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References cited:
- GB-A- 2 062 421
  (HITACHI LTD ) 23 May 1987
  MOTOR CO LTD ) 17 October 1987
  (OLYMPUS OPTICAL CO LTD ) 5 March 1986
  (MATSUSHITA DENKI SANGYO K.K. ) 23 January 1985

Digital information reproducing apparatus
Gerät zur Wiedergabe von Digitalinformationen
Appareil de reproduction d'informations numériques

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Description

BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates to a digital information reproducing apparatus such as a magneto-optical disk reproducing apparatus.

2. Description of the Related Art

In a digital information storing apparatus for recording and reproducing or just reproducing digital information along tracks on a recording medium included in an optical disk unit, a magnetic disk unit and a magnetic tape unit, an amplitude detection or a peak detection is carried out for the purpose of converting a reproduction signal output from a reproduction head into a digital binary signal.

For example, the optical disk unit employs an amplitude detecting system or a peak detecting system according to a recording system of the digital data. The amplitude detecting system and the peak detecting system respectively correspond to an NRZI system and a RZ system. In the NRZI system, the information to be detected is a front edge and a rear edge of a recorded mark.

Hence, the reproduction signal is converted into a digital signal by binary-comparing the reproduction signal with a proper threshold value. In the RZ system, the center of a recorded mark is information to be detected. It is, therefore, necessary to differentiate the reproduction signal for seeking a zero cross point.

This method, however, is unable to separate a signal pulse for the center of the recorded mark from a noise pulse generated on the noise of the reproduction signal. Under this method, a data error is more likely to take place. To overcome this shortcoming, even in a peak detecting system, it is necessary to detect an amplitude for distinguishing the signal from the noise according to the amplitude of the reproduction signal.

Herein, in the amplitude detecting system, it is necessary to transmit the reproduced waveform containing DC components. On the other hand, because of the problems such as low frequency variation due to reflectivity variation of a recording medium, occurrence of DC offset, and difficulty in arranging a wide-band amplifier, in a reproducing circuit system, it is necessary to take the steps of removing low frequency components from the reproduction signal and constantly controlling a threshold value of a comparator for converting the reproduction signal into a digital signal to be a center value of the reproduction signal according to the reproduction signal level. The prior art of such a reproduction circuit is shown in Fig. 13.

In Fig. 13, the reproduction signal converted into an electric signal through the effect of the reproduction head is amplified up to a proper amplitude level in an AC-coupled preamplifier 500 and then is guided to a low-pass filter 501. The low-pass filter 501 operates to remove unnecessary high-frequency components of the reproduction signal. The output of the low-pass filter 501 is guided to an equalizing circuit 502 for compensating for a restricted bandwidth caused according to resolution of the reproduction head. The output of the equalizing circuit 502 is guided to one input terminal of the comparator 503 and the first envelope detecting circuit 504 and the second envelope detecting circuit 505.

A first envelope detecting circuit 504 is provided for detecting a positive side envelope of the output of the equalizing circuit 502 and is arranged to have a diode 610, a resistor 611, a capacitor 612 and a buffer 613. A second envelope detecting circuit 505 is provided for detecting a negative side envelope of the output of the equalizing circuit 502 and is arranged to have a diode 620, a resistor 621, a capacitor 622 and a buffer 623.

The outputs of the envelope detecting circuits 504 and 505 are guided to an operating unit 506 in which the outputs are added to one signal and then the resulting signal is subject to filtering for removing a ripple. The output of the operating unit 506 is guided to the other input terminal of the comparator. According to the above arrangement, the reproduction signal is constantly controlled to keep the threshold value of the comparator for converting into a digital signal at the center of the reproduction signal according to the reproduction signal level.

In such a reproducing circuit, however, the envelope detecting circuit operates to follow even a disordered envelope of the reproduction signal caused by a minute defect on the recording medium. That is, the capacitors 612 and 622 store lots of charges corresponding to an amplitude of the reproduction signal obtained when the reproduction head passes along a minute defect. After it passes along the minute defect, the charges are discharged on the time constant defined by the combination of the capacitor 612 or 622 and the resistor 611 or 621. Hence, a considerably long time is taken in returning the storage into the storage amount of charges for the original amplitude of the reproduction signal. During this period, it is impossible to obtain a proper threshold value for the reproduction signal level, thereby bringing about a reproduction data error. That is, even a minute defect causes the reproduction data error to be like a burst, thereby disadvantageously lowering reliability of the data.

JP-A-62-112230 relates to a threshold level automatic control circuit. The control circuit comprises positive and negative envelope detecting circuits and a control circuit for selecting suitable time constants for the envelope detecting circuits.

GB-A-2062421 relates to a data extracting circuit comprising a level comparator having a first input terminal supplied with a delayed input signal and a second input terminal supplied with a threshold voltage and operative to derive a digital binary data signal.
SUMMARY OF THE INVENTION

It is therefore an object of the present invention to provide a digital information reproducing apparatus which reduces an adverse effect given by a disordered portion of an envelope resulting from a minute defect or the like.

According to the invention there is provided a digital information reproducing apparatus as set out in claim 1.

When the reproduction head scans along the track, normally, by setting the threshold value of the second comparator as a larger level than the amplitude level of the reproduction signal obtained from an information-recorded part, the output of the second comparator is a signal corresponding to a disordered portion of a reproduction signal amplitude level (the disordered portion in which the amplitude level is larger than the normally obtained reproduction signal amplitude level) caused by a minute defect.

Based on the output signal of the second comparator, the switching circuit operates to reduce the time constant on which the charges stored in the capacitors of the first and second envelope detecting circuits are discharged. After the reproduction head passes along the minute defective part, it takes a less time to return the storage amount of the charges into the charges for the original amplitude of the reproduction signal.

There may be provided a subtracting circuit for obtaining a difference signal between the outputs of the first and second envelope detecting circuits; a differentiating circuit for differentiating the difference signal; a switching circuit for switching time constants of the time constant circuits provided in the first and second envelope detecting circuits, based on an output signal of the differentiating circuit.

The output of the differentiating circuit is a signal indicating an abrupt change of an envelope of the reproduction signal caused by a minute defect or the like. Based on the output signal of the differentiating circuit, the switching circuit operates to reduce the time constant on which the charges stored in the capacitors of the first and second envelope detecting circuits are discharged. After the reproduction head passes along the minute defective part, it takes a less time to return the storage amount of charges into the charges for the original amplitude of the reproduction signal.

Further objects and advantages of the present invention will be apparent from the following description of the preferred embodiments of the invention as illustrated in the accompanying drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is an electric circuit diagram showing a digital information reproducing apparatus according to a first embodiment of the present invention.

Fig. 2 is a waveform view for explaining an operation of the first embodiment.

Fig. 3 is an electric circuit diagram showing a digital information reproducing apparatus according to the second embodiment of the present invention.

Fig. 4 is a waveform view for explaining the operation of the second embodiment.

Fig. 5 is an electric circuit diagram showing a digital information reproducing apparatus according to a third embodiment of the present invention.

Fig. 6 is a waveform view for explaining the operation of the third embodiment.

Fig. 7 is an electric circuit diagram showing a digital information reproducing apparatus according to a fourth embodiment of the present invention.

Fig. 8 is a waveform view for explaining the opera-
tion of the fourth embodiment.

Fig. 9 is an electric circuit diagram showing a digital information reproducing apparatus according to a fifth embodiment of the present invention.

Fig. 10 is a waveform for explaining the operation of the fifth embodiment.

Fig. 11 is an electric circuit block diagram showing a concrete circuit of a pulse generating circuit included in the second embodiment.

Fig. 12 is an electric circuit block diagram showing a head position detecting circuit included in the third to the fifth embodiments.

Fig. 13 is an electric circuit block diagram showing the conventional digital information reproducing apparatus.

DESCRIPTION OF THE PREFERRED EMBODIMENT

The description will be directed to a first embodiment of the present invention as referring to the drawings.

Fig. 1 is a block diagram showing an arrangement of a reproducing circuit of this invention. As shown in Fig. 1, a reproduction signal converted into an electric signal by a reproduction head is amplified by an AC-coupled pre-amplifier 100 and then is guided to a low-pass filter 101. The output of the low-pass filter 101 is guided to an equalizing circuit 102 in which the waveform equalizing treatment is carried out with respect to the output.

The output of the equalizing circuit 102 is guided to one input terminal of a first comparator 103 as well as a first envelope detecting circuit 104 and a second envelope detecting circuit 105. The first envelope detecting circuit 104 for detecting an envelope on the positive side of the output of the equalizing circuit 102 is arranged of a diode 210, a resistor 211, a capacitor 212, an analog switch 213, a resistor 214 and a buffer amplifier 215.

The second envelope detecting circuit 105 for detecting an envelope on the negative side of the output of the equalizing circuit 102 is arranged of a diode 220, a resistor 221, a capacitor 222, an analog switch 223, a resistor 224 and a buffer amplifier 225. The outputs of the first and the second envelope detecting circuits 104 and 105 are guided to an operating unit 106 in which these outputs are added. The added signal is filtered for removing a ripple. The output of the operating unit 106 is guided to the other input terminal of the comparator 103. The outputs of the envelope detecting circuits 104 and 105 are guided to an operating unit 107 in which a difference between these outputs are derived. The output of the operating unit 107 is guided to one input terminal of a second comparator 108. The other input terminal of the comparator is kept at a predetermined potential \( V_p \). The output of the comparator 108 is connected to the control input terminals of the analog switches 213 and 223.

The operation of the above-mentioned arrangement will be discussed below as referring to a model view shown in Fig. 2. Herein, the analog switches 213 and 223 are normally made open and the resistors have the relations of resistor 211 > resistor 214 and resistor 221 > resistor 224.

A real line of Fig. 2a indicates an output signal of the equalizing circuit 102. The amplitude of the signal is amplified up to a proper level so that the high-pass noise components are removed from the signal through the effect of the pre-amplifier 100 at the previous stage, the low-pass filter 101 and the equalizing circuit 102. Further, the degrade of a frequency characteristic resulting from the reproduction head is compensated by these components.

In Fig. 2, the periods A and C indicate normal states in which no disorder caused by a defect takes place in an envelope. The period B indicates a portion in which the amplitude of the reproduction signal is swung too much by means of a minute defect existing on the disk.

The output signals of the first and the second envelope detecting circuits 104 and 105 for such a reproduction signal are indicated by Fig. 2b and Fig. 2c, respectively. Further, the output signal of the operating unit 107 is indicated by Fig. 2d and the output signal of the comparator 108 is indicated by Fig. 2e.

For a head portion of the succeeding period B corresponding to a minute defect, as indicated by the output signal Fig. 2b, the output signal of the first envelope detecting circuit 104 abruptly rises. Hence, the output signal of the operating unit 107 is higher than the potential \( V_p \) as shown by the output signal Fig. 2d (D of Fig. 2).

At this time, the output signal of the comparator 108 is at high level. Since the output signal of the comparator 108 is at high level, the analog switches 213 and 223 are made closed and the charges stored in the capacitor 212 (222) are discharged through the resistors 211 and 221 and 224. The time constant for this purpose is defined by (resistor 211) (resistor 214) /{(resistor 211) + (resistor 214)} (resistor 221) (resistor 224) / (resistor 221) + (resistor 224)) and the capacitor 212 (222). By setting the resistance of the resistor 214 (224) to a far smaller value than that of the resistor 211 (221), the time constant on which the charges are discharged from the capacitor 212 (222) when the analog switch 213 (223) is closed is sufficiently small and is substantially defined by the resistor 214 (224) and the capacitor 212 (222).

In this embodiment, it is observed that an envelope peak appears on the positive polarity side of the repro-
duction signal, the output signal of the first envelope detecting circuit 104 represents the envelope with fidelity during the period when successive minute defects appear as indicated in the output signals Fig. 2b and Fig. 2c. On the other hand, the output signal of the second envelope detecting circuit 105 is returned to a zero potential based on the time constant defined by the resistor 224 and the capacitor 222. As stated above, the output signal of the operating unit 107 is as shown by the output signal Fig. 2d corresponding to the appearance of minute defects but keeps the potential Vₕ or more. The output of the comparator 108 is maintained at high level (output signal (e)).

As shown in Fig. 2a, in the head portion of C where the original reproduction signal is obtained after the passage of the minute defective portion, the output signal of the first envelope detecting circuit 104 is started to return to the zero potential based on the time constant defined substantially by the resistor 214 and the capacitor 212. The output signal of the operating unit 107 lowers its signal level to the potential Vₕ or less (E of Fig. 2).

At this time, the output signal of the comparator 108 is at low level. Since the output signal of the comparator 108 is at low level, the analog switches 213 and 223 are made open and the normal circuit state takes place where the charges stored in the capacitor 212 (222) are discharged through the resistor 211 (221).

In this embodiment, immediately after the passage of a minute defective portion, a peak on the negative polarity side appears. Hence, as indicated by F of Fig. 2, the output signal of the operating unit 107 goes up to the potential Vₕ or higher again. The analog switches 213 and 223 are made closed, so that the time constant on which the charges stored in the capacitor 212 (223) are discharged becomes smaller. Finally, at the amplitude level of the normal reproduction signal, the analog switches 213 and 223 are toggled from an open state to a closed state or vice versa until the output signal of the operating unit 107 does not reach the potential Vₕ or higher.

In accordance with the foregoing operation of the circuit components, the output signal of the operating unit 106 is as shown by a broken line of Fig. 2a and quickly follows the normal reproduction signal after the passage of the minute defective portion.

The second embodiment of the present invention will be discussed as referring to the drawings.

Fig. 3 is a block diagram showing a reproducing circuit according to the second embodiment of the present invention. The same components as those shown in Fig. 1 have the same reference numbers. In this embodiment, based on the output signal of the comparator 108, a pulse generating circuit 109 is newly provided for generating a control signal for opening or closing the analog switches 213 and 223. The concrete arrangement of the pulse generating circuit 109 has as a component a retriggerable and monostable multi-vibrator 300 which operates at the leading portion of the input signal as shown in Fig. 11.

The operation of the foregoing arrangement will be discussed below as referring to Fig. 4.

Herein, it is assumed that the analog switches 213 and 223 are normally made open and the resistors have the relations of resistor 211 > resistor 214 and resistor 221 > resistor 224.

A real line indicated by Fig. 4a indicates the output signal of the equalizing circuit 102. This is true to the real line of Fig. 2a. With respect to this kind of reproduction signal, the output signals of the first and the second envelope detecting circuits 104 and 105 are indicated by Fig. 4b and Fig. 4c respectively. The output signal of the operating unit 107 is indicated by Fig. 4d and the output of the pulse generating circuit 109 is indicated by Fig. 4e.

In the A portion of Fig. 4a, the operations of the envelope detecting circuits 104 and 105 are analogous to those in the embodiment shown in Figs. 1 and 2 (output signals (b) and (c)).

In the head portion of the B for the successive minute defect, as indicated by the output signal Fig. 4b, the output signal of the first envelope detecting circuit 104 indicates an abrupt rise. Hence, the output signal of the operating unit 107 is kept at the potential Vₕ or more as indicated by Fig. 4d (D of Fig. 4).

At this time, the output signal of the comparator 108 goes up to a high level so that the pulse generating circuit 109 may output a pulse signal having a predetermined time width (output signal (e)). While the output signal of the pulse generating circuit 109 is at high level, the analog switches 213 and 223 are made closed so that the charges stored in the capacitor 212 (222) are discharged through the resistors 211 and 214 (221 and 224). Herein, by setting the resistance of the resistor 124 (224) to a far smaller value than that of the resistor 211 (221), like the foregoing embodiment, the time constant on which the charges are discharged from the capacitor 212 (222) in the closed state of the analog switches 213 and 223 is decreased.

Assuming that the width of a pulse output from the pulse generating circuit 109 is set to a larger value than a length of time for a minute defect, at the head portion of C for obtaining the original reproduction signal after the passage of a minute defective portion, the output signal (b) of the first envelope detecting circuit 104 starts to return to a zero potential with a small time constant. The output signal (d) of the operating unit 107 lowers the signal level to the potential Vₕ or less (E of Fig. 4).

Then, while the output signal of the pulse generating circuit is kept at high level, the output signal of the operating unit 107 follows the reproduction signal.

By shifting the output signal (e) of the pulse generating circuit 109 to a low level, the analog switches 213 and 223 are made open, so that the overall circuit may enter into the normal state. According to the operations of the above circuit components, the output signal of the operating unit 106 is as indicated by a broken line of Fig. 2a.
4a and swiftly follows the normal reproduction signal after the passage of a minute defective portion.

The third embodiment of the present invention will be discussed as referring to Fig. 5.

Fig. 5 is a block diagram showing a reproducing circuit according to the third embodiment of the present invention. For the explanation's sake, the same components as those shown in Figs. 1 and 3 have the same reference numbers. In this embodiment, the output of the operating unit 107 for operating a difference between the outputs of the envelope detecting circuits 104 and 105 is guided to the differentiating circuit 110 in which the output is differentiated. The output signal of the differentiating circuit 110 is guided to one input terminal of the second comparator 111. The other input terminal of the comparator 111 is kept at the potential Vc.

The output signal of the comparator 111 is guided to the pulse generating circuit 112. Based on the output signal of the comparator 111, the pulse generating circuit 112 serves to generate a control signal for opening and closing the analog switches 213 and 223. The output of the pulse generating circuit 112 passes through a gate circuit 113 in order to control the analog switches 213 and 223. The pulse generating circuit 112 may be arranged of a retriggerable and monostable multi-vibrator, for example, like the foregoing embodiment. A circuit block 114 indicated by a broken line of Fig. 5 operates to generate a control signal for masking an output signal of the pulse generating circuit 112.

In this arrangement, by differentiating the output signal of the operating unit 107, no adverse effect is brought about by the variation of the amplitude level of the reproduction signal. The variation of the amplitude level takes place due to variation between disks, environmental conditions like a temperature, and an uneven recording and reproducing characteristic of a disk.

The operation of the foregoing arrangement will be discussed as referring to a model view of Fig. 6. Herein, it is assumed that the analog switches 213 and 223 are normally made open, the resistors have the relations of resistor 211 > resistor 214 and resistor 221 > resistor 224 and the output of the circuit block 114 is normally maintained at high level.

A real line of Fig. 6a indicates an output signal of the equalizing circuit 102. This is true to Fig. 2a. With respect to the reproduction signal, the output signals of the first envelope detecting circuit 104 and the second envelope detecting circuit 105 are indicated by Fig. 6b and Fig. 6c, respectively. The output signal of the operating unit 107 is indicated by Fig. 6d and the output signal of the differentiating circuit 110 is indicated by Fig. 6e. Further, the output signal of the pulse generating circuit 112 is indicated by Fig. 6f.

In the A portion of Fig. 6a, the operations of the envelope detecting circuits 104 and 105 are the same as those of the embodiment shown in Figs. 1 and 2 (output signals (b) and (c)).

In the head portion of the B for the successive minute defect, as indicated by the output signal (Fig. 6b), the output signal of the first envelope detecting circuit 104 indicates an abrupt rise. According to the abrupt rise, the output signal of the operating unit 107 rises (output signal (d)). Hence, the differentiating circuit 110 operates to output a pulsewise signal as corresponding to the abrupt rise of the output signal of the operating unit 107 as indicated by an output signal (Fig. 6e). The signal level of the pulsewise signal is maintained at the potential Vc or more (D of Fig. 6). At this time, the output signal of the comparator 111 rises to a high level and the pulse generating circuit 112 serves to output a pulse signal having a predetermined time width (output signal (f)).

While the output signal of the pulse generating circuit 112 is at high level, the analog switches 213 and 223 are made closed so that the charges stored in the capacitor 212 (222) are discharged through the resistors 211 and 214 (221 and 224). Herein, by setting the resistance of the resistor 214 (224) to a far smaller value than that of the resistor 211 (221), the time constant on which the charges of the capacitor 212 (222) are discharged in the closed state of the analog switches 213 and 223 is sufficiently small. This is the same as the foregoing embodiment.

Assuming that the width of the pulse output from the pulse generating circuit 112 is set to a larger value than a length of time for a minute defective portion, after the passage of the minute defective portion, at the head portion of C for obtaining an original reproduction signal, the output signal (b) of the first envelope detecting circuit 104 starts to return to zero potential with a small time constant. The output signal (d) of the operating unit 107 lowers its level signal. Then, the output signal (d) of the operating unit 107 follows the reproduction signal while the output signal (f) of the pulse generating circuit 112 is kept at high level.

By shifting the output signal (f) of the pulse generating circuit 112 to a low level, the analog switches 213 and 223 are made open and thus the overall circuit may return to the normal state.

In a case that the optical disk reproducing apparatus employs a system of dividing the optical disk into a plurality of blocks along the information tracks and recording data for each of the blocks, the reproduction signal converted into an electric signal by the reproduction head is an intermittent signal in each block. If the reproduction signal is intermittent, at the head portion of the reproduction signal, the output signal (d) of the operating unit 107 abruptly rises so that the differentiating circuit 110 may output a pulsewise signal. If the output signal (e) of the differentiating circuit 110 goes up to a predetermined potential Vc or higher, the output signal of the comparator 111 is shifted to a high level signal so that the pulse generating circuit 112 may output a pulse for closing the analog switches 213 and 223 even at the head portion of the reproduction signal.

A circuit block 114 indicated by a broken line of Fig.
5 serves to generate a control signal for removing a pulse signal output by the pulse generating circuit 112 at the head portion of the reproduction signal if the reproduction signal is intermittent. In this embodiment, a head position detecting circuit 230 and an inverter 231 are provided for receiving the output signal (d) of the operating unit 107 as its input and sensing the head portion of the intermittent reproduction signal and generating a pulse having a predetermined time width. The output signal of the inverter 231 is led to a gate circuit 113. The head position detecting circuit is, as shown in Fig. 12, arranged of a comparator 310 for comparing the output of the operating unit 107 with the predetermined potential \( V_a \) and a monostable multi-vibrator 311 for generating a pulse having a predetermined width of time when the output signal of the comparator 310 is shifted to a high level signal.

In an optical disk reproducing apparatus which is operated to obtain a serial reproduction signal given while the reproduction head moves along the information tracks, the foregoing description clearly indicates that the circuit block 114 and the gate circuit 113 are not necessary.

According to the operations of the above circuit components, the output signal of the operating unit 106 is indicated by a broken line of Fig. 6a. After the passage of a minute defective portion, the output signal swiftly follows the normal reproduction signal.

The fourth embodiment of the present invention will be described as referring to the drawings.

Fig. 7 is a block diagram showing a reproducing circuit included in an optical disk reproducing apparatus to which the present invention applies. The same components as those shown in Figs. 1 and 5 have the same reference numbers. In this embodiment, the output signal of the operating unit 106 is guided to one input terminals of the comparator 103 as well as to the comparators 115 and 116. The other input terminals of the comparators 115 and 116 are maintained at the potentials \( V_a \) and \(-V_a\), respectively.

The output signals of the comparators 115 and 116 are made logical OR in a gate circuit 117. The signal made logical OR passes to the analog switches 213 and 223 through a gate circuit 118 for the purpose of opening and closing those switches. A circuit block 114 indicated by a broken line of Fig. 7 operates to generate a control signal for masking the output signal of the gate circuit 117.

In this arrangement, the peak portion of the output signal of the operating unit 106 which is input to the comparator 103 is directly detected for opening or closing the analog switches 213 and 223.

The operation of this arrangement will be described below as referring to Fig. 8. Herein, it is assumed that the analog switches 213 and 223 are normally made open and the resistors have the relations of resistor 211 > resistor 214 and resistor 221 > resistor 224 and the output of the circuit block 114 is normally kept at high level.

A real line indicated by Fig. 8a indicates an output signal of the equalizing circuit 102. This is the same as that shown in Fig. 2a. With respect to this kind of reproduction signal, the output signals of the first envelope detecting circuit 104 and the second envelope detecting circuit 105 are indicated by Fig. 8b and Fig. 8c, respectively. The output signals of the operating units 106 and 107 are indicated by Fig. 8d and Fig. 8e, respectively. The output signal of the gate circuit 117 is indicated by Fig. 8f.

In the A portion of Fig. 8a, the operations of the envelope detecting circuits 104 and 105 are analogous to those of the embodiments shown in Figs. 1 and 2 (output signals (b) and (c)).

In the head portion of B corresponding to the successive minute defect, as indicated by the output signal (b), the output signal (b) of the first envelope detecting circuit 104 indicates an abrupt rise. Hence, the output signal (d) of the operating unit 106 becomes an potential \( V_a \) or higher (D of Fig. 8). At this time, the output signal of the comparator 115 rises to a high level. By shifting the output signal of the comparator 115 to the high level, the analog switches 213 and 223 are made closed through the effect of the gate circuits 117 and 116 so that the charges stored in the capacitor 212 (222) are discharged through the resistors 211 and 214 (221 and 224). Herein, by setting the resistance of the resistor 214 (224) to a far smaller value than that of the resistor 211 (221), like the foregoing embodiment, the time constant on which the charges of the capacitor 212 (222) are discharged in the closed state of the analog switches 213 and 223 is made far smaller.

In the head portion of C for obtaining the original reproduction signal after the passage of the portion for the minute defect, the output signal (b) of the first envelope detecting circuit 104 is started to return to a zero potential on the time constant substantially defined by the resistor 214 and the capacitor 212. At a time, the output signal (d) of the operating unit 106 lowers its signal level to the potential \( V_a \) or less (E of Fig. 8). At this time, the output signal of the comparator 115 indicates a low level. Since the output signal of the comparator 115 indicates the low level, the analog switches 213 and 223 are made open and the overall circuit enters into the normal state.

In a case that the optical disk reproducing apparatus employs a system of dividing the optical disk into a plurality of blocks along the information tracks and recording data for each of the blocks, the reproduction signal converted into an electric signal by the reproduction head is an intermittent signal in each block. If the reproduction signal is intermittent, at the head portion of the reproduction signal, a sag takes place because of the AC coupling at the first stage shown in Fig. 7. Hence, any one of the first envelope detecting circuit 104 and the second envelope detecting circuit 105 abruptly rises or falls and the output signal (d) of the operating unit 106
abruptly rises or falls accordingly. When the output signal of the operating unit 106 reaches a predetermined potential \( V_f \) or more (\(-V_f \) or less), the output signal of the comparator 115 (116) is shifted to a high level so that the gate circuit 117 may output a pulse for closing the analog switches 213 and 223 even in the head portion of the reproduction signal. The circuit block 114 indicated by a broken line of Fig. 7 operates to generate a control signal for removing the pulse signal output from the gate circuit 117 in the head portion of the reproduction signal if the reproduction signal is intermittent. This embodiment has the circuit block 114 arranged like the embodiment shown in Fig. 5.

In an optical disk reproducing apparatus which is operated to obtain a serial reproduction signal given while the reproduction head moves along the information tracks, the foregoing description clearly indicates that the circuit block 114 and the gate circuit 118 are not necessary.

According to the operations of the above circuit components, the output signal of the operating unit 106 is indicated by a broken line of Fig. 8a. After the passage of a minute defective portion, the output signal swiftly follows the normal reproduction signal.

The description will be oriented to a fifth embodiment of the present invention as referring to the drawings.

Fig. 9 is a block diagram showing a reproducing circuit included in an optical disk reproducing apparatus to which the present invention applies. The same components as those shown in Figs. 1, 5 and 7 have the same reference numbers. In this embodiment, the output signal of the operating unit 106 is guided to one input terminals of the comparator 103 as well as the differentiating circuit 119 in which the signal is differentiated. The output signal of the differentiating circuit 119 is guided to one input terminal of the second comparators 120 and 121. The other input terminals of the comparators 120 and 121 are maintained at the potentials \( V_f \) and \(-V_f \) respectively.

The output signals of the comparators 120 and 121 are made logical OR in the gate circuit 122 and the signal made logical OR is guided to the pulse generating circuit 123. The output signal of the pulse generating circuit 123 is led to the analog switches 213 and 223 through a gate circuit 124 for the purpose of opening and closing the analog switches 213 and 223. The pulse generating circuit 123 may be arranged of a retriger-able and monostable multi-vibrator, for example, like the foregoing embodiment. A circuit block 114 indicated by a broken line of Fig. 9 operates to generate a control signal for masking the output signal of the gate circuit 124.

In this arrangement, the differentiation of the output signal of the operating unit 106 stops the adverse effect of an envelope of the reproduction signal. In the optical disk unit, normally, a data train to be recorded is modulated by a modulator before recording it on an information track. In such a data modulation, the frequency band of the reproduction signal is limited so as to match to the frequency characteristic of the transmission path. This modulating system often includes relatively low frequency components. On the other hand, as shown in Fig. 9, such low frequency components are removed by the AC coupling at the first stage of the reproducing system. Hence, the variation of an envelope (variation corresponding to the low frequency components of a bit train to be recorded) takes place.

The operation of this arrangement will be discussed below as referring to Fig. 10. Herein, it is assumed that the analog switches 213 and 223 are normally made open and the resistors have the relations of resistor 211 > resistor 214 and resistor 221 > resistor 224 and the circuit block 114 is normally kept at high level.

A real line shown by Fig. 10a indicates an output signal of the equalizing circuit 102. This is the same as that shown by Fig. 2a. With respect to this type of reproduction signal, the output signals of the first envelope detecting circuit 104 and the second envelope detecting circuit 105 are indicated by Fig. 10b and Fig. 10c, respectively. Further, the output signal of the operating unit 106 is indicated by Fig. 10d and the output signal of the differentiating circuit 119 is indicated by Fig. 10e. Moreover, the output signal of the gate circuit 122 is indicated by Fig. 10f and the output signal of the pulse generating circuit 123 is indicated by Fig. 10g.

The operations of envelope detecting circuits 104 and 105 in the A portion of Fig. 10a are the same as that of the embodiment shown in Figs. 1 and 2 (the output signals (b) and (c)).

In the head portion B corresponding to a successive minute defect, as indicated by the output signal (b), the output signal (b) of the first envelope detecting circuit 104 abruptly rises and the output signal (d) of the operating unit 106 rises accordingly. Hence, as indicated by the output signal (e), the output signal (e) of the differentiating circuit 119 changes according to the rise of the output signal of the operating unit 106 and keeps the signal level at the potential \( V_f \) or higher (D of Fig. 10).

At this time, the output signal of the comparator 120 rises to a high level so that the pulse generating circuit 123 may output a pulse signal (g) having a predetermined time width. While the output signal of the pulse generating circuit 123 is at high level, the analog switches 213 and 223 are made closed so that the charges stored in the capacitor 212 (222) are discharged through the resistors 211 and 214 (221 and 224). Herein, by setting the resistance of the resistor 214 (224) to a far smaller value than that of the resistor 211 (221), like the foregoing embodiment, the time constant on which the charges stored in the capacitor 212 (222) are discharged when the analog switches 213 and 223 are made closed becomes far smaller.

Assuming that the width of a pulse output from the pulse generating circuit 123 is set to a larger value than the time length corresponding to a minute defective por-
tion, in the head portion of C for obtaining an original reproduction signal after the passage of the portion for the minute defective portion, the output signal of the first envelope detecting circuit 104 is started to return to a zero potential with a time constant substantially defined by the resistor 214 and the capacitor 212, resulting in lowering the signal level of the output signal of the operating unit 106 accordingly. Hence, the output signal (e) of the differentiating circuit 119 changes according to the fall of the output signal of the operating unit 106 and keeps its signal level at the potential -Vf or less (E of Fig. 10). At this time, the output signal of the comparator 121 rises to a high level so that the pulse generating circuit 123 may be restarted (output signals (f) and (g)).

By shifting the output signal of the pulse generating circuit 123 to a low level, the analog switches 213 and 223 are made open and the overall circuit enters into the normal state.

Herein, the arrangement and the operation of the circuit block 114 are the same as that of the foregoing embodiment.

According to the operations of the foregoing circuit components, the output signal of the operating unit 106 is indicated by a broken line of the signal (Fig. 10a) and swiftly follows the normal reproduction signal after the passage of the minute defective portion.

According to the present invention, in a peak portion of the envelope caused by a defect or the like, by switching the time constant on which the charges stored in the capacitor of the envelope detecting circuit are discharged, after the reproduction head passes the minute defective portion, it is possible to reduce the time taken in returning to the storage amount of charges corresponding to the amplitude of the original reproduction signal. Therefore, it is possible to prevent a burst-like data error from being caused by a minute defect.

According to the present invention, if an intermittent reproduction signal is input to the AC-coupling reproducing system, it is easier to detect the peak portion of the envelope caused by the minute defect.

According to the present invention, it is possible to reduce the adverse effect given by the variation of an amplitude level of the reproduction signal caused by some factors such as variation between disks, environmental conditions like a temperature, and unevenness of the recording and reproducing characteristics of a disk and easier to detect the peak portion of the envelope caused by a minute defect.

Further, according to the present invention, it is also possible to reduce the adverse effect given by the low frequency components lost by the AC-coupled reproducing circuit system and easier to detect the peak portion of the envelope.

Claims

1. A digital information reproducing apparatus comprising readout means (102) for reproducing information recorded on tracks of a recording medium by scanning the track, first (104) and second (105) envelope detecting circuits for detecting respective polarity envelopes of the reproduced information, an adding circuit (106) for summing the output of said first and second envelope detecting circuits for producing a sum signal and a first comparator (103) for comparing the reproduced information with said sum signal, characterized in that said apparatus further comprises time constant control means (107, 108, 110, 115, 119, 109, 112, 123) for varying the time constants of said first and second envelope detecting circuits including switching circuit (213, 223) for switching the time constants in accordance with a time constant control signal derived from the outputs of both of said first and second envelope detecting circuits.

2. A digital information reproducing apparatus as claimed in claim 1, characterized in that said apparatus further comprises subtracting circuit (107) for deriving a difference signal between the outputs of said first and second envelope detecting circuits and a second comparator (108) for comparing said difference signal with a signal of predetermined level.

3. A digital information reproducing apparatus as claimed in claim 1, characterized in that said apparatus further comprises a second comparator (115) for comparing said sum signal with a signal of predetermined level.

4. A digital information reproducing apparatus as claimed in claim 1, characterized in that said apparatus further comprises subtracting circuit (107) for deriving a difference signal between the outputs of said first and second envelope detecting circuits and a differentiating circuit (110) for differentiating said difference signal.

5. A digital information reproducing apparatus as claimed in claim 1, characterized in that said apparatus further comprises differentiating circuit (119) for differentiating said sum signal.

6. A digital information reproducing apparatus as claimed in any one of claims 2 to 5, characterized in that said apparatus further comprises a pulse generating circuit (109, 112, 123) for outputting said signal.

7. A digital information reproducing apparatus as
claimed in claim 6, characterized in that said pulse generating circuit includes a retriggerable and monostable multi-vibrator (300) which operates at the leading portion of the input signal.

Patentansprüche


3. Wiedergabegerät für digitale Information nach Anspruch 1, dadurch gekennzeichnet, dass die Zeitkonstanten-Steueereinrichtung einen zweiten Komparator (115) zum Vergleichen des Summensignals mit einem Signal von vorbestimmtem Pegel aufweist.

4. Wiedergabegerät für digitale Information nach Anspruch 1, dadurch gekennzeichnet, dass die Zeitkonstanten-Steueereinrichtung eine Subtrahierschaltung (107) zum Herleiten des Differenzsignals zwischen den Ausgangssignalen der ersten und der zweiten Hüllkurven-Erfassungsschaltung sowie eine Differenzierschaltung (110) zum Differenzieren des Differenzsignals aufweist.

5. Wiedergabegerät für digitale Information nach Anspruch 1, dadurch gekennzeichnet, dass die Zeitkonstanten-Steueereinrichtung eine Differenzierschaltung (119) zum Differenzieren des Summensignals aufweist.


7. Wiedergabegerät für digitale Information nach Anspruch 6, dadurch gekennzeichnet, dass die Impulserzeugungsschaltung ein umtriggerbares Monoflop (300) aufweist, das im vorderen Abschnitt des Eingangssignals arbeitet.

Revendications

1. Appareil de reproduction d'informations numériques comprenant des moyens de lecture (102) pour reproduire des informations enregistrées sur des pistes d'un support d'enregistrement par balayage de la piste, des premier (104) et second (105) circuits de détection d'enveloppe pour détecter des enveloppes de polarité respectives des informations reproduites, un circuit de sommation (106) pour additionner la sortie desdits premier et second circuits de détection d'enveloppe afin de produire un signal de somme, et un premier comparateur (103) pour comparer les informations reproduites avec un signal de somme, caractérisé en ce que ledit appareil comprend également des moyens de commande de constantes de temps (107, 108, 110, 115, 119, 109, 112, 123) pour modifier les constantes de temps desdits premier et second circuits de détection d'enveloppe incluant un circuit de commutation (213, 223) pour commuter les constantes de temps en fonction d'un signal de commande de constantes de temps déduit à partir des sorties dudit premier et dudit second circuit de détection d'enveloppe.

2. Appareil de reproduction d'informations numériques selon la revendication 1, caractérisé en ce que lesdits moyens de commande de constantes de temps comprennent un circuit de soustraction (107) pour déduire un signal de différence entre les sorties desdits premier et second circuits de détection d'enveloppe, et un second comparateur (108) pour comparer ledit signal de différence à un signal de niveau prédéterminé.

3. Appareil de reproduction d'informations numériques selon la revendication 1, caractérisé en ce que
lesdits moyens de commande de constantes de temps comprennent un second comparateur (115) pour comparer ledit signal de somme à un signal de niveau prédéterminé.

4. Appareil de reproduction d'informations numériques selon la revendication 1, caractérisé en ce que lesdits moyens de commande de constantes de temps comprennent un circuit de soustraction (107) pour déduire un signal de différence entre les sorties desdits premier et second circuits de détection d'enveloppe, et un circuit différentiateur (110) pour différentier ledit signal de différence.

5. Appareil de reproduction d'informations numériques selon la revendication 1, caractérisé en ce que lesdits moyens de commande de constantes de temps comprennent un circuit différentiateur (119) pour différentier ledit signal de somme.

6. Appareil de reproduction d'informations numériques selon l'une quelconque des revendications 2 à 5, caractérisé en ce que lesdits moyens de commande de constantes de temps comprennent également un circuit de génération d'impulsions (109, 112, 123) pour délivrer ledit signal de commande de constantes de temps.

7. Appareil de reproduction d'informations numériques selon la revendication 6, caractérisé en ce que ledit circuit de génération d'impulsions inclut un multivibrateur monostable redéclenchable (300) qui fonctionne sur la portion avant du signal d'entrée.
Fig. 2
Fig. 4
Fig. 6

- a
- b
- c
- d
- e
- f
Fig. 8
Fig. 10
Fig. 11

MONOSTABLE MULTI-VIBRATOR

300

Fig. 12

MONOSTABLE MULTI-VIBRATOR

310
311