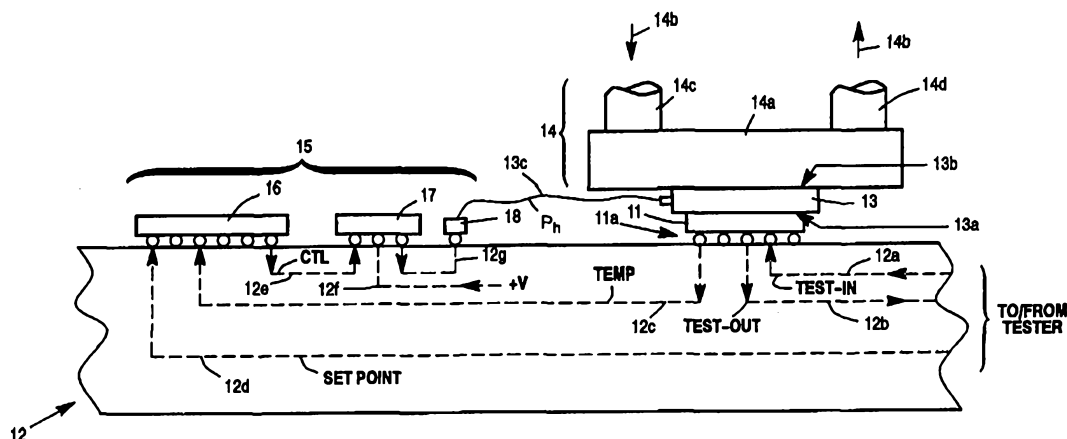


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(54) Title: TEMPERATURE CONTROL SYSTEM FOR AN ELECTRONIC DEVICE**(57) Abstract**

A temperature control system comprises: an electric heater (13) that has a first face (13a) which makes contact with an electronic device (11), and a second face (13b) which is opposite the first face; a heat sink (14), coupled to the second face of the heater, which absorbs heat from an electronic device (11) through the heater's second face; and a temperature sensor (12c), coupled to the electronic device which senses the device temperature T_d . A control circuit (15) is coupled to the device temperature sensor and to the heater; and it decreases the power to the heater when the sensed temperature of the electronic device is above the set point, and vice-versa. When the heater temperature T_h is less than T_d , then heat flows from the electronic device through the heater to the heat sink; and the rate of heat flow increases as $T_d - T_h$ increases. When T_h is more than T_d , then heat flows to the electronic device from the heater; and the rate of heat flow increases as $T_h - T_d$ increases.

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TITLE: TEMPERATURE CONTROL SYSTEM FOR AN ELECTRONIC DEVICE

BACKGROUND OF THE INVENTION:

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This invention relates to temperature control systems which maintain the temperature of an electronic device near a constant set point temperature while the device is being tested; and it relates to subassemblies which comprise key portions of such temperature control systems.

Two specific examples of electronic devices which have a need to be tested near a constant temperature are packaged integrated chips, or bare chips which are unpackaged. Any type of circuitry can be integrated into the chips, such as digital logic circuitry or memory circuitry or analog circuitry. Also, the circuitry in the chips can be comprised of any type of transistors, such as field effect transistors or bipolar transistors.

One reason for trying to keep the temperature of a chip constant while it is tested is that the speed with which the chip operates may be temperature dependent. For example, a chip comprised of complementary field effect transistors (CMOS transistors) typically increases its speed of operation by about 0.3% per °C drop in chip temperature.

A common practice in the chip industry is to mass produce a particular type of chip, and then speed sort them and sell the faster operating chips at a higher price. CMOS memory chips and CMOS microprocessor chips are processed in this fashion. However, in order to determine the speed of such chips properly, the temperature of each chip must be kept nearly constant while the speed test is performed.

Maintaining the chip temperature near a constant set point is quite simple if the instantaneous power dissipation of the chip is constant or varies in a small range while the speed test is being performed. In that case, it is only necessary to couple the chip through a fixed thermal resistance to a thermal mass which is at a fixed temperature. For example, if the maximum chip power variation is ten watts, and the coupling between the chip and the thermal mass is 0.2°C/watt, then the chip temperature will vary a maximum of 2°C.

But, if the instantaneous power dissipation of the chip varies up and down in a wide range while the speed test is being performed, then maintaining the chip temperature near a constant set point is very difficult. Each time the device power dissipation makes a big change, its temperature and its speed will also make a big change.

The above problem is particularly severe in CMOS chips because their instantaneous power dissipation increases as the number of CMOS transistors which are switching ON or OFF increases. During the speed test of a CMOS chip, the number of transistors which are switching is always changing; thus, the chip's power dissipation and temperature and speed are always changing. Also, the magnitude of these changes increases as more transistors get integrated into a single chip, because the number of transistors which are switching at any particular instant will vary from none to all of the transistors on the chip.

In the prior art, several temperature control systems for integrated circuit chips have been disclosed by the following U.S. patents:

5,420,521	5,297,621	5,104,661
5,315,240	5,205,132	5,125,656
5,309,090	5,172,049	4,734,872.

However, none of the temperature control systems which are disclosed in the above patents are capable of enabling speed tests to be run on chips which widely vary their power dissipation during the test.

5 This is because the above temperature control systems cannot react fast enough to compensate for such instantaneous power variations.

In the above patent `656 and `661 and `090 and `240, no means are provided for removing heat from the
10 chip; they merely include various heaters for adding heat to the chip. These control systems are suitable only for raising the chip's temperature to an elevated level, such as above 200°C, at which "burn-in" tests are performed. At any elevated temperature, the failure of weak or
15 faulty components in a chip is accelerated, and the "burn-in" test causes such components to fail after several hours.

In the above patents `872 and `132 and `621, the chip temperature is increased or lowered by directing
20 a temperature controlled gas jet at the chip or immersing the chip in a temperature controlled liquid. But, these control systems are limited by the speed at which the temperature of the gas jet or liquid can be increased or decreased.

25 In patent `521, a control system is disclosed in Fig. 7 which includes both a heater for heating the chip and a liquid cooled aluminum block for cooling the chip. However, the Fig. 7 control system still is not capable of enabling speed tests to be run on chips which
30 widely vary their power dissipation during the test, because it cannot react fast enough to compensate for such power variations. Why this deficiency occurs is explained herein in detail in conjunction with Figs. 8, 9, and 10.

Accordingly, a primary object of the inventions which are disclosed herein is to provide novel temperature control systems, and novel subassemblies for
5 such systems, which react quickly to large variations in power dissipation within an electronic device and thereby maintain the device temperature near a constant set point temperature while the device is being tested.

BRIEF SUMMARY OF THE INVENTION:

10 In accordance with the invention of Docket 550,577, a temperature control system which maintains the temperature of an electronic device near a set point comprises: an electric heater that has a first face which makes contact with the electronic device, and a second
15 face which is opposite the first face; a heat sink, coupled to the second face of the heater, which absorbs heat from the electronic device through the heater's second face; a temperature sensor, coupled to the electronic device which senses the device temperature T_d ;
20 and, a control circuit, coupled to the temperature sensor and to the heater, which decreases the power to the heater when the sensed temperature of the electronic device is above the set point, and vice-versa. When the heater temperature T_h is less than T_d , then heat flows
25 from the electronic device through the heater to the heat sink; and the rate of heat flow increases as $T_d - T_h$ increases. When T_h is more than T_d , then heat flows to the electronic device from the heater; and the rate of heat flow increases as $T_h - T_d$ increases. By electrically
30 controlling just the heater power, heat flow to/from the electronic device is quickly adjusted; and that in turn quickly regulates the device temperature.

BRIEF DESCRIPTION OF THE DRAWINGS:

Fig. 1 shows a temperature control system which constitutes a preferred embodiment of the invention
5 having Docket No. 550,577.

Fig. 2 is a schematic diagram which helps explain the operation of the Fig. 1 temperature control system.

Fig. 3 replicates the schematic diagram of
10 Fig. 2 and depicts the operation of the Fig. 1 control system for the case where the power dissipation in the electronic device steps up.

Fig. 4 replicates the schematic diagram of
Fig. 2 and depicts the operation of the Fig. 1 control
15 system for the case where the power dissipation in the electronic device steps down.

Fig. 5 is a set of differential equations which precisely defines the behavior of the Fig. 1 control system.

20 Fig. 6 shows how several parameters vary in the Fig. 1 control system when that system is simulated via the Fig. 5 equations.

Fig. 7 shows how several additional parameters vary in the Fig. 1 control system when that system is
25 simulated via the Fig. 5 equations.

Fig. 8 is a schematic diagram, similar to the schematic diagram of Fig. 2, which helps explain the operation of a prior art temperature control system that is described in U.S. Patent 5,420,521.

30 Fig. 9 is a set of differential equations which precisely defines the behavior of the Fig. 8 temperature control system and help explain a distinguishing feature between the operation of the control systems of Figs. 1 and 8.

Fig. 10 is a set of differential equations which helps explain another distinguishing feature between the operation of the control systems of Figs. 1 and 8.

5 Fig. 11 is a set of differential equations which forms the basis of the temperature control system having Docket 550,578.

 Fig. 12 shows a preferred embodiment of the temperature control system which is based on the Fig. 11
10 equations.

 Fig. 13 is a set of curves which is obtained by a simulation of the Fig. 12 temperature control system.

 Fig. 14 is another set of curves which is obtained by a second simulation of the Fig. 12
15 temperature control system in which a particular parameter is injected with an error.

 Fig. 15 lists the steps of a process by which the error in the parameter of Fig. 14 can be detected and corrected.

20 Fig. 16 illustrates how the Fig. 15 process is carried out.

 Fig. 17 shows an initial stage of fabricating a subassembly which constitutes the invention having Docket No. 550,579.

25 Fig. 18 shows how the subassembly of Fig. 17 is completed and incorporated into the temperature control systems of Figs. 1 and 12.

 Fig. 19 shows how a film on an electronic device in the subassembly of Fig. 18 is removed.

30 Fig. 20 shows a modification for a heater component within the temperature control systems of Figs. 1 and 12.

 Fig. 21 shows a modification for the subassembly of Fig. 18.

DETAILED DESCRIPTION:

In Fig. 1, reference numeral 11 identifies an electronic device which is to be tested while its temperature is maintained near a constant set point. Two
5 specific examples of this electronic device 11 are an integrated chip which is packaged in ceramic or plastic, or a bare chip which is unpackaged. Any type of circuitry which needs to be tested can be integrated into the electronic device 11, such as digital logic circuitry or
10 memory circuitry or analog circuitry; and that circuitry can be comprised of any type of transistors, such as N-channel field effect transistors or P-channel field effect transistors or complimentary field effect transistors or bi-polar transistors.

15 Included on the electronic device 11 are dozens of input/output terminals 11a; but for simplicity in Fig. 1, only a few of those terminals are shown. These terminals 11a are coupled to three sets of signal lines 12a, 12b, and 12c. On the signal lines 12a, input signals
20 "TEST-IN" are received which test the electronic device 11; and on the signal lines 12b, output signals "TEST-OUT" are sent from the electronic device 11 as a response to the test. Conventional test equipment (not shown) is coupled to the signal lines 12a and 12b to
25 respectively send and receive the signals on those lines. On the signal lines 12c, output signals "TEMP" are sent from the electronic device 11 which indicate its temperature. These TEMP signals originate from a temperature sensor, such as a temperature sensitive
30 resistor or diode, which is integrated into the electronic device.

All of the remaining components 12-17 in Fig. 1 constitute a temperature control system which keeps the temperature of the electronic device 11 near the set

point while that device is being tested. Each component in this control system is described below in TABLE 1.

TABLE 1

COMPONENT	DESCRIPTION
12 ...	Component 12 is a printed circuit board which physically holds the components 11, 16, 17 and 18. Also, the printed circuit board 12 contains several sets of signal and power lines 12a-12g.
13 ...	Component 13 is a thin, flat electric heater which has two major faces 13a and 13b that lie against components 11 and 14 respectively. A variable amount of electrical power P_h is supplied to the heater 13 via two wires 13c, and that power is dissipated as heat within the heater. One embodiment of the heater 13 is comprised of aluminum nitride ceramic in which electrical resistors (not shown) are uniformly integrated for converting the power from the conductors 13c to heat.
14 ...	Component 14 is a liquid cooled heat sink that has a hollow base 14a in which cooling fins (not shown) are disposed. A liquid coolant 14b enters the base 14a from a tube 14c, and that liquid coolant exits the base via another tube 14d. This coolant 14b is circulated through the base 14a at a constant flow rate by a pump (not shown) and held at a constant temperature T_L .
15 ...	Component 15 is an electronic control circuit which sends the variable amount of electrical power P_h to the electric heater 13. This control

circuit 15 consists of a power regulator 16 and a variable power supply 17.

16 ... Component 16 is a power regulator which is coupled to three sets of signal lines 12c, 12d, and 12e. The TEMP signals which indicate the present temperature T_d of the electronic device 11 are received on the signal lines 12c, and SET-POINT signals which indicate the set point temperature for the device 11 are received on the signal lines 12d. Based on those two temperatures and their rate of change, power regulator 16 generates control signals CTL on the signal lines 12e which indicate the amount of power that should be sent to the heater 13 such that the temperature of the device 11 stays near the set point.

17 ... Component 17 is a variable power supply which is coupled to the signal lines 12e and two sets of power lines 12f and 12g. On the signal lines 12e, the control signals CTL from the power regulator 16 are received, and on the power lines 12f, a supply voltage +V and ground are received. In response to the CTL signals, the power supply 17 sends the variable amount of power P_h on the power lines 12g as a portion of the power which is available from the supply voltage +V.

18 ... Component 18 is a connector which inter couples the heater wires 13c to the variable power supply.

In operation, the electronic device 11 varies in its power dissipation as it is being tested by the TEST-IN signals which it receives on the signal lines 12a. This variation in power

occurs because the transistors in the electronic device 11 turn on and off, and thereby vary in their power dissipation, in response to the TEST-IN signals. Consequently, the temperature of the electronic device 11 tends to increase as its power dissipation increases; and vice-versa.

Within the electronic device 11, a temperature sensor generates the TEMP signals on the signal lines 12c which indicate the present temperature of the device. Those TEMP signals are sent to the power regulator 16 where they are compared with the SET-POINT signals on the signal lines 12d. If the temperature of the device 11 is less than the set point temperature, then the regulator 16 generates the control signals CTL such that the variable power P_h to the heater 13 is increased. Conversely, if the temperature of the device 11 is more than the set point temperature, then the regulator 16 generates the control signals CTL such that the variable power PWR to the heater 13 is decreased.

Turning now to Figs. 2, 3 and 4, the operation of the Fig. 1 control system will be described in further detail. Those figures are schematic diagrams of the Fig. 1 control system wherein the electronic device 11 is coupled to the electric heater 13 through a thermal resistance θ_{d-h} ; the heater 13 is coupled to the heat sink base 14a through a thermal resistance θ_{h-s} ; and the heat sink base is coupled to the liquid coolant through a thermal resistance θ_{s-L} . Also in the Figs. 2-4, the electronic device 11 receives and dissipates a variable amount of power P_d , and the electric heater receives and

dissipates a variable amount of power P_h .
Further in the

Figs. 2-4, the electronic device 11 has a temperature T_d and a thermal mass M_d ; the heater 13 has a temperature T_h and a thermal mass M_h ; and the heat sink base 14a has a temperature T_s and a thermal mass M_s .

5 Preferably, the heater's thermal mass M_h is made as small as is practical. This constraint, as will be shown below, helps to improve the speed with which the heater 13 can raise or lower its temperature, and thereby maintain the temperature T_d of the electronic device near
10 the set point. One suitable upper limit for the thermal mass M_h is that it be no larger than three times the thermal mass M_d of the electronic device; and preferably, M_h is less than M_d .

 Also preferably, the heater's thermal
15 resistance θ_{d-h} is made as small as practical. This constraint, as will be shown below, helps to increase the amount of heat that is transferred from the heater 13 to the electronic device 11, relative to the amount of heat that is transferred from the heater 13 to the heat sink
20 14. A suitable upper limit for the thermal resistance θ_{d-h} is that it be no more than three times the thermal resistance θ_{h-s} between the heater and the heat sink base; and preferably, θ_{d-h} is less than θ_{h-s} .

 Under steady state conditions, heat flows from
25 the device 11 to the liquid coolant along a thermal path 21 as shown in Fig. 2, and heat flows from the heater 13 to the liquid coolant along a thermal path 22 as shown in Fig. 2. Also under steady state conditions, the device temperature equals the set point temperature, and the
30 heater temperature equals the device temperature minus

$P_d \theta_{d-h}$. If the device power dissipation is zero, then the heat flow on path 21 is zero and T_d equals T_h .

Suppose now that the electronic device 11 increases its power dissipation P_d and thereby causes its temperature T_d to rise above the set point. This is shown in Fig. 3 as occurring at time instants t_1 and t_2 . In response to the rise in the device temperature T_d above the set point, the power P_h to the heater 13 is reduced by the regulator 16; and this is shown in Fig. 3 as occurring at time t_3 . Heater 13 has a low thermal mass M_h ; and consequently, the heater's temperature T_h drops rapidly when the heater's power is reduced, as indicated at time t_4 .

As the heater temperature is reduced, the amount of heat which is transferred from the device 11 along the thermal path 21 to the liquid coolant increases. Thus, the device temperature T_d starts to cool down, as indicated at time t_5 . As the device temperature T_d approaches the set point, the heater power P_h ramps up as indicated at time t_6 . Thus the heater temperature T_h rises and thereby reduces the flow of heat from the device 11 along the thermal path 21. By raising the heater power P_h to the proper level, a steady state returns where the heat flow is as shown in Fig. 2 and the device temperature T_d is at the set point.

Conversely, suppose that the electronic device 11 decreases its power dissipation P_d and thereby causes its temperature T_d to drop below the set point. This is shown in Fig. 4 as occurring at time instants t_1 and t_2 . In response to the drop in the device temperature T_d below the set point, the power P_h to the heater 13 is increased by the regulator 16; and this is shown in Fig. 4 as occurring at time t_3 . Heater 13 has a low thermal mass M_h ; and consequently, the heater's

temperature T_h rises rapidly when the heater's power is increased as indicated at time t_4 .

If the heater temperature T_h rises so high that it exceeds the device temperature T_d , one portion of the heat from the heater travels along the thermal path 22a in Fig. 4 to the electronic device 11, and another portion of the heat from the heater travels along the thermal path 22b to the liquid coolant. Also, the amount of heat which travels on path 22a to the electronic device 11 increases as θ_{d-h} decreases relative to θ_{h-s} .

In response to the rise in heater temperature at time t_4 , the device temperature T_d starts to rise, as indicated at time t_5 . As the device temperature T_d approaches the set point, the heater power P_h ramps down as indicated at time t_6 . Thus the heater temperature T_h ramps down and thereby reduces the flow of heat from the heater to the device 11 along the thermal path 22a. By dropping the heater power P_h to the proper level, a steady state returns where the heat flow is as shown in Fig. 2 and the device temperature T_d is at the set point temperature.

From the above description of Figs. 2-4, it is seen that the heater 13 controls the amount of heat that flows on the paths 21 and 22a. When T_h equals T_d , no heat flows on path 21 or 22a. When T_h is less than T_d , then heat flows on path 21 from the electronic device through the heater 13 to the heat sink; and the rate of heat flow increases as $T_d - T_h$ increases. When T_h is more than T_d , then heat flows on path 22a from the heater 13 to the electronic device; and the rate of heat flow increases as $T_h - T_d$ increases. By controlling the heater power P_h , heat flow to/from the electronic device is regulated; and that in turn regulates the device temperature.

To further explain the speed with which the Fig. 1 control system operates, reference should now be made to Figs. 5, 6 and 7. In Fig. 5, three differential equations are provided which are labeled equation 1,
5 equation 2, and equation 3. These equations are based on the principle that the sum of the thermal power which is received by a body is equal to the thermal mass of the body times the rate of change of the body's temperature.

Equation 1 contains three terms which are
10 identified by reference numerals 31a, 31b and 31c. Term 31a is the power which the electronic device 11 receives and dissipates in response to the TEST-IN signals; term 31b is the power which is transferred to the electronic device 11 through the thermal resistance θ_{d-h} ; and term
15 31c is the thermal mass of the electronic device 11 times the rate of change of its temperature.

Similarly, equation 2 contains four terms which are identified by reference numerals 32a, 32b, 32c, and 32d. Term 32a is the power which is supplied to the
20 electric heater 13 from the variable power supply 17; term 32b is the power which is transferred to the electric heater 13 through the thermal resistance θ_{h-s} ; term 32c is the power which is transferred to the heater 13 through the thermal resistance θ_{d-h} ; and term 32d is
25 the thermal mass of the heater times the rate of change of the heater's temperature.

Likewise, equation 3 contains three terms which are identified by reference numerals 33a, 33b, and 33c. Term 33a is the power which is transferred to the heat
30 sink base 14a through the thermal resistance θ_{s-L} ; term 33b is the power which is transferred to the heat sink base through the thermal resistance θ_{h-s} ; and term 33c is the mass of the heat sink base times the rate of change of its temperature.

Using the three equations of Fig. 5, a simulation on a digital computer was performed; and the results of that simulation are shown in Figs. 6 and 7. In performing this simulation, the various parameters in the Fig. 5 equations had the values which are shown below in TABLE 2.

TABLE 2

$\theta_{d-h} = 0.1^{\circ}\text{C per watt}$	$M_d = 0.478 \text{ joules per } ^{\circ}\text{C}$
$\theta_{h-s} = 0.05^{\circ}\text{C per watt}$	$M_h = 0.80 \text{ joules per } ^{\circ}\text{C}$
$\theta_{s-L} = 0.1^{\circ}\text{C per watt}$	$M_s = 5.60 \text{ joules per } ^{\circ}\text{C}$

MAXIMUM $P_h = 500 \text{ watts}$

$T_L = 40^{\circ}\text{C below set point}$

Also, the simulation assumed that the electronic device 11 was initially at the set point temperature and was dissipating no power. Thereafter, at a time equal to two seconds, the power dissipation of the electronic device 11 was stepped up in response to the TEST-IN signals, to a power of 100 watts. This power dissipation remained constant in the electronic device 11 for three seconds at which time the power dissipation in the device 11 went back to zero.

By changing the power dissipation of the electronic device 11 with a pair of one hundred watt steps as described above, the ability of the Fig. 1 control system to maintain the device temperature near the set point is tested under a worst case condition. Also, the thermal mass M_d in TABLE 2 is the thermal mass of a bare integrated circuit chip; and thus, it constitutes another worst case test condition because its temperature will change temperature faster than the temperature of a packaged chip (which inherently has a larger thermal mass).

Curve 41 in Fig. 6 shows how the power dissipation in the electronic device 11 changes with time, as described above. Also in Figs. 6 and 7, curve 42 shows how the temperature of the electronic device 11 changes with time; curve 43 shows how the heater power changes with time; curve 44 shows how the heater's temperature changes with time; and curve 45 shows how the temperature of the heat sink base changes with time.

Inspection of curve 42, at time equal to two seconds, shows that the maximum deviation in the temperature of the electronic device 11 is approximately 4°C due to the step increase in power dissipation within that device. After reaching that maximum temperature deviation, the temperature of the electronic device 11 then ramps down and returns to the set point within about 0.4 seconds.

Similarly, inspection of curve 42, at time equal to five seconds, shows that the maximum deviation in the temperature of the electronic device 11 is approximately 3.6°C due to the negative step in power dissipation within that device. After reaching that maximum temperature deviation, the temperature of the electronic device 11 ramps up and returns to the set point within 0.3 seconds.

One primary feature of the above-described control system is the speed with which the control system corrects a deviation in the device temperature from the set point, and thus keeps the maximum deviation relatively small. How this speed of operation is achieved can be seen from the equations in Fig. 5 as follows.

Each deviation in the device temperature from the set point is corrected by a change in the heater power P_h as given by term 32a in equation 2. When term

32a changes, the temperature of the heater changes as indicated by term 32d in equation 2. A change in heater temperature then causes term 31b to change in equation 1. Then the change in term 31b causes the device temperature
5 to change as indicated by term 31c in equation 1. In Fig. 5, the above chain of events is indicated by a dashed line 51.

By comparison, in the prior art patent 5,420,521 the chain of events which occur between a
10 change in the heater power and a resulting change in the device temperature is much longer; and this is shown by Figs. 8 and 9. Fig. 8 is a schematic diagram for the control system which occurs in Figs. 7 and 8 of the prior art patent '521. In this schematic diagram, the
15 reference numerals 140, 116, 124, 126, and 82 are the same reference numerals which occur in Fig. 7 of patent '521.

Component 140 is a heater block which contacts the electronic device 11 that is to be cooled; component
20 116 is an aluminum block which contacts the heater block; component 82 is a coolant tube which carries a liquid coolant at a temperature T_L ; component 124 is a heater assembly which contacts the aluminum block 116; and component 126 is a heating element in the heater
25 assembly. Each component in Fig. 8 has a respective thermal mass M and a respective temperature T as shown in the figure. Also, all of the components in Fig. 8 are intercoupled through respective thermal resistance's θ as shown.

30 From the schematic diagram of Fig. 8, a set of six differential equations 11-15 can be derived as shown in Fig. 9. These Fig. 9 equations are similar to the equations of Fig. 5 is that they are based on the principle that the sum of the thermal power which is

received by a body is equal to the thermal mass of the body times the rate of change of the body's temperature.

When the heater power P_h in equation 15 changes, that causes the temperature T_3 to change as indicated by term 61 in equation 15. A change in the temperature T_3 then causes term 62 to change in equation 14. Then the change in term 62 causes the temperature T_2 to change as indicated by term 63 of equation 14. A change in the temperature T_2 then causes term 64 to change equation 13. Then the change in term 64 causes the temperature T_1 to change as indicated by term 65 of equation 13. A change in the temperature T_1 then causes term 66 to change in equation 12. Then the change in term 66 causes the temperature T_h to change as indicated by term 67 of equation 12. A change in the temperature T_h then causes term 68 to change in equation 11. Then the change in term 68 causes the device temperature T_d to change as indicated by term 69 of equation 11.

In Fig. 9, the above chain of events is indicated by reference numeral 51'. In this chain of events 51', each of the temperatures T_3 , T_2 , T_1 , T_h , and T_d must sequentially change in order to regulate the device temperature by varying the heater power P_h . Consequently, varying the heater power only indirectly changes the device temperature. By comparison, in the chain of events 51 in Fig. 5, only the temperatures T_h and T_d must sequentially change in order to regulate the device temperature by varying the heater power. Consequently, varying the heater power in the Fig. 1 control system affects the device temperature much more directly, and thus more quickly, than it does in patent '521.

Another reason why varying the heater power in the Fig. 1 control system effects the device temperature

more quickly than it does in patent `521 can be seen from equations 20 and 21 in Fig. 10. Equation 20 is formed by adding equations 2 and 3 of Fig. 5 together, and by rearranging the resulting terms such that only term 32d
5 lies to the left of the equal sign. Similarly, equation 21 is obtained by adding equations 12-16 of Fig. 9 together, and by rearranging the resulting terms such that only term 67 lies to the left of the equal sign.

10 In term 32d of equation 20, the thermal mass M_h is the thermal mass of the Fig. 1 heater 13 which is in direct contact with the electronic device 11 that is to be cooled. Similarly, in term 67 of equation 21, the thermal mass M_h is the thermal mass of the Fig. 8 heater
15 block 140 which is in direct contact with the electronic device 11 that is to be cooled. Thus, in order to regulate the temperature of the electronic device 11 quickly, it is critical that the temperature T_h of the mass M_h can be changed quickly by varying the heater
20 power P_h .

When the heater power P_h increases in equation 20, the effect of that power increase on the temperature T_h of the mass M_h is diminished by the negative term 33c which occurs in the right-hand side of
25 equation 20. By comparison, when the heater power P_h in equation 21 is increased, the effect of that power increase on the temperature T_h of the mass M_h is diminished by four negative terms 60, 61, 63, and 65 in the right-hand side of equation 21.

30 Term 60 in equation 21 is similar to term 33c in equation 20; and thus their negative effects on changing the temperature T_h by varying the heater power P_h are similar. But the remaining negative terms 61, 63, and 65 in equation 21 are completely eliminated in

equation 20. Thus, changing the heater power P_h has a much larger effect on the rate of change of the temperature T_h in the Fig. 1 control system than it does in patent `521.

5 Turning now to Figs. 11 and 12, a second temperature control system will be described. A major point of distinction in this second temperature control system over the Fig. 1 system is that it regulates the temperature of an electronic device 11' which does not
10 include a temperature sensor. Consequently, the electronic device 11' is unable to generate the temperature signals TEMP which indicate how its temperature is varying in response to the TEST-IN signals on the signal lines 12a. This distinguishing point is
15 very significant since many, if not most, integrated circuit chips in the prior art do not include a temperature sensor.

Equation 2, as previously described, provides the principle on which this second temperature control
20 system is based; and for convenience equation 2 is reproduced in Fig. 11. By algebraically rearranging the terms of equation 2, another equation 31 is obtained wherein the device temperature T_d is expressed in terms of several parameters which are on the right of the
25 equals sign.

In equation 31, the parameters T_h , \dot{T}_h , P_h , and T_s vary with time as each particular device 11' is being tested; and the remaining parameters θ_{d-h} , M_h , and θ_{h-s} are essentially constant. Thus, from equation 31, the device
30 temperature T_d can be estimated by measuring the constant parameters θ_{d-h} , M_h , and θ_{h-s} before the device 11' is tested and by sensing the parameters T_h , \dot{T}_h , P_h , and T_s while the device 11 is tested. Then, if the estimated

device temperature is above the set point, the heater temperature is decreased; and vice-versa.

Fig. 12 shows all of the components of a temperature control system which operates on the above principle. In Fig. 12, the heater 13' is the same as the heater 13 of Fig. 1, except that it includes a sensor 13d which senses the heater temperature T_h . Similarly in Fig. 12, the heat sink 14' is the same as heat sink 14 of Fig. 1 except that it includes a sensor 14f which senses the heat sink temperature T_s .

Component 12' in Fig. 12 is the same as the printed circuit board 12 in Fig. 1, except that it includes five different sets of signal lines 12c', 12e', 12h, 12i and 12j, plus two connectors 71 and 72. Signal lines 12h carry the sensed heater temperature T_h ; signal lines 12i carry the sensed heat sink temperature T_s ; signal lines 12j carry the measured parameters θ_{d-h} , M_h , and θ_{h-s} ; signal lines 12e' carry the control signals CTL which indicate the heater power P_h ; and signal lines 12c' carry the estimated device temperature T_d .

Component 73 in Fig. 12 estimates the device temperature based on equation 31 and all of the parameters which it receives on the signal lines 12e', 12i, 12j and 12h. Component 73 can be a digital microprocessor chip which digitally calculates the device temperature from equation 31; or alternatively, component 71 can be an analog circuit which calculates the device temperature from equation 31 in an analog fashion.

Components 16, 17 and 18 are identical to those components in Fig. 1 which have the same reference numeral. Components 16, 17 and 18 together constitute a control circuit 15' which receives an estimated device temperature on the signal lines 12c', and receives the set point temperature on the signals lines 12d. When the

estimated device temperature exceeds the set point temperature, the control circuit 15' reduces the power P_h to the heater 13'; and vice-versa.

A digital simulation of the Fig. 12 temperature control system is shown in Fig. 13. In this simulation, the electronic device 11' had a step increase in power when the simulation time equaled three seconds, and it had a step decrease in power when the simulation time equaled six seconds. Curve 91 in Fig. 13, (which is comprised of a series of small circles), shows how the estimated device temperature varied with time during the simulation; and curve 92 in Fig. 13, (which is a solid line), shows how the actual device temperature would be sensed if the device had a temperature sensor.

When the above simulation of Fig. 13 was performed, it was assumed that all of the parameters on the right-hand side of equation 31 were sensed or measured with no errors. This is a valid assumption for θ_{h-s} and M_h because they are essentially fixed and can be measured accurately. Also T_h and T_s can be made as accurate as desired by using proper temperature sensors, and P_h is known from the control signal CTL. But in a mass production environment, where many electronic devices are tested one at a time, the parameter θ_{d-h} may vary within some range from an average value due to microscopic variations in physical contact between the heater 13' and each device 11'.

If an average value of θ_{d-h} is sent to the estimator component 73, but the actual value of θ_{d-h} differs from that average value, then an error in the estimated device temperature will occur. As a result, an offset will occur between the actual temperature of the device 11' as regulated by the Fig. 12 control system and the set point temperature. This is evident from Fig. 14

where a curve 91', (which is comprised of a series of small circles), shows how the estimated device temperature varies with time when the actual θ_{d-h} is 20% smaller than the average θ_{d-h} ; and a curve 92', (which is
5 a solid line), shows how the actual device temperature varied with time.

When certain types of devices 11' are tested, some offset between the actual device temperature and set point temperature may be acceptable, in which case it is
10 sufficient to send the average value of θ_{d-h} to the estimator 73. If, however, the amount of offset is not acceptable, then the offset can be reduced by measuring the actual θ_{d-h} for each device that is tested.

One preferred method by which the actual θ_{d-h}
15 can be measured for each device is shown in Figs. 15 and 16. This method, which includes steps 1-3, is based on equation 2, which again is reproduced in Fig. 15. If θ_{d-h} is smaller than its average value, then $T_d - T_h$ will decrease. Thus term 32c will decrease, and so the rate
20 of change of heater temperature in term 32d will decrease. Conversely, if θ_{d-h} is larger than its average value, then the rate of change of heater temperature in term 32d will increase.

In step 1 of the Fig. 15 process, the power
25 dissipation of the device 11' is set to zero. Then in step 2, the heater power P_h is varied in some predetermined fashion. Suitably, the heater power is stepped up, or stepped down, or varied as a sine wave. Next in step 3, the effect which varying the heater power
30 has on \dot{T}_h is sensed and correlated to θ_{d-h} .

If \dot{T}_h varies faster than average, then the actual θ_{d-h} is larger than its average value; whereas if \dot{T}_h varies slower than average, then θ_{d-h} is smaller than its average value. This is indicated in Fig. 16 by curves

93, 94, and 95. Curve 93 was obtained by setting θ_{d-h} equal to its average value, raising P_h to the set point and waiting until the device temperature reached the set point, and then dropping P_h to zero. Curves 94 and 95 were similarly obtained but the actual θ_{d-h} deviated from the average θ_{d-h} by 20%.

To measure the actual value of θ_{d-h} for any deviation from the average, the heater temperature is sensed at some time after P_h drops to zero, such as at time equal to 0.04 seconds. Then by extrapolation from the curves 93, 94 and 95, the actual θ_{d-h} is obtained.

Next, with reference to Figs. 17-19, a preferred process for minimizing the thermal resistance θ_{d-h} in the temperature control systems of Figs. 1 and 12 will be described. This process is significant because as θ_{d-h} is reduced, the amount of heat that can be transferred from the heater to the electronic device increases relative to the amount of heat that is transferred from the heater to the heat sink. Such heat transfers occur when the heater temperature exceeds the device temperature, as was explained previously in conjunction with Fig. 4.

If a dry contact is made between the electronic device 11 (or 11') and the heater 13 (or 13'), and the contacting surfaces are very flat (i.e., less than 5 μ m deviation from a perfect plane), then the typical resistivity which can be obtained between the contacting surfaces is about 1.3°C cm²/watt. This resistivity will be larger if the contacting surfaces are less flat. Placing a thermal pad between two non-flat surfaces reduces the resistivity between them, but a practical resistivity which can be obtained through a thermal pad is about 1.0°C cm²/watt. By comparison, with the method of Figs. 17-19, the resistivity from the electronic

device to the heater has been reduced to $0.07^{\circ}\text{C cm}^2/\text{watt}$. This is an improvement of $1.0 \div 0.07$, which is more than 1000%.

Initially in the process of Figs. 17-19, a drop
5 of a particular type of liquid 101 is dispensed onto the surface of the electronic device 11 (or 11') which is to make contact with the heater 13 (or 13'). This step is shown in Fig. 17. Two limiting characteristics of the liquid 101 are - a) that it is a good thermal conductor
10 relative to air, and b) that it evaporates, without leaving any residue, at a temperature which is too low to damage the electronic device. Preferably, the liquid 101 evaporates at a temperature of less than two hundred degrees centigrade.

15 One preferred embodiment of the liquid 101 is water. A second embodiment of the liquid 101 is a mixture of water and a volatile material (e.g. ethylene glycol) which elevates the boiling point of the mixture above the boiling point of water. A third embodiment of
20 the liquid 101 is a mixture of water and a volatile material (e.g. methanol) which suppresses the freezing point of the mixture below the freezing point of water. A fourth embodiment of the liquid 101 is a mixture of water and a volatile material (e.g. ethanol) which
25 enhances the wettability of the mixture above the wettability of water.

After the liquid 101 is dispensed on the electronic device as described above, surface 13a of the heater is pressed against that liquid. This step is
30 shown in Fig. 18. As a result, some of the originally dispensed liquid 101 is squeezed out, and the remaining liquid 101' fills the microscopic gaps which inherently exist between the electronic device and the heater. These gaps occur because surface 13a of the heater and

the mating surface of the electronic device are not perfectly flat.

Due to the presence of the liquid 101', the thermal resistivity from the heater 13 (or 13') to the electronic device 11 (of 11') is substantially reduced over that which it would otherwise be without the liquid. When the liquid 101' is water and the flatness of the mating surfaces is less than $5\mu\text{m}$, then the above-described resistivity of about $0.07^\circ\text{C cm}^2/\text{watt}$ is obtained.

One way to reduce the size of the gaps between the mating surfaces of the electronic device and the heater is to increase the pressure with which those two surfaces are pushed together; and this in turn will tend to decrease θ_{d-h} . But, if too much pressure is applied, the electronic device or an interconnect to it can crack. Thus, for the case where the electronic device is a bare integrated circuit die, the pressure between it and the heater preferably is less than 10 psi.

Fig. 18 also shows that the mating surfaces between the heater and the heat sink are not perfectly flat. But those two surfaces preferably are joined in a permanent fashion, for example by a layer 102 of thermally conductive epoxy. A suitable thickness for the layer 102 is $50\mu\text{m}$ - $80\mu\text{m}$.

While the electronic device and the heater and the heat sink are all coupled together as shown in Fig. 18, the electronic device is tested and its temperature is kept near the set point by the control systems of Figs. 1 or 12. Thereafter, when the test is complete, the electronic device is separated from the heater. Then the liquid 101' is removed from the electronic device by evaporation; and this is indicated by the arrows 103 in Fig. 19.

The total time which it takes to evaporate all of the liquid 101' is proportional to the area of the liquid 101' which is exposed. If the electronic device is square with a dimension L on each side, then the exposed area in Fig. 19 is L^2 . By comparison, when the electronic device is being tested in the Fig. 18 structure, the exposed area of the liquid 101' is greatly reduced to $(4L)(5\mu\text{m})$.

If the electronic device is a typical integrated circuit chip, then L is less than one-half inch. In that case, the liquid 101' evaporates more than 500 times slower in the Fig. 18 structure than it does in Fig. 19. Thus, during short tests (i.e., less than one hour) no significant amount of liquid 101' will be lost from the Fig. 18 structure due to evaporation.

One feature of the above process is that after the liquid 101' is evaporated, no residue remains on the electronic device. Consequently, the electronic device can be put into an end product without any additional clean-up steps.

Another feature of the above process is that it is very easy and inexpensive to perform. Only one step is needed to dispense a drop of the liquid 101 on the electronic device as shown in Fig. 17, and only one step is needed to push the heater against the electronic device as shown in Fig. 18.

Three separate, but related, inventions have now been described in detail in conjunction with Figs. 1-10, Figs. 11-16, and Figs. 17-19 respectively. In addition, various changes and modifications can be made to the details of these Figs. without departing from the nature and spirit of the inventions.

For example, Fig. 20 shows a modification wherein the temperature of the electronic device 11 (or

11') is kept near the set point by the use of a laser. This Fig. 20 modification can be incorporated into the temperature control systems of the previously described Figs. 1, 12, and 18.

5 In Fig. 20, reference number 13" identifies a modified electric heater; and reference number 14" identifies a modified heat sink. Included in the modified heater 13" is a laser 13a" which generates an infrared laser beam 13b" , and a thin, flat piece of
10 infrared laser beam absorbing material 13c" . This material 13c" has two major faces 13d" and 13e" which respectively lie against the electronic component 11 (or 11') and the heat sink 14" .

 The modified heat sink 14" is the same as the
15 previously described heat sinks of Figs. 1 or 12, except that the base 14a" includes two windows 14e" and 14f" , and no cooling fins are disposed in the base between the windows. Window 14e" is transparent to the infrared laser beam 13b" and is thermally conductive. Window
20 14f" also is transparent to the laser beam 13b" , but it need not be thermally conductive. Suitably, the windows are made of monocrystalline silicon.

 In operation, the temperature of the electronic device 11 (or 11') is either sensed directly as per the
25 Fig. 1 control system, or estimated as per the Fig. 12 control system. Then, if the device temperature T_d is above the set point, the power P_h to the laser on lines 13f" is reduced by the control circuit 15 of Fig. 1, or by the control circuit 15' of Fig. 12. Conversely if the
30 device temperature is below the set point, the power P_h to the laser is increased.

 As another modification, the enhanced thermal interface of Fig. 18 can be incorporated into a temperature control system other than the one of Figs. 1

and 12. An example of this modification is shown in Fig. 21. In the Fig. 1 control system the heater 13 (or 13' or 13'') and the corresponding circuitry for controlling the heater power P_h is eliminated; and the
5 heat sink 14 (or 14') is pressed against the electronic device with the liquid film 101' disposed there between.

With the Fig. 21 modification, the temperature of the electronic device 11 (or 11') will vary substantially more from the set point than it does when
10 the heater 13 (or 13' or 13'') is present. But on the other hand, the temperature of the electronic device will vary substantially less with the Fig. 21 modification than the variation which occurs if the liquid film 101' is absent. For particular tests on particular electronic
15 devices, such as chips with a small variation in power dissipation, the amount of temperature regulation which is achieved with the Fig. 21 modification will be sufficient; and all costs associated with the heater and its control circuits are eliminated.

20 As another modification, the temperature control system of Fig. 1 can be changed to an open loop control system in which the electronic device 11 does not send any TEMP signals to the power regulator 16. With this modification, the tester sequentially sends a series
25 of predicted temperatures on the signals lines 12d to the power regulator 16 for each particular TEST-IN signal which it sends to the electronic device 11. In response, the power regulator 16 compares the predicted temperature of electronic device 11 with the set point temperature.
30 If the predicted temperature is less than the set point temperature, then the regulator 16 generates the control signals CTL such that the variable power P_h to the heater 13 is increased. Conversely, if the predicted temperature is more than the set point temperature, then

the regulator 16 generates the control signals CTL such that the variable power P_h to the heater 13 is decreased.

As still another modification, the electric heater 13" of Fig. 20 can be changed such that the
5 infrared laser beam absorbing material 13c" is eliminated. With this modification, the window 14e" lies against the electronic device 11 (or 11'); and the laser beam 13b" is absorbed directly by the electronic device.

10 As another modification, the Fig. 1 temperature control system can be changed such that the TEMP signals on the signal lines 12c originate from a temperature sensor that is attached to the external surface of the electronic device 11. This modification is useful when
15 a temperature sensor is not integrated into the electronic device 11.

As another modification, the Fig. 1 and Fig. 12 temperature control systems can be changed such that the set point temperature on the signal lines 12d varies with
20 time. For example, the set point temperature can be stepped from one level to another, or it can vary continuously in a ramped or sinusoidal fashion.

Accordingly, it is to be understood that the present inventions are not limited to the details of any
25 one embodiment in any one figure, but are defined by the appended claims.

WHAT IS CLAIMED IS:

1. A temperature control system (Fig 1) which maintains the temperature of an electronic device (11) near a set point temperature; said system comprising:
5 an electric heater (13) that has a first face (13a) for making contact with said electronic device, and a second face (13b) which is opposite said first face (13a);
10 a heat sink (14), coupled to said second face (13b) of said heater (13), which absorbs heat from said electronic device (11) through said second face (13b) of said heater (13);
a temperature sensor (12c), coupled to said
15 electronic device (11); and,
a control circuit (15), coupled to said temperature sensor (12c) and to said heater (13), which decreases the power to said heater (13) when the sensed temperature of said electronic device (11) is above said
20 set point, and vice-versa.
2. A temperature control system according to claim 1 wherein said temperature sensor is integrated into said electronic device.
- 25 3. A temperature control system according to claim 1 wherein a temperature sensor is attached to the exterior of said electronic device.
4. A temperature control system according to claim 1 wherein said heater includes electric resistors
30 between or on said first and second face, and wherein said control circuit sends current to said resistors.

5. A temperature control system according to claim 1 wherein said heater includes a laser absorbing material between or on said first and second face, and wherein said control circuit heats said material with a laser.
6. A temperature control system according to claim 1 wherein said heater is thin in extent from said first face to said second face.
7. A temperature control system according to claim 1 wherein said heater is flat and less than one-quarter centimeter thick.
8. A temperature control system according to claim 1 wherein the thermal mass of said heater is no more than three times the thermal mass of said electronic device.
9. A temperature control system according to claim 1 wherein the thermal mass of said heater is less than the thermal mass of said electronic device.
10. A temperature control system according to claim 1 wherein the thermal resistance from said heater to said electronic device is less than three times the thermal resistance from said heater to said heat sink.
11. A temperature control system according to claim 1 wherein the thermal resistance from said heater to said electronic device is less than the thermal resistance from said heater to said heat sink.

12. A temperature control system according to claim 1 wherein the power dissipation of said electronic device is varied by sending input signals to said
5 electronic device which switch transistors ON and OFF in said electronic device and simultaneously test said electronic device.

13. A temperature control system according to claim 1 wherein lying between said first face of said
10 heater and said electronic device is a film of a liquid which evaporates, without leaving any residue, at a temperature that is too low to damage the electronic device.

14. A temperature control subassembly, for use in a temperature control system which regulates the temperature of an electronic device; said subassembly comprising:

an electric heater that has a first face for
20 making contact with said electronic device, and a second face which is opposite said first face; and,

a heat sink which is coupled to said second face of said heater and which cools said electronic device by absorbing heat from said electronic device
25 through said heater.

15. A temperature control subassembly according to claim 14 wherein said heater includes electric resistors between or on said first and second face.

16. A temperature control subassembly according to claim 14 wherein said heater includes a laser absorbing material between or on said first and second face.

17. A temperature control subassembly, for use in a temperature control system which regulates the temperature of an electronic device; said subassembly
5 comprising;

an electric heater that generates an infrared laser beam; and,

a heat sink which contacts said electronic device and which has windows that pass said laser beam
10 through said heat sink to said electronic device.

18. A temperature control system which maintains the temperature of an electronic device near a set point temperature; said system comprising:

an electric heater that has a first face for
15 making contact with said electronic device, and a second face which is opposite said first face;

a heat sink, coupled to said second face of said heater, which absorbs heat from said electronic device through said second face of said heater;

20 respective temperature sensors coupled to said heater and said heat sink;

an estimator circuit, coupled to said temperature sensors, which estimates the temperature of said electronic device as a function of the sensed heater
25 temperature and the sensed heat sink temperature; and,

a control circuit, coupled to said estimator circuit and said heater, which decreases the power to said heater when the estimated temperature of said electronic device is above said set point, and vice-
30 versa.

19. A temperature control system according to claim 18 wherein said function by which said estimator

circuit estimates the device temperature T_d is
 $T_d = T_h + \theta_{d-h} [M_h \dot{T}_h - P_h + (T_h - T_s) \div (\theta_{h-s})]$ where: T_h is the
sensed heater temperature and \dot{T}_h is the rate of change of
 T_h ; T_s is the sensed heat sink temperature; θ_{d-h} is a
5 thermal resistance between the electronic device and the
heater; θ_{h-s} is a thermal resistance between the heater
and the heat sink; P_h is the power to the heater; and M_h
is the thermal mass of the heater.

10 20. A temperature control system according to
claim 18 wherein said estimator circuit selects a
magnitude for a thermal resistance, between said
electronic device and the first face of said heater,
based on the rate at which the temperature of said heater
15 changes when a change in heater power occurs; and wherein
said estimator circuit uses the selected thermal
resistance in said function for estimating said device
temperature.

20 21. A temperature control system according to
claim 18 wherein said estimator circuit includes a memory
that stores respective values of: a) a thermal
resistance from the first face of said heater to said
device, b) a thermal resistance from the second face of
25 said heater and said heat sink, and c) a thermal mass of
said heater; and wherein said estimator circuit uses said
stored values in said function for estimating device
temperature.

30 22. A temperature control system according to
claim 18 wherein said estimator circuit is a digital
microprocessor.

23. A temperature control system according to claim 18 wherein said estimator circuit is an analog circuit.

5

24. A temperature control system according to claim 18 wherein said heat sink is coupled to said second face of said heater through a thermal resistivity which is larger than $0.1^{\circ}\text{C cm}^2/\text{watt}$.

10

25. A temperature control system according to claim 18 wherein said heater includes electric resistors between or on said first and second face, and wherein said control circuit sends current to said resistors.

15

26. A temperature control system according to claim 18 wherein said heater includes a laser absorbing material between or on said first and second face, and wherein said control circuit heats said material with a laser.

20

27. A temperature control system according to claim 18 wherein the thermal resistance from said heater to said electronic device is less than three times the thermal resistance from said heater to said heat sink.

25

28. A temperature control system according to claim 18 wherein the thermal mass of said heater is no more than three times the thermal mass of said electronic device.

30

29. A temperature control system according to claim 18 wherein said heater is thin in extent from said first face to said second face.

5

30. A temperature control system according to claim 18 wherein lying between said first face of said heater and said electronic device is a film of a liquid which evaporates, without leaving any residue, at a
10 temperature that is too low to damage the electronic device.

31. A temperature control subassembly, for use in a temperature control system which regulates the
15 temperature of an electronic device; said subassembly comprising:

an electric heater that makes contact with said electronic device, and a heat sink that makes contact with said heater;

20 a pair of temperature sensors coupled to said heater and said heat sink; and,

an estimator circuit, coupled to said temperature sensors, which estimates the temperature of said electronic device as a function of the sensed heater
25 temperature and the sensed heat sink temperature.

32. An electro-mechanical subassembly having an enhanced thermal interface, comprising:

an electronic device which has a face that
30 dissipates heat;

a heat exchange member which has a face that mates with said face on said electronic device; and,

a film of liquid, lying between said mating faces of said heat exchange member and said electronic device, which evaporates without leaving any residue at a temperature that is too low to damage the electronic device.

33. A system according to claim 32 wherein said liquid is water.

34. A system according to claim 32 wherein said liquid is a mixture of water and a volatile material which elevates the boiling point of said mixture above the boiling point of water.

35. A system according to claim 32 wherein said liquid is a mixture of water and a volatile material which suppresses the freezing point of said mixture below the freezing point of water.

36. A system according to claim 32 wherein said liquid is a mixture of water and a volatile material which enhances the wettability of said mixture above the wettability of water.

37. A system according to claim 32 wherein said liquid evaporates at a temperature of less than two hundred degrees centigrade.

38. A system according to claim 32 wherein said heat exchange member is cooler than said electronic device.

5 39. A system according to claim 32 wherein said heat exchange member is cooled with a liquid.

40. A system according to claim 32 wherein said heat exchange member is hotter than said electronic
10 device.

41. A system according to claim 32 wherein said heat exchange member is an electronic heater.

15 42. A system according to claim 32 wherein said heat exchange member has a thermal mass which is at least ten times larger than the thermal mass of said electronic device.

20 43. A method of thermally coupling an electronic device to a heat exchange member; said method including the steps of:

providing said heat exchange member with a face that mates with a corresponding face on said electronic
25 device; and,

disposing a film of liquid between said mating faces of said heat exchange member and said electronic device, which evaporates without leaving any residue at a temperature that is too low to damage the electronic
30 device.

44. A method according to claim 43 and further including the step of evaporating said liquid completely off of said electronic device, after said electronic device has dissipated power for a predetermined time interval while being coupled to said heat exchange member.

45. A method according to claim 43 wherein said predetermined time interval is less than one hour.

46 A method according to claim 43 wherein said evaporating step is performed by the substeps of separating said mating faces and exposing the liquid film to a temperature which is too low to damage said electronic device.

47. A method according to claim 43 wherein said liquid is water.

20

48. A method according to claim 43 wherein said liquid is a volatile mixture.

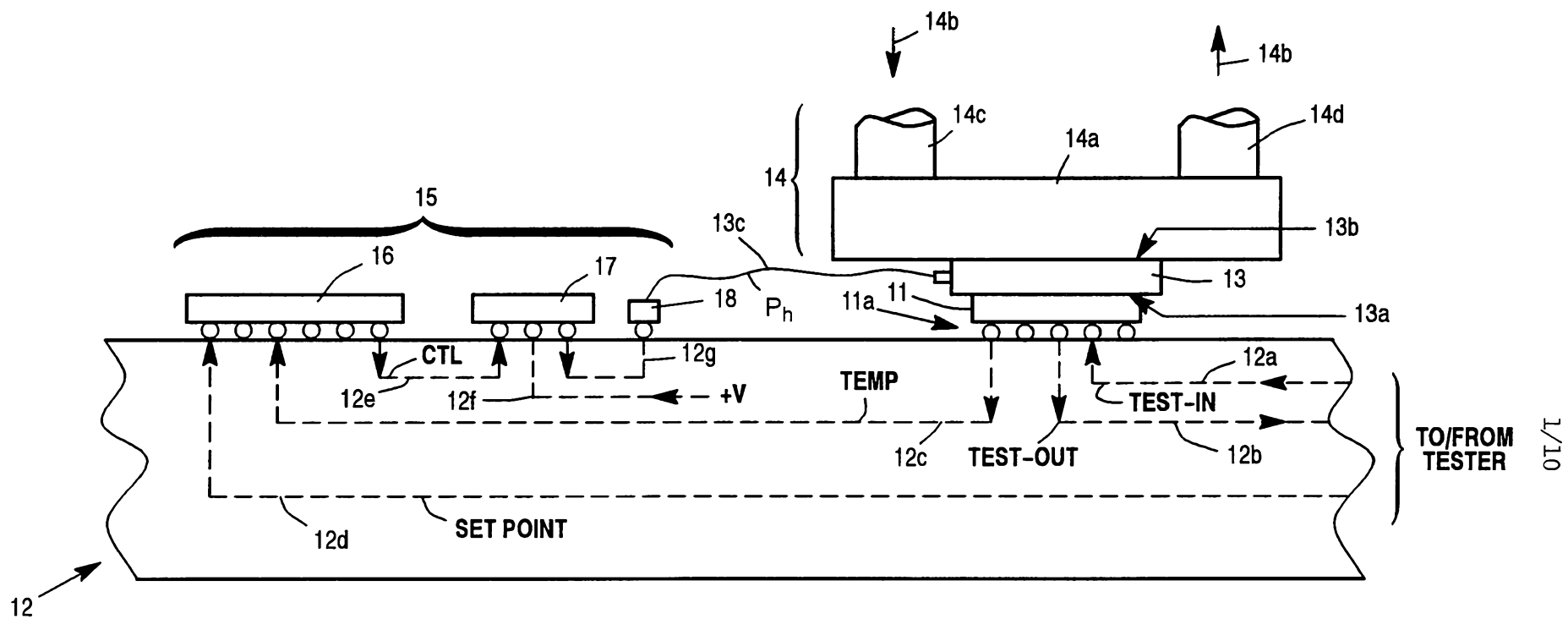
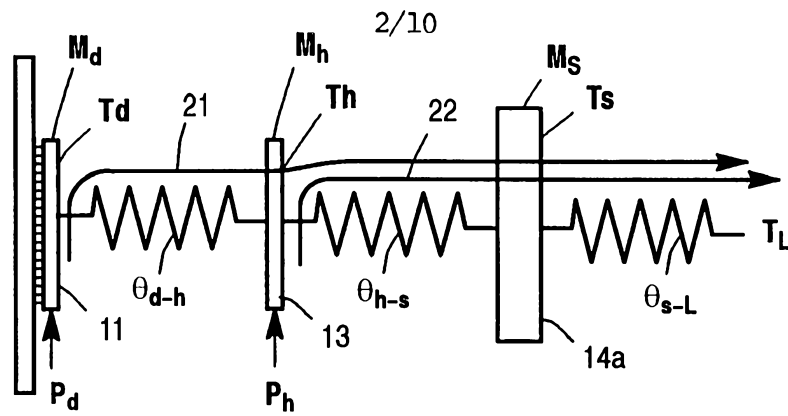
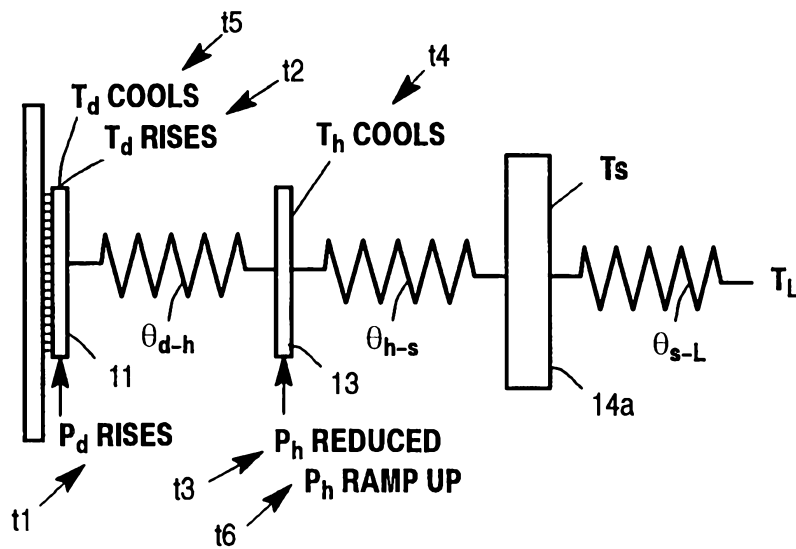
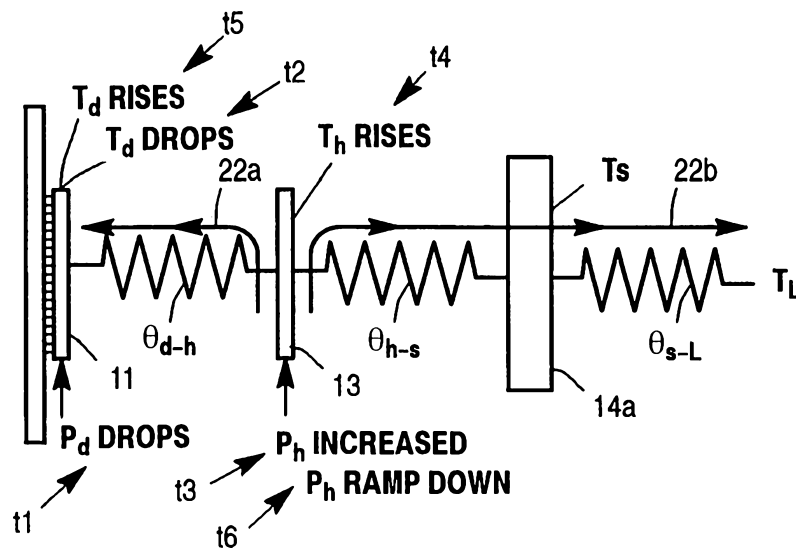


Figure 1

**Figure 2****Figure 3****Figure 4**

3/10

EQUATION 1

31c 31b 31a

$$\dot{M}_d \dot{T}_d = (T_h - T_d) / \theta_{d-h} + P_d$$

EQUATION 2

32d 32c 32b 32a

$$\dot{M}_h \dot{T}_h = (T_d - T_h) / \theta_{d-h} + (T_s - T_h) / \theta_{h-s} + P_h$$

EQUATION 3

33c 33b 51 33a

$$\dot{M}_s \dot{T}_s = (T_h - T_s) / \theta_{h-s} + (T_L - T_s) / \theta_{s-L}$$

Figure 5

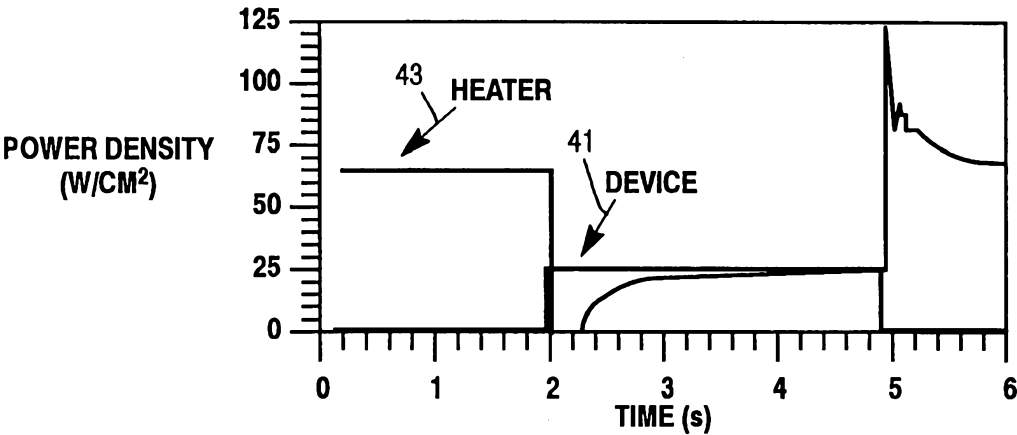


Figure 6

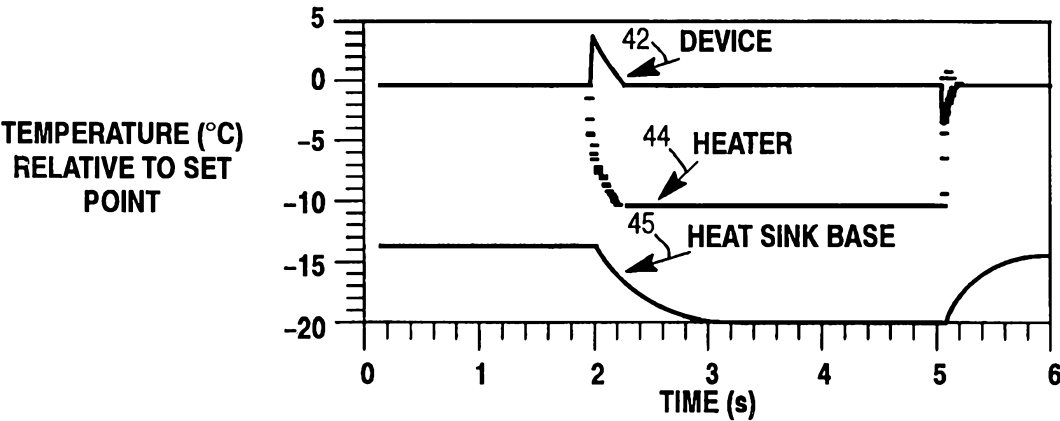


Figure 7

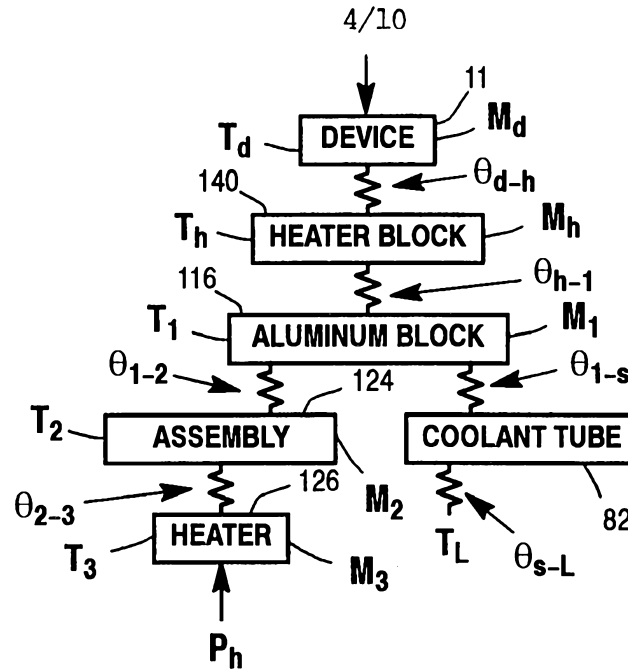


Figure 8
(Prior Art)

EQUATION 11

$$\dot{M}_d T_d = (T_h - T_d) / \theta_{d-h} + P_d$$

EQUATION 12

$$\dot{M}_h T_h = (T_d - T_h) / \theta_{d-h} + (T_1 - T_h) / \theta_{h-1}$$

EQUATION 13

$$\dot{M}_1 T_1 = (T_h - T_1) / \theta_{h-1} + (T_2 - T_1) / \theta_{1-2} + (T_s - T_1) / \theta_{1-s}$$

EQUATION 14

$$\dot{M}_2 T_2 = (T_1 - T_2) / \theta_{2-1} + (T_3 - T_2) / \theta_{2-3}$$

EQUATION 15

$$\dot{M}_3 T_3 = (T_2 - T_3) / \theta_{2-3} + P_h$$

EQUATION 16

$$\dot{M}_s T_s = (T_1 - T_s) / \theta_{1-s} + (T_L - T_s) / \theta_{s-L}$$

Figure 9
Prior Art

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$$\text{EQUATION 20} \quad \overset{32d}{\dot{M}_h \dot{T}_h} = - \overset{33c}{\dot{M}_s \dot{T}_s} + \overset{32c}{(T_d - T_h) / \theta_{d-h}} + \overset{33a}{(T_L - T_s) / \theta_{s-L}} + \overset{32a}{P_h}$$

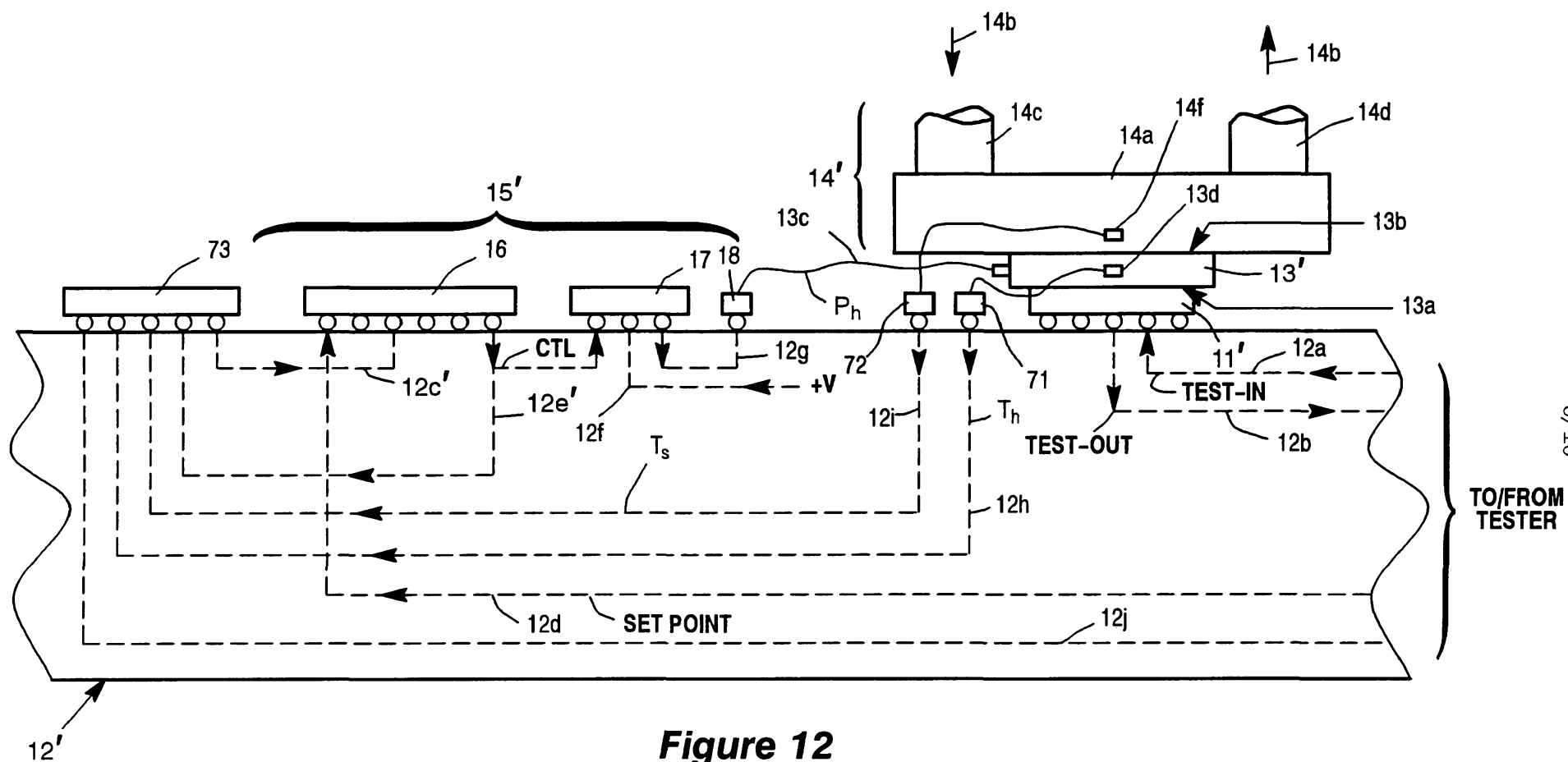
$$\text{EQUATION 21} \quad \overset{67}{\dot{M}_h \dot{T}_h} = - \overset{65}{\dot{M}_1 \dot{T}_1} - \overset{63}{\dot{M}_2 \dot{T}_2} - \overset{61}{\dot{M}_3 \dot{T}_3} - \overset{60}{\dot{M}_s \dot{T}_s} + (T_d - T_h) / \theta_{d-h} + (T_L - T_s) / \theta_{s-L} + P_h$$

Figure 10

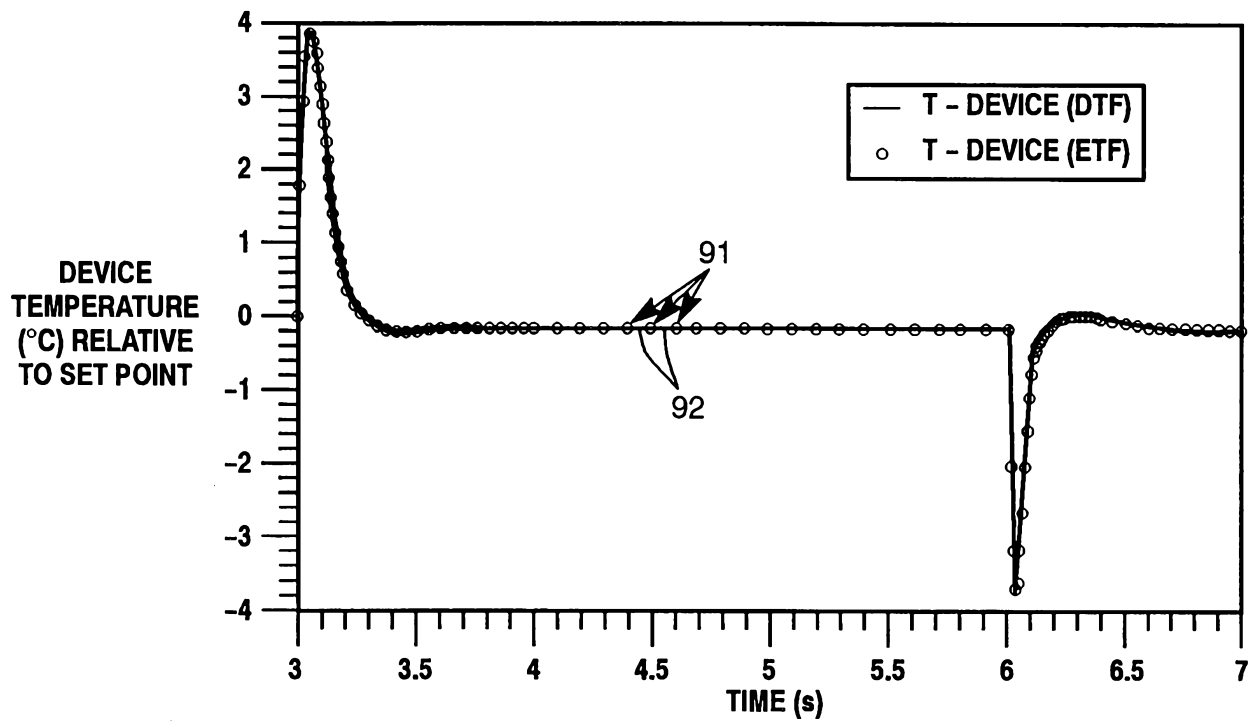
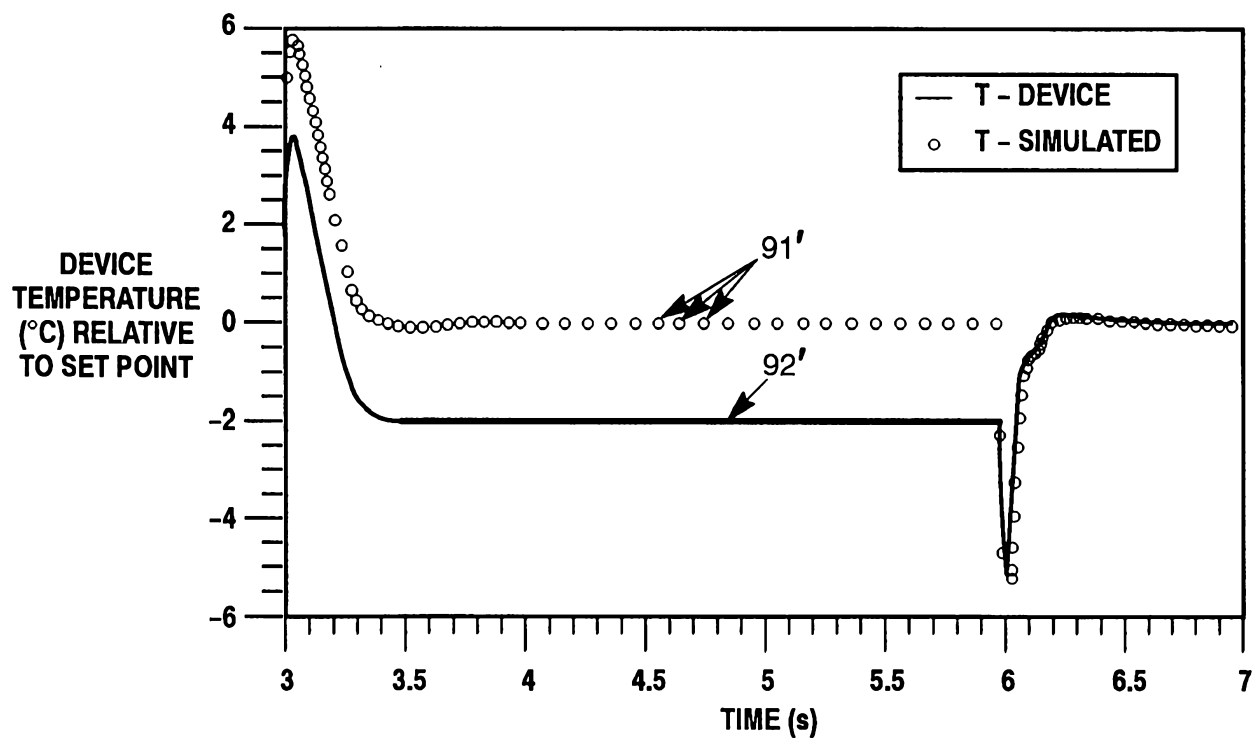
$$\text{EQUATION 2} \quad \dot{M}_h \dot{T}_h = (T_d - T_h) / \theta_{d-h} + (T_s - T_h) / \theta_{h-s} + P_h$$

$$\text{EQUATION 31} \quad T_d = T_h + \theta_{d-h} [\dot{M}_h \dot{T}_h - P_h + (T_h - T_s) \div (\theta_{h-s})]$$

Figure 11



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Figure 13Figure 14

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$$\text{EQUATION 2} \rightarrow M_h \dot{T}_h = (T_d - T_h) / \theta_{d-h} + (T_s - T_h) / \theta_{h-s} + P_h$$

32d
32c
32b
32a

STEP 1 \rightarrow let $P_D = 0$

STEP 2 \rightarrow change P_h

STEP 3 \rightarrow $\begin{cases} \dot{T}_h \text{ varies "slow" if } \theta_{d-h} \text{ is "small"} \\ \dot{T}_h \text{ varies "fast" if } \theta_{d-h} \text{ is "large"} \end{cases}$

Figure 15

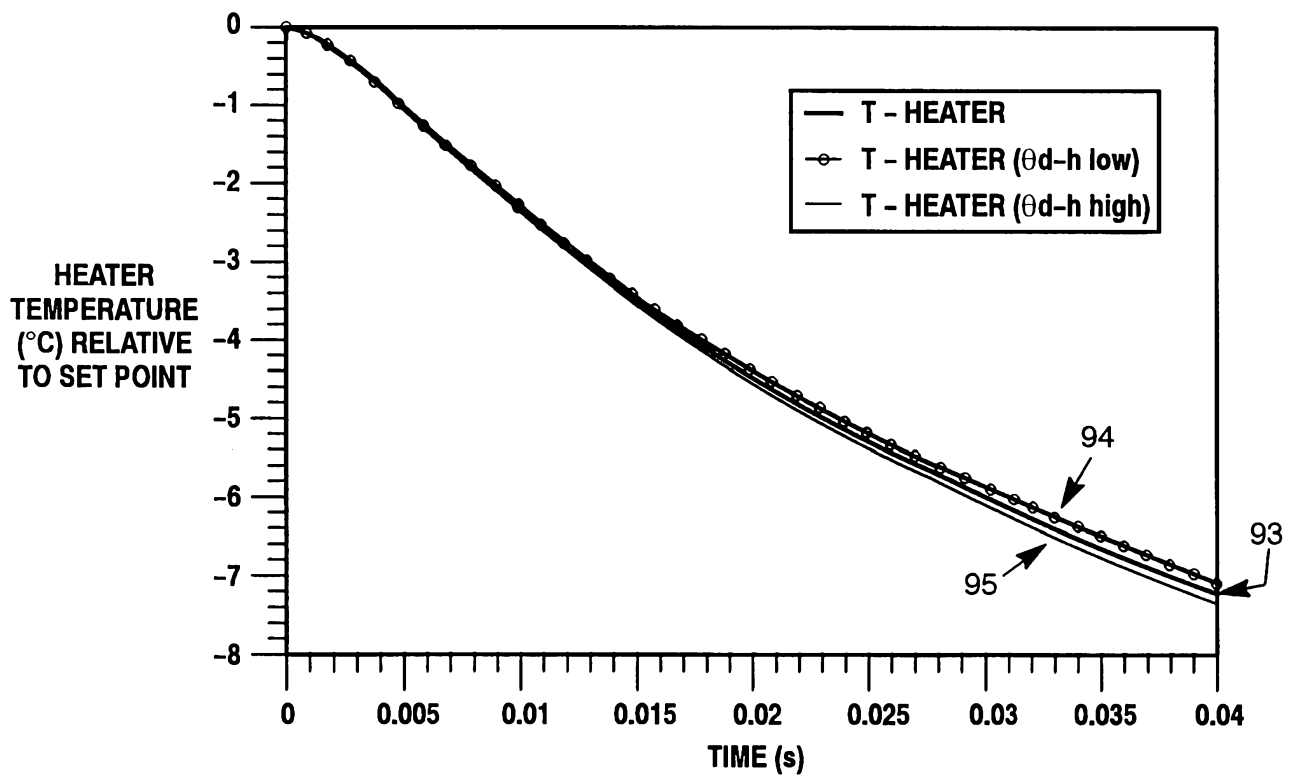


Figure 16

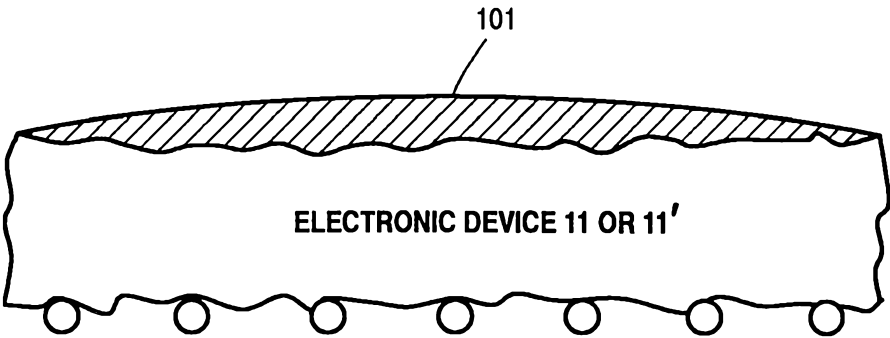


Figure 17

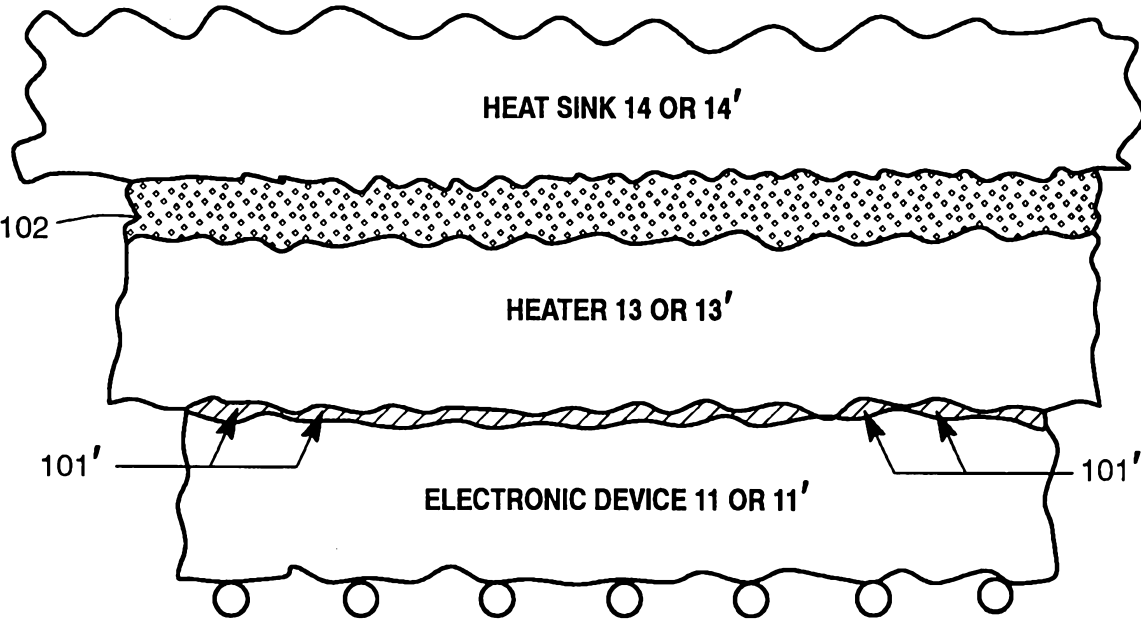


Figure 18

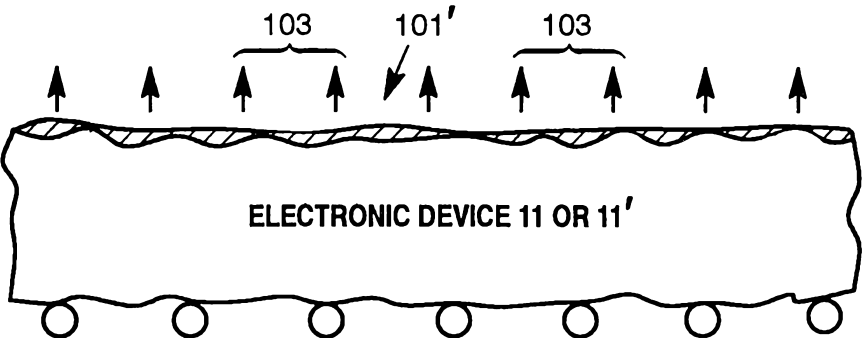


Figure 19

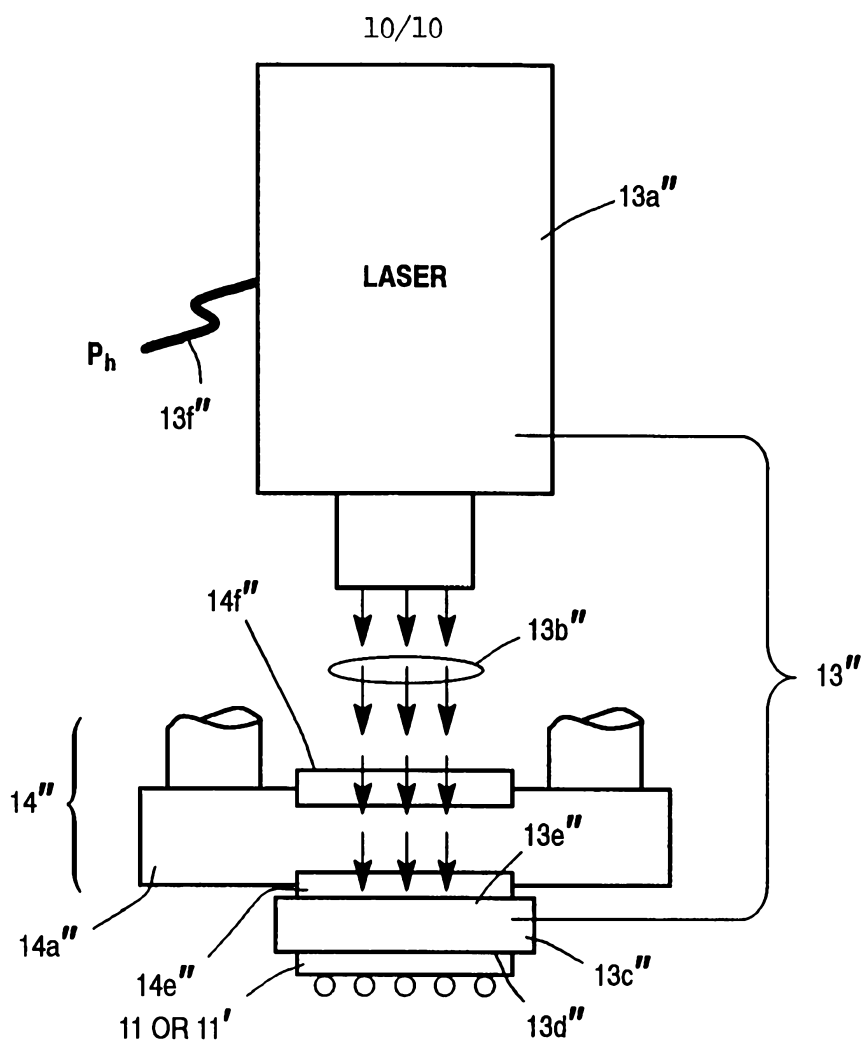


Figure 20

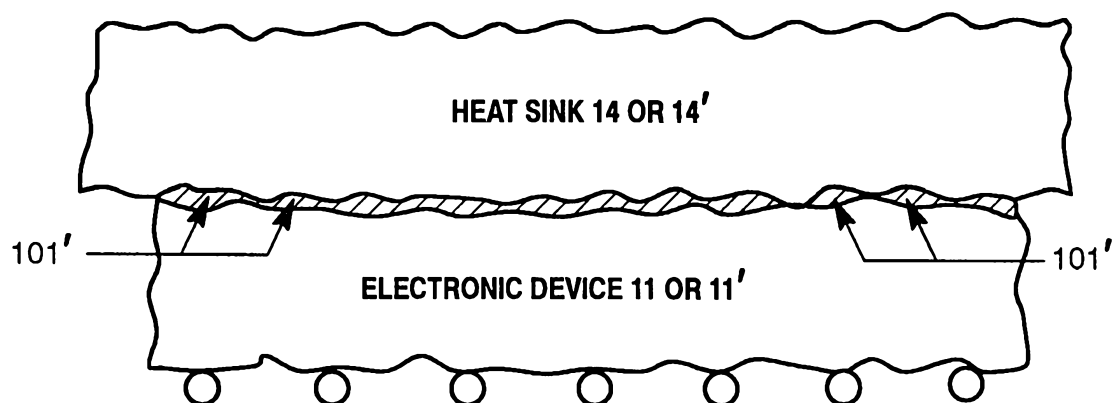


Figure 21

INTERNATIONAL SEARCH REPORT

1. National Application No

PCT/US 98/06701

A. CLASSIFICATION OF SUBJECT MATTER

IPC 6 H05K7/20 H01L21/66 G05D23/19

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 6 H05K H01L G05D

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category °	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
P, X	GB 2 316 196 A (THE COURT OF THE UNIVERSITY OF PAISLEY) 18 February 1998 see abstract see page 1, line 19 - page 3, line 1 see page 7, line 29 - page 9, line 20; figure 1	1
Y	US 4 848 090 A (PETERS) 18 July 1989 see abstract see column 2, line 61 - column 4, line 16; figures 2-4	1-4, 6-9, 12, 14, 15
Y	PATENT ABSTRACTS OF JAPAN vol. 13, no. 23 (E-705), 19 January 1989 & JP 63 226936 A (TOKYO ELECTRON LTD), 21 September 1988, see abstract	1-4, 6-9, 12, 14, 15
	--- -/--	

☒ Further documents are listed in the continuation of box C.

☒ Patent family members are listed in annex.

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"&" document member of the same patent family

Date of the actual completion of the international search

15 July 1998

Date of mailing of the international search report

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INTERNATIONAL SEARCH REPORT

National Application No

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C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
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A	EP 0 320 297 A (CANON KABUSHIKI KAISHA) 14 June 1989 see abstract see column 4, line 5-55; figure 2B ---	1
A	US 4 854 726 A (LESLEY ET AL.) 8 August 1989 see abstract see column 4, line 39 - column 5, line 45; figures 1,11 ---	1
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A	PATENT ABSTRACTS OF JAPAN vol. 97, no. 3, 31 March 1997 & JP 08 286551 A (RICOH CO LTD), 1 November 1996, see abstract ---	18,31
A	US 4 975 766 A (UMEZAWA) 4 December 1990 see abstract see column 2, line 40 - column 3, line 15; figure 3 -----	32

INTERNATIONAL SEARCH REPORT

Information on patent family members

I. International Application No

PCT/US 98/06701

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