Title: MOTOR CONTROL SYSTEM FOR DYNAMICALLY CHANGING MOTOR ENERGIZATION CURRENT WAVEFORM PROFILES

Abstract: A control system is provided for a multiphase motor having a plurality of stator phase components, each stator phase component comprising a phase winding formed on a core element, and a rotor. A plurality of motor control schemes are stored in memory or calculated dynamically during motor operation. Subject to user or system selection, one of the motor control schemes is selected by a controller for implementation. The controller generates control signals that are applied to energization circuitry for supplying current to the phase windings with a particular current waveform profile in accordance with the selected motor control scheme. The controller has an input terminal for receiving a user initiated torque command signal representing a desired motor torque. Each motor control scheme provides motor driving current that corresponds to torque command signals received at the controller input terminal. In an automatic mode, the controller accesses the appropriate control scheme to meet the torque tracking requirements of the system.
Published:
without international search report and to be republished upon receipt of that report

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MOTOR CONTROL SYSTEM FOR DYNAMICALLY CHANGING MOTOR ENERGIZATION CURRENT WAVEFORM PROFILES

Related Applications

This application contains subject matter related to copending U.S. Application number 09/826,423 of Maslov et al., filed April 5, 2001, copending U.S. Application number 09/826,422 of Maslov et al., filed April 5, 2001, U.S. Application number 09/966,102, of Maslov et al., filed October 1, 2001, U.S. Application number 09/993,596 of Pyntikov et al., filed November 27, 2001, U.S. Application number 10/173,610, of Maslov et al., filed June 19, 2002, and U.S. Application number 10/290,537, of Maslov et al., filed November 8, 2002, all commonly assigned with the present application. The disclosures of these applications are incorporated by reference herein.

Field of the Invention

The present invention relates to control of electric motors, more particularly to implementation, individually, of a plurality of motor control schemes to effect associated stator current waveform profiles.

Background

The above-identified copending patent applications describe the challenges of developing efficient electric motor drives. Electronically controlled pulsed energization of motor windings offers the prospect of more flexible management of motor characteristics. By control of pulse width, duty cycle, and switched application of an energy source to appropriate stator windings, greater functional versatility can be achieved. The use of permanent magnets in conjunction with such windings is advantageous in limiting current consumption.

In a vehicle drive environment, wherein power availability for a traction motor is limited to an on-board supply, it is highly desirable to attain a high torque output capability at minimum power consumption while maintaining high efficiency in all conditions of traction motor operation. Motor structural arrangements described in the copending applications contribute to these objectives. As described in those applications, electromagnet core segments may be configured as isolated magnetically permeable structures in an annular ring to provide increased flux concentration. Isolation of the
electromagnet core segments permits individual concentration of flux in the magnetic cores, with a minimum of flux loss or deleterious transformer interference effects occurring from interaction with other electromagnet members.

The above-identified copending application 10/173,610 describes a control system for a multiphase motor that compensates for variations in individual phase circuit elements. A high degree of precision controllability is obtained with each phase control loop closely matched with its corresponding winding and structure. Successive switched energization of each phase winding is governed by a controller that generates signals in accordance with parameters associated with the respective stator phase components and selected driving algorithms. The phase windings are energized with current of sinusoidal waveform for high efficiency operation. The control system varies the output current to respond to, and accurately track, the user's torque command input.

The sinusoidal current waveform profile obtained with this commutation strategy can extend battery life through efficient operation. However, in vehicle driving operation there may be a need for torque capability in excess of that available from the most efficient control scheme. Typically, the power supply is rated for a maximum current discharge rate, for example, 10.0 amps. If the user of the system requests a torque command that correlates to this maximum current draw, then the motor torque output for a sinusoidal current waveform profile is limited, for example, to approximately 54.0 Nm in a motor with a configuration such as described above. In vehicle drive applications, torque input commands are associated by users with commands for change of speed. In typical driving operation, user torque requests are subject to wide variability with little, if any, long term predictability. A driver may demand higher acceleration or greater speed than the system can accommodate at maximum torque with a sinusoidal current waveform. Driving conditions, such as steep uphill grade or heavy vehicle load or the like, may impose other limitations on available speed and acceleration. Other non-vehicular applications may have similar high torque requirements.

The need thus exists for a motor control system that is capable of performing with high efficiency yet can deliver increased torque output when required by the user. The above-identified application 10/290,537 addresses this need by making available a plurality of motor control schemes for a motor drive, each of which can provide a unique current waveform profile. One of the motor control schemes may be selected by the user to obtain a current waveform profile that has the greatest capability to meet operating objectives.
For example, a control scheme may be selected that yields high efficiency operation, such as a sinusoidal waveform, while another control scheme may be selected that provides higher torque, albeit with less operating efficiency. Selection among motor control schemes may be made in accordance with the user's needs or objectives with respect to torque and efficiency, or other factors, e.g., low torque ripple and noise, etc., at any particular time. A selected motor control scheme will be implemented to generate control signals to produce motor energization current having the associated waveform profile.

In a vehicle traction application, for example, user profile selection provides the driver of a vehicle flexibility to adjust operation to meet objectives. For example, if the driver seeks to reach the destination in minimum time, a high torque profile can be selected and maintained throughout a trip to provide maximum speed and acceleration capability. If, however, a greater concern is to conserve an on-board energy source for a relatively long trip, the high efficiency profile can be selected throughout, possibly with the user's selection of the high torque profile at various points on a limited basis. Reference is made to the ([identify]) application for a more detailed description of exemplified waveforms, particularly high efficiency, sinusoidal waveforms, and high torque, square wave shaped waveforms.

The variable conditions and changing requirements of vehicle operation, however, may call for a change in profile more frequently or rapidly than the driver can, or would desire to, keep pace with. A driver's torque requests may be adequately met with selection of the high efficiency profile mode except for relatively transient instances such as passing situations, uphill grades, etc. In those instances, the driver may not be sufficiently responsive to the changing conditions to obtain optimum advantage of a change in selection from a high efficiency profile to a high torque profile. When the high torque requirement conditions diminish, return to the high efficiency profile may be delayed until the user realizes that the high torque profile is no longer necessary, thus drawing unnecessary current from the battery. Thus, it would be desirable to use the high torque mode only when torque greater than that available from the high efficiency mode is required.

The need remains for a system in which a motor control scheme is automatically selected on a dynamic basis to provide an appropriate energization current waveform profile.
Disclosure of the Invention

The present invention fulfills this need by providing a plurality of motor control schemes for a motor drive, each of which can produce a unique motor energization current waveform profile. One or more conditions are monitored continuously throughout motor operation. One of the motor control schemes is automatically selected on a dynamic basis in accordance with criteria associated with the monitored conditions. The motor is energized in accordance with the selected motor control scheme with the appropriate current waveform profile. The present invention provides additional advantages in motors having ferromagnetically autonomous stator electromagnets.

The motor control schemes may comprise a high efficiency motor control scheme that provides a current waveform profile for relatively optimum operating efficiency and a high torque motor control scheme that provides a current waveform profile for relatively high operating torque response. The system is responsive to a user input signal that represents a torque request. The user input signal is sensed and the ability of the system to meet the torque request is monitored to select the appropriate motor control scheme accordingly. The ability of the system to meet torque request is a function of motor speed, which is continuously sensed to facilitate the torque demand monitoring function. The high efficiency motor control scheme is implemented unless the corresponding current waveform profile is unable to meet the torque demand; otherwise, the high torque motor control scheme is implemented.

An additional advantage of the invention is that the user may be given the option to disengage the automatic profile selection function by inputting a manual selection. For example, the user may elect to employ the high efficiency mode throughout operation to preserve the power supply as much as possible on longer trips, even if occasional sacrifices are made in maximum available torque.

The present invention may be manifested in a control system for a multiphase motor having a plurality of stator phase components, each stator phase component comprising a phase winding formed on a core element, and a permanent magnet rotor. Preferably, each of the stator core elements comprises ferromagnetic material separated from direct contact with the other core elements, each stator phase component thereby forming an autonomous electromagnet unit. Stator energization current is provided by a direct current power supply through circuitry coupled to a controller. The controller can
access any of a plurality of stored motor control schemes to implement stator energization current having a corresponding waveform profile. The controller is dynamically responsive to one or more monitored conditions to effect selection of the motor control schemes. The stored motor control schemes are determinative of the current waveform profiles and, when accessed, are incorporated into controller operation. The motor control schemes may comprise, for example, an efficiency motor control scheme that provides a current waveform profile for relatively optimum operating efficiency, such as a substantially sinusoidal waveshape, and a high torque motor control scheme that provides a current waveform profile for a relatively high operating torque response, such as a substantially rectangular waveshape.

A user input is coupled to the controller from which a torque request signal is received. Another input to the controller receives a signal representing motor speed, derived from continuous sensing of rotor position. Based on these signals, the controller can derive the motor torque demands and the control voltages necessary to meet the torque demands on a real time basis. If the voltage required to sustain sinusoidal waveform profile mode exceeds the power supply voltage, then the controller selects the high torque profile mode operation, accessing the data therefor from the profile memory. As an alternative to repeated real time calculation for profile selection, a lookup table that correlates profile selection with torque request input and monitored speed may be stored in memory.

Additional advantages of the present invention will become readily apparent to those skilled in this art from the following detailed description, wherein only the preferred embodiment of the invention is shown and described, simply by way of illustration of the best mode contemplated of carrying out the invention. As will be realized, the invention is capable of other and different embodiments, and its several details are capable of modifications in various obvious respects, all without departing from the invention. Accordingly, the drawings and description are to be regarded as illustrative in nature, and not as restrictive.
Brief Description of Drawings

The present invention is illustrated by way of example, and not by way of limitation, in the figures of the accompanying drawing and in which like reference numerals refer to similar elements and in which:

Fig. 1 is an exemplary view showing rotor and stator elements in a configuration that may be employed in the present invention.

Fig. 2 is a block diagram of a motor control system in accordance with the present invention.

Fig. 3 is a block diagram that illustrates torque controller methodology for use in the control system of Fig. 2.

Fig. 4 is a flow chart for operation of the profile selection functionality in accordance with the present invention.

Fig. 5 is a curve representing profile mode selection for ranges of torque and speed in accordance with the present invention.

Detailed Description of the Invention

The present invention is applicable to a motor such as disclosed in the copending Application 09/826,422, although the invention can be used with various other motors, including permanent magnet motors. Fig. 1 thus is an exemplary view showing rotor and stator elements as described in that application, the disclosure of which has been incorporated herein. Rotor member 20 is an annular ring structure having permanent magnets 21 substantially evenly distributed along cylindrical back plate 25. The permanent magnets are rotor poles that alternate in magnetic polarity along the inner periphery of the annular ring. The rotor surrounds a stator member 30, the rotor and stator members being separated by an annular radial air gap. Stator 30 comprises a plurality of electromagnet core segments of uniform construction that are evenly distributed along the air gap. Each core segment comprises a generally U-shaped magnetic structure 36 that forms two poles having surfaces 32 facing the air gap. The legs of the pole pairs are wound with windings 38, although the core segment may be constructed to accommodate a single winding formed on a portion linking the pole pair. Each stator electromagnet core
structure is separate, and magnetically isolated, from adjacent stator core elements. The stator elements 36 are secured to a non-magnetically permeable support structure, thereby forming an annular ring configuration. This configuration eliminates emanation of stray transformer flux effects from adjacent stator pole groups. The stator electromagnets are thus autonomous units comprising respective stator phases. The concepts of the invention, more fully described below, are also applicable to other motor structures, including a unitary stator core that supports all of the phase windings.

Fig. 2 is a block diagram of a motor control system in accordance with the present invention. A plurality of multiphase motor stator phase windings 38 are switchably energized by driving current supplied from d-c power source 40 via hybrid power block 42. The power block may comprise electronic switch sets that are coupled to controller 44 via a pulse width modulation converter and gate drivers. Each phase winding is connected to a switching bridge having control terminals connected to receive pulse modulated output voltages from the controller. Alternatively, the switching bridges and gate driver components may be replaced by amplifiers linked to the controller output voltages. Reference is made to the above identified application 10/290,537 for a more detailed description of the winding power circuitry.

Current in each phase winding is sensed by a respective one of a plurality of current sensors 45 whose outputs are provided to controller 44. The controller may have a plurality of inputs for this purpose or, in the alternative, signals from the current sensors may be multiplexed and connected to a single controller input. Rotor position and speed sensor 46 provides rotor position and speed feedback signals to the controller. The sensor may comprise a well known resolver, encoder or their equivalents and a speed approximator that converts the position signals to speed signals in a well known manner. The controller is connected to the supply 40 by a primary power supply bus. The controller is also provided with user inputs, including a torque request input 47 and a profile selection input 48.

The controller may comprise a microprocessor or equivalent microcontroller, such as Texas Instrument digital signal processor TMS320LF2407APG. Coupled to the controller are program RAM memory 50, program ROM 52, DATA RAM 54 and profile memory 56. These illustrated units are merely representative of any well known storage arrangements by which the controller may access stored random data and program data. Profile memory 56 is shown separately in the drawing for purposes of illustration of the
inventive concepts. The profile memory may comprise a ROM in which are stored the portions of the motor control scheme programs that dictate the motor current waveform profiles obtained with implementation of the associated control schemes. The profile memory data may be stored in the form of a profile functions library and/or lookup tables. The profile memory data structure can be in a form of real-time calculations and optimization routines. As an alternative, or in addition, to ROM, a unit can be provided that calculates values during real-time motor operation.

In the vehicle drive application example, the torque request input 47 represents torque required by the user's throttle. An increase in throttle is indicative of a command to increase speed, which can be realized by an increase in torque. Alternatively, it may be indicative of a command to increase torque in order to maintain the same speed of a vehicle under heavy load conditions, such as uphill driving. In operation, the control system torque tracking functionality should maintain steady state torque operation for any given torque request input through varying external conditions, such as changes in driving conditions, load gradient, terrain, etc., and should be responsive to changes at the torque request input to accommodate the driver's throttle commands. The manner in which the control system responds to torque input requests is dependent upon the particular motor control scheme implemented. A plurality of motor control schemes are available to obtain an appropriate response. Each control scheme effects a particular motor current waveform profile having unique characteristics with respect to efficiency, torque capacity, response capability, power losses, etc.

Fig. 3 is a block diagram that illustrates a motor control scheme utilized in the copending 10/173,610 application identified and incorporated herein above. Reference is made to that application for detailed description of operation. In order to develop the desired phase currents the following per-phase voltage control expression is applied to the driver for the phase windings:

\[ V_i(t) = L_i dI_{th}/dt + R_i I_i + E_i + k_e e_i \]

Fig. 3 represents the methodology, generally indicated by reference numeral 60, by which the controller derives the components of this voltage expression in real time, utilizing the torque request input and the signals received from phase current sensors, position sensor and speed detector. Functional block 70 represents the formulation and addition of the components of the above expression to obtain the control voltage in real time. Each of the functional blocks 62, 64, 66, 68, 72, 74 and 76, shown as inputs to block
70, represents the generation of the various elements of the components obtained from real time inputs received by the controller or parameter constants. Block 62 represents a precision torque tracking functionality, the per-phase desired current trajectories being selected according to the following expression:

\[ I_a = \left( \frac{2\tau_d}{N_t K_{ri}} \right) \sin(N_t \theta_i) \]

where \( I_{th} \) denotes per-phase desired current trajectory, \( \tau_d \) denotes the user's requested torque command, \( N_t \) represents the total number of phase windings, \( K_{ri} \) denotes a per-phase torque transmission coefficient and \( \theta_i \) represents rotor position signal for the \( i^{th} \) phase. The per-phase current magnitude is dependent upon the per-phase value of the torque transmission coefficient \( K_{ri} \).

In operation, controller 44 successively outputs control signals \( V_i(t) \) to the hybrid power block for individual energization of respective phase windings in a sequence established in the controller. Each successive control signal \( V_i(t) \) is related to the particular current sensed in the corresponding phase winding, the immediately sensed rotor position and speed, and also to model parameters, \( K_{ai} \) and \( K_{di} \), that have been predetermined specifically for the respective phases. The computations illustrated in Fig. 3 are performed successively in real time. The expression shown in block 62 in this motor control scheme provides the desired current component for the tracking torque output control signal \( V_i(t) \) with a sinusoidal waveform profile. The sine wave current trajectory \( I_{\sin}(t) \) is generated from the following equation

\[ I_{\sin} = I_m \sin(N_t \theta) \]

where \( I_m \) denotes the phase current magnitude, \( N_t \) denotes the number of permanent magnet pairs and \( \theta \) denotes the measured per phase rotor position signal. As described in the above identified copending application 10/290,537, this sinusoidal current waveform profile provides efficient motor operation.

Different expressions for block 62 can be used for the torque tracking functionality of Fig. 3 to obtain different current waveform profiles for manifesting other operational aspects, although sacrificing some of the efficiency achieved with the sinusoidal waveform profile. For higher torque operation, the expression of block 62 shown in Fig. 3 can be replaced with an expression yielding a square wave current waveform trajectory \( I_{sq}(t) \), such as

\[ I_{sq} = I_m \text{ sgn} \left( \sin(N_t \theta) \right) \]
where sgn (x) denotes the standard signum function and is defined as 1 if x > 0, 0 if x = 0, and -1 if x < 0. Reference is again made to copending application 10/290,537 for a more detailed description comparing the efficiency and torque features of the two different current waveform profiles obtained from the motor control schemes discussed above.

Profile memory 56 stores data that are used by the controller to obtain the current values that satisfy the expressions exemplified above. For the square wave profile, the expression \( L, dI_a/dt \) may be prestored. The data may be stored as lookup tables in a profile functions library, each motor control scheme having a corresponding lookup table. Each entry in a lookup table represents a value of current, shown as the output of block 62 in Fig. 3, for a particular combination of torque request value and rotor position for the corresponding motor control scheme. If a control scheme is selected for which the sinusoidal waveform is produced, the corresponding profile memory data is accessed. Square wave profile memory data would be accessed if the corresponding control scheme is selected. Alternatively, the profile memory may store data for each profile with which the desired current value \( I_a \) is repeatedly computed by the controller in real time. While expressions for sinusoidal and square wave waveforms have been set forth above for purposes of illustration, other waveform profiles, such as sawtooth, etc., may be utilized for different operational purposes.

Selection of profile data can be made by the controller automatically as appropriate during motor operation. Alternatively, a user can select an operational mode corresponding to one of the profiles by inputting a profile selection signal at controller input 48. Profile selection operation is described with reference to the flow chart shown in Fig. 4. The description pertains to a specific example in which the profile memory contains data for implementing a high efficiency profile motor control scheme, such as a control scheme for producing a sinusoidal motor current waveform and for implementing a high torque profile, such as a control scheme for producing a square wave motor current waveform. This example is merely illustrative, as data for other profiles may be stored in the profile memory and accessed under operating conditions for which different current waveforms are appropriate.

In the absence of a profile select signal detected by the controller, an automatic profile selection mode is invoked. At step 100, the controller detects whether a user profile selection signal has been received at input 48 to determine whether the automatic mode is to be invoked. If the determination in step 100 is negative, the controller
determines at step 102 whether the received profile select signal is a high torque profile selection. If not, the controller, after any appropriate delay, accesses the profile memory to retrieve data from the high efficiency profile lookup table at step 104. The retrieved data yields the desired current value $I_d$ for the instantaneous values of the torque request and the sensed rotor position levels. If, instead, the high torque profile has been selected, as determined in step 102, the corresponding lookup table is accessed at step 106 and the appropriate value of $I_d$ for this table is obtained. The process flow from both steps 104 and 106 returns to step 100 for determination of whether there is still a user profile selection received and the nature of such selection to continue in the above described manner. Operation at steps 104 and 106 occurs after the selection in step 102 for a period sufficiently long to overcome transient effects in profile changeover. Thus, an appropriate delay for return of the process flow to step 100 may extend for a number of successive feedback samplings.

If no user profile selection input signal is present and the system has not been switched off, the controller determines at step 100 that the waveform profile is to be automatically selected. In this automatic mode, determination is made by the controller at step 108 of whether or not the system, in the high efficiency profile motor control scheme, has the capability of meeting the torque tracking requirements for the user requested torque input. Such determination can be made with reference to the value of the controller $V_i(t)$ from the output of block 70 of Fig. 3 that would be derived from values of the user requested torque input and the motor speed. The torque demands can be met if the derived control voltages do not exceed the voltage level of the power supply. If the derived level of this output does not exceed the power supply voltage as determined in step 108, the controller can apply the voltage required by the high efficiency motor control scheme for torque tracking. The controller, after any appropriate delay, will then access the profile memory to retrieve data from the high efficiency profile lookup table at step 110. If, instead, the derived voltage level is higher than the power supply voltage, determination is made at step 108 that the power supply capacity is exceeded. The controller, after any appropriate delay, will then access the profile memory to retrieve data from the high torque profile lookup table at step 112. The process flow from both steps 110 and 112 returns to step 100 to continue in the above described manner. The delays discussed above are appropriate if operation is to change from one operational profile mode to another.
While the automatic mode profile selection represented by step 108 can be performed by repeated calculation of a torque capacity threshold on a real time basis, calculations of voltage for various combinations of torque request and motor speed can be made in advance and linked with the appropriate profile in a lookup table in the profile memory. Fig. 5 is a curve that represents a boundary in such a lookup table between ranges for high efficiency profile mode selection and high torque profile mode selection for values of torque and speed in accordance with the illustrated example. The lookup table is formulated by making the above described controller voltage/supply voltage comparison for a multitude of speed/requested torque combinations. With the abscissa of the graph representing speed and the ordinate representing requested torque, the curve is asymptotic to both axes with speed/torque combinations above the curve being beyond the capacity of the system to track torque in the high efficiency profile operational mode. The controller upon accessing this lookup table will select the high efficiency profile mode for speed/torque combinations represented by points below the curve and select the high torque profile mode for the remaining points.

In this disclosure there is shown and described only preferred embodiments of the invention and but a few examples of its versatility. It is to be understood that the invention is capable of use in various other combinations and environments and is capable of changes or modifications within the scope of the inventive concept as expressed herein. For example, various other current waveform profiles may be utilized. The profile memory thus may store a plurality of profiles accessible by the controller in response to receipt of specific profile selection commands. Various lookup tables of varying complexities can be formulated for appropriate profile mode selection by the controller.
CLAIMS

What is claimed is:

1. A method for adaptive control of a multiphase motor comprising the steps of:
   identifying a plurality of motor control schemes for effecting respective different
   operating profiles;
   inputting a command;
   sensing a motor operational condition;
   selecting one of the motor control schemes in accordance with the input command
   and the sensed motor operational condition; and
   generating motor control signals to provide motor energizing current from a power
   supply for an operating profile corresponding to the selected motor control scheme.

2. A method as recited in claim 1, wherein the input signal represents a user
   request and is subject to variation, and the steps of sensing, selecting and generating are
   repeatedly performed throughout motor operation for adaptive control of the motor.

3. A method as recited in claim 2, wherein the input signal represents a torque
   request and the motor operational condition is motor speed.

4. A method as recited in claim 3, wherein the stored motor control schemes
   comprise an efficiency motor control scheme that provides a current waveform profile for
   relatively optimum operating efficiency and a high torque motor control scheme that
   provides a current waveform profile for relative high operating torque response.

5. A method as recited in claim 4, wherein the relatively optimum operating
   efficiency current waveform profile has a substantially sinusoidal waveshape and the
   relatively maximum torque response current waveform profile has a substantially
   rectangular waveshape.
6. A method as recited in claim 5, wherein the selecting step comprises:

determining whether a threshold is exceeded based on the torque request and the
motor speed; and

accessing a stored motor control scheme, the motor control scheme accessed being
dependent upon the determination made in the determining step.

7. A method as recited in claim 6, wherein the torque motor control scheme is
selected when the threshold is exceeded and the efficiency motor control scheme is
selected when the threshold is not exceeded.

8. A method as recited in claim 7, wherein the threshold is the power supply
voltage level and the determining step comprises deriving a control voltage required for
operation using the efficiency motor control scheme.

9. A method as recited in claim 7, the determining step comprises accessing a
lookup table having entries based on torque request and motor speed.

10. A method as recited in claim 1, wherein the motor stator comprises a plurality
of ferromagnetically autonomous electromagnets, each electromagnet having wound
thereon one of the phase windings.

11. A motor control system for a multiphase permanent magnet motor having a
stator with a plurality of phase windings, said system comprising: energization circuitry
for providing energization current to the motor stator windings from a power supply;
a controller coupled to the energization circuitry, the controller having a plurality
of inputs;

storage means coupled to the controller, the storage means having stored therein a
plurality of different motor control schemes for respectively effecting different stator
current waveform profiles; and
motor condition sensing means coupled to one input of the controller;
wherein the controller is dynamically responsive to the motor condition sensing means to access the motor control schemes from the storage means to energize the motor with corresponding current waveform profiles in accordance with criteria associated with the sensed motor condition.

12. A motor control system as recited in claim 11, wherein said stored motor control schemes comprise an efficiency motor control scheme that provides a current waveform profile for relatively optimum operating efficiency and a high torque motor control scheme that provides a current waveform profile for a relative high operating torque response.

13. A motor control system as recited in claim 12, wherein the relatively optimum operating efficiency current waveform profile has a substantially sinusoidal waveshape and the relatively maximum torque response current waveform profile has a substantially rectangular waveshape.

14. A motor control system as recited in claim 13, wherein another input to the controller is a torque request input for receiving a user signal; and
wherein a motor control scheme is accessed by the controller in accordance with a threshold dependent upon a signal received at the torque request input and a signal received from the motor condition sensing means.

15. A motor control system as recited in claim 14, wherein the high torque motor control scheme is accessed when the threshold is exceeded and the efficiency motor control scheme is accessed when the threshold is not exceeded.

16. A motor control system as recited in claim 14, wherein the motor condition sensing means comprises a motor speed sensor.
17. A motor control system as recited in claim 16, wherein the storage means comprises a lookup table having entries based on torque request and motor speed, each entry designating one of the motor control schemes.

18. A motor control system as recited in claim 11, wherein the motor stator comprises a plurality of ferromagnetically autonomous electromagnets, each electromagnet having wound thereon one of the phase windings.

19. A motor control system as recited in claim 11, wherein one of the controller inputs is a user select input for receiving a signal designating one of the stored motor control schemes, and
   wherein, in response to receipt of a signal at the user select input, a motor control scheme is accessed for implementation regardless of the motor condition sensing means.
\[ \theta_1(t) = \text{mod}\left\{ \text{mod}\left( \frac{\theta, 2\pi}{N_T} \right) + (i-1) \cdot \frac{2\pi}{N_S N_T} , \frac{2\pi}{N_T} \right\} \]

\[ \tau_d(t) \]

\[ \left( \frac{2 \cdot \tau_d}{N_S \cdot K_{ri}} \right) \sin(N_T \theta) \]

\[ \left( K_{vi} \omega \right) \sin(N_T \theta) \]

\[ \frac{2L_i}{N_S \cdot K_{ri}} \left( N_T \omega \right) \cdot \tau_d \cdot \cos(N_T \theta) \]

\[ L_i \frac{d}{dt} i_d + R_i i + E_i + k_s e_i \]

\[ V_i(t) \]

**Fig. 3**