of the test object, is attenuated by the test object, and is received by the detector array. Each detector element in the detector array produces a separate electrical signal indicative of the attenuated x-ray beam intensity, the beam projected from the x-ray source to the particular detector element, incident at its sensor surface. The electrical signals from all the detector elements are collated by circuitry within the rotating frame to produce a projection data set at each gantry angle or projection. Each projection data set is referred to as a "view", and a "scan" is a set of such views from the different gantry angles during one revolution of the x-ray source and detector array. The scan is then processed by a computer in the stationary frame to reconstruct the projection data sets into a CT image of the slice or cross-section of the test object.

[0003] To transfer the projection data sets from the rotating frame to the stationary frame for image reconstruction, various communication links such as an umbilical cable, an optical data link, a slipring with a contacting brush, and a slipring with a contactless coupler are currently available. Newer CT systems typically utilize a slipring disposed on the rotating frame with a contactless coupler, having a certain air gap with respect to the slipring, disposed on the stationary frame as the communication link between the rotating and stationary frames. The slipring comprises a broken circle of wire or transmission line encircling the aperture of the rotating frame such that each half of the broken circle of wire forms an arc of exactly the same length. Data signals, e.g., the projection data sets, are encoded and transmitted from the first ends of the two wires to the second ends of the two wires at the opposite side of the broken circle such that both data signals arrive at the second ends, generally referred to as the termination gap, at the same time. The contactless coupler disposed on the stationary frame lies close to the slipring and captures the transmitted encoded data signals via electromagnetic coupling. Because each projection data set is transmitted as they are acquired (after encoding), i.e. while the rotating frame is still rotating to acquire the next projection data set for the next gantry angle, propagation of the data signals along the wires of the slipring and electromagnetic coupling from the

slipring to the contactless coupler occurs while the rotating frame and thus the slipring are in rotation.

[0004] Unfortunately, although the slipring and contactless coupler provides many advantages over other types of communication links such as higher data rate transfer, shorter image acquisition time, increased patient comfort, and less mechanical stress and wear, it also suffers from problems associated with controlling the data signal strength present at the contactless coupler. Data signals received by the contactless coupler can have a power variation of up to 15 to 20 dB around its center operating power level. It has been found that numerous sources contribute to the signal strength variability, including, but not limited to: (1) axial and/or radial rotation runouts of the rotating assembly; (2) axial and/or radial misalignment of the contactless coupler; (3) variability of the printed circuit board within the channel groove; (4) nominal air gap losses; (5) accuracy of the dimensions of the various components of the rotating assembly such as the channel groove; (6) skin effect and dielectric losses in ring circuit board traces; and (7) amplifier gain variability. Moreover, although attempts have been made to correct signal variation resulting from any one contributing source, it is difficult, time consuming, and costly to correct the variations caused by all of these contributing sources given the interaction between these sources in CT systems.

[0005] Thus, there is a need for an apparatus and method capable of accommodating wide variations in the data signal strength received by a contactless coupler in a CT system. Moreover, there is a need for such an apparatus and method to be efficient, cost effective, flexible to the range of signal variations, and to enhance, or at least not counteract, the advantageous capabilities of a communication link comprised of a slipring with a contactless coupler in a CT system.

BRIEF SUMMARY OF THE INVENTION

[0006] One embodiment of the invention relates to an apparatus for converting a digital modulated data signal having a wide range of power levels and a modulation component to its pre-modulated state. The apparatus includes means for variably attenuating the amplitude of the digital modulated data signal to generate an amplitude stabilized signal, and means for digitizing the amplitude of the amplitude stabilized signal to generate a digitized signal. The apparatus further includes means for filtering the digitized signal to generate a reconstructed digital data signal representative of the modulated digital data signal in its pre-modulated state, and means for controlling the amount of attenuation of the amplitude of the digital modulated data signal in the means for variably attenuating by receiving an amplitude signal from the means for filtering and completing a feedback loop to the means for variably attenuating by generating a feedback control voltage signal to be received therein.

[0007] Another embodiment of the invention relates to a computerized tomography (CT) system. The CT system includes an x-ray source and a detector for receiving a detected image signal representative of the x-ray intensity thereon, and a transmitter coupled to encode the detected image signal with a carrier signal to provide a modulated data signal. The CT system further includes a receiver coupled to the transmitter and having an automatic radio

frequency (RF) attenuation loop for receiving a coupled modulated data signal and converting the coupled modulated data signal having a wide range of power levels to its pre-modulated state.

[0008] Another embodiment of the invention relates to a method of reconstructing a digital image data in a computerized tomography (CT) system. The method includes receiving an encoded digital RF amplitude modulated data signal having a wide range of power levels, bandwidth filtering the encoded data signal causing undesirable frequency components to be filtered out, and attenuating the amplitude of the encoded data signal partly in response to a control voltage. The method further includes amplifying the amplitude of the encoded data signal causing amplification of the desired data component of the encoded data signal, digitizing the encoded data signal by filtering out the RF component therein, and filtering the encoded data signal to remove remaining undesirable signal components from the RF amplitude modulation encoding. The digital image data has thus been reconstructed from the encoded data signal for CT image processing.

[0009] Still another embodiment of the invention relates to an apparatus for demodulating a modulated digital data signal having a wide range of power levels and a modulation component to its pre-modulated state. The apparatus includes a variable attenuation circuit, a conversion circuit, a filter, and an attenuation controller.

BRIEF DESCRIPTION OF THE DRAWINGS

[0010] The invention will become more fully understood from the following detailed description, taken in conjunction with the accompanying drawings, wherein like reference numerals refer to like parts, in which:

[0011] FIG. 1 is a perspective view of a CT system which employs the present invention;

[0012] FIG. 2 is a schematic block diagram of portions of the CT system of FIG. 1; and

[0013] FIG. 3 is a block diagram of a portion of the receiver which forms part of the CT system of FIG. 1.

DETAILED DESCRIPTION OF THE INVENTION

[0014] Referring to FIG. 1, a computerized tomography (CT) system 10 includes a generally annular rotating frame 12 or gantry, and a stationary frame 13 that supports rotating frame 12. Rotating frame 12 includes an x-ray source 14 for emitting a highly collimated x-ray beam 16 toward a detector array 18 positioned on the opposite side of an aperture 19. Aperture 19 permits a test object 20, such as a patient, to be placed on a platform 21 which may be movable, for example, by translation, along rotating frame's 12 rotational axis 22. Movement of platform 21 allows different cross-sectional portions of interest of test object 20 to be positioned within the imaging plane of rotating frame 12.

[0015] Once test object 20 has been positioned within aperture 19, as desired, by movement of test object 20 and/or platform 21, rotating frame 12 rotates about the rotational axis 22, and at each of a plurality of angular positions along the rotational path, x-ray source 14 emits x-ray beam 16 which passes through test object 20 and is incident on the

receiving surfaces of a plurality of detector elements (not individually shown) of the detector array 18. In response, each of the detector elements of detector array 18 produces an electrical signal at a magnitude proportional to the intensity of the received rays, and thus to the amount of attenuation of the x-ray beam after passing through test object 20. As described in further detail hereinafter, the signals from each of the detector elements of detector array 16, which represent the projection data, are presented through lines 23 to a control and array processor 24 that processes the received projection data into a radial image of test object 20 at the selected radial or angular position, which is referred to as a view. Then the aggregate of the views taken over a full revolution of the rotating frame 12, generally referred to as a scan, are further processed, using known image processing algorithms, into a cross-sectional image of the portion of interest of test object 20 that was within the imaging plane.

[0016] Referring now to FIG. 2, there is shown a schematic block diagram of a portion of the CT system 10 of FIG. 1. It should be understood that only functional elements necessary for a teaching of the present invention are shown in FIG. 2 and that only the relative connections between functional elements are shown. The signals from the detector array 18 are provided through lines 26 to a data acquisition system (DAS) 28 disposed on rotating frame 12, which converts each signal from a detector element of detector array 18 from an analog signal format into a digital binary signal format, typically a 16 bit digital value representing the attenuated x-ray intensity. DAS 28 multiplexes the converted detector channel signals, together with a data clock signal and an error checking signal function, into a serial digital bit signal. The serial digital bit signal is then received by a transmitter 32, disposed on rotating frame 12, via lines 30. In the preferred embodiment, transmitter 32 digitally encodes the serial digital bit signal with amplitude modulation using a radio frequency (RF) carrier signal to produce a RF amplitude modulated digital serial data signal. For example, such an encoded data signal can be in a non-return to zero inverted (NRZI) format.

[0017] Next, the encoded data signal is presented from transmitter 32 to transmission line segments 36, 38 of a RF slipring 34. As described in further detail hereinafter, the encoded data signal is propagated along transmission line segments 36, 38 to be electromagnetically coupled into a coupler 50 disposed on the stationary frame 13. RF slipring 34 may be configured to include one or more transmission lines disposed on rotating frame 12. Depending on the distance, generally referred to as the air gap, between the coupler 50 and the transmission line segments 36, 38, more transmission line segments may be required to ensure that coupler 50 is always in sufficient spatial proximity to at least one of the transmission line segments to receive the encoded data signal. When more than one segment is required, each segment will have a length that is a fractional portion of the arc length of the rotating frame's 12 rotational path. The segments are cascaded, end-to-end around the rotating frame's 12 rotational axis 22 (see FIG. 1), typically along the circumference of aperture 19, such that the aggregate segment length provides a substantially 360 degree arc, e.g., fully encircling the rotating frame 12.

[0018] In the preferred embodiment, two transmission line segments 36, 38 having first ends 40, 41 and second ends 42,

43, respectively, are contiguously positioned on rotating frame 12 encircling aperture 19 such that substantial continuity of electromagnetic coupling along the full rotational path of rotating frame 12 is possible. The first ends 40, 41 are connected to the transmitter 32, and the second ends 42, 43 are connected through terminal impedances 44, 46, respectively, to an electrical ground 48. The terminal impedances 44, 46 have a predetermined resistance value chosen to minimize reflection of energy in each of the transmission line segments 36, 38.

[0019] The coupler 50 is positioned on the stationary frame 13 (not shown in FIG. 2) such that physical proximity between coupler 50 and at least one of the transmission line segments 36, 38 will be maintained during rotation of rotating frame 12. In the preferred embodiment, the air gap between coupler 50 and the transmission lines 36, 38 of slipping 34 is in the range of 0.050 to 0.080 inches, and coupler 50 is a short piece of wire or transmission line approximately 2 inches in length. Coupler 50 may alternately be a pickup antenna, a RF shoe, a contactless brush, or an electromagnetic coupling apparatus capable of receiving the encoded data signal from the RF slipping 34 across a transmitting distance in the range of 0.050 to 0.080 inches.

[0020] Transmitter 32 transmits the encoded data signal to the first ends 40, 41 of the transmission lines 36, 38, respectively, and the encoded data signal propagates from the first ends 40, 41 to the second ends 42, 43 to be terminated at the electrical ground 48. Before the encoded data signal propagates to ground 48, however, the encoded data signal will be electromagnetically coupled to coupler 50, thereby completing the transfer from the rotating frame 12 to the stationary frame 13 for signal processing.

[0021] On the stationary frame side, the coupled encoded data signal, also referred to as the coupled modulated data signal, is transmitted to a receiver 54 located in the control and array processor 24 via lines 52. As described in further detail hereinafter in FIG. 3, the receiver 54 decodes the coupled modulated data signal to its pre-encoded state and provides this decoded signal to a signal processor 58 via lines 56. The signal processor 58 includes a computer and signal memory (not shown) for storing the program algorithms which govern the CT processing of the received data in response to operator commands and scanning parameters received via an operator console 60, such as a keyboard, mouse, trackball, or switches. Although not shown, the operator commands and parameters are used by the signal processor 58 to provide control signals and information to the DAS 28, the x-ray controller (not shown), the gantry motor controller (not shown), as well as controlling movement of platform 21. In this manner, the signal processor 58 collates the decoded signal, i.e. the projection data, into a composite view corresponding to a particular angular position of the rotating frame 12. Each composite view is stored in a mass storage device 62 and is retrieved as necessary during processing of other composite views to be further processed to provide a final image of the desired cross-section of test object 20. This final image, also referred to as a reconstructed image, can then be displayed on a display 64, for example as a conventional cathode ray tube (CRT) display, a liquid crystal display, or other display devices, or it may be converted to a film or printed record by means of an appropriate computer controlled camera or printer (not shown). Moreover, reconstructed images may be stored in

the mass storage 62 and stored reconstructed images and/or other data may be retrieved as commanded by the operator via the operator console 60 and the signal processor 58.

[0022] Referring now to FIG. 3, there is shown a block diagram of a portion of the receiver which forms part of the CT system of FIG. 1. The coupled encoded data signal, also referred to as a coupled modulated data signal 102, is provided on lines 52 (see FIG. 2) to an automatic RF attenuation loop 100 within the receiver 54. In the preferred embodiment, the attenuation loop includes a RF bandwidth filter 104, a RF voltage controlled variable attenuator 108, RF high gain amplifier stages 112, a RF amplitude modulation (AM) digital envelope detector 116, a digital hysteresis circuit 120, a digital signal level detector 126, a digital fiber optic output circuit 130, and a digital signal level detector status light emitting diode (LED) circuit 136. Also in the preferred embodiment, the coupled modulated data signal 102 has a power level range of up to 50 dBm (decibels referenced at 1 milliWatts) around a center operating power level of -13.6 dBm. Alternately, the coupled modulated data signal 102 can have a power level range of up to 40 dBm around a center operating power level.

[0023] The coupled data signal 102 is received by the RF bandwidth filter 104 and produces a bandwidth filtered signal 106. The bandwidth filter 104 filters the coupled data signal 102 in a bandwidth range of 500 MHz (megahertz) to 1 GHz (gigahertz) to remove signal components from frequencies outside the bandwidth range such as noise.

[0024] The bandwidth filtered signal 106 is received by the controlled variable attenuator 108 to produce an amplitude stabilized signal 110. The amount of attenuation applied to the bandwidth filtered signal 106 is determined by a feedback control voltage signal 128 also received by the controlled variable attenuator 108. Depending on the feedback signal 128, the amplitude stabilized signal 110 will be the same amplitude or up to 40 dB (decibels) smaller in amplitude than the bandwidth filtered signal 106. Thus, the controlled variable attenuator 108 selectively attenuates the amplitude of the received signal but preserves the frequency components therein.

[0025] The amplitude stabilized signal 110 is received by the amplifier stages 112 to produce an amplitude gain signal 114. The amplifier stages 112 amplifies the amplitude of the amplitude stabilized signal 110 by a gain of up to approximately 45 dB. Similar to the controlled variable attenuator 108, the amplifier stages 112 preserves the frequency components of the received signal. In the preferred embodiment, amplifier stages 112 comprises three amplifiers connected in series, each amplifier providing a gain of up to approximately 15 dB, for a total gain of approximately 45 dB. Such amplifiers can be a chip device made by Minicircuits and it is well-known in the art that an amplifier chip device will have various other driving components such as capacitors, other circuits, coils, etc. to provide biasing and other functionally necessary inputs.

[0026] The amplitude gain signal 114 is received by the digital envelope detector 116 to produce a digitized signal 118. The digital envelope detector 116 removes the RF component of the amplitude gain signal 114, thereby digitizing signal 114. Prior to the digital envelope detector 116, the signal is still amplitude modulated with a RF carrier signal and although a digital signal, contains sine wave-

forms. But after the digital envelope detector **116**, signal **114** has essentially been converted to a digital square wave. The digital envelope detector **116** is available as a Minicircuits device and alternatively could be a modulator circuit implemented as an amplitude modulation detector.

[0027] The digitized signal **118** is received by the digital hysteresis circuit **120** to produce a noise-filtered digitized signal **124** and an amplitude signal **122**. The digitized signal **118** entering the hysteresis circuit **120** is a square wave containing some noise components because the digital envelope detector **116** is typically unable to remove all the RF components of the amplitude modulation. Thus the hysteresis circuit **120** serves to filter out the residual undesirable signal components, such as noise, remaining in the signal after the digital envelope detector **116**, by adding hysteresis to the digitized signal **118**, as is well-known in the art, to generate the noise filtered digitized signal **124**. Although not shown, hysteresis circuit **120** will have associated resistors and capacitors connected thereto to provide a certain turn on and turn off hysteresis. Alternatively, the hysteresis circuit **120** may be a digital buffer gate, a digital noise filter circuit, or a NRZI data screen. The hysteresis circuit **120** also serves a dual function in that the digitized signal **118** is used to generate the amplitude signal **122**, which is a voltage signal having its amplitude representative of the amplitude of the digitized signal **118**.

[0028] The amplitude signal **122** is received by the digital signal level detector **126** to produce the feedback control voltage signal **128** and a status signal **134**. The digital signal level detector **126** converts the amplitude signal **122** into the feedback signal **128** which is a low-frequency voltage signal (almost a direct current (DC) signal) representative of the digitized signal **118**. The response rate of the feedback signal **128** is around 100 milliseconds which is orders of magnitude slower than the response rates of the rest of the attenuation loop **100**. The feedback signal **128** is received by the controlled variable attenuator **108**, completing the feedback loop to maintain the amplitude of the digitized signal **118** at roughly the same level or range by controlling the amount of attenuation provided in the controlled variable attenuator **108**, the amount of attenuation representative of the level of the feedback signal **128**.

[0029] The status signal **134** received by the digital signal level detector status LED circuit **136** is an indicator signal of the voltage level within the attenuation loop **100**, particularly the level of the digitized signal **118**. The status circuit **136** includes indicia means for displaying or indicating the relative range of this operating voltage level. For example, status circuit **136** can include a set of bar graph LEDs (not shown) that displays the relative range of this operating voltage level. When the voltage is at the maximum limit, which means that the controlled variable attenuator **108** is letting the bandwidth filtered signal **106** through with no attenuation, all the bar graph LEDs will be illuminated. Conversely, when the voltage is very low, which means that the controlled variable attenuator **108** is providing maximum attenuation to the bandwidth filtered signal **106**, only one or two LEDs at one end will be illuminated. Optimally the voltage level will be a mid-gain level such that the bar graph LEDs comprised of 10 LEDs in a row will have the fifth or sixth such LED illuminated. In this manner, the number of LEDs illuminated in the status circuit **136** can increase or decrease to continually reflect the signal level operating with

the attenuation loop **100**. Alternatively, the status circuit **136** may include a numerical display, a LCD indicator, or other indicator means capable of providing the relative signal level information. In the preferred embodiment, the bar graph LEDs within status circuit **136** are housed within the receiver **54** to provide diagnostic-type information to field service personnel such that the CT system operator or end-user typically will not be aware of or view these bar graph LEDs.

[0030] Lastly, the noise filtered digitized signal **124** is received by the digital fiber optic output circuit **130** for converting signal **124**, which is now in its pre-encoded or pre-modulated state similar to the serial digital data signal outputted by DAS **28** (see FIG. 2), into a digital fiber optic signal **132** suitable for transmission in a fiber optic cable. In the preferred embodiment, the digital fiber optic output circuit **130** is a chip device which is well-known in the art. Then digital fiber optic signal **132** may be transmitted to the signal processor **58** (see FIG. 2) via lines **56** or signal **132** may be further processed before being transmitted to the signal processor **58** via lines **56**. Thus, the present invention provides a data signal receiver having an extensive dynamic input range and which also provides an elegant means for correcting or compensating for various signal variations based on the received signal strength. In this manner, the quality of RF communication is much improved without the need for additional mechanical fine tuning such as special test equipment to repeatedly calibrate and center the center operating power level in the slirping assembly.

[0031] While the embodiments and application of the invention illustrated in the FIGs. and described above are presently preferred, it should be understood that these embodiments are offered by way of example only. For example, the RF carrier signal described herein may be at a different frequency or the modulation scheme described herein may be a modulation type different from amplitude modulation. The slirping and coupler may communicate using a capacitive type coupling, which also suffers from signal variability to some degree, instead of electromagnetic coupling such that the receiver described herein may be implemented by including other combinations or types of signal processing devices. Still further, although all of the communication between the rotating frame and the stationary frame has been serialized in the preferred embodiment, i.e., converted from parallel to serial data for transmission and vice versa on reception employing well-known multiplexing techniques, this was done so that only a single data stream need be transmitted. It should be understood that multiple parallel paths could be employed using multilevel or multibase encoding techniques to further increase the maximum data rate transfer,

[0032] Even still further, the encoded data signal may have a even wider signal variation range than the 40 dBm or 50 dBm described herein such that the controlled variable attenuator would correspondingly have to provide an appropriate range of attenuation and the amplifier stages would have to provide greater gain than described herein. Furthermore, it is contemplated that the invention may be applied to systems other than medical systems which can benefit from the use of feedback-based signal attenuation. Accordingly, the present invention is not limited to a particular embodiment, but extends to various modifications that nevertheless fall within the scope of the appended claims.

What is claimed is:

1. An apparatus for demodulating a modulated digital data signal having a wide range of power levels and a modulation component, comprising:

means for variably attenuating the amplitude of the modulated digital data signal partly in response to a feedback control voltage signal, causing an amplitude stabilized signal to be generated therefrom;

means for digitizing the amplitude stabilized signal to generate a digitized signal therefrom by removing the modulation component in the amplitude stabilized signal;

means for filtering the digitized signal to generate a demodulated digital data signal representative of the modulated digital data signal in its pre-modulated state, wherein the digitized signal is filtered to remove undesirable residual signal components therein, and to generate an amplitude signal representative of the amplitude of the digitized signal; and

means for controlling the amount of attenuation of the amplitude of the modulated digital data signal in the means for variably attenuating by receiving the amplitude signal from the means for filtering and completing a feedback loop to the means for variably attenuating by generating the feedback control voltage signal to be received therein.

2. The apparatus of claim 1, further comprising:

means for bandwidth filtering the modulated digital data signal causing a bandwidth filtered signal to be generated therefrom to the received by the means for variably attenuating; and

means for amplifying the amplitude of the amplitude stabilized signal to produce an amplitude gain signal to be received by the means for digitizing.

3. The apparatus of claim 2, wherein the means for bandwidth filtering provides a bandwidth filtering range of 500 MHz to 1 GHz.

4. The apparatus of claim 2, wherein the means for amplifying provides a gain of up to approximately 45 dB.

5. The apparatus of claim 1, further comprising means for converting the reconstructed digital data signal to a digital filter optic signal for transmission along a fiber optic cable.

6. The apparatus of claim 1, further comprising means for displaying the digitized signal as a relative signal level.

7. The apparatus of claim 6, wherein the means for displaying is at least one of a bar graph light emitting diodes (LEDs), a numerical display, and a liquid crystal display (LCD) indicator.

8. The apparatus of claim 1, wherein the modulated digital data signal comprises a signal having a power level range of up to 40 dBm around a center operating power level.

9. The apparatus of claim 1, wherein the modulated digital data signal comprises a signal having a power level range of up to 50 dBm around a center operating power level.

10. The apparatus of claim 1, wherein the modulated digital data signal comprises a signal having an operating power centered at approximately -13.6 dBm.

11. The apparatus of claim 1, wherein the modulated digital data signal comprises a serial digital bit data signal amplitude modulated with a radio frequency (RF) carrier signal.

12. The apparatus of claim 1, wherein the modulated digital data signal comprises a signal in a non return to zero inverted (NRZI) format.

13. The apparatus of claim 1, wherein the means for variably attenuating comprises an attenuation range of -40 dB to 0 dB.

14. The apparatus of claim 1, wherein the means for digitizing comprises a radio frequency (RF) amplitude modulation envelope detector.

15. The apparatus of claim 1, wherein the means for filtering comprises a digital hysteresis circuit.

16. The apparatus of claim 1, wherein the means for controlling the amount of attenuation comprises a digital signal level detector.

17. A computerized tomography system comprising:

an x-ray source and a detector each disposed to pass x-rays through a test object toward the detector, wherein the detector provides a detected image signal of the intensity of the x-rays received thereby, at each of several positions along a predefined path;

a transmitter disposed to encode the detected image signal with a carrier signal to provide a modulated data signal; and

a receiver coupled to the transmitter and including an automatic radio frequency (RF) attenuation loop configured to receive a coupled modulated data signal and to convert the coupled modulated data signal to its pre-modulated state, wherein the coupled modulated data signal comprises the modulated data signal potentially having a wide range of power levels.

18. The computerized tomography system of claim 17, wherein the automatic RF attenuation loop comprises:

a controlled variable attenuator configured to receive two inputs, the coupled modulated data signal and a feedback control voltage signal, to produce an amplitude stabilized signal;

a digital envelope detector configured to receive the amplitude stabilized signal and to output a digitized signal, wherein the digitized signal comprises the amplitude stabilized signal digitized by filtering out the RF component therein;

a digital noise filter stage configured to receive the digitized signal from the digital envelope detector and to output a noise filtered digitized signal and an amplitude signal, wherein the noise filtered digitized signal comprises the digitized signal filtered to remove the residual noise therein and the amplitude signal is responsive in part to the amplitude of the digitized signal; and

a digital signal level detector configured to receive the amplitude signal from the digital noise filter circuit and to produce the feedback control voltage signal, wherein the feedback control voltage signal is a low-frequency signal having its amplitude proportional to the digitized signal whereby the feedback control voltage signal completing the feedback loop into the controlled variable attenuator controls the amount of amplitude attenuation provided by the controlled variable attenuator.

19. The computerized tomography system of claim 18, wherein the automatic RF attenuation loop further comprises:

a RF bandwidth filter configured to receive the coupled modulated data signal from the second element and to output a bandwidth filtered signal, wherein the bandwidth filter has a bandwidth range of 500 MHz to 1 GHz; and

an amplifier stage configured to receive the amplitude stabilized signal from the controlled variable attenuator and to produce an amplitude gain signal.

20. The computerized tomography system of claim 19, wherein the automatic RF attenuation loop further comprises a digital fiber optic stage configured to receive the noise filtered digitized signal from the digital noise filter circuit for causing the digital fiber optic stage to output a digital fiber optic signal for transmission in a fiber optic cable.

21. The computerized tomography system of claim 19, wherein the automatic RF attenuation loop further comprises a status indicator configured to receive a status signal from the digital signal level detector and to relatively display the digitized signal level within the attenuation loop.

22. The computerized tomography system of claim 21, wherein the status indicator comprises a circuit having means for displaying visible indicia representing the relative signal level, the visible indicia being mapped to locations within the status indicator using common scaling.

23. The computerized tomography system of claim 19, wherein the amplifier stage provides a gain of up to approximately 45 dB.

24. The computerized tomography system of claim 19, wherein the amplifier stage comprises a plurality of amplifiers, wherein each amplifier provides a gain of up to approximately 15 dB.

25. The computerized tomography system of claim 18, wherein the digital noise filter stage comprises a digital hysteresis circuit that adds hysteresis to the digitized signal.

26. The computerized tomography system of claim 18, wherein the controlled variable attenuator attenuates the amplitude of the coupled modulated data signal by a range of -40 dB to 0 dB partly in response to the feedback control voltage signal.

27. The computerized tomography system of claim 18, wherein the digital envelope detector is a radio frequency (RF) amplitude modulation digital envelope detector.

28. The computerized tomography system of claim 17, wherein the coupled modulated data signal comprises a signal having a power level range of approximately 50 dBm around a center operating power level.

29. The computerized tomography system of claim 17, wherein the coupled modulated data signal comprises a signal having a power level range of approximately 40 dBm around a center operating power level.

30. The computerized tomography system of claim 17, wherein the modulated data signal comprises a signal having its operating power centered at approximately -13.6 dBm.

31. The computerized tomography system of claim 17, wherein the detected image signal comprises a serial digital bit data signal, the modulated data signal comprises an amplitude modulated data signal, and the carrier signal comprises a RF carrier signal.

32. The computerized tomography system of claim 17, wherein the modulated data signal comprises a signal in a non return to zero inverted (NRZI) format.

33. The computerized tomography system of claim 17, wherein the first and second elements are separated by an air gap in the range of 0.050 to 0.080 inches.

34. The computerized tomography system of claim 17, wherein the first element comprises a RF slippage.

35. The computerized tomography system of claim 17, wherein the second element comprises at least one of a RF shoe, a pickup shoe antenna, a contactless brush, and a coupler.

36. The computerized tomography system of claim 17, further comprising

a stationary frame;

a generally annular rotating frame, in fixed relation to the stationary frame, having an aperture adapted to rotationally encircle a test object placed in an image plane therein, and having the x-ray source and the detector each disposed on opposite sides of the aperture; and

a coupling system have first and second elements disposed on the rotating frame and the stationary frame, respectively, the first element receiving the modulated data signal from the transmitter for providing electromagnetic coupling thereof to the second element.

37. A method for reconstructing a digital image data in a computerized tomography (CT) system, the digital image data being radio frequency (RF) amplitude modulation encoded to be electromagnetically coupled from a RF slippage disposed on a rotating frame of the CT system to a contactless coupler disposed on a stationary frame of the CT system, comprising:

(a) receiving an encoded RF amplitude modulated digital serial data signal having a wide range of power levels from the RF slippage;

(b) bandwidth filtering the encoded data signal causing undesirable frequency components to be filtered out;

(c) attenuating the amplitude of the encoded data signal partly in response to a control voltage;

(d) amplifying the amplitude of the encoded data signal causing amplification of the desired data component of the encoded data signal;

(e) digitizing the encoded data signal by filtering out the RF component therein; and

(f) filtering the encoded data signal to remove remaining undesirable signal components from the RF amplitude modulation encoding, whereby the digital image data has been reconstructed from the encoded data signal for CT image processing at the stationary frame for generating a CT image thereof.

38. The method of claim 37, further comprising the step of converting the reconstructed digital image data in step (f) to a digital fiber optic signal for transmission along a fiber optic cable.

39. The method of claim 37, further comprising the steps of:

generating a digital signal level proportional to the amplitude of the digitized encoded data signal in step (e); and

controlling the amount of attenuation of the amplitude of the encoded data signal in step (c) by a feedback loop such that the reconstructed digital image data of step (f) will be within a desired signal level.

40. The method of claim 39, further comprising repeating the generating and controlling steps to iteratively maintain the desired signal level.

41. The method of claim 37, further comprising the step of displaying the amount of attenuation provided in the attenuating step (c) as a relative signal level.

42. The method of claim 37, wherein step (a) of receiving an encoded digital RF amplitude modulated data signal comprises receiving the encoded data signal having a power level range of 40 dBm around a center operating power level.

43. The method of claim 37, wherein step (a) of receiving an encoded digital RF amplitude modulated data signal comprises receiving the encoded data signal having a power level range of 50 dBm around a center operating power level.

44. The method of claim 37, wherein step (a) of receiving an encoded digital RF amplitude modulated data signal comprises receiving the encoded data signal having a center operating power level of approximately -13.6 dBm.

45. The method of claim 37, wherein the bandwidth filtering step (b) comprises filtering between a bandwidth range of 500 MHz to 1 GHz.

46. The method of claim 36, wherein the attenuating step (c) comprises variably attenuating the amplitude of the encoded data signal by an attenuation range of -40 dB to 0 dB.

47. The method of claim 37, wherein the amplifying step (d) comprises amplifying the amplitude of the encoded data signal by a gain of up to approximately 45 dB.

48. The method of claim 37, wherein the amplifying step (d) comprises a plurality of amplifying steps, each of the plurality of amplifying steps providing a gain of up to approximately 15 dB.

49. The method of claim 37, wherein the filtering step (f) comprises adding hysteresis to the encoded data signal to remove remaining undesirable signal components from the RF amplitude modulation encoding.

50. An apparatus for demodulating a modulated digital data signal having a wide range of power levels and a modulation component to its pre-modulated state, comprising:

a variable attenuation circuit coupled to the source of a modulated digital data signal to receive and attenuate the amplitude of the modulated digital data signal in

response to a feedback control signal to generate an amplitude stabilized signal;

a conversion circuit coupled to the variable attenuation circuit to remove the modulation component from the amplitude stabilized signal to generate a digitized signal therefrom;

a filter coupled to the conversion circuit to generate a reconstructed digital data signal representative of the modulated digital data signal in its pre-modulated state, wherein the filter is configured to remove undesirable residual signal components from the digitized signal and generate an amplitude signal representative of the amplitude of the digitized signal; and

an attenuation controller coupled to the filter and the variable attenuation circuit, wherein the attenuation controller generates the feedback control signal based up the amplitude signal and applied the feedback control signal to the variable attenuation circuit.

51. The apparatus of claim 50, further comprising:

a bandwidth filter coupled to the variable attenuation circuit to filter the modulated digital data signal, wherein the bandwidth filter has a bandwidth range of 500 MHz to 1 GHz; and

an amplifier coupled to the variable attenuation circuit and the conversion circuit to receive the amplitude stabilized signal and generate an amplitude gain signal, wherein the amplitude gain signal is amplified by a gain of up to 45 dB.

52. The apparatus of claim 51, further comprising a fiber optic circuit coupled to the filter to receive the reconstructed digital data signal and generate a digital fiber optic signal for transmission in a fiber optic cable.

53. The apparatus of claim 50, wherein the filter comprises a digital hysteresis circuit that adds hysteresis to the digitized signal and the conversion circuit comprises a radio frequency (RF) amplitude modulation digital envelope detector.

54. The apparatus of claim 50, wherein the modulated digital data signal comprises a signal having a power level range of up to 50 dBm around a center operating power level.

55. The apparatus of claim 50, wherein the modulated digital data signal comprises a signal having a power level range of up to 40 dBm around a center operating power level.

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