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- (71) Applicant: ROXAR FLOW MEASUREMENT AS  
[NO/NO]; P.O.Box 112, NO-4065 Stavanger (NO).
- (72) Inventor: SVIHUS, Jone; Postveien 151a, 4307 Sandnes (NO).
- (74) Agent: PROTECTOR IP CONSULTANTS AS; Oscarsgate 20, NO-0352 Oslo (NO).
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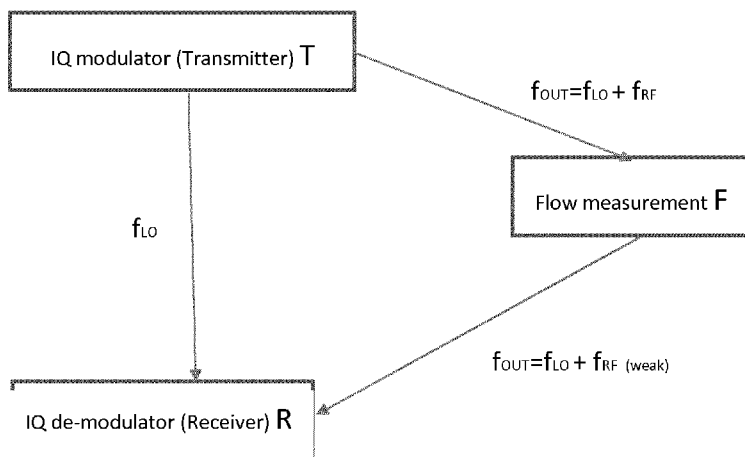


Fig. 4

(57) Abstract: The present invention relates to a measuring system for measuring dielectric properties of a multiphase fluid flow in a pipe. The system includes a microwave signal generator connected to the multiphase fluid flow for transmitting signals within a predetermined frequency range into said flow, a signal receiver adapted to receive signals within said range from said flow, and analyzing means for calculating the dielectric properties based on the transmitted and received signals. The signal generator includes: an IQ modulator being coupled to a high frequency oscillator and a low frequency synthesizer, the high frequency oscillator generating a high frequency signal  $f_{LO}$ , and the low frequency synthesizer is adapted to generate signals  $f_{RF}$  at a number of frequencies within a chosen frequency range, said IQ-modulator being adapted to generate an output signal  $IQ_{OUT}$  constituted by the combination of said signals from the oscillator and the synthesizer, the predetermined frequencies of the oscillator and the synthesizer being chosen based on the required output frequency range for the IQ-modulator.



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## FLOW MEASURING SYSTEM

The present invention relates to a measuring system for measuring dielectric properties of a multiphase fluid flow in a pipe. The system includes a microwave signal generator  
5 for transmitting signals within a predetermined frequency range into said flow and a signal receiver adapted to receive signals from said flow and analyzing means for calculating the dielectric properties.

There are many types of microwave sensors in use in meters for various kinds of  
10 applications. In a meter there is always an electronics module connected to the sensor for measuring the properties of the sensor. The present invention relates to the use of resonator sensors within or close to the microwave range. The use of microwave resonators for measuring fluid characteristics in flow meters is well known in the oil and gas industry, for example as discussed in WO2014/122093 and WO2016/169847  
15 providing salinity measurements in a fluid flow.

As an example a resonator sensor displays a resonance peak, when the insertion loss is measured. The peak has two main properties, which are the primary measurement results, the resonant frequency and the Q-factor, i.e. at which frequency the peak  
20 appears and how wide or sharp it is. The electronics needs to be able to find the peak within a given frequency range, which might be quite broad, e.g. an octave or more. In on-line applications the peak might in addition change position and shape quite fast. Then the measurement must be performed so fast that it will in practice represent an instantaneous snapshot of the peak. It is therefore an object of the present invention to  
25 provide a sufficiently fast measurement. This is obtained by utilizing a system as described in the accompanying claims.

The most accurate method of measuring a resonance peak is to sample the insertion loss at a number of frequencies, and perform a curve-fit to find the resonant frequency and  
30 Q-factor. But if the frequencies are sampled one by one, the movement of the peak during the sampling process creates an error especially in the Q-factor. The sampling process must be fast compared to the dynamics in the resonance peak. Another

alternative is to use a multi-tone signal and sample all frequencies simultaneously, as enabled by one embodiment of this invention.

Another type of microwave sensor is a transmission sensor. Then a signal is transmitted  
5 from one small antenna, often called a probe, to another one. On the path the signal passes through the medium to be measured and both the phase shift and the attenuation of the signal are affected by the properties of the medium. Also in this case the properties to be measured may vary fast in on-line applications. In addition, it is often of interest to measure at several, often widely separated, frequencies to gather information  
10 on frequency-dependent properties. The phase shift and/or the attenuation may therefore be measured at more than one frequency. Also in this case it may be an advantage to sample all the measurement frequencies simultaneously, or within a very short time interval, to capture data at all the frequencies corresponding to the same measurement situation. E.g. if the sensor is used for measuring a flow of oil, water and gas in a pipe,  
15 and the flow moves so fast that while the measurement at one frequency is performed when a gas slug passes the sensor, the measurement at another frequency is performed when there is substantially more liquid at the sensor, the two measurements correspond to different situations and are not directly comparable. In such cases averaging over a much longer time is needed to get useful results. If all the measurement frequencies can  
20 be sampled simultaneously, or almost simultaneously, as enabled by one embodiment of the invention, this improves the quality of the results, and shortens the time to reach a high accuracy for the measured and derived properties, which may be e.g. the salinity of the water component, the GVF (Gas Void Fraction), or the water to oil ratio in a flow of oil, water and gas.

25

Therefore one aspect of the present invention is aimed at detecting the dielectric properties of a flow by using an IQ converter to combine a high frequency range signal from a high frequency generator such as a PLL with an easy to manipulate lower frequency signal from a frequency synthesizer such as a DAC (or DDS) or similar  
30 electronic circuit for generating any of a range of frequencies from a single fixed timebase or oscillator, to generate an easy to manipulate, high frequency range signal. The high frequency signal is coming from the PLL (up to several GHz for the use in

multiphase flow meters) but as it is difficult to manipulate this signal a lower frequency signal from a DAC is provided manipulating the higher frequency signal using a IQ modulator.

- 5 The invention will be described more in detail below with reference to the accompanying drawings illustrating the invention by way of examples.  
 Figure 1a-h illustrates the details of IQ-modulators used according to the invention.  
 Figure 2a illustrates a first embodiment of the invention.  
 Figure 2b illustrates an example of an IQ de-modulator  
 10 Figure 3a, b illustrates and alternative embodiment of the invention.  
 Figure 4 Illustrates a measuring system including the resonator to be monitored.

As illustrated in figures 1a-h an IQ modulator is based on mixing signals. Figure 1a shows a standard frequency multiplier M where two sinus signals I, LO<sub>i</sub> mixed together  
 15 will form two new signals I·LO<sub>i</sub> with respectively the sum of and the difference between the two signals given by:

$$\begin{aligned} I \cdot LO_i &= A \cdot (\sin(\omega_1 t) \sin(\omega_2 t)) \\ &= \frac{A}{2} \cos(\omega_1 - \omega_2)t - \frac{A}{2} \cos(\omega_1 + \omega_2)t \end{aligned}$$

- 20 In figure 1b the amplitude vs frequency is illustrated and as can be seen in the spectrum the sum of the frequencies,  $\omega_1 + \omega_2$ , has negative amplitude while the difference,  $\omega_1 - \omega_2$ , has a positive amplitude.

If both signals I and LO<sub>i</sub> are shifted 90 degrees ( $\sin + 90$  degrees is  $\cos$ ), it changes to  
 25 the situation as is shown in figures 1c and 1d, where Q and LO<sub>q</sub> are mixed shifting the sum of the frequencies,  $\omega_1 + \omega_2$ , to be positive. Thus as illustrated in figure 1d.

$$\begin{aligned} Q \cdot LO_q &= A \cdot \cos(\omega_1 t) \cdot \cos(\omega_2 t) \\ &= \frac{A}{2} \cos(\omega_1 - \omega_2)t + \frac{A}{2} \cos(\omega_1 + \omega_2)t \end{aligned}$$

Since the goal of an IQ modulator is to mix two signals and achieve a signal with a single frequency, the above circuits are combined and a summing device S is added. Both multipliers M and the summer S are basic electronic devices that will not be  
 5 described here.

Figure 1e illustrates a basic sketch for a mixer where the I and Q signals are mixed with LO<sub>i</sub> and LO<sub>q</sub> signals, respectively, and the outputs from these mixers are added to provide the sum  $Q \cdot LO_q + I \cdot LO_i$ . The result of this operation may be expressed as  
 10

$$\begin{aligned}
 Q \cdot LO_q + I \cdot LO_i &= A \cdot \cos(\omega_1 t) \cdot \cos(\omega_2 t) \\
 &+ A \cdot \sin(\omega_1 t) \cdot \sin(\omega_2 t) \\
 &= \frac{A}{2} \cos(\omega_1 - \omega_2)t + \frac{A}{2} \cos(\omega_1 + \omega_2)t \\
 &+ \frac{A}{2} \cos(\omega_1 - \omega_2)t - \frac{A}{2} \cos(\omega_1 + \omega_2)t \\
 &= A \cdot \cos(\omega_1 - \omega_2)t
 \end{aligned}$$

And can be seen both vector representation in figure 1f. One of the frequency components is cancelled and the output signal consist of a single frequency  $\omega_1 - \omega_2$  equal  
 15 to the difference between the frequencies of the input signals.

This principle is widely used in for instance communication systems, in most systems the I and Q signals are low frequency signals and the LO<sub>i</sub> and LO<sub>q</sub> are at a much higher frequency. Also the I and Q signals are complex signals (multiphase and multi  
 20 frequency) and the LO<sub>i</sub> and LO<sub>q</sub> are single frequency carrier signals. If the I and Q have several frequencies, each of the frequencies needs to be shifted 90 degrees. Since the I and Q signals in most applications are complex signals with different phases and frequencies they are generated with DAC's combined with a processing unit (for instance a DSP, Digital Signal Processor). Provided that the DAC's are synchronized  
 25 (using a single clock) a processor can calculate both the I signal and the corresponding 90 degree shifted Q signal for all the different frequencies in the I signal.

The generation of the LO<sub>i</sub> and LO<sub>q</sub> signals (to be mixed with the I and Q signals respectively) from one LO signal may be performed in different ways, where figure 1g illustrates the use of an RC filter. If the two R and C's are matched the two signal will have exactly 90 degrees phase difference, the amplitude will however differ which can  
5 be restored by running both the LO<sub>i</sub> and LO<sub>q</sub> through a comparator and generate a square wave with a set amplitude.

Figure 1h illustrates an alternative way to generating the LO<sub>i</sub> and LO<sub>q</sub> signals using a frequency divider. Starting with a frequency twice the desired frequency  $f$  and inverting  
10 it will result in a 90 degrees phase shift when both signal are used to trigger a frequency divider. In most applications the starting point is a differential LO signal and two identical frequency dividers, cross connecting the positive and negative part of the LO signal on the two frequency dividers enables a precise 90 degree phase shift.

15 Figure 2a shows an IQ modulator setup based on the discussion related to figures 1e and 1h. The illustrated embodiment includes a PLL, preferably capable of covering a wide frequency range, an IQ modulator and two DAC's (Digital to Analog Converter) I<sub>DAC</sub>, Q<sub>DAC</sub>. In the shown embodiment the signal from the PLL is divided as illustrated in figure 1h in order to produce the two LO inputs, LO<sub>i</sub>, LO<sub>q</sub> to the IQ modulator.

20

The PLL in figure 2a may provide a frequency within a wide range from a few megahertz up to several Gigahertz, but the frequency cannot change instantly (typically 1  $\mu$ s to settle on a new frequency) and the PLL cannot be used to generate multi-tunes. The DAC's, however, are generating a lower frequency signal typically from a few  
25 hundred kilohertz up to some 10's of Megahertz (limited by the clock frequency of the ADC's in the IQ demodulator, for example 30 MHz).

The advantage of using the two DAC's is that they can generate several tunes  $f_{RF}$  at the same time and even white noise if needed, provided that the DAC's are clocked by a  
30 frequency much higher than the frequency of the output signal, for example 300 MHz. Note that a DDS (Direct Digital Synthesis) is also considered to be a DAC (a DDS is a DAC with a dedicated control system).

In figure 2a the output from the DAC's is illustrated as a signal with number of distinct frequency components  $f_{RF}$  at a relatively low frequency compared to the PLL signal. The IQ modulator needs both the I and Q version of both the signal from the PLL( $LO_I$   
5  $LO_Q$ ) and from the DAC's ( $RF_I$ ,  $RF_Q$ ) and provides signal  $IQ_{OUT}$  at a corresponding number of output frequencies with known amplitudes that are applied to the microwave sensor, in this example illustrated as a resonator. This is illustrated with the  $f_R$ , a multitone signal at a high frequency. This process is usually referred to as an IQ modulation or a Single Side Band mixing (SSB).

10

According to the invention, when tracking a resonator, the PLL in this embodiment is aiming to find the resonance frequency  $\omega_R$  of the resonator in a certain range by applying signals at chosen frequencies within the range expected to have a resonance peak. Using an IQ modulator generating the signals and then measuring the amplitude  
15 and the phase of the signals from the resonator the information can be used to determine the frequency of the resonance peak without sweeping over the frequency range.

20

This may be used to provide an instant measurement to determine the peak and possibly the Q-factor of the resonance  $\omega_R$  by means of curve fitting the information based on the  
20 amplitudes of the multiple frequency points  $f_R$  illustrated in figure 2a as the Resonator<sub>out</sub>. In the drawing four frequencies are used but other numbers may be used depending on the required accuracy in the calculated resonance frequency and the characteristics of the DACs and receiver system.

25

As an alternative to the plurality of fixed frequencies a fast sweeping over a limited frequency range (up to a few MHz) may be performed using the DAC's for providing the varying frequency, while the slower PLL may provide a constant base frequency. This way it is possible to make sure that one of the frequencies moves into the resonance frequency, thus being detected at the receiver means.

30

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Using the limited frequency sweep the responsiveness of the system may be increased, but will require an increase in the processing power and advantageously increased amplification of the measured signals since:

$$P_{total} = \sqrt{P_{f1} + P_{f2} + P_{f3} \dots \dots \dots}$$

- 5 There are 4 different frequencies in figure 2a, in reality it is also possible to use more, such as 20, 100 or even white noise, depending on the device being measured. Depending on how well known the shape of the resonance peak is it may also be possible to use three frequencies.
- 10 The  $f/2$  part in the illustrated example is used to divide the frequency in order to generate two signals with identical frequency and these signals needs to be exactly 90 degrees phase shifted to suppress the mirror signal (if we are generating  $f1 + f2$  the mirror signal has  $f1 - f2$ ). If the LO signal goes to 2 different circuits, one divides the frequency in 2 (a D flip-flop for instance), the other inverts the signal and then divides
- 15 the frequency, then the result is 2 signals 90 degrees phase shifted, as is described in figure 1h. The alternative is an RC network as illustrated in figure 1g, but this is less precise.

As an alternative to DAC's a Digital Direct Synthesis (DDS) with I and Q output exists,

20 being able to be programmed to produce frequency patterns that could be used directly.

In figure 3a an embodiment is shown using the IQ modulator for rapid scanning of a large frequency range. A wide band PLL is generally slow, after a frequency shift the PLL can take typically 1  $\mu$ s to stabilize in the new frequency setting. Rapid scanning of

25 a wide frequency range with high resolution is therefore a slow process with a PLL. A DAC on the other hand (or a DDS which is practically the same) changes frequency instantaneously and this allows for a rapid, high resolution scan. The disadvantage of using a DAC is that the clock frequency needs to be many times higher than the output frequency in order to avoid dominant spurious frequencies. This limits the max

30 frequency in these components and makes it hard to use them in the GHz range.

This embodiment of the invention allows for both rapid, high resolution scanning and high frequency output signals with low spurious levels. The PLL produces a high frequency signal, (GHz range) and this is added with the lower frequency signal from the 2 DAC's (MHz range). The 2 DAC's produces identical signals with a 90 degrees phase between them (I and Q), this allows for the frequency of the signal from the PLL (5  $f_{LO}$ ) and the frequency of the signals from the DAC's ( $f_{RF}$ ) to be added producing a signal with a frequency equal to the sum of frequency of  $f_{RF}$  and  $f_{LO}$ .

By using the DAC's to scan between 0 and the max DAC frequency  $f_{RF\ max}$  we can perform a rapid scan around the PLL frequency  $f_{LO}$ . (changing the order of the I and Q signal from the DAC will cause the frequency to be subtracted instead of added to the PLL frequency). After this operation we will have scanned (rapidly with high resolution) a band around the PLL frequency. To continue the scanning the PLL frequency may be increases with  $2 \times f_{RF\ max}$ . As illustrated in figure 3b a scan is performed from  $f_{LO1}-f_{RF\ max}$  to  $f_{LO1}+f_{RF\ max}$  by changing the frequency from the DAC but without changing the PLL frequency ( $f_{LO1}$ ), after the scan from  $f_{LO1}-f_{RF\ max}$  to  $f_{LO1}+f_{RF\ max}$  has been completed the PLL frequency is increased with  $2 \times f_{RF\ max}$  to  $f_{LO2}$ , then the band from  $f_{LO2}-f_{RF\ max}$  to  $f_{LO2}+f_{RF\ max}$  is scanned so that all intermediate frequencies between  $f_{LO1}-f_{RF\ max}$  and  $f_{LO2}+f_{RF\ max}$  are covered.

20

The advantages with this system are as follows. Few and large steps may be provided by the PLL, and therefore slow settling of the PLL will have less impact on the scanning speed. The DAC's (or DDS) may be operated with a high ratio between the clock signal and the RF signal, securing low spurious levels. The PLL signal can be used in the demodulator ensuring that there is a frequency difference between the signal used for measurement and the reference signal this will help avoiding leakage.

25

In figure 4 an implementation of the system is illustrated. The output from the IQ modulator  $f_{OUT}=f_{LO}+f_{RF}$  is transmitted to the resonator in the flow measuring system, e.g. of the types discussed in WO2014/122093 and WO2016/169847. The output from the resonator  $f_{OUT}=f_{LO}+f_{RF(WEAK)}$  is transmitted to an IQ demodulator of any per se known type An example of an IQ de-modulator is illustrated in figure 2b. Figure 2a

30

shows a IQ modulator and a resonator, figure 2b illustrates how the signals coming out of the resonator can be demodulated. The IQ modulator up converts the signals to the frequency needed to analyze the resonator, the IQ demodulator down converts the signals to a frequency that within the range of an ADC. The I and Q ADC's provides  
5 information containing the phase and gain of the frequencies in the signal to a processing unit that can extract this. The IQ demodulator also receives the LO signal as a reference from the IQ modulator. As the frequency of the attenuated output signal from the resonator is different from the reference signal leakage between the reference and measuring signals is avoided.

10

To summarize the invention relates to a measuring system for measuring dielectric properties of a multiphase fluid flow in a pipe. The system including a microwave signal generator connected to the multiphase fluid flow for transmitting signals within a predetermined frequency range into said flow through an antenna or similar. In a similar  
15 way the signal receiver may include an antenna or similar and is adapted to receive signals within said range from said flow. It also includes analyzing means for calculating the dielectric properties based on the transmitted and received signals. It should be noted that the "microwave range" in this case is meant in a broad sense covering the frequency range in or close to the usual definition, as ranges needed for  
20 measuring the dielectric characteristics may vary. Also, the phrase multiphase fluid may also include wet gas.

The signal generator unit used according to the invention for generating high frequency signals for a flow meter including a high frequency oscillator such as a Phase Locked  
25 Loop (PLL) oscillator generating a first, high frequency signal  $f_{LO}$ , the output of the oscillator being connected to the first I and Q inputs of an IQ modulator. The system also comprises a low frequency oscillators constituted by a frequency synthesizer adapted to generating signals in several frequencies sequentially or simultaneously within a chosen frequency range  $f_{RF}$ , preferably within a short time window so as to  
30 sample a measurement in the frequency range within the same flow volume, the duration of the time window depending on the known flow velocity. The output signal frequency  $f_{RF}$  of the frequency synthesizer is significantly lower than the high frequency

signal  $f_{LO}$ , each being coupled the I and Q modulator inputs of the IQ modulator. The IQ-modulator thus generates output signals constituted by the mix of the input signals, i.e. a sum or difference between the high frequency signals  $f_{LO}$  and the low frequency signal  $f_{RF}$  at the modulator input, wherein the output range of the IQ modulator is  
5 defined by a range above and below the frequency of the oscillator.

The system according to the invention may therefore provide a rapid frequency scan over a predetermined frequency range where the speed of the scan is driven by the frequency synthesizer and the carrier frequency is given by the oscillator. The scanning  
10 may alternatively be provided in two frequency ranges simultaneously as the frequency synthesizer may be capable of operating at more than one frequency at the same time.. Also it is possible to extend the scanning range by changing the first, oscillator frequency

15 The pipe containing the flow may constitute or include a resonator within the frequency range. As an example frequency synthesizer may be adapted to emit signals at at least three fixed frequencies having known amplitudes, the receiver means being adapted to analyze the amplitudes of the received signal at each frequency and finding resonance frequency of the resonator by interpolating the amplitude curve between said  
20 frequencies. Alternatively the synthesizer may be adapted to emit a rapid frequency scan having known amplitude and the analyzing means may analyse the signal so as to find resonance frequency. In both cases the Q-factor may also be calculated.

When used in a microwave resonator measuring system the generator is adapted to emit  
25 the output signal into a resonator for finding the resonance frequency of the resonator. In a preferred embodiment the output of said IQ modulator may include at least four different frequencies. The system also comprising receiver means for measuring the amplitude of the signals at said output frequencies and analyzing means for calculating the resonance frequency and preferably also the Q factor based on the amplitude and  
30 frequencies of the received signals.

Alternatively, the IQ modulator output signal including a signal having a varying frequency, the system including a receiver means and an analyzing means for analyzing the amplitudes of the received signal and finding the resonance frequency and preferably also the Q-factor of the resonator.

5

The system according to the invention may also be adapted to emit the output of the signal generator into an antenna in a transmission sensor system, measuring the received signal at another antenna, the IQ modulator output signal including a signal having a varying frequency, the system including a receiver means and an analyzing means for  
10 analyzing e.g. the phase shift and attenuation between the two antennas. The properties of the medium between the antennas can thereby be measured over a broad frequency range within a very short time window.

C l a i m s

1. Measuring system for measuring dielectric properties of a multiphase fluid flow in a pipe, the system including a microwave signal generator connected to the multiphase fluid flow for transmitting signals within a predetermined frequency range  
5 into said flow, signal receiver adapted to receive signals within said range from said flow, and analyzing means for calculating the dielectric properties based on the transmitted and received signals, wherein the signal generator includes: an IQ modulator being coupled to a high frequency oscillator and a low frequency synthesizer,  
10 the high frequency oscillator generating a high frequency signal  $f_{LO}$ ,  
the low frequency synthesizer is adapted to generate signals  $f_{RF}$  at a number of frequencies within a chosen frequency range, said IQ-modulator being adapted to generate an output signal  $IQ_{OUT}$  constituted by the combination of said signals from the oscillator and the synthesizer, the predetermined frequencies of the oscillator and the synthesizer being chosen based on the required output frequency  
15 range for the IQ-modulator.
2. System according to claim 1, wherein the frequency synthesizer simultaneously emits at least two different frequency signals, the output of the IQ modulator thus being constituted by a mix of said high frequency signal and said signals  
20 from the second oscillator.
3. System according to claim 1, wherein the frequency synthesizer a sequence of different frequencies within said range.
- 25 4. System according to claim 3, wherein said sequence constitutes a frequency scan over a predetermined frequency range.
5. System according to claim 1, wherein said frequency synthesizer is a DAC adapted to control the output frequency and amplitude..  
30
6. System according to claim 1, wherein said pipe constitutes a resonator in said frequency range, where said frequency synthesizer is adapted to emit signals at at

least three fixed frequencies having known amplitudes, the receiver means being adapted to analyze the amplitudes of the received signal at each frequency and finding resonance frequency of the resonator by interpolating the amplitude curve between said frequencies , the analyzing means being adapted to calculate the dielectric properties on this basis..

7. System according to claim 1, wherein said pipe constitutes a resonator in said frequency range, where said frequency synthesizer is adapted to emit a rapid frequency scan having known amplitude, the receiver means being adapted to analyze the received signal so as to find resonance frequency of the resonator, the analyzing means being adapted to calculate the dielectric properties on this basis..

8. System according to claim 6 or 7, wherein the Q-factor of the resonance peak at said resonance frequency is calculated.

9. System according to claim 7, wherein the oscillator is adapted to change first frequency, thus allowing to change the scanning range of the signal generator.

10. System according to claim 1, wherein the receiver means is adapted to calculate the transmission time and attenuation of the transmitted signal through the flow, the analyzing means being adapted to calculate the dielectric properties on this basis.

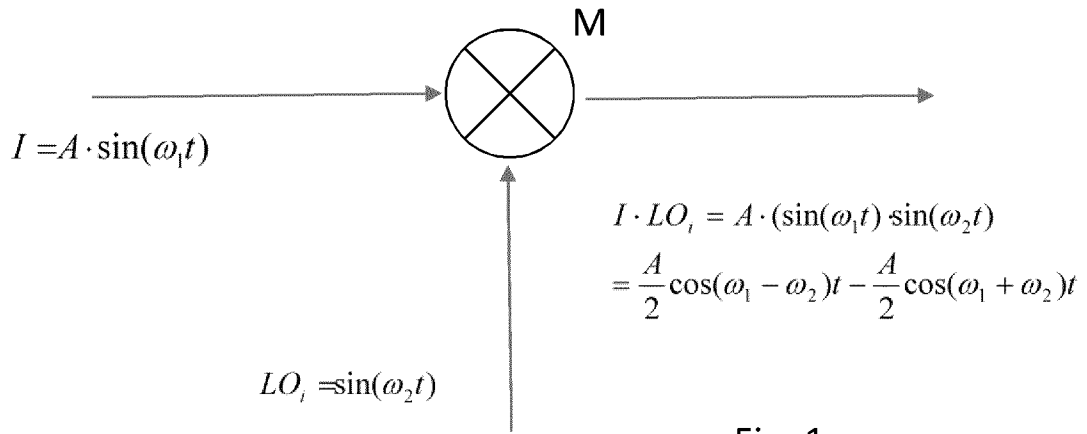


Fig. 1a

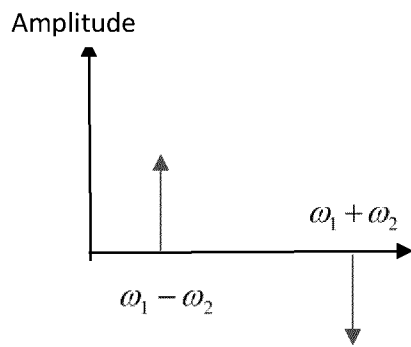


Fig. 1b

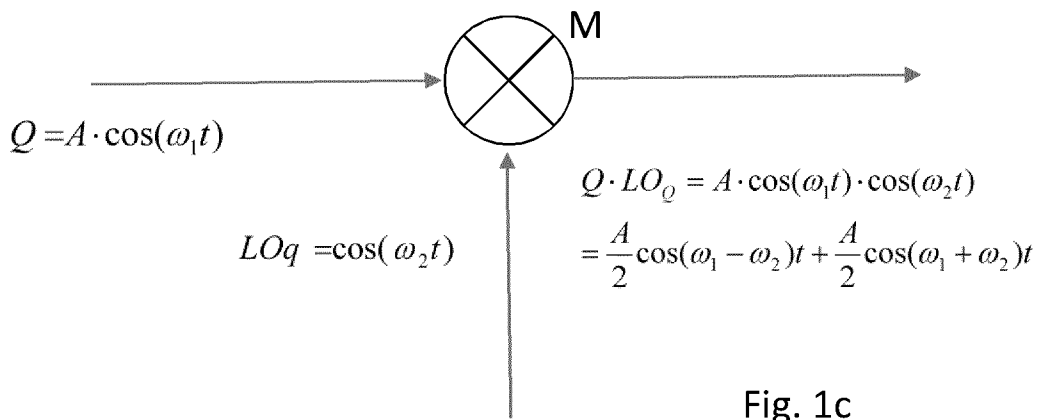


Fig. 1c

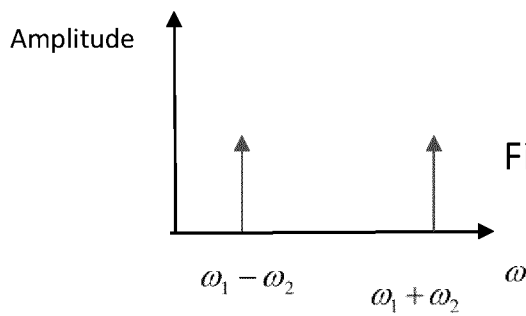
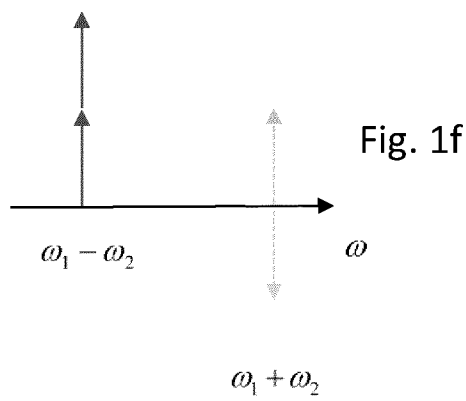
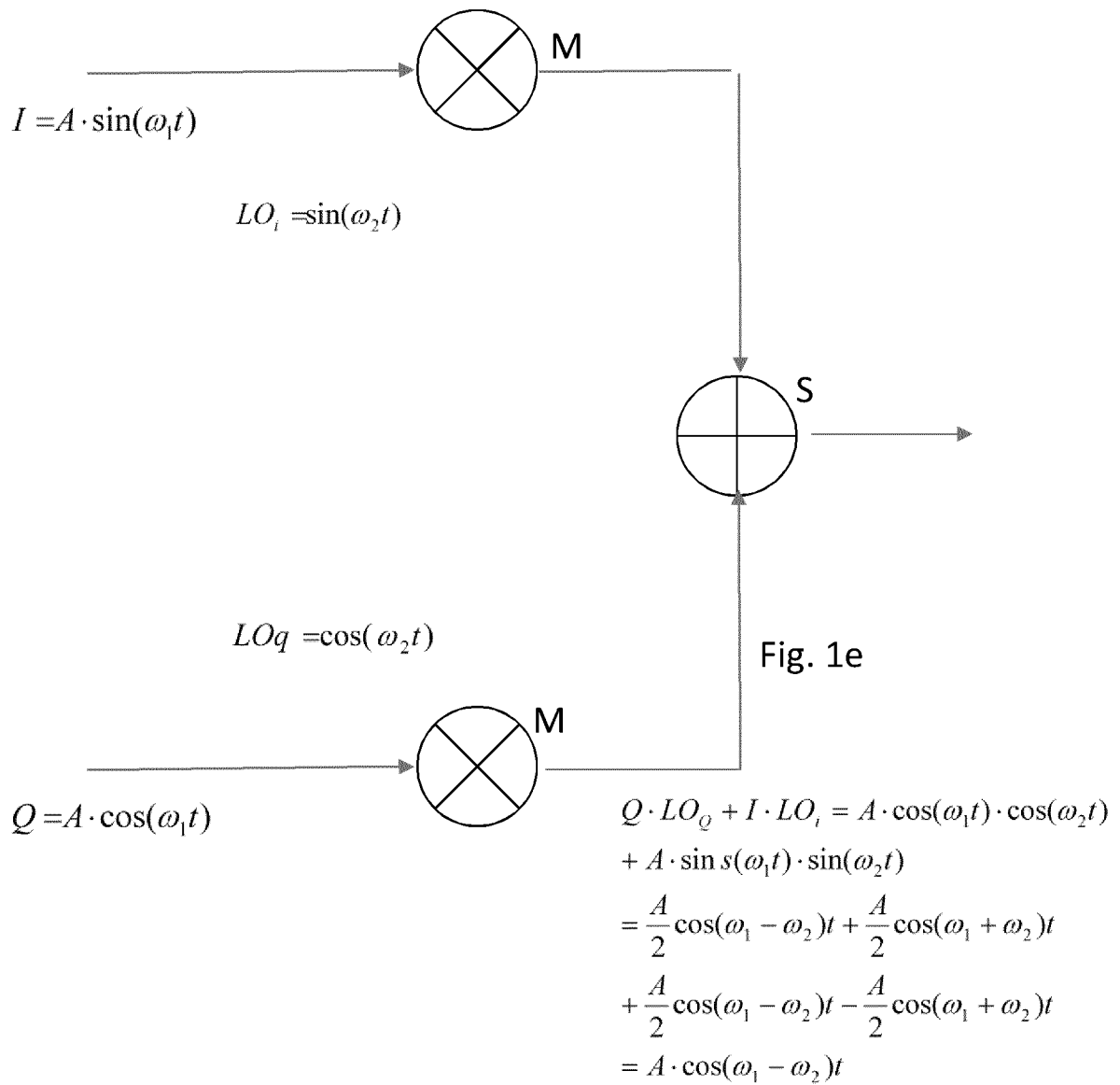


Fig. 1d





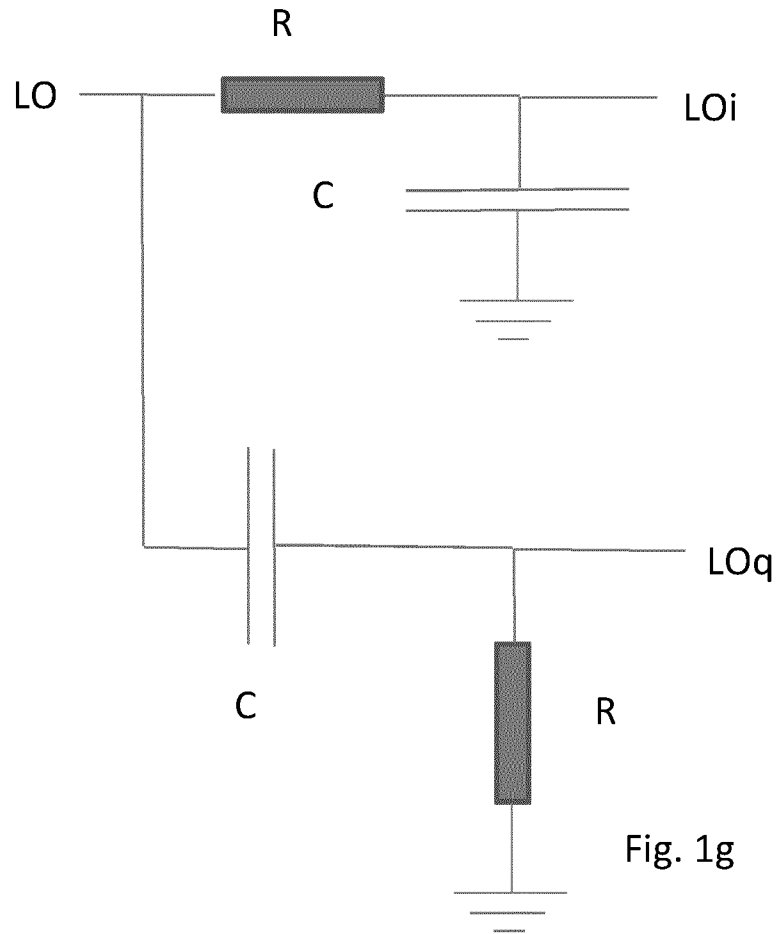


Fig. 1g

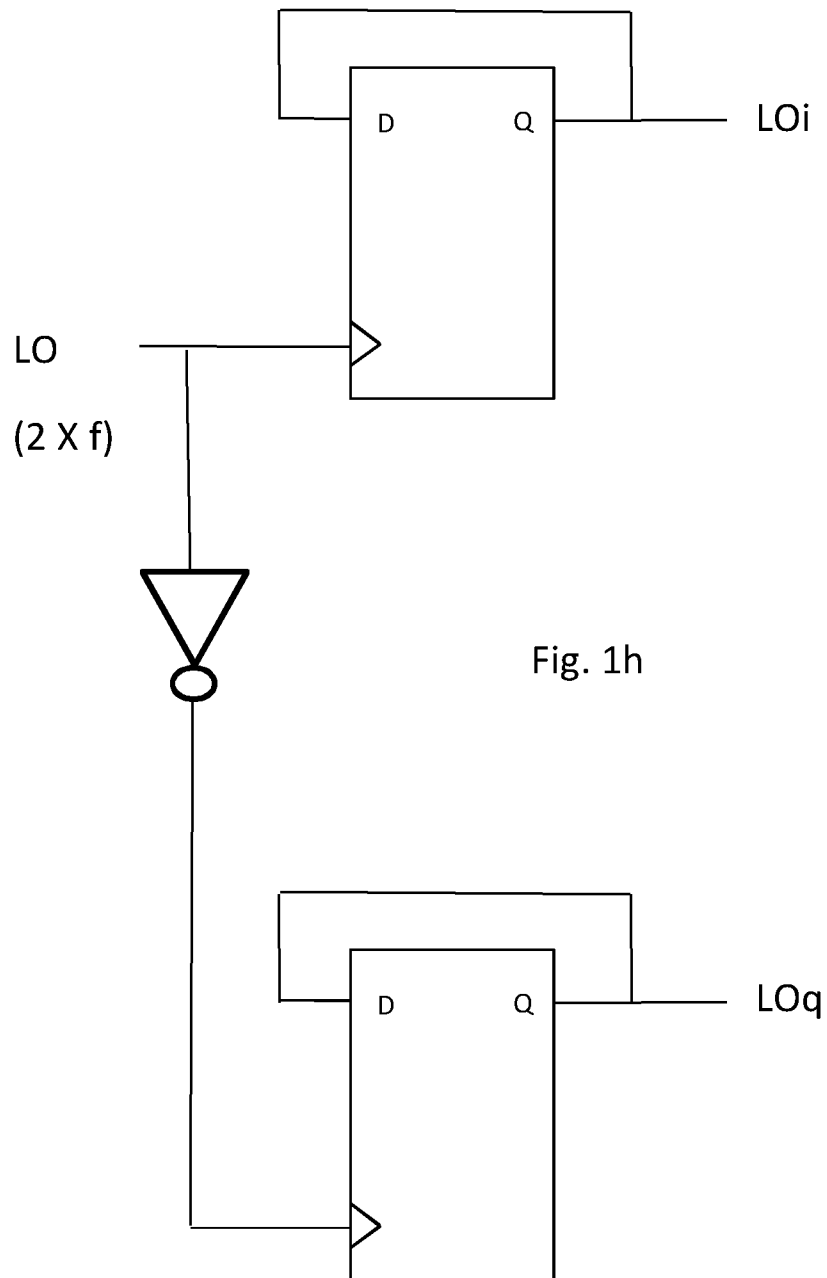


Fig. 1h

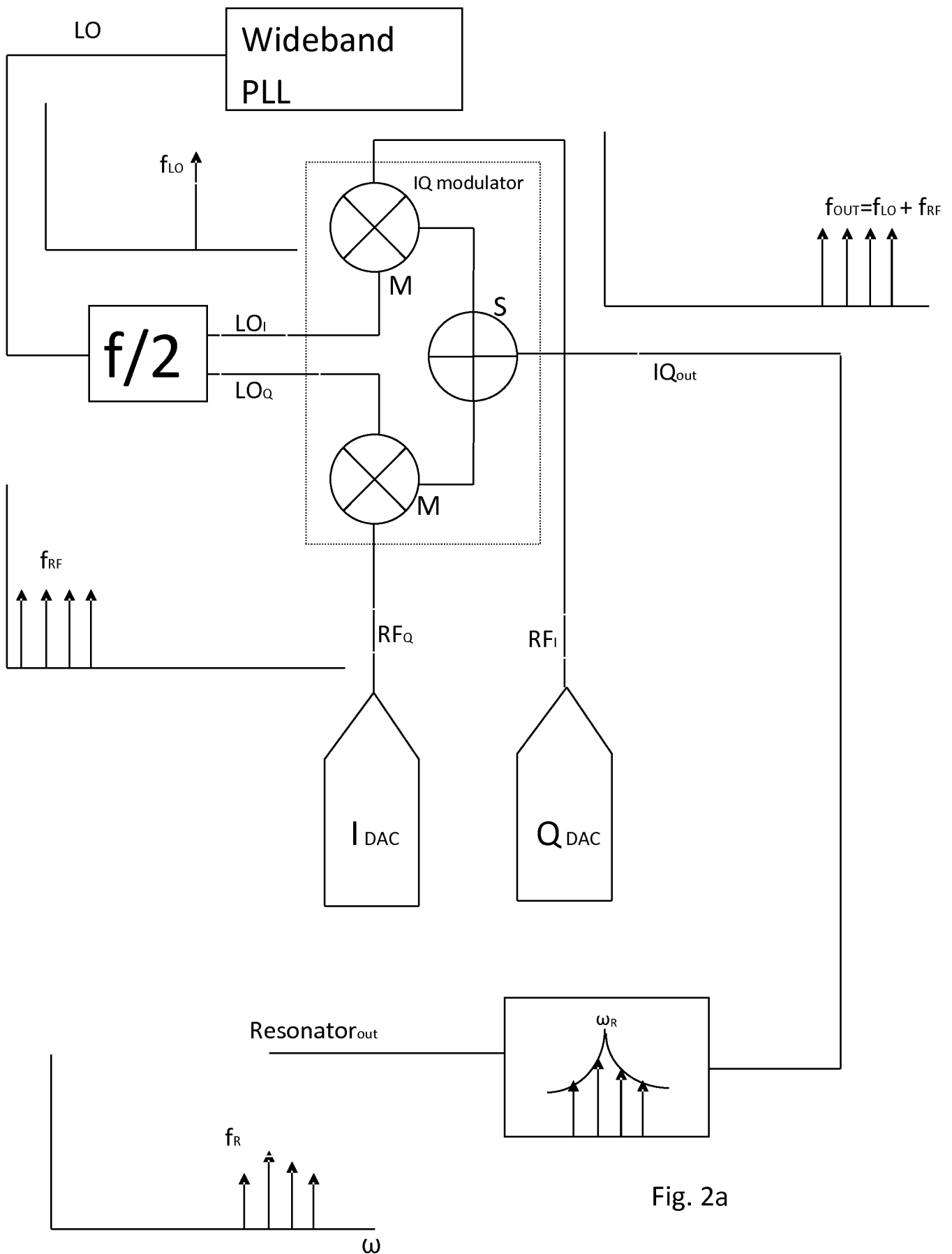


Fig. 2a

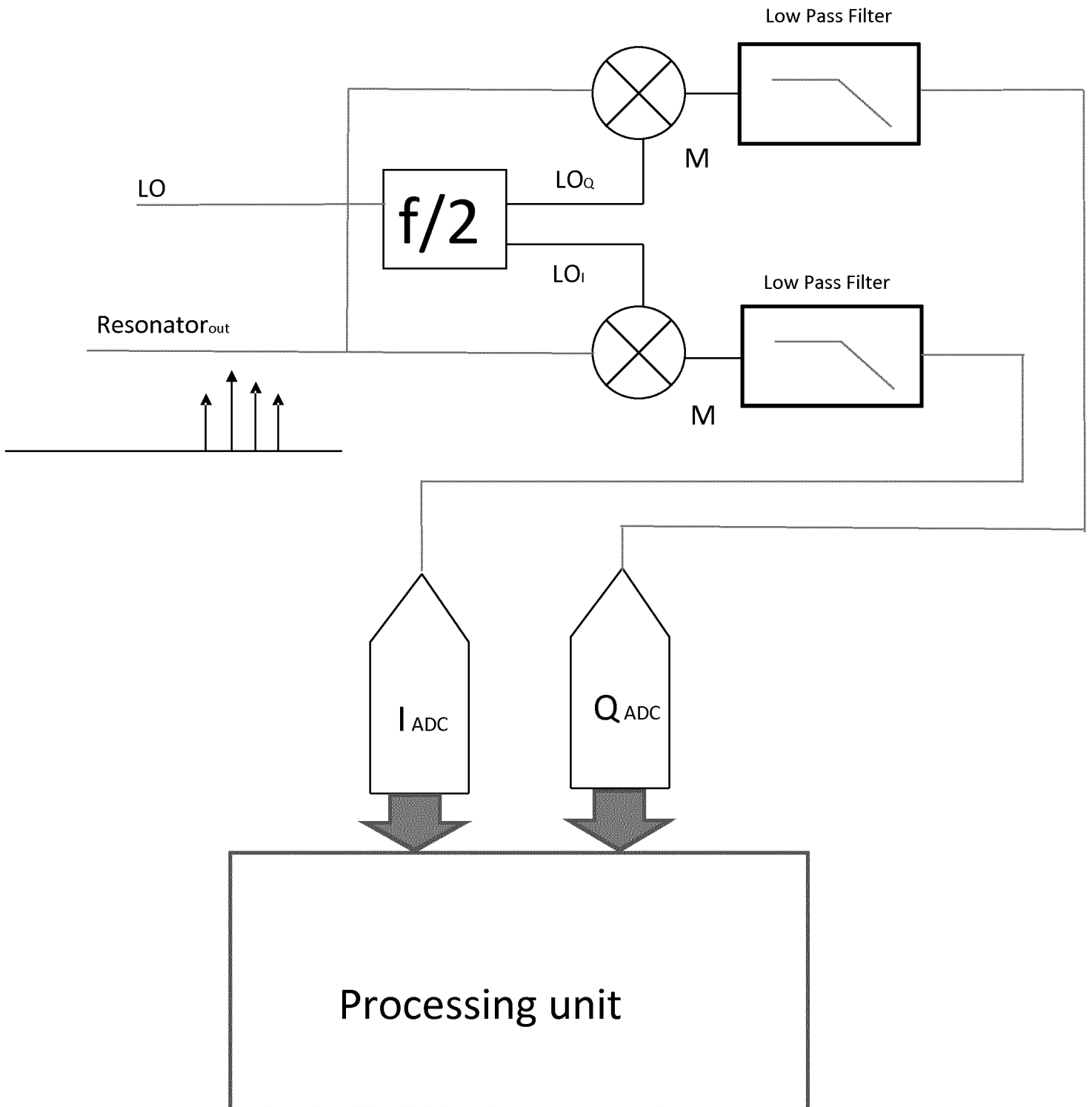


Fig. 2b

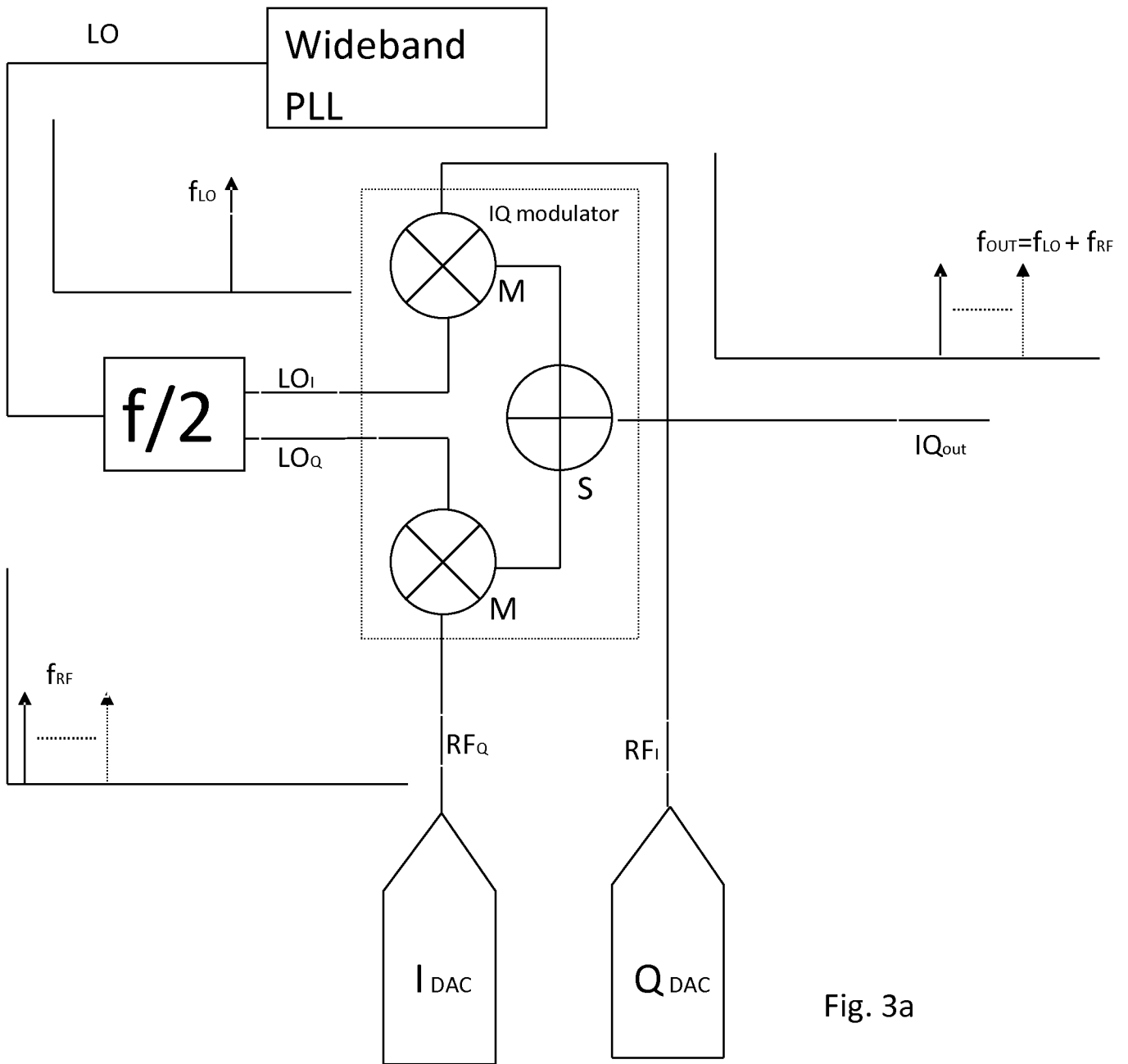


Fig. 3a

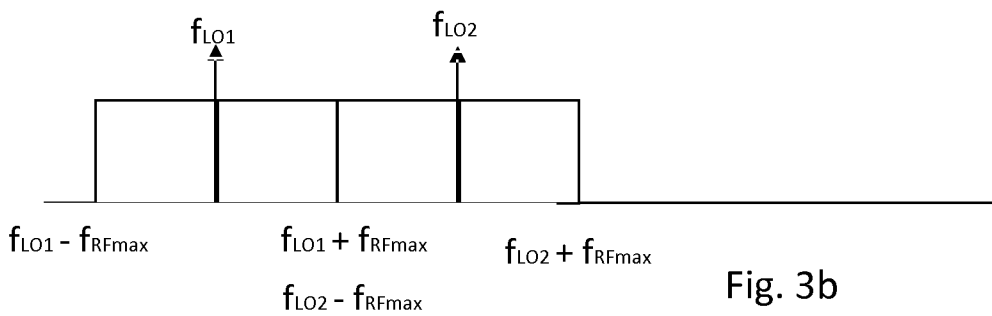


Fig. 3b

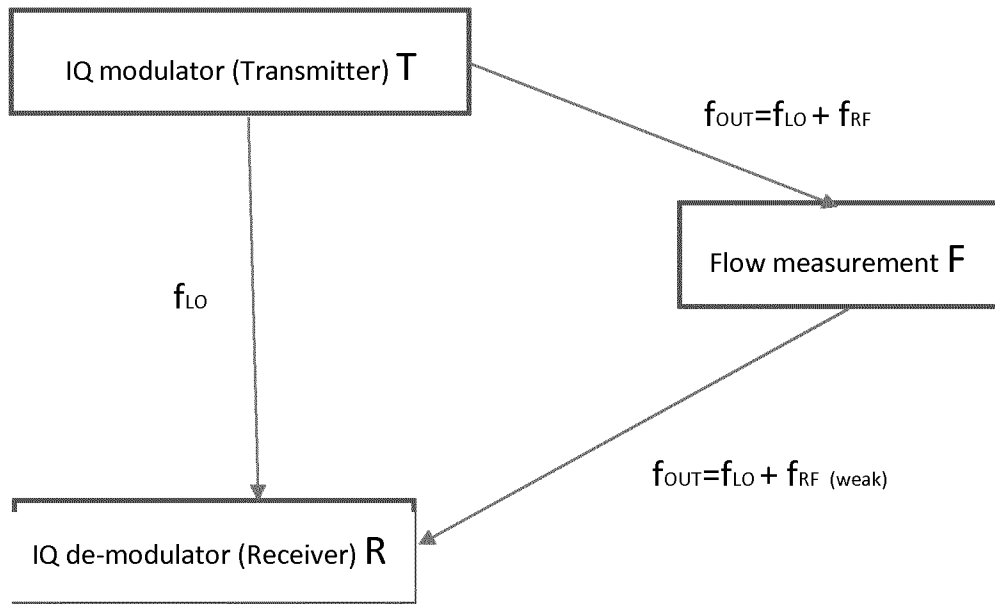


Fig. 4

INTERNATIONAL SEARCH REPORT

International application No  
PCT/EP2018/057639

A. CLASSIFICATION OF SUBJECT MATTER  
INV. G01N22/00 G01N22/04 G01N33/28  
ADD.  
According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED  
Minimum documentation searched (classification system followed by classification symbols)  
G01N

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practicable, search terms used)  
EPO-Internal, WPI Data

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	US 5 841 288 A (MEANEY PAUL M [US] ET AL) 24 November 1998 (1998-11-24) column 9 - column 10; figure 1	1-4,6,7, 9
Y	-----	5,8,10
X	TUAN A NGUYEN ET AL: "Instantaneous high-resolution multiple-frequency measurement system based on frequency-to-time mapping technique", OPTICS LETTERS, OPTICAL SOCIETY OF AMERICA, US, vol. 39, no. 8, 15 April 2014 (2014-04-15) , pages 2419-2422, XP001588124, ISSN: 0146-9592, DOI: 10.1364/OL.39.002419 [retrieved on 2014-04-10] page 2422	1-10
	----- -/--	

Further documents are listed in the continuation of Box C.

See patent family annex.

\* Special categories of cited documents :

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"&" document member of the same patent family

Date of the actual completion of the international search  4 July 2018	Date of mailing of the international search report  10/07/2018
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Name and mailing address of the ISA/ European Patent Office, P.B. 5818 Patentlaan 2 NL - 2280 HV Rijswijk Tel. (+31-70) 340-2040, Fax: (+31-70) 340-3016	Authorized officer  Martin, Hazel
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## INTERNATIONAL SEARCH REPORT

International application No  
PCT/EP2018/057639

C(Continuation). DOCUMENTS CONSIDERED TO BE RELEVANT		
Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
Y	US 7 078 913 B1 (PELLETIER MATHEW G [US]) 18 July 2006 (2006-07-18)	5
A	column 12 -----	1-4,6-10
Y	US 2010/145636 A1 (NYFORS EBBE GUSTAF [NO]) 10 June 2010 (2010-06-10)	8,10
A	paragraphs [0048], [0069] -----	1-7,9
X	CHENG-GANG XIE: "Measurement Of Multiphase Flow Water Fraction And Water-cut", AIP CONFERENCE PROCEEDINGS, vol. 914, 1 January 2007 (2007-01-01), pages 232-239, XP055427281, NEW YORK, US	1-5,7-10
A	ISSN: 0094-243X, DOI: 10.1063/1.2747436 page 234 - page 235 -----	6

# INTERNATIONAL SEARCH REPORT

Information on patent family members

International application No

PCT/EP2018/057639

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US 7078913	B1	18-07-2006	NONE
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			EP 2104837 A1 30-09-2009
			NO 328801 B1 18-05-2010
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