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[Continued on next page]

- (54) Title:** LC-TYPE VCO

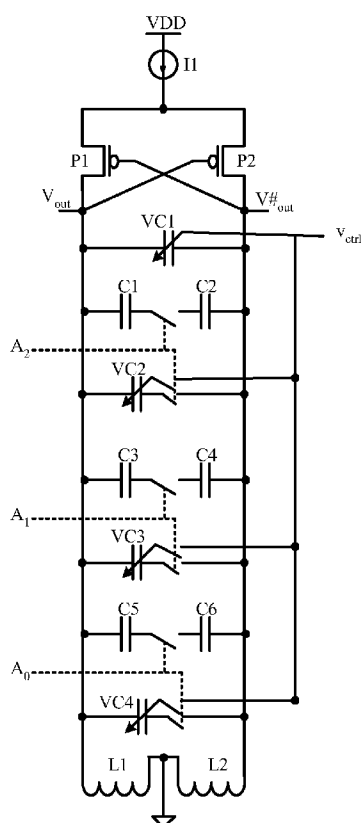


FIGURE 2

- (57) Abstract:** Disclosed herein are embodiments of an LC-type VCO with multiple operational frequency bands having reasonably similar frequency vs. control signal slopes.



Published:

— *with international search report*

LC-TYPE VCO

BACKGROUND

The present invention relates generally to voltage controlled oscillators (VCOs) and in particular, to LC-type VCOs.

- 5 Voltage controlled oscillators (VCOs) are used for a wide variety of applications including but not limited to phase locked loops (PLLs), delay locked loops (DLLs) and mixer circuits, to mention just a few. There are different types of VCOs including ring oscillator and LC (inductor-capacitor) types. LC VCOs typically use one or more amplifiers coupled to a tank circuit formed from one or more inductors and capacitors.
- 10 usually, some of the capacitors and/or inductors are controllably variable (e.g., based on an applied control voltage) so that the LC VCO output frequency can be varied in response to the control voltage.

- Figure 1 shows a conventional LC-type VCO. It generally comprises a differential amplifier, formed from P-type transistors P1, P2, and current source I1, and a tank circuit
- 15 formed from selectably engageable fixed capacitors C1-C6, variable capacitor (or varactor) VC1, and inductors L1, L2. The fixed capacitors are grouped into three separately engageable sets: C1-C2, C3-C4, and C5-6. Different combinations of these capacitor groups can be selected resulting in eight different operating frequency bands. The variable capacitor VC1 is coupled to a control voltage V_{ctrl} (e.g., from a charge pump
- 20 or loop filter in a PLL), which controls the output frequency by controlling the capacitance of the variable capacitor VC1. The VCO generates a differential output signal at nodes V_{out} , $V_{out}^{\#}$ whose frequency is based on the selected capacitor groups and the control voltage (V_{ctrl}) level.

- Unfortunately, an undesirable aspect of this circuit is that the slopes (K_{VCO}) of the
- 25 output frequency vs. control voltage curve is different for each frequency band. Accordingly, an improved solution is desired.

BRIEF DESCRIPTION OF THE DRAWINGS

- Embodiments of the invention are illustrated by way of example, and not by way of limitation, in the figures of the accompanying drawings in which like reference
- 30 numerals refer to similar elements.

Figure 1 is a schematic diagram of a conventional LC type VCO.

Figure 2 is a schematic diagram of an LC-type VCO in accordance with some embodiments.

Figure 3 is a schematic diagram of a selectably engageable variable capacitor in accordance with some embodiments.

Figure 4 is a block diagram of a computer system with an LC VCO in accordance with some embodiments.

Figure 5 is a schematic diagram of an LC-type VCO in accordance with some other embodiments.

10

DETAILED DESCRIPTION

At least for purposes of this disclosure, VCO gain (K_{vco}) is defined as oscillation frequency change for a given control signal (e.g., control voltage) change. This can be represented as:

$$(Eq. 1) K_{vco} = \frac{d\omega_o(V_{ctrl})}{dV_{ctrl}}$$

15

For an LC-type VCO (such as is shown in Figure 2 or Figure 5, discussed below), K_{vco} can be derived from the tank resonant frequency:

$$(Eq. 2) K_{vco} = \frac{d\omega(V_{ctrl})}{dV_{ctrl}} = \frac{d}{dV_{ctrl}} \left[\frac{1}{\sqrt{LC(V_{ctrl})}} \right]$$

20

Where $C(V_{ctrl}) = C_{fixed} + C_{var}(V_{ctrl})$ is the total tank capacitance. (In this type of LC VCO, the fixed and variable capacitors are in parallel thereby resulting in their net capacitance being additive. It should be appreciated, however, that other inductor-capacitor configurations could be implemented and are within the scope of the invention.) Accordingly, the overall tank capacitance is a function of V_{ctrl} because the variable capacitance depends on V_{ctrl} . K_{vco} can be expressed as:

Taking the derivative yields an equation for K_{vco} :

25

$$(eq. 3) K_{vco} = \frac{1}{\sqrt{L}} \cdot \frac{d}{dV_{ctrl}} \left[\frac{1}{\sqrt{C(V_{ctrl})}} \right] = -\frac{1}{2\sqrt{L}} \cdot \frac{1}{[C(V_{ctrl})]^{3/2}} \cdot \frac{dC(V_{ctrl})}{dV_{ctrl}}$$

because: $\omega(V_{ctrl}) = \frac{1}{\sqrt{LC(V_{ctrl})}}$, the K_{vco} equation can be rearranged to get:

$$(Eq. 4) \quad K_{VCO,rad} = - \frac{L \cdot \omega_O (V_{ctrl})^3}{2} \cdot \frac{dC_{var}(V_{ctrl})}{dV_{ctrl}}$$

It has been determined that in order to attain an LC-type VCO with a wide-band capability with a constant (or reasonably approaching constant) K_{vco} for the different frequency bands, the change in frequency should be compensated by either changing inductance (which is typically impractical) or by changing the variable capacitance slope. Different embodiments disclosed herein involve changing the varactor slope, e.g., by appropriately changing the variable capacitance in conjunction with changes in the fixed capacitance, when moving between frequency bands. By adjusting the variable capacitor sizes (and, consequently, the slope of the combined tank varactor) to accommodate the change in frequency (i.e., change in total tank capacitance), K_{vco} can be made to be reasonably consistent over the different frequency bands.

Figure 2 shows an LC-type VCO with compensated variable capacitance in accordance with some embodiments. The depicted VCO is similar to the VCO circuit of Figure 1, except that each fixed capacitor group additionally has an associated variable capacitor that is selectably engageable with an associated fixed capacitor group. The depicted VCO has a variable capacitor VC1 that is always engaged, along with three additional variable capacitors VC2-VC4, each associated with a different fixed capacitor group. In particular, the VCO has a first capacitor group formed from fixed capacitors C1, C2, and variable capacitor VC2; it has a second capacitor group formed from fixed capacitors C3, C4, and variable capacitor VC3; and it has a third capacitor group formed from fixed capacitors C5, C6, and variable capacitor VC4.

In accordance with some embodiments, the total fixed capacitance of the first group is 2 times larger than that of the second group, which is two times larger than that of the third group. In other words, the first group is the most significant group, and the third group is the least significant group, with the groups being binary weighted relative to each other. With such a configuration, eight separate, equally spaced apart frequency bands are available, depending on the values of selection control inputs A_0 , A_1 , and A_2 . For example, in some embodiments, C5 and C6 are each 0.2 pF, resulting in the third group fixed capacitance being 0.1 pF. C3 and C4 are each 0.4 pF, resulting in the second group capacitance being 0.2 pF; and C1 and C2 are 0.8 pF, resulting in the first group

capacitance being at 0.4 pF. The inductors L1, L2 are each 0.5 nH, and the variable capacitors are each at capacitor ranges that result in a near-constant KVCO for the eight different frequency bands. In accordance with Equation 3 (above), the variable capacitors are chosen so that the product of $1/[C(V_{ctrl})^{3/2}]$ and $dC(V_{ctrl})/dV_{ctrl}$ is maintained

- 5 substantially constant over the different frequency bands. (Note that the $dC(V_{ctrl})/dV_{ctrl}$ term is only a function of the engaged variable capacitors since the derivative of the constant engaged fixed capacitors is zero. However, the first term, $1/[C(V_{ctrl})^{3/2}]$, is a function of both the engaged fixed and variable capacitors.)

- It should be appreciated that different capacitor and/or inductor configurations may be implemented to attain reasonably constant K_{VCO} , along with desired design parameters, as taught herein. In the depicted embodiment, each selectable capacitor group comprises two fixed capacitors, in series with each other, with the different selectable groups in parallel with each other, as well as with the inductors. However, persons of skill will appreciate that there are numerous other ways to achieve an LC-type VCO and retain
- 15 reasonably constant K_{VCO} as taught herein. One or more fixed and/or variable capacitors could be used in each group, and the groups could be arranged in different combinations relative to one another other than simply being in parallel, although since capacitors in parallel are additive, it makes maintaining the product of $1/[C(V_{ctrl})^{3/2}]$ and $dC(V_{ctrl})/dV_{ctrl}$ simpler to implement. Along these lines, the fixed and variable capacitors may be
- 20 separately controllable, i.e., not all controlled by selection lines A_i . Moreover, their values may be distributed in any suitable manner. for example, the fixed capacitors may be binary weighted, while the variable capacitors could be thermo-coded. the opposite could be true or both could be weighted the same, e.g., binary or thermo-coded.

- Figure 3 shows a circuit for implementing a selectable capacitor group in accordance with some embodiments. It comprises variable capacitors VC_B , VC_C , switch transistors N1, N2, N3 and fixed capacitors C_B and C_C , all coupled together as shown. The variable capacitors, VC_B and VC_C , are formed from PMOS transistors with their drains and sources coupled together, as shown. They are coupled together in a back-to-back configuration, thereby allowing them to be biased by the frequency control voltage
- 30 V_{ctrl} which is substantially decoupled from the output rails V_{out} , $V_{\#out}$. With such variable capacitor implementations, as V_{ctrl} increases, so to does the variable capacitance thereby lowering the output frequency.

The selectable capacitor group is coupled between terminals B and C, which correspond to the output rails V_{out} and $V_{\#out}$ from the VCO of Figure 2. A select control input A_i is coupled to the gates of switch transistors N1, N2, N3, and the frequency control voltage node V_{ctrl} is coupled to the drain/sources of P-type MOS transistors used to form the variable capacitors VC_B and VC_C . (Note that in this implementation, there are two N-type switches, N1 and N2, to isolate the MOS-implemented variable capacitors from the control voltage (V_{ctrl}), as well as from the output rails. This may be desirable since the control voltage node, which may substantially be DC in nature, may function as an AC ground. In other embodiments, for example, with other types of variable capacitor implementations, this may or may not be desired. Moreover, while MOS-transistor variable capacitors are shown, any suitable variable capacitor solution (e.g., PN junction, etc.) could be used, depending on design objectives.)

(The term "PMOS transistor" refers to a P-type metal oxide semiconductor field effect transistor. Likewise, "NMOS transistor" refers to an N-type metal oxide semiconductor field effect transistor. It should be appreciated that whenever the terms: "transistor", "MOS transistor", "NMOS transistor", or "PMOS transistor" are used, unless otherwise expressly indicated or dictated by the nature of their use, they are being used in an exemplary manner. They encompass the different varieties of MOS devices including devices with different VTs and oxide thicknesses to mention just a few. Moreover, unless specifically referred to as MOS or the like, the term transistor may include other suitable transistor types, e.g., junction-field-effect transistors, bipolar-junction transistors, and various types of three dimensional transistors, known today or not yet developed.)

With reference to Figure 4, one example of a computer system is shown. The depicted system generally comprises a processor 402 that is coupled to a power supply 404, a wireless interface 408, and memory 406. It is coupled to the power supply 404 to receive from it power when in operation. The wireless interface 408 is coupled to an antenna 410 to communicatively link the processor through the wireless interface chip 408 to a wireless network (not shown). The processor 402 comprises a communications interface 403, having one or more LC-type VCOs 403 such as are disclosed herein, to communicatively link the processor 402 to the memory 406.

It should be noted that the depicted computer system could be implemented in different forms. That is, it could be implemented in a single chip module, a circuit board,

or a chassis having multiple circuit boards. Similarly, it could constitute one or more complete computers or alternatively, it could constitute a component useful within a computing system.

The invention is not limited to the embodiments described, but can be practiced
5 with modification and alteration within the spirit and scope of the appended claims. For example, Figure 5 shows another embodiment of an LC-type VCO in accordance with embodiments of the invention. It generally comprises variable capacitor groups, formed from variable capacitors, VC1 to VCM and controlled by control lines A_{V1} to A_{VM} , and fixed capacitor groups formed from fixed capacitors C1 to CN and separately controlled
10 by control lines A_{F1} to A_{FN} . Thus, it is shown that the fixed and variable capacitors may be controlled in any suitable manner, either in association with each other or independently. This embodiment also employs cross-coupled NMOS transistors N1, 2 and a single inductor L1 coupled between the output rails as shown.

In addition, it should be appreciated that the present invention is applicable for use
15 with all types of semiconductor integrated circuit ("IC") chips. Examples of these IC chips include but are not limited to processors, controllers, chip set components, programmable logic arrays (PLA), memory chips, network chips, and the like.

Moreover, it should be appreciated that example sizes/models/values/ranges may have been given, although the present invention is not limited to the same. As
20 manufacturing techniques (e.g., photolithography) mature over time, it is expected that devices of smaller size could be manufactured. In addition, well known power/ground connections to IC chips and other components may or may not be shown within the FIGS. for simplicity of illustration and discussion, and so as not to obscure the invention. Further, arrangements may be shown in block diagram form in order to avoid obscuring
25 the invention, and also in view of the fact that specifics with respect to implementation of such block diagram arrangements are highly dependent upon the platform within which the present invention is to be implemented, i.e., such specifics should be well within purview of one skilled in the art. Where specific details (e.g., circuits) are set forth in order to describe example embodiments of the invention, it should be apparent to one skilled in
30 the art that the invention can be practiced without, or with variation of, these specific details. The description is thus to be regarded as illustrative instead of limiting.

CLAIMS

What is claimed is:

1. A chip, comprising:
an LC-type VCO having two or more operational frequency bands with
5 substantially equivalent frequency gains.
2. The chip of claim 1, in which the VCO comprises one or more selectable capacitor groups each having at least one fixed capacitor and at least one variable capacitor to be controlled by a frequency control voltage.
3. The chip of claim 2, in which the one or more selectable capacitor groups comprise a
10 plurality of selectable capacitor groups that are controllably coupled in parallel with one another.
4. The chip of claim 3, in which the plurality of selectable capacitor groups are controllably coupled to one or more inductors to form a resonant tank circuit.
5. The chip of claim 2, in which the at least one variable capacitor comprises a MOS
15 transistor configured into a capacitor.
6. The chip of claim 5, in which the at least one variable capacitor comprises first and second MOS capacitors configured as capacitors and coupled together in a back-to-back configuration at a common frequency control voltage node.
7. The chip of claim 2, in which the VCO comprises one or more variable capacitors
20 coupled in parallel to the one or more selectable capacitor groups and to be controlled by the frequency control voltage.
8. The chip of claim 7, in which the product of $1/[C(V_{ctrl})^{3/2}]$ and $dC(V_{ctrl})/dV_{ctrl}$, where V_{ctrl} is the frequency control voltage and $C(V_{ctrl})$ is the total engaged capacitance as a function of the frequency control voltage, is maintained substantially constant for the two
25 or more operational frequency bands.
9. An integrated circuit, comprising:
an amplifier to generate a frequency signal based on a resonant tank circuit that comprises a plurality of selectable fixed and variable capacitors to provide two or more operational bands each having a reasonably equivalent frequency vs. control voltage slope.
- 30 10. The integrated circuit of claim 9, in which the amplifier is a differential amplifier having complementary output nodes to provide the frequency signal.

11. The integrated circuit of claim 10, in which the plurality of fixed and variable capacitors make up selectable capacitor groups selectably coupled in parallel with each other and to the complementary output nodes.
12. The integrated circuit of claim 11, in which the tank circuit comprises first and second
5 inductors coupled between the complementary output nodes and a supply reference.
13. The integrated circuit of claim 12, in which the variable capacitors are formed from capacitor-configured MOS transistors.
14. The integrated circuit of claim 13, comprising a variable capacitor fixedly coupled between the complementary output nodes.
- 10 15. A computer system, comprising:
 a processor having one or more LC-type VCOs to implement at least one communication interface, the one or more LC-type VCOs comprising: two or more operational frequency bands with substantially equivalent frequency gains;
 a network interface device coupled to the processor via the at least one
15 communication interface; and
 an antenna coupled to the network interface device to communicatively link the processor to a wireless network.
16. The computer system of claim 15, in which the VCO comprises one or more selectable capacitor groups each having at least one fixed capacitor and at least one
20 variable capacitor to be controlled by a frequency control voltage.
17. The computer system of claim 16, in which the one or more selectable capacitor groups comprise a plurality of selectable capacitor groups that are controllably coupled in parallel with one another.
18. The computer system of claim 17, in which the plurality of selectable capacitor groups
25 are controllably coupled to one or more inductors to form a resonant tank circuit.
19. The computer system of claim 16 in which the at least one variable capacitor comprises a MOS transistor configured into a capacitor.
20. The computer system of claim 19, in which the at least one variable capacitor comprises first and second MOS capacitors coupled together in a back-to-back
30 configuration at a common frequency control voltage node.

21. The chip of claim 16, in which the VCO comprises one or more variable capacitors coupled in parallel to the one or more selectable capacitor groups and to be controlled by the frequency control voltage.

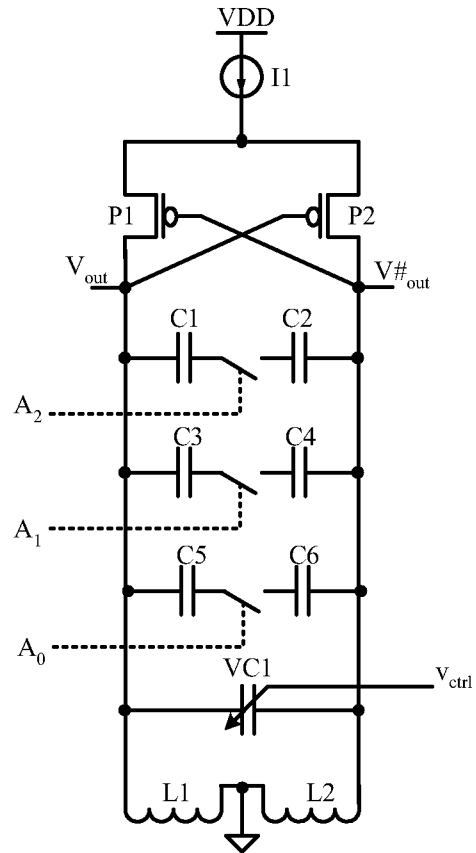


FIGURE 1
(PRIOR ART)

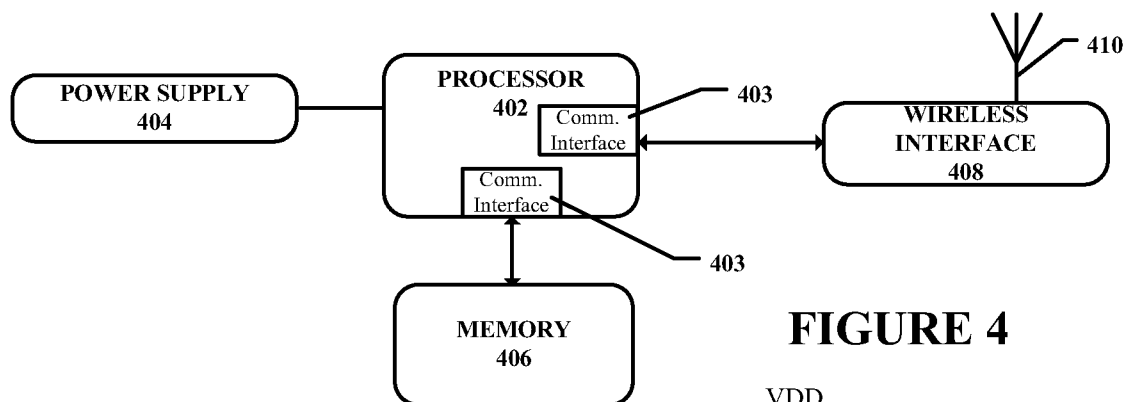


FIGURE 4

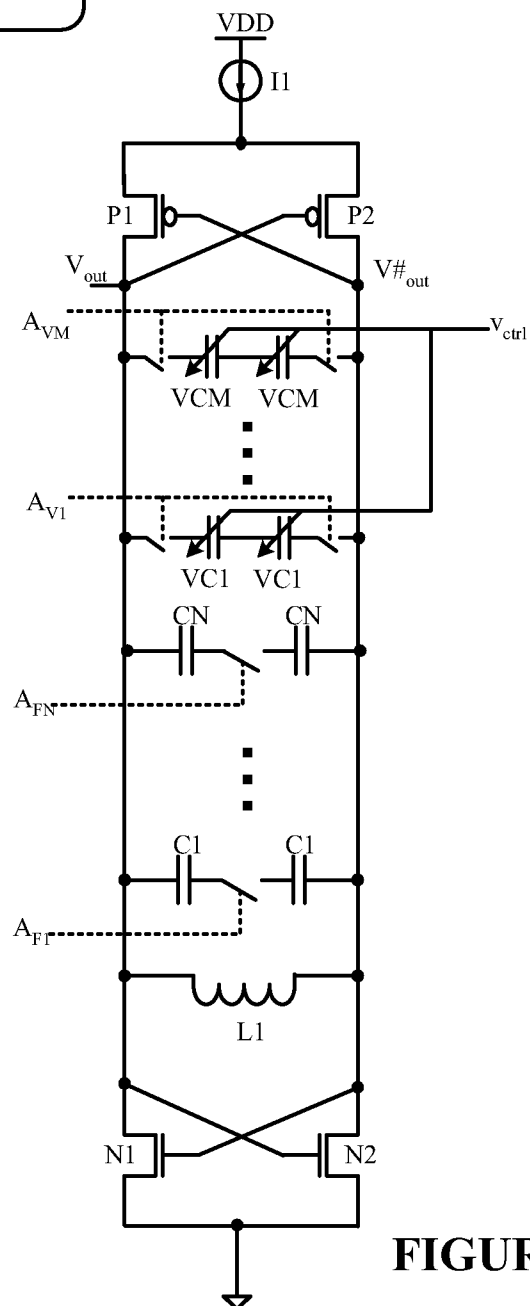


FIGURE 5

A. CLASSIFICATION OF SUBJECT MATTER**H03B 5/12(2006.01)i, H03L 7/099(2006.01)i**

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 8: H03B

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Korean Utility models and applications for Utility models since 1975

Japanese Utility models and applications for Utility models since 1975

Electronic data base consulted during the international search (name of data base and, where practicable, search terms used)

eKIPASS(KIPO internal), IEEE & Keyword: VCO," "OSCILLATOR," "VARIABLE CAPACITOR," "CAPACITOR BANK," "MULTI-BAND," and "QUADRATURE"

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	KR 10-2006-0115193 A (SAMSUNG ELECTRONICS CO., LTD.) 08 November 2006 See the abstract; figures 4-11; claims 1-48; pages 3-7	1-7, 9-14
X	KR 10-2004-0069822 A (SAMSUNG ELECTRONICS CO., LTD.) 06 August 2004 See the abstract; figures 7-8; claims 1-9; pages 2-7	1-7, 9-14
X	US 6833769 B2 (Pauli Seppinen et al.) 21 December 2004 See the abstract; figures 4-6, 8 & 9; claims 1-16; column 2 line 29-column 3 line 19 & column 4 line 22-column 8 line 51,	1-7, 9-21
A	JP 2004-147310 A (MATSUSHITA ELECTRIC IND CO LTD) 20 May 2004 See the abstract; figures 1, 8, 9, 11, 14-22; claims 1-18	1-21
A	JP 2006-229266 A (MATSUSHITA ELECTRIC IND CO LTD) 31 August 2006 See the abstract; figures 1, 4-11; claims 1-9	1-21
A	KR 10-2006-0048020 A (SONY ERICSSON MOBILECOMMUNICATIONS JAPAN INC) 18 May 2006 See the abstract; figures 1-3; claims 1-8	1-21



Further documents are listed in the continuation of Box C.



See patent family annex.

* Special categories of cited documents:

"A" document defining the general state of the art which is not considered to be of particular relevance

"E" earlier application or patent but published on or after the international filing date

"L" document which may throw doubts on priority claim(s) or which is cited to establish the publication date of citation or other special reason (as specified)

"O" document referring to an oral disclosure, use, exhibition or other means

"P" document published prior to the international filing date but later than the priority date claimed

"T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention

"X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

"Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art

"&" document member of the same patent family

Date of the actual completion of the international search

28 AUGUST 2008 (28.08.2008)

Date of mailing of the international search report

29 AUGUST 2008 (29.08.2008)

Name and mailing address of the ISA/KR

Korean Intellectual Property Office
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Facsimile No. 82-42-472-7140

Authorized officer

Cho Sung Chan

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INTERNATIONAL SEARCH REPORT

International application No.
PCT/US2008/061217

Box No. II Observations where certain claims were found unsearchable (Continuation of item 2 of first sheet)

This international search report has not been established in respect of certain claims under Article 17(2)(a) for the following reasons:

1. ☐ Claims Nos.:
because they relate to subject matter not required to be searched by this Authority, namely:

2. ☐ Claims Nos.:
because they relate to parts of the international application that do not comply with the prescribed requirements to such an extent that no meaningful international search can be carried out, specifically:

3. ☐ Claims Nos.:
because they are dependent claims and are not drafted in accordance with the second and third sentences of Rule 6.4(a).

Box No. III Observations where unity of invention is lacking (Continuation of item 3 of first sheet)

This International Searching Authority found multiple inventions in this international application, as follows:

Claims 1-8 are directed to a chip comprising an LC type VCO. Claims 9-14 are directed to an integrated circuit comprising an amplifier and a resonant tank circuit. Claims 15-21 are directed to a computer system comprising a processor having an LC type VCO. The only common technical feature between claims 1-8, claims 9-14 and claims 15-21 is LC type VCO (or amplifier and resonant tank circuit having a plurality of selectable fixed and variable capacitors to provide two or more operational frequency bands).

However, this technical feature lacks novelty and inventive step with respect to the following documents cited in this ISR:

(a) KR 10-2006-0115193 A, (b)KR 10-2004-0069822 A, (c) US 6833769 B2

Thus there is no technical relationship left over the prior art among the claimed inventions, leaving the claims without a single general inventive concept. The application, hence, does not meet the requirements of Unity of Invention as defined in PCT Rules 13.1 and 13.2.

1. ☐ As all required additional search fees were timely paid by the applicant, this international search report covers all searchable claims.
2. ☒ As all searchable claims could be searched without effort justifying an additional fee, this Authority did not invite payment of any additional fee.
3. ☐ As only some of the required additional search fees were timely paid by the applicant, this international search report covers only those claims for which fees were paid, specifically claims Nos.:

4. ☐ No required additional search fees were timely paid by the applicant. Consequently, this international search report is restricted to the invention first mentioned in the claims; it is covered by claims Nos.:

Remark on Protest

- ☐ The additional search fees were accompanied by the applicant's protest and, where applicable, the payment of a protest fee.
- ☐ The additional search fees were accompanied by the applicant's protest but the applicable protest fee was not paid within the time limit specified in the invitation.
- ☐ No protest accompanied the payment of additional search fees.

INTERNATIONAL SEARCH REPORT

Information on patent family members

International application No.

PCT/US2008/061217

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