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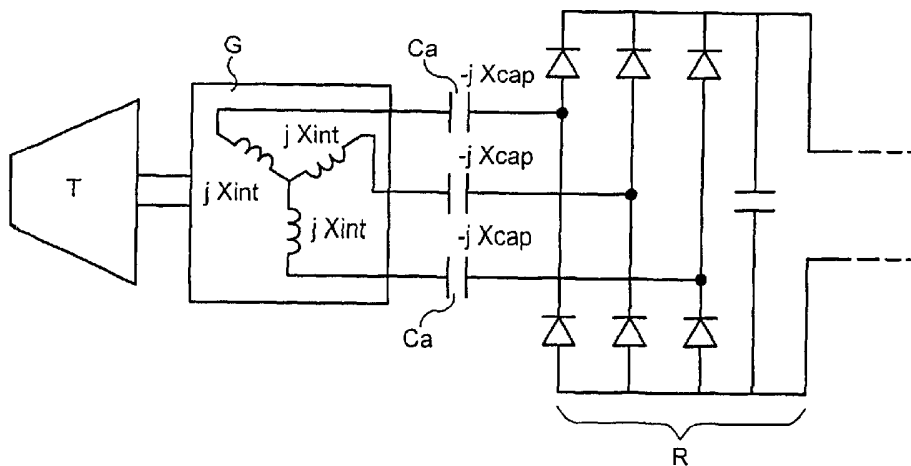
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(54) Title: HIGH-FREQUENCY GENERATOR



(57) Abstract: The invention provides an electrical generating system comprising an AC electrical generator (G) having an output, the system being characterised by a capacitor arrangement (Ca) which is provided at the output of the generator and which is arranged so as to offset a drop of voltage from no-load to full-load occurring at the output of the generator, whereby to permit increased power to be drawn from the generator without an unacceptable drop in output voltage and without exceeding permissible temperature limits for the generator winding. As described, the generator (G) is a permanent-magnet generator having a plurality of terminals and associated output lines, and the capacitor arrangement comprises a respective capacitor connected in series in one or more of the output lines, the value of the capacitance being selected such that a drop of voltage from no-load to full-load occurring at the associated generator terminal is substantially offset at an output terminal of the said capacitor.



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High-Frequency Generator

5 The invention relates to an electrical generating system which comprises a high frequency generator as one component of the system in the main power chain, and is particularly concerned with improvements to the output of such a high-frequency generator. In a preferred embodiment, the generator is a rotating permanent-magnet generator.

10

The principle of the invention is applicable to any such system, but in practice the invention is most valuable for systems in which the generator either rotates at high speed or has many poles, or a combination of the two, so as to produce electrical power at the generator terminals at a relatively high frequency. The best frequency is
15 found to be a function of the rated power output of the generator. With present technology and as a very rough guide-line, it may be said that substantial practical and economic benefit may be demonstrated for the invention if the product of frequency in kilohertz and rated machine power in kilowatts is of the order of a few hundred, or anywhere in a band from 100 to 1000 and perhaps higher. However this guide-line
20 should not be taken to define limits outside which the invention is not applicable with advantage.

A particularly important example of such a generating system is known as a 'micro-turbine-generator', or MTG, which is designed to provide moderate amounts of power
25 from around a few kW to a few MW. Examples of MTG applications are: to provide power to specific local loads, or to a multiplicity of points in a distribution network supplying a large number of local loads dispersed over a region, or to a parallel combination of local loads and a distribution network.

A conventional MTG system is shown and described with reference to Figures 1 to 3 of the accompanying drawings, in which:

Figure 1 is a circuit diagram of a typical MTG system;

5 Figure 2 is a cross sectional view of a four-pole permanent-magnet generator employed in the system of figure 1; and

Figure 3 is a graph of generator terminal voltage versus output power for the system of Figure 1.

10 Figure 1 shows an example of a typical MTG system that comprises a prime mover in the form of a gas turbine (T), mechanically coupled to a permanent-magnet poly-phase electrical generator (G). Electrically connected to the generator is a power conditioning unit (PCU) for converting the output voltages at the terminals of the generator to the voltage waveforms required by the consumer of the generated
15 electricity at the output of the MTG.

The power conditioning unit (PCU) may typically comprise an input stage that may be an uncontrolled rectifier (R), as shown in Figure 1, in which the rectifying elements are simple diodes followed by a DC/DC converter (C), that takes the alternator
20 voltages as input and produces a stable DC link voltage as output. Alternatively the input stage may embody controllable switched devices such as IGBTs in place of the diodes, and the full function is accomplished in one step by control of the switch timing of the devices. Connected to this input stage is the output stage of the power conditioning unit which may typically be a pulse width modulated (PWM) inverter (I)
25 plus output filter (F); the output stage takes the DC link voltage as input and produces at its output sinusoidal three-phase voltages usually of 50 or 60 Hz. Figure 1 shows a typical configuration based on the former option for the first stage; the invention is equally applicable to either option.

It is found in the design process for any permanent-magnet generator, that the machine possesses substantial internal inductive reactance, which will be denoted X_{int} , that affects the performance at the terminals in an important way. (Depending on the geometry of the magnets and the underlying steel rotor hub, there may as is well known be some difference in the values of the per-phase two-axis reactances of the machine, commonly denoted by the d-axis and q-axis values X_d and X_q respectively, this effect being known as saliency. Since saliency is typically slight in the type of machine that forms the main focus of discussion, and since X_d has the greater influence on the relevant operating circumstances, that is accordingly the value implied herein when speaking of 'the reactance' of the machine.)

A rotating drum-type machine, having radial magnetic flux across an annular airgap with a heteropolar permanent-magnet rotor may be taken for illustration, as shown in Figure 2, described in detail later. This is a widely adopted configuration because of various advantages: simplicity and robustness of construction; high electrical efficiency; good achievement in terms of capability for high speed when designed with a prestressed retaining sleeve around the outer diameter of the rotor; and competitive manufacturing cost. When the design has been optimised in terms of the dimensions of the sleeve for the speed requirement, together with the proportions of the magnetic circuit to give the highest possible flux density at the airgap of the machine, and therefore the greatest possible voltage and power output at the terminals, it is common to find that the resultant steady-state internal inductive reactance, when expressed in terms of the well-known per-unit system (in which rated full-load voltage and current form the two bases or units for those variables, their quotient forms the base of impedance, and their product multiplied by the number of phases forms the base of power) is roughly of the order of 0.4 pu.

A generator that possesses approximately such a level of reactance has a degree of reduction of terminal voltage from no-load to full-load, termed voltage regulation, which can be accepted though it is a disadvantage to the performance of the machine. Figure 3 shows the characteristic of terminal voltage V versus output power P_o , in per-

unit terms, for a machine with internal per-unit reactance of 0.39 pu. The regulation at rated power may be readily determined from the usual phasor diagram (as exemplified later) for steady-state operation, and is found to be just under 12% as shown. (As shown in Figure 3, this result includes a small percentage effect on output rms voltage due to harmonic voltages caused by a rectifier bridge that is assumed to be connected at the terminals). It should be noted that, as is well-known, rated power in the per-unit system has numerical value equal to the power factor, which in this example is 0.933. However, as can be seen from the Figure, if the power output is increased by 15% to 1.07 pu, the voltage regulation increases quite rapidly from just under 12% to about 16%. Worse still, at about 43% increase in power output above rated value (1.34 pu) the regulation is very high at about 35%, and this condition actually represents the operating limit of the machine; at that point the voltage is about to fall away to zero and no higher power level can be obtained. Any operating condition that approaches close to this limit is only marginally stable and highly undesirable.

15

In practical terms, allowing for variability in material properties of permanent magnets, machining tolerances on critical dimensions of the magnetic circuit, and the general desirability of having not too much voltage regulation from no-load to full-load together with reasonable stability of the voltage/power working point, it is fair to say that it is desirable to limit the internal inductive reactance to something less than 0.4 pu, and at the most about 0.45 pu, with correspondingly full-load regulation preferably not exceeding 12% and at most 14%.

The problem of internal reactance can, and generally does, place a constraint on the available power from a permanent-magnet generator, and this is for two distinct reasons. To exemplify the first constraint, suppose that a generator is designed to operate with a stipulated output power, and is found to have internal reactance of 0.45 pu. At that level of reactance the voltage regulation is around 14%, which means that the full-load output voltage is only 86% of what it was at no-load. If that drop of voltage could be avoided, so that the voltage remained approximately constant from no-load to full-load, then with the same current in the windings, the voltage-current

product (which is the main determinant of power output assuming, as is typically the case, that power factor remains close to unity) would increase by $\times 1/0.86$ or about 16%. The ohmic losses inside the machine remain the same (because current magnitude is unchanged), as do other principal components of loss, and so internal temperatures remain the same. Moreover these losses represent a smaller fraction of the increased output power, and so machine efficiency has risen. Thus internal reactance and consequent voltage regulation are seen to have the effect of reducing both available output power and efficiency of operation.

10 To exemplify the second constraint, it must first be appreciated that as rated current is increased in value, the effect is to increase the per-unit value of the internal reactance X_{int} , because the base of impedance has decreased due to the increase in the base of current, and therefore there is a corresponding increase in the percentage voltage regulation. In the design of a permanent-magnet generator, it may consequently turn
15 out that the rated current has to be chosen, not to correspond to the highest permissible value that would raise internal temperatures to their highest safe levels, but rather to some lower value of current that limits the per-unit value of X_{int} to an acceptable level as discussed previously in relation to Figure 3. This situation is quite commonly found to occur in the design of high-speed electrical generators, particularly in the larger
20 power sizes (say, 100 kW and up) where due to dimensional scale effects the proportionate heating effect of stator current in the windings tends to be less, and therefore the winding tends to run cool relative to the temperature limits set by insulation properties. Alternatively, it may be the case that the degree of internal cooling has been adjusted in design so as to allow temperatures to rise to their highest
25 safe levels, by economising to some extent on cooling effort, but that cooling could be increased if there were the possibility of increasing the power output by so doing. In either case, it is the need to limit the effects of inductive reactance that is effectively imposing the limit on the current rating of the machine, and therefore on the rated power output, rather than the need to limit internal heating.

Thus, in general, inductive reactance may restrict the available output power from a permanent-magnet generator by two effects: firstly, by reducing the rated terminal voltage at rated current, which reduces both power output and efficiency; and secondly, by imposing a limit on the permissible current, this limit having to be
5 accepted in order to restrict the per-unit value of the reactance to a suitable level. The first effect is always present in any machine that embodies significant internal inductive reactance. The second effect comes into play in any design in which rated current is determined not primarily by the need to limit internal heating, but rather by the need to hold the effects of reactive voltage drop down to an acceptable level. Both
10 effects are however caused by inductive reactance being undesirably high.

The present invention concerns means for reducing or eliminating the effects of internal inductive reactance, thereby substantially increasing the permissible output power of the machine and its operating efficiency.

15

In accordance with the present invention, there is provided an electrical generating system comprising an AC electrical generator having an output, the system being characterised by a capacitor arrangement which is provided at the output of the generator and which is arranged so as to offset a drop of voltage from no-load to full-
20 load occurring at the output of the generator, whereby to permit increased power to be drawn from the generator without an unacceptable drop in output voltage and without exceeding permissible temperature limits for the generator winding.

According to a preferred embodiment of the present invention the generator is a poly-
25 phase generator and there is provided in each output line of the generator a series capacitor, whose magnitude of capacitive reactance is so chosen as to have a substantial offsetting effect against the internal inductive reactance of the machine. Suppose that the machine has total internal effective inductive reactance per line equal to X_{int} . (If the machine is star-connected then X_{int} equals the per-phase value of
30 internal reactance, X_{phase} ; if delta-connected, then by the well-known equivalence of

delta- and star-connected systems it is easily seen that $X_{int} = X_{phase}/3$.) The capacitance value per line may be so chosen as to have a reactance X_{cap} approximately equal in magnitude to X_{int} at rated speed. Alternatively, X_{cap} may be chosen to be less than but a substantial fraction of X_{int} at rated speed, or it may (with
5 due caution to avoid generating excessive voltages) be chosen to be somewhat greater than X_{int} . Considerations influencing this choice are discussed later.

Assuming that the capacitance value has been so sized as to counter-balance substantially the internal inductive reactance at rated speed, then the net series
10 reactance in the line, which is now $(X_{int} - X_{cap})$, is much smaller than the previous value, X_{int} . The machine/capacitor combination behaves as though it were a simple generator with small internal reactance. Thus the voltage at the terminals of the combination remains substantially constant from no-load to full-load. Moreover if the power output of the machine is increased above its previous rating, the voltage remains
15 approximately constant. It will be clear that both the constraints on power, discussed previously, have been removed. There is negligible loss of power output capability due to terminal voltage falling with load, and current is not constrained by the need to limit inductive reactive voltage drop within the machine to an acceptable level. Power losses in the external capacitors are very small, and efficiency accordingly improves as
20 the power output is increased.

The attractiveness of this technique owes much to the superior performance of modern metallised-film capacitors. These can be made quite cheaply, and can offer a uniquely
25 good combination of the following properties: capacitance per unit volume; ripple current tolerance; voltage withstand; long life; acceptability of case temperature in the order of 70°C. Use of the capacitors typically does not compromise the life of the equipment, and their cost may be considerably less than the extra cost of alternatively building a larger alternator to offer the same increase of power without capacitors – thus the cost per kilowatt of generated power is reduced. Moreover it is readily
30 possible to have capacitors designed and manufactured in moderate numbers to match,

within the constraints of present technology, any specific application in terms of capacitance value and current and voltage ratings, without incurring a large cost penalty.

5 In this regard it may be noted that the voltage rating of the capacitor is determined by the current through it, I , and the highest value of IX_{int} voltage drop that can consequently occur across it in service; it needs to be typically only a fraction of the rated phase voltage of the machine. It is the limits to the physical properties of the capacitors that can be manufactured with metallised film technology that give rise to
10 the preferred relationship between electrical frequency and machine power rating, that was indicated earlier. Particularly significant here is the way in which these limits of manufacturing technology cause capacitor current rating to be effectively linked to capacitance value, and hence restrict the way in which the ohmic (non-pu) value of X_{int} is related to current at a particular frequency.

15

As a useful guide-line, it typically turns out in many cases that the apparent power rating of the total capacitor bank used with the machine may be up to about half its real power rating.

20 Clearly by setting the inductive and capacitive reactances approximately equal at rated speed, a resonance condition (where the two reactance values are precisely equal although of course opposite in sign) is being established near to rated speed, and it is natural to enquire whether this produces any over-voltage or over-current effects, as are commonly encountered with resonance phenomena in other applications, that
25 might be troublesome. However, this is a series resonance (because in the preferred embodiment the inductive and capacitive reactances are connected in series) as opposed to a parallel resonance, and consideration shows that in this case no over-voltage effects occur at resonant frequency, nor is there any over-current since current is controlled by the demand load connected to the complete generating system.
30 Moreover, higher harmonic currents that are drawn from the generator by non-linear

loads, such as a rectifier bridge, are negligibly affected, since the impedance to these currents is dominated by the internal inductive reactance of the machine, the capacitive reactance being small at the higher frequency.

5 One feature that calls for consideration at the design stage is operation at reduced speed. A moderate degree of speed reduction at rated power generally does not pose any problem. However, as speed and frequency reduce, capacitive reactance increases, and it is necessary to check that the voltage drop across the capacitor is not exceeding its permissible value due to the current passing through it. In this regard, the MTG
10 application is particularly well suited. It is common practice to start the MTG set using the generator in motor mode to accelerate the shaft up to a speed at which the turbine can become self-sustaining, which typically occurs at around 40% rated speed. As part of the switching operations to establish the motoring mode, it is convenient to switch the capacitors out of circuit, switching them back in when the mode changes over to
15 generating. At the lower end of the speed range above this changeover point, the turbine has very restricted capability to generate power. Consequently high line currents at low speed are not experienced. It is typically found with an MTG system that if the capacitor is rated for voltage in accordance with the highest IX_{cap} voltage drop that occurs in any defined operating condition at or near rated speed, then no
20 other condition occurs during start-up or other transient circumstance that calls for a higher rating.

The operating power factor of the machine/capacitor combination is the same as for the machine alone, being determined by the characteristics of the connected load.
25 However the phase relationship between voltage and current at the machine terminals is changed by the addition of capacitors, and the magnitude of machine voltage on load is generally increased as a result. Without capacitors this phase relationship generally gives a power factor of around 0.93 or better; with the capacitors added, the power factor typically improves. Detailed study shows the voltage at the machine
30 terminals in any likely operating condition typically to be less than the rated no-load voltage of the machine.

The inclusion of the capacitors reduces the total series impedance in the output lines of the generator to a very low level. Consequently any short circuit at or near the terminals of the machine/capacitor combination will cause an extremely high current
5 that burns out the windings very rapidly. To protect the machine against such an external short circuit, it may be preferable to provide fuses or switches in its output lines with appropriate $I^2 t$ characteristic, so that thermal capacity can limit internal temperature rise to a safe level.

10 It is often satisfactory to size the capacitance value so that capacitive reactance compensates most but not all the inductive reactance, and voltage regulation at the output terminals is reduced to a few percent. By reducing the capacitance and so increasing reactance, an approximately level voltage characteristic can of course be obtained. In general it is found to be perfectly permissible to reduce capacitance still
15 further so that there is a moderate degree of rise of voltage at the generator/capacitor output terminals. This can be used to offset the fall of voltage with load that typically occurs across a following rectifier stage, so that the final output DC voltage is substantially constant, independent of load. It may thus be possible to avoid the cost and extra losses associated with a further DC/DC converter, which would otherwise
20 typically be needed to maintain a controlled, constant DC voltage. However, in the MTG case for example, it is common to design for some variation of speed with load, so as to optimise the efficiency of the turbine on part load. In that case the provision of a constant DC level for all operating conditions simply by suitably sizing the capacitance is of course not possible.

25

Preferred features of the generating system according to the present invention are set out in the following paragraphs.

According to these preferred features, a permanent-magnet AC electrical generator has
30 capacitors connected in series in one or more of the output lines of the generator, the

value of capacitance being so chosen that the drop of voltage from no-load to full-load that would occur at the terminals of the machine without the effect of the capacitors is substantially offset in respect of the voltage occurring at the output terminals of the capacitors, thus permitting increased power to be drawn from the machine/capacitor combination without unacceptable drop in output voltage and without exceeding permissible temperature limits for the machine winding.

The permanent-magnet AC electrical generator may be a poly-phase AC generator, with the capacitors connected in series in each of the poly-phase output lines.

10

Advantageously, the electrical generator operates at a fairly high frequency such that the combination of required capacitance value and required voltage rating and current rating of the capacitor forms a good match to what is naturally available with metallised film or other high-output capacitor technology, thus enabling the cost of the capacitors per kW of increased power output to be advantageously less than the saved cost per kW in the generator by virtue of its reduced size for the given power rating.

For example, the generator may be driven either directly or through gearing by the turbine in a micro-turbine-generator system.

20

In the preferred embodiment, the machine/capacitor combination is connected to a rectifier bridge and the capacitance value is so chosen that the DC output voltage from the rectifier bridge varies approximately in a required manner with variation of load. For example, the capacitance value may be so chosen that the DC output voltage from the rectifier bridge is substantially constant, independent of the load.

One possible advantage of the use of such capacitors is to enable a smaller electrical machine for the given power rating to be fitted into a smaller void space than would be possible or convenient without the use of the capacitors.

- The invention at least in its preferred form described below thus provides an arrangement for obtaining greater power from a permanent-magnet generator, with improved efficiency and in many cases lower cost per generated kilowatt, by connecting
- 5 capacitors in series with the poly-phase output lines of the generator. The arrangement shows greatest benefit in, but is not in principle restricted to, generators that produce their output power at fairly high frequency, defined by the relationship f (in kilohertz) multiplied by rated power (in kilowatts) is close to or greater than 100.
- 10 Consequently a particularly important application of the invention is to micro-turbine-generators, where power output can usually be increased by at least 15%, and quite often by up to 35% or even more. The value of capacitance must be appropriately chosen, so as substantially to offset the adverse effects of internal inductive reactance, and power output is consequently increased due to two distinct effects. Firstly, the
- 15 voltage at the terminals of the machine/capacitor combination shows little or no voltage drop from no-load to full-load, differing from the action of the machine alone, where substantial voltage drop occurs. The power output of the combination is correspondingly greater for the same generator current magnitude. Secondly, it is no longer necessary to restrict the value of rated current in order to hold the aforesaid
- 20 voltage drop at an acceptable level. If therefore previously this restriction was causing the machine to be rated at a current magnitude less than that which could be tolerated on grounds of internal temperature rise alone, then it now becomes possible to up-rate the current and so further increase power output.
- 25 The invention will now be described by way of example only and with reference to the accompanying drawings, in which:

Figures 1 to 3 show a conventional MTG system, as discussed above;

Figure 4 is a circuit diagram showing a modification of the system of Figure 1 in which the generator is formed as a star-connected generator with series connected capacitors;

5 Figure 5 is a phasor diagram for the generator of Figure 4 including the series connected capacitors; and

Figure 6 is a phasor diagram for the generator of Figure 1 without the series connected capacitors.

Referring firstly to Figure 2, this shows a cross-section of an electrical generator rated
10 84 kW output power at 80,000 r/min, for use in an MTG system as shown in Figure 1. The proportions and specifications of the various parts are satisfactory for the present purpose of demonstrating the effect of the invention, though they are not optimised and the performance described does not represent best practice for the specified application. A stator lamination (1) is provided with 24 slots (2) into which is inserted
15 a 3-phase, 4-pole winding (not shown). A prestressed cylindrical sleeve (3) which may be made of inconel or carbon fibre retains four rare-earth permanent-magnet rotor poles (5) which are bonded to an inner steel hub (6). Interstices under the sleeve between magnets contain epoxy filler (7). Between the stationary stator and the rotating sleeve there is a radial airgap (4). The stator is cooled by a close fitting water
20 jacket (not shown) surrounding the OD of the laminations. Air cooling may be supplied axially along the airgap as necessary to control principally the operating temperature of the rotor.

Principal design details (all dimensions in mm) are as follows: outer diameter of stator
25 lamination = 105.0; diameter to slot bottoms = 80.0; diameter at stator bore = 60.6; diameter over sleeve = 59.6; diameter over magnets = 51.3; width of square hub = 34.5; width of main part of tooth = 4.2; slot opening = 2.2; axial length (not shown) of laminated core and magnets = 125.0.

Stator winding comprises: 24 coils in 2-layer form; the coils of one phase connected to form two parallel paths; each coil wound with 3 series turns; stator conductor = 19 strands of 25 American Wire Gauge (AWG) enamelled wire. Insulation specification permits 180°C maximum winding temperature. Coolant inlet temperature = 50°C.

5

A person skilled in the art of electrical machine design may verify the following: the machine generates 298 rms V/phase on no-load; total inductance on the d-axis = 57.8 μH /phase, giving reactance = 0.968 Ω /phase at an electrical frequency = 2.67 kHz; when delivering 84 kW output power into an uncontrolled 3-phase rectifier bridge the terminal voltage falls to 260 rms V/phase, current = 117 rms A/line; per-unit impedance base = 2.23 Ω ; per-unit d-axis reactance = 0.434 pu; voltage regulation = 13.5%; efficiency = 97.8%; with 0.1 litre/sec of water flow and adequate supply of air to the airgap the maximum temperature in the winding = 155°C.

15 This machine therefore represents an example towards the high-regulation end of the advisable range recommended earlier, having reactance of 0.434 pu (close to the advisable limit of 0.45 pu) and regulation correspondingly of 13.5% (close to the advisable limit of 14%). Clearly, with this machine, it would not be advisable (or indeed, possible) to increase the power rating significantly. Note however that the winding temperature is comfortably less than the maximum of 180°C that is in fact permissible according to the insulation specification. This is a machine, therefore, that suffers from both the effects described earlier: the drop of voltage on full load substantially reduces the available power, and the current cannot be increased further to take advantage of the permissible maximum temperature because it will bring the operating point too close to the absolute power limit that is depicted (for a slightly lower value of per-unit reactance) by the curve of Figure 3.

In accordance with the invention, capacitors may now be introduced as shown in Figure 4. The general arrangement shown in Figure 4 depicts the generator (G) and rectifier stage (R) of Figure 1, assuming for example a star-connected generator, and

30

shows series capacitors (Ca) added.

In the example of Figure 4, capacitance value is 80 $\mu\text{F}/\text{line}$, rated 150 rms V max, ripple current = 150 rms A max, permissible case temperature = 70°C, 100,000 hour
 5 life minimum at fully rated condition, considerably more if running cooler. The choice is now made to increase the power output of the machine by 25% to 105 kW. Again, a person skilled in the art of electrical machine design may verify the following: no-load voltage and inductance in $\mu\text{H}/\text{phase}$ unchanged; when delivering 105 kW output
 10 power into an uncontrolled 3-phase rectifier bridge the terminal voltage of the machine/capacitor combination falls to 294 V/phase; current = 129 A/line; voltage regulation = 1.4%; efficiency = 97.9%; maximum temperature in the winding = 173°C.

This arrangement is now producing 25% more power out of the same machine.

Voltage regulation is almost negligible, efficiency has slightly improved, and winding
 15 temperature is still comfortably within specification. In other examples, it is possible to demonstrate an even greater percentage increase in output power, more marked improvement in efficiency, and generally lower temperature rises than are characteristic of this case.

20 The phasor diagram in Figure 5 depicts this operating condition, for the fundamental sinusoidal components of voltage and current. (Minor apparent discrepancies in quoted numbers and between numbers and diagram are due to second order effects introduced by voltage and current harmonics and saliency.) The quantities V_{cap} and V_{ind} are the voltages dropped across the capacitor and across the internal machine inductance,
 25 respectively, V_{term} is the output terminal voltage of the machine/capacitor combination, V_{mc} is the voltage at the terminals of the machine, and V_{nl} is the particular value of V_{mc} on no-load. Phasor relationships are: $V_{\text{ind}} = j I X_{\text{int}}$; $V_{\text{cap}} = -j I X_{\text{cap}}$; $V_{\text{term}} = V_{\text{nl}} - (V_{\text{ind}} + V_{\text{cap}})$; $V_{\text{mc}} = V_{\text{nl}} - j I X_{\text{int}}$. It will be clear that the output voltage is similar in magnitude to the no-load voltage when the capacitors
 30 are present, showing almost negligible reduction (low voltage regulation). Also it can

be seen that the voltage at the machine terminals is similar in magnitude to the no-load voltage, again showing a small reduction, and that the phase angle between V_{mc} and I is small, giving an internal power factor (equal to the cosine of this angle, neglecting minor harmonic effects in the current and voltage waveforms) that is close to unity.

5

Figure 6 shows the situation if the capacitors are removed and it is attempted to work still at 105 kW. V_{mc} (which now is the same as V_{term}) is much reduced at 228 V (high regulation). The current I increases to the unacceptably high value of 166 A, which will rapidly over-heat the machine, and V_{ind} is increased correspondingly. A
10 further important point, not immediately apparent from inspecting the diagram, is that this operating condition is now undesirably close to the absolute limit of power output.

The following discussion makes clear why a value of frequency appropriately related to power rating is to be preferred. If the power, voltage and current data were as
15 specified above, but related to a 4-pole machine running at only one tenth speed = 8,000 r/min, for the sake of argument, then 10 times as much capacitance would clearly be required. This might be achieved by having 10 units in parallel of the same capacitor as before, which would cost 10 times as much and offer 10 times the current capability – which is not called for in this example. Current capability is being wasted,
20 and the cost of capacitors per kW of power rating is therefore multiplied by 10, and that would probably exceed the saving in cost achieved by the reduced size of the machine for the given power. Alternatively the increased capacitance might be achieved in a single larger unit, which would tend to be about 10 times the volume and again roughly 10 times the price, but because of connection difficulties would not offer
25 so much increase in current rating. Considerations of this kind lead to the general conclusion that, for the greatest advantage, the frequency must be sufficiently high so that the combination of capacitance, voltage and current ratings, tend to match well to the natural optimum of what can be achieved by metallised film capacitor technology for the given ratings. In this regard, MTG units tend to be an example of a particularly

good match, and the cost savings achieved by applying the invention in this embodiment are very substantial.

However, each case must be judged on its merits. The embodiment above is described
5 by way of example and is only to be considered preferred and illustrative of the inventive concepts disclosed. The scope of the invention is not to be restricted to the embodiment. Various and numerous other arrangements may be devised by one skilled in the art without departing from the spirit and scope of this invention.

10 For example, the above description relates to an embodiment that includes a series connection for the capacitors, but an alternative possibility is to connect capacitors across the machine terminals in parallel with the external load. This can achieve improvement in voltage regulation and output power capability in a manner that is similar in principle to the action of series capacitors. However, there are features of the
15 parallel connection that may be less desirable in a practical embodiment. In particular: a substantial increase occurs in the terminal voltage of the machine/capacitor combination at no-load (whereas series capacitance has no effect in this condition), and the machine may need to be re-designed in order to bring this voltage down to a desirable level; the effectiveness of the added capacitors tends to decrease with
20 increase of load basically because the magnitude of the capacitor current that is being drawn through the machine is becoming smaller relative to the magnitude of the demanded load current (whereas the effectiveness of series capacitors is sustained well up to high load levels as has been discussed); it may typically be found that the combined voltage/current/frequency requirements for parallel capacitors do not match
25 as well to what is available within the limits of manufacturing technology, compared with the requirements for series capacitors. However, with some system specifications, it may nonetheless be possible that the parallel configuration is to be preferred.

Claims

- 1 An electrical generating system comprising an AC electrical generator having an
5 output, the system being characterised by a capacitor arrangement which is provided at the output of the generator and which is arranged so as to offset a drop of voltage from no-load to full-load occurring at the output of the generator, whereby to permit increased power to be drawn from the generator without an unacceptable drop in output voltage and without exceeding permissible
10 temperature limits for the generator winding.
- 2 A system according to Claim 1 characterised in that the generator is a permanent-magnet generator having a plurality of terminals and associated output lines, and in that the capacitor arrangement comprises a respective capacitor connected in series
15 in one or more of the output lines, the value of the capacitance being selected such that a drop of voltage from no-load to full-load occurring at the associated generator terminal is substantially offset at an output terminal of the said capacitor.
- 3 A system according to Claim 1 or 2, characterised in that the generator is a poly-
20 phase AC generator, and in that a respective capacitor is connected in series in each of the poly-phase output lines.
- 4 A system according to any of Claims 1 to 3, characterised in that the generator is a
high frequency generator, for example in which the product of frequency in
25 kilohertz and rated machine power in kilowatts is close to or exceeds 100.
- 5 A system according to any preceding Claim which is a micro-turbine-generator system, characterised in that the generator is driven either directly or through gearing by the turbine.

- 6 A system according to any preceding Claim characterised in that the capacitor arrangement is connected to a rectifier bridge and in that the capacitance value and/or arrangement of the or each capacitor is selected so as to control a DC output
5 voltage from the rectifier bridge to vary in a predetermined manner with variation of load.
- 7 A system according to Claim 6, characterised in that the DC output voltage from the rectifier bridge is arranged to be substantially constant independent of the load.
10
- 8 A system according to any preceding Claim characterised in that the capacitor arrangement comprises at least one metallised film capacitor.
- 9 A system according to any preceding Claim characterised by means for switching
15 the capacitor arrangement in and out of connection into the system at some speed in the speed range from standstill to full speed respectively.
- 10 A system according to any preceding Claim characterised by a fuse arrangement provided at the output of the generator for limiting current flow in terminals of the
20 generator.

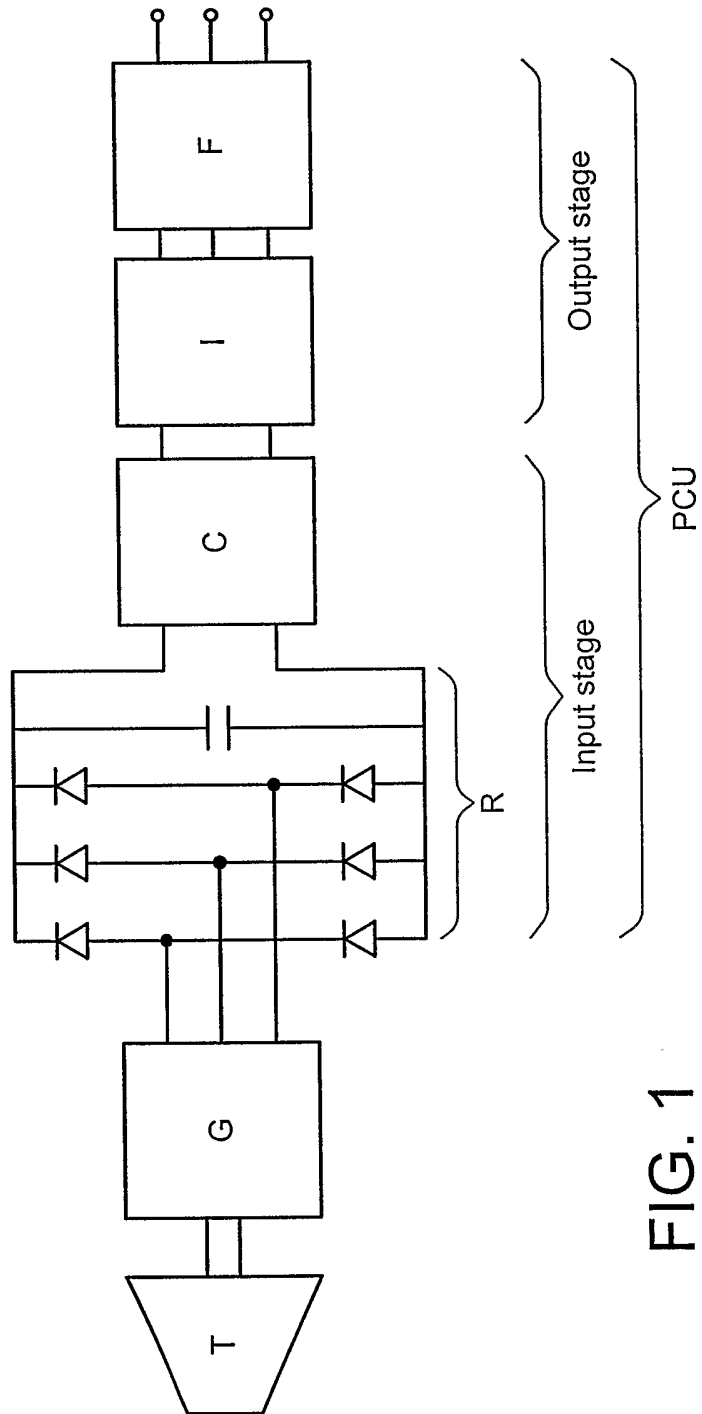


FIG. 1

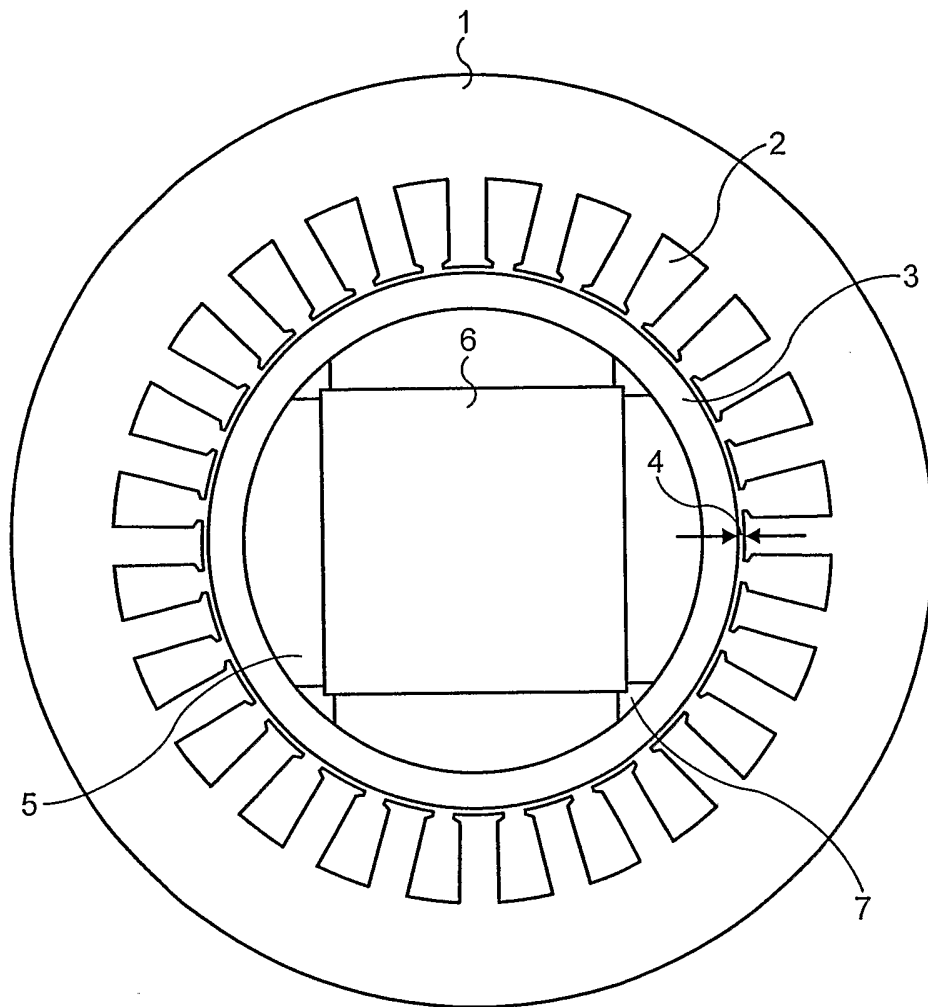


FIG. 2

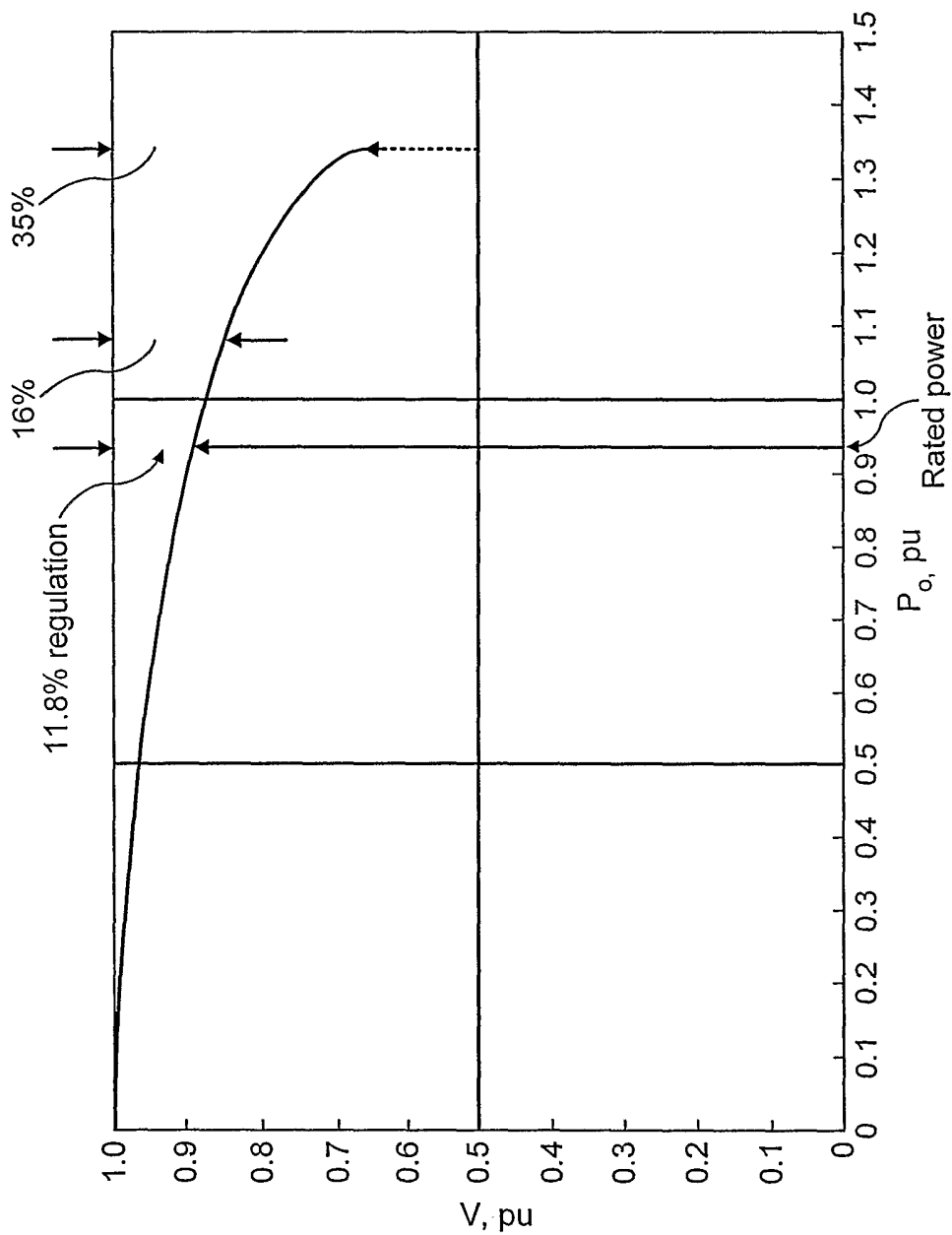


FIG. 3

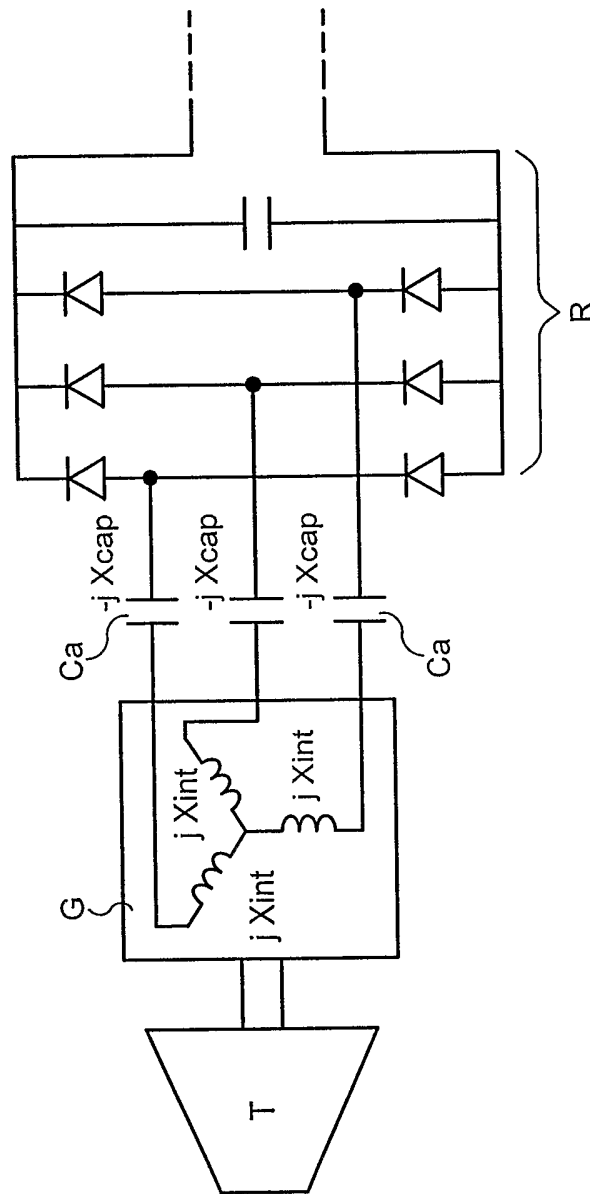


FIG. 4

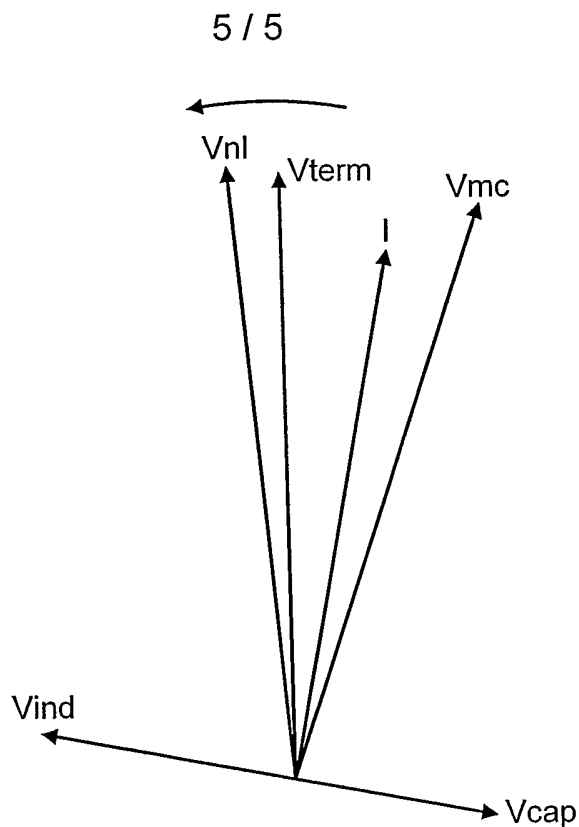


FIG. 5

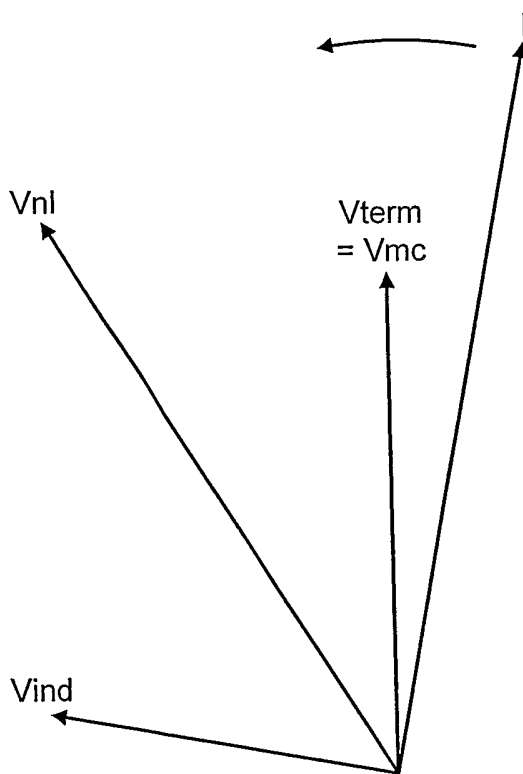


FIG. 6

INTERNATIONAL SEARCH REPORT

International Application No
PCT/GB 03/02349

A. CLASSIFICATION OF SUBJECT MATTER
IPC 7 H02J3/18 H02M5/45

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)
IPC 7 H02J H02M

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	WO 94 24622 A (ELECTRIC POWER RES INST) 27 October 1994 (1994-10-27) page 2, line 13-15; figure 1 page 5, line 18-21 page 6, line 2 page 7, line 6-23	1
Y	page 5, line 22-34 page 6, line 2	2-5
Y	figure 1	6
Y	page 10, line 11-35 page 11, line 25-27 page 14, line 21-29	7-10
X	US 6 166 929 A (MA DAMING ET AL) 26 December 2000 (2000-12-26) column 2, line 15-22; figures 1,2 column 4, line 50-62	1
	-/--	

Further documents are listed in the continuation of box C.

Patent family members are listed in annex.

* Special categories of cited documents:

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Date of the actual completion of the international search

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INTERNATIONAL SEARCH REPORT

International Application No

PCT/GB 03/02349

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT		
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Y	figures 3,4	6-10
A	US 5 198 744 A (SCHRAMM GUENTER ET AL) 30 March 1993 (1993-03-30) figures 1,2	1
A	US 5 627 738 A (LUBOMIRSKY VADIM ET AL) 6 May 1997 (1997-05-06) column 6, line 33-55; figure 6	8

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