(54) Title: PLASMA PROCESSOR
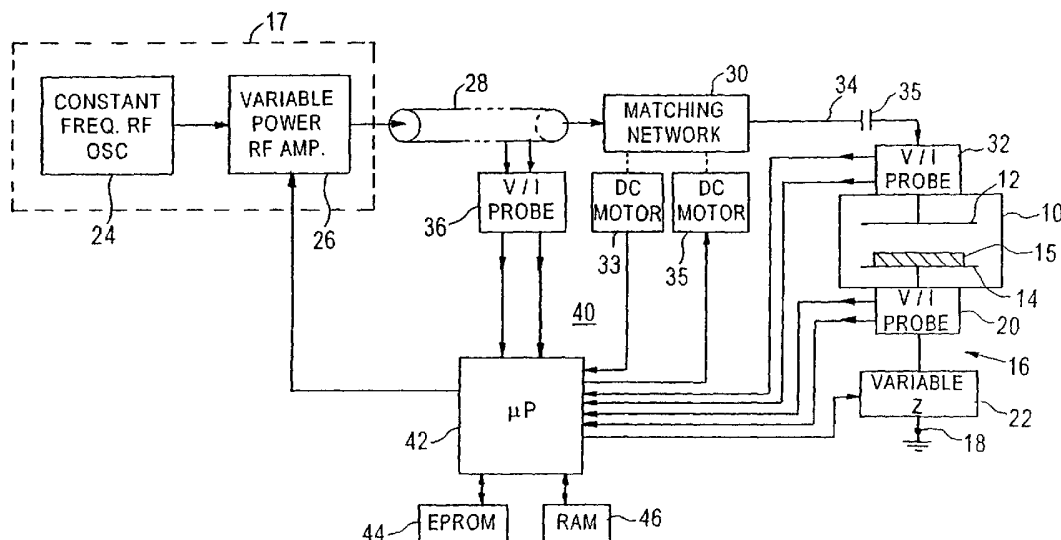
(57) Abstract: A vacuum plasma processor includes a voltage-current detector (32) connected between a matching network (30) and a reactive impedance for exciting gas in a chamber (10) to a plasma for processing a workpiece (15). A constant non-zero RF parameter is maintained in a connection between an electrode (14) in the chamber and ground. The electrode and connection to ground are such that no RF energizing source is directly coupled with the electrode.

PLASMA PROCESSOR

Relation to Co-Pending Application

[1] The present application is a continuation-in-part
of my co-pending, commonly assigned provisional
application Serial Number 60/296,832 filed June 6,
2001, entitled "Apparatus for Accurate RF Delivery to a
Plasma Chamber."

Field of Invention

[2] The present invention relates generally to vacuum
plasma processor methods and apparatus and more
particularly to a vacuum plasma processor method and
apparatus wherein a constant, non-zero RF parameter is
maintained between an electrode having a connection to
ground such that no RF energizing source is directly
coupled with the electrode.

Background Art

[3] Vacuum plasma processors are used to deposit
materials on and etch materials from workpieces that
are typically semiconductor, dielectric and metal
substrates. A gas is introduced into a vacuum plasma
processing chamber where the workpiece is located. The
chamber pressure is typically in the range of 0.1 to
1000 torr. The gas is ignited into an RF plasma in
response to an RF electric or electromagnetic field.

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The RF field is provided by a reactive impedance
element, usually either an electrode array or a coil
which couples both magnetic and electrostatic RF fields
to the gas.   The reactive impedance element is
connected to an RF source having an RF frequency and
sufficient power such that the gas is ignited into the
plasma.   Connections between the source and reactive
impedance element are usually by way of a relatively
long cable, connected directly to the RF source.   A
resonant matching network connected between the cable
and reactive impedance element usually includes at
least one variable reactance adjusted to match the
impedance of the source to the load it is driving.

[4] The load seen by the source is subject to
substantial random, unpredictable variations.   The load
has a relatively high impedance prior to ignition of
the gas into a plasma state.   In response to the plasma
being ignited, the load impedance drops substantially
due to the presence of charge carriers, i. e.,
electrons and ions, in the excited plasma.   The ignited
plasma impedance also changes substantially during
processing of workpieces due to variations in the
plasma flux, i.e., the product of the plasma density
and the plasma charge particle velocity.

[5] The RF losses are also influenced by the physical
makeup of the RF delivery path of the plasma processor,

the hardware of the processor chamber and ground path impedance, which is subject to considerable variation at the RF frequencies exciting the plasma. The ground path impedance is determined by the physical makeup of the parts forming the ground, as well as the associated impedance of the ground path at the RF excitation frequency. The load is subject to these variations during processing of a single workpiece. In addition, the load is subject to these variations while the same processor is processing different workpieces. Further, the load is subject to these variations among different processors having the same nominal design, such as the same model number, because different processors have different unpredictable characteristics. The losses and impedance differences have a large effect on the performance of the processor, for example etch and deposition rates.

[6] Previously, it was thought that control of the matching network variable reactance and the output power of the RF source could provide adequate compensation for these random, unpredictable variations. The matching network variable reactance is controlled to maintain an impedance match and resonance between the RF source output impedance and the load impedance.

[7] In addition, control is frequently provided in response to a voltage probe connected between the matching network and the reactive impedance element. The voltage probe derives a signal indicative of the RF voltage between the reactive impedance element and a reference potential, such as ground, the potential at which a metal wall of the chamber is maintained. The signal indicative of the RF voltage between the reactive impedance element and the reference potential is coupled to a controller for an output parameter of the RF source, which is remote from the chamber and matching network. The controller is usually part of the RF source and includes a monitor for the current and voltage the RF source applies to an end of the cable connected to the RF source. The monitored current and voltage, either from the voltage probe connected between the matching network and reactive impedance element or the voltage monitor of the RF source, are combined to control output power of the RF source to a desired setpoint.

[8] The assumption has been that the current monitored in the controller of the RF source is an accurate replica of the current flowing in the reactive impedance element and load. I have realized that this is an invalid assumption because of the loss effects of the cable connected between the RF source and the

matching network, as well as other associated losses. In addition, the foregoing load variations have an adverse effect on this assumption. Because of this invalid assumption, desired power has not actually been supplied in many instances to the plasma during processing, with a resulting adverse effect on processor performance.

Summary of the Invention

[9] In accordance with the present invention, a plasma processor includes a vacuum plasma chamber for processing a workpiece, wherein the chamber includes a reactive impedance element for electrical coupling with gas in the chamber and an electrode having a connection to RF ground such that no RF energizing source is directly coupled with the electrode. The connection of the electrode to RF ground is such that a finite, non-zero RF voltage has a tendency to be developed between the electrode and RF ground. Sufficient power is supplied to the reactive impedance element to excite the gas in the chamber to a plasma while a constant finite, non-zero RF parameter, preferably voltage, is maintained between the electrode and RF ground in the connection.

[10] Preferably the constant voltage is maintained between the electrode and RF ground in the connection by detecting RF voltage between the electrode and

ground and/or RF current flowing between the electrode and ground. In response to the detected voltage and/or current in the connection, RF impedance between the electrode and ground in the connection is controlled to provide the substantially constant, finite non-zero RF parameter.

[11] In one preferred embodiment, the RF impedance includes a variable reactance, preferably an inductor and/or resistor, and/or capacitor having a value controlled by the detected voltage and/or current in the connection. The detected voltage and/or current in the connection are used to control (1) whether the inductor, capacitor or resistor, or which combination thereof, is part of the connection and (2) the value(s) of the connected impedance(s).

[12] As a result of the foregoing, ground impedance is maintained at a constant value to assist in accurately controlling power delivered to the plasma. Together, accurate control of delivered power and maintaining ground impedance at a constant value enable a user of the processor to have almost total control over many RF parameters having a large effect on processor performance. This is especially advantageous when matching multiple processors running the same process and maintaining process stability on the same processors over long time periods. The principles of

accurately controlling delivered power and maintaining ground impedance at a constant value can also be used to maintain constant characteristics during the manufacture of processors of the same nominal type. The RF characteristics of the processor can be accurately measured and adjusted to guarantee performance during manufacturing of the processors, prior to shipping the processors to the end-user.

[13] Plasma processors that maintain a constant finite non-zero RF parameter between the electrode and RF ground in the connection are preferably, but not necessarily, employed with arrangements for controlling an RF source driving a reactive impedance element maintaining constant power across a load including the reactive impedance and the plasma it excites.

[14] The above and still further objects, features and advantages of the present invention will become apparent upon consideration of the following detailed description of a specific embodiment thereof, especially when taken in conjunction with the accompanying drawings.

<u>Brief Description of the Drawing</u>

[15] Figure 1 is a block diagram of a preferred embodiment of a vacuum plasma processor of the invention; and

[16] Figure 2 is a schematic diagram of a variable impedance included in the processor of Figure 1.

Detailed Description of the Drawing

[17] Reference is now made to Figure 1 of the drawing wherein vacuum plasma processing chamber 10 is illustrated as including parallel metal electrode plates 12 and 14 between which an RF plasma discharge is excited. The plasma is derived by plate 12, which is mounted at the top of chamber 10 and functions as an RF excitation reactance that supplies RF energy to gas from a gas source (not shown). Chamber 10 is maintained at a vacuum in the range of 0.1 to 1000 torr by a suitable vacuum pump (not shown). Workpiece 15, typically a semiconductor, glass or metal substrate, is mounted on plate 14 at the bottom of chamber 10. Workpiece 15 is processed by charge particles, i.e., electrons and ions, and neutral particles in the plasma so the workpiece is etched and/or material is deposited in thereon. The plasma discharge and plate 12 form a load for RF source 17 and the circuit elements driven by it. If it is necessary to cool workpiece 16 by supplying it with a cooling gas that flows through plate 14, plate 14 includes a chuck for the workpiece. If a chuck is employed, it is preferably an electrostatic chuck responsive to a DC chucking voltage source (not shown).

[18] An RF connection 16 having a low RF impedance is provided between plate 14 and ground terminal 18. Connection 16 has no RF excitation source directly connected or coupled to it, but can include the DC chucking voltage source, if such a source is provided. Despite the fact that connection 16 has no RF excitation source directly connected or coupled to it and connection 16 has a relatively high conductivity path between plate 14 and ground terminal 18, I found that random, variable voltages have a tendency to be developed in prior art processors in connection 16 between plate 14 and ground terminal 18. This random, variable voltage has adverse effects in providing consistent processing of workpiece 15.

[19] To overcome this problem, a constant voltage is maintained in connection 16 between plate 14 and ground terminal 18. The constant voltage is maintained by mounting a voltage-current probe 20 on the outside of chamber 10 immediately below plate 14. Probe 20 is mounted so that it is electrically and magnetically coupled to a lead of connection 16. Probe 20 derives first and second signals respectively replicas of the RF voltage between plate 14 and ground terminal 18 and the RF current flowing in connection 16 between plate 14 and ground terminal 18. In addition, connection 16 is provided with variable RF impedance 22 which is

adjusted in response to a function of the first and second signals to maintain a constant finite, non-zero RF voltage between plate 14 and ground terminal 18.

[20] RF source 17 includes constant frequency RF oscillator 24, typically having a frequency of 13.56 MHz, and variable power RF amplifier 26, having a power input terminal connected to be responsive to the output of oscillator 24 and a variable gain control input terminal. Amplifier 26 derives a relatively high power RF output, having a desired setpoint value, typically in the range of 1-4 kilowatts. The output of amplifier 26, at the frequency of oscillator 24, is connected to one end of a relatively long (for example, 13 feet) coaxial cable 28, having a second end connected to an input terminal of matching network 30, having an output terminal connected to supply RF power to plate 12 .

[21] Thus, cable 28 has a length that is at least several wavelengths of the frequency of oscillator 24 and causes substantial RF losses to be introduced between its two opposite ends. As a result of the length of cable 28 and these losses, the RF current flowing from RF source 17 into cable 28 differs from the RF current matching network 30 supplies to plate 12.

[22] To accurately monitor the RF current that source 17 supplies to plate 12, as well as the RF voltage

between plate 12 and ground terminal 18, voltage-current probe 32 is mounted on or in very close proximity (within 1/8 of a wavelength of the frequency of oscillator 24) to the outer surface of chamber 10 immediately above plate 12. Voltage-current probe 32 is electrically and magnetically coupled to lead 34, connected between the output of matching network 30 and plate 12, to derive third and fourth signals respectively replicas of the RF voltage between plate 12 and ground terminal 18 and the RF current flowing in lead 34.

[23] Matching network 30 includes at least a pair of variable reactances adjusted so there is an impedance match between the output impedance of variable power RF amplifier 26 and the impedance of the load formed by plate 12 and the plasma excited by the plate. The variable reactances of matching network 30 are adjusted by DC motors 33 and 35 in response to a function of output signals of voltage-current probe 36, electrically and magnetically coupled to be responsive to the voltage between the interior conductor and ground of cable 28 and the current flowing in the cable. Probe 36, like probes 20 and 32, derives output signals that are replicas of the RF voltage and current the probe detects.

[24] In one preferred embodiment, resonant matching network 30 is mounted on the exterior of chamber 10 and includes a transformer having a variable inductance primary winding, connected in series with a fixed capacitor. The transformer includes a secondary winding magnetically coupled to a variable tuning vane. The RF output of matching network 30 is coupled to plate 10 via DC blocking capacitor 37, serially connected in line 34. Capacitor 37 is also preferably mounted on the exterior of chamber 10. DC motors 33 and 35 respond to output signals of controller 40 to control the matching network tuning vane and the inductance of the primary winding of matching network 30.

[25] Controller 40, including microprocessor 42, electronically programmable read-only memory (EPROM) 44 and random access memory (RAM) 46, responds to the output signals of probes 20, 32 and 36 to derive signals for controlling (1) the constant RF voltage to be maintained across variable impedance 22 and/or the constant RF current that flows through variable impedance 22 or the RF power dissipated in connection 16, (2) the power gain (and thus the output power) of variable power RF amplifier 26, and (3) the voltages applied to DC motors 33 and 35. The signals that controller 40 applies to DC motors 33 and 35 control

the variable reactances of matching network 30 and thus, resonance of network 30 and the impedance match between the output impedance of amplifier 26 and the load it drives.

[26] Controller 40 also derives signals to control impedance 22 between plate 12 and ground terminal 18 so a constant predetermined RF parameter, usually voltage, is maintained in connection 16. EPROM 44 stores setpoint signals for (1) the desired power RF load from plate 12 to ground terminal 18, and (2) the desired constant RF voltage to be maintained between plate 14 and ground terminal 18 in connection 16. Alternatively, EPROM 44 stores a setpoint signal for a constant RF current flowing in connection 16 between plate 14 and ground terminal 18 or setpoint signals for constant RF voltage and both constant RF current associated with connection 16.

[27] Microprocessor 42 responds to the voltage and current replicas that probe 32 derives to calculate the power dissipated in the load between plate 12 and ground terminal 18 in accordance with: $V_1 I_1 \cos\theta$

where:

V_1 is the voltage probe 32 detects,

I_1 is the current probe 32 detects, and

θ is the phase angle between the voltage and current probe 32 detects.

[28] Microprocessor 42 compares the calculated power with a setpoint value of power, as stored in EPROM 44, to derive a control signal for the power gain and output power of amplifier 26. Thereby, the RF power that amplifier 26 supplies to cable 28 is controlled to maintain the RF power between plate 12 and ground terminal 18 constant, at the set point EPROM 44 stores.

[29] Alternatively, microprocessor 42 can respond to the voltage magnitude and phase angle of the replicas probe 32 derives to adjust the voltage RF source 17 applies to one end of cable 28. Or, microprocessor 42 can adjust the RF current that RF source 17 applies to cable 28. If a variable current amplifier is controlled, microprocessor 42 responds to the current replica probe 32 derives, as well as the phase angle between the replicas the probe derives. If the voltage or current that RF source 17 applies to cable 28 is controlled, rather than the power applied to the cable, variable voltage or variable current amplifiers respectively replace the variable power amplifier 26 of RF source 17.

[30] Microprocessor 42 responds to the output signals of probe 36 to control DC motors 33 and 35 and therefore the variable reactances of matching network 30 in a conventional manner. Microprocessor 42 also responds to the magnitude as well as the phase angle of

the voltage and current replicas that probe 20 derives to control the nature and magnitude of variable impedance 22. Microprocessor 42 responds in an iterative manner to the voltage and current replicas that probe 20 derives to connect one or more appropriate impedances of variable impedance 22 in connection 16 and to control the values of the one or more impedances.

[31] Figure 2 is a schematic diagram of a preferred configuration of variable impedance 22. Variable impedance 22 includes four separate branches, respectively including switch contacts 47 (incorporated in switch 60), variable inductor 48, variable capacitor 50 and variable resistor 52. Inductor 48, capacitor 50 and resistor 52 have values respectively controlled by DC motors 54, 56 and 58 in response to signals from microprocessor 42, while contacts 47 are initially closed when a determination is made as to which branch(es) is/are to be connected in connection 16. One terminal of each of contacts 47, inductor 48, capacitor 50 and resistor 52 is connected to ground and the other terminal of these elements is selectively connected by switch 60 to terminal 62, in turn connected to plate 14 via voltage-current probe 20.

[32] Switch 60 includes contacts 47, as well as other contacts (not shown), and three input terminals

responsive to a three-bit signal microprocessor 42
derives to provide eight different combinations for the
connections of contacts 47, inductor 48, capacitor 50
and resistor 52 to terminal 62. In particular, each of
impedance elements 48, 50 and 52 can be separately
connected one at a time to terminal 62 by the contacts
of switch 60 other than contacts 47. The impedance
elements can also be connected in parallel pairs to
terminal 62 or all three of the impedance elements can
be connected in parallel to terminal 62. Initially,
switch 60 is set so impedance 22 is a short circuit by
virtue of contacts 47 being closed.

[33] EPROM 44 stores a desired setpoint value for an RF
parameter associated with connection 16; usually the
parameter is the RF voltage between plate 14 and ground
terminal 18, but the parameter can be the RF current
flowing in connection 16 or the RF power dissipated in
the connection. A typical value of the setpoint is
(20 + j0) volts and the specific embodiment is
described in connection with RF voltage.

[34] Initially, microprocessor 42 responds to the
replicas probe 20 derives when impedance 22 is a short
circuit to calculate the initial complex voltage
between plate 14 and ground terminal 18, based on the
magnitude of the voltage replica and the phase angle
between the voltage and current replicas that probe 20

derives. Microprocessor 42 determines from the initial complex voltage whether the initial phase angle is zero, or whether the voltage is leading the current or the current is leading the voltage.

[35] After microprocessor 42 makes the phase determination, it removes the short circuit of impedance 22 by opening contacts 47. If the initial phase angle between the voltage and current replicas that probe 20 derives is zero, microprocessor 42 supplies a signal to activate switch 60 to connect only resistor 52 in circuit between terminal 62 and ground. This connection is made since the setpoint voltage between plate 14 and ground terminal 18 can be achieved with no reactive component in impedance 22. If the initial voltage and current replicas that probe 20 derives indicate the voltage is leading the current, microprocessor 42 activates switch 60 to connect capacitor 50 and perhaps resistor 52 between terminal 62 and ground. If, however, the replicas that probe 20 derives indicate the voltage lags the current, microprocessor 42 controls switch 60 to connect inductor 48 and perhaps resistor 52 in circuit between terminal 62 and ground.

[36] After the appropriate connection(s) between impedance elements 48, 50 and 52 have been established through switch 60, microprocessor 42 supplies signals

to at least one of motors 54, 56 and 58 to control the value(s) of the impedance element(s) 48, 50 or 52 connected in circuit between terminal 62 and ground. The signals are iteratively supplied to switch 60 and motors 54, 56 and 58 until probe 20 supplies microprocessor 42 with current and voltage replicas that cause the microprocessor to calculate a complex voltage equal to the desired, setpoint voltage between plate 14 and ground terminal 18.

[37] Control of amplifier 26 and variable impedance 22 is, for example, performed during processing of a single workpiece at a foundry. In addition, control of amplifier 26 and variable impedance 22 is performed while different workpieces are being processed at a foundry. In addition, control of amplifier 26 and variable impedance 22 can be provided for different processors during processor manufacturing, at the manufacturer's facility, to assure that different processors of the same nominal type, as delivered, have consistent characteristics.

[38] While there has been described and illustrated a specific embodiment of the invention, it will be clear that variations in the details of the embodiment specifically illustrated and described may be made without departing from the true spirit and scope of the invention as defined in the appended claims.

I claim:

1. In combination,

a vacuum plasma chamber for processing a workpiece, the chamber including a reactive impedance element for electrical coupling with gas in the chamber;

an RF electric source having sufficient power to excite the gas in the chamber to a plasma connected to the reactive impedance element; and

the chamber including an electrode having a connection to RF ground such that no RF energizing source is directly coupled with said electrode, the connection of the electrode to RF ground being such that a finite, non-zero RF voltage has a tendency to be developed between the electrode and RF ground, the connection of the electrode to RF ground including an RF impedance arranged so that there is a substantially constant finite, non-zero RF parameter maintained between the electrode and RF ground.

2. The combination of claim 1 further including a detector arrangement for detecting at least one RF parameter in the connection between the electrode and ground, and a controller connected to be responsive to the detector arrangement for controlling the RF impedance to provide the substantially constant, finite non-zero RF parameter.

3.   The combination of claim 2 wherein the RF impedance includes a variable reactance, the controller being arranged to be responsive to the detector arrangement for controlling the value of the variable reactance.

4.   The combination of claim 3 wherein the variable reactance comprises a variable inductor.

5.   The combination of claim 3 wherein the variable reactance comprises a variable capacitor.

6.   The combination of claim 3 wherein the variable reactance comprises a variable inductance and a variable capacitor, the controller being arranged to be responsive to the detector arrangement for selectively connecting at least one of the variable inductor and the variable capacitor in the connection of the electrode to ground.

7.   The combination of claim 2 wherein the detector arrangement includes a voltage-current probe coupled with the connection.

8.   The combination of claim 2 wherein the detector arrangement is mounted on the chamber.

9.   The combination of claim 8 wherein the RF impedance includes a variable resistor, the controller being arranged to be responsive to the detector arrangement for selectively connecting the variable

resistor in the connection of the RF electrode to RF ground and for controlling the value of the variable resistor.

10. The combination of claim 1 wherein the chamber includes a reactive impedance element for electrical coupling with gas in the chamber;

an RF electric source having sufficient power to excite the gas in the chamber to a plasma;

a matching network connected between the source and the reactive impedance element;

a cable having one end connected to the RF electric source and a second end connected to the matching network;

a probe connected between the matching network and the reactive impedance element for deriving at least one signal indicative of a parameter applied by the matching network to the reactive impedance element and the plasma load coupled to the reactive impedance element; and

a controller connected to be responsive to the at least one signal for controlling a parameter that influences the power the RF electric source applies to said one end of the cable.

11. The combination of claim 10 wherein the probe is arranged for deriving first and second signals respectively indicative of the voltage and current

applied by the matching network to the reactive impedance element and the plasma load coupled to the reactive impedance element and the controller is connected to be responsive to the first and second signals.

12. The combination of claim 10 wherein the probe is mounted within 1/8 of a wavelength of the frequency of the RF source of the vacuum plasma chamber.

13. The combination of claim 11 wherein the probe is mounted on the chamber.

14. The combination of claim 10 wherein the matching network includes a variable reactance, and further including a detector arrangement for deriving signals indicative of the voltage and current supplied by the cable to the matching network, the controller being connected to be responsive to signals indicative of the voltage and current supplied by the cable to the matching network for controlling the variable reactance of the matching network.

15. A method of processing a workpiece with a plasma in a vacuum plasma processing chamber including a reactive impedance element for electrical coupling with gas in the chamber and an electrode having a connection to RF ground such that no RF energizing source is directly coupled with said electrode, the connection of the electrode to RF ground being such

that a finite, non-zero RF voltage has a tendency to be developed between the electrode and RF ground, the method comprising supplying sufficient power to the reactive impedance element to excite the gas in the chamber to a plasma while maintaining in the connection a constant finite, non-zero RF parameter between the electrode and RF ground.

16.   The method of claim 15 wherein the parameter is voltage and the connection of the electrode to RF ground includes an RF impedance, and the constant finite non-zero RF voltage between the electrode and RF ground being provided by maintaining a constant RF voltage across the RF impedance in the connection.

17.   The method of claim 15 wherein the method is performed in connection with a plurality of workpieces and the same constant, non-zero RF voltage is maintained between the electrode and ground while the plural workpieces are being processed in the vacuum plasma chamber.

18.   The method of claim 15 wherein the method is performed in connection with a plurality of vacuum plasma chambers having substantially the same characteristics, the characteristics of the different vacuum plasma chambers differing from each other sufficiently to cause a tendency for different non-zero RF voltages to be developed in the connection between

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the electrode and ground of the different chambers, and maintaining the same constant, non-zero RF voltages between the electrode and ground in the plural vacuum plasma chambers operating under the same nominal conditions.

19. The method of claim 15 wherein the method is performed by connecting an RF impedance in the connection between the electrode and RF ground and maintaining a constant finite, non-zero RF voltage across the RF impedance.

20. The method of claim 19 wherein the constant finite, non-zero RF voltage between the electrode and ground is maintained by detecting at least one parameter indicative of RF voltage between the electrode and ground and RF current flowing between the electrode and ground, and controlling the RF impedance in response to the detected at least one parameter.

21. The method of claim 20 wherein the finite non-zero RF voltage across the RF impedance is maintained constant by controlling the value of the RF impedance in response to the detected at least one parameter.

22. The method of claim 21 wherein the RF impedance includes a variable capacitor and a variable inductor, and the controlling step includes connecting

one or both of the variable capacitor and variable inductor in the connection between the electrode and RF ground.

23.   The method of claim 21 wherein the RF impedance includes a variable reactance and a variable resistor, and the controlling step includes connecting one or both of the variable reactance and the variable resistor in the connection between the electrode and RF ground, and controlling the value of the variable reactance or the variable resistor.

24.   The method of claim 15 wherein the chamber includes a reactive impedance element electrically coupled with gas in the chamber, a matching network connected between the source and the reactive impedance element, a cable having one end connected to the RF electric source and a second end connected to the matching network, the method further comprising

detecting at least one parameter applied by the matching network to the reactive impedance element and the plasma load coupled to the reactive impedance element; and

controlling a parameter that influences the RF power the RF electric source applies to said one end of the cable in response to the detected at least one parameter.

25. The method of claim 24 wherein the at least one parameter includes voltage and current.

26. The method of claim 24 wherein the voltage and current are detected within 1/8 of a wavelength of the frequency of the RF source of the vacuum plasma chamber.

27. The method of claim 24 wherein the voltage and current are detected at the chamber.

28. The method of claim 24 wherein the controlling step includes controlling a plurality of parameters that influence the RF power the RF electric source applies to said one end of the cable.

29. The method of claim 24 wherein the matching network includes a variable reactance, the method further comprising the steps of detecting the voltage and current supplied by the cable to the matching network, and controlling the variable reactance of the matching network in response to the detected voltage and current supplied by the cable to the matching network.
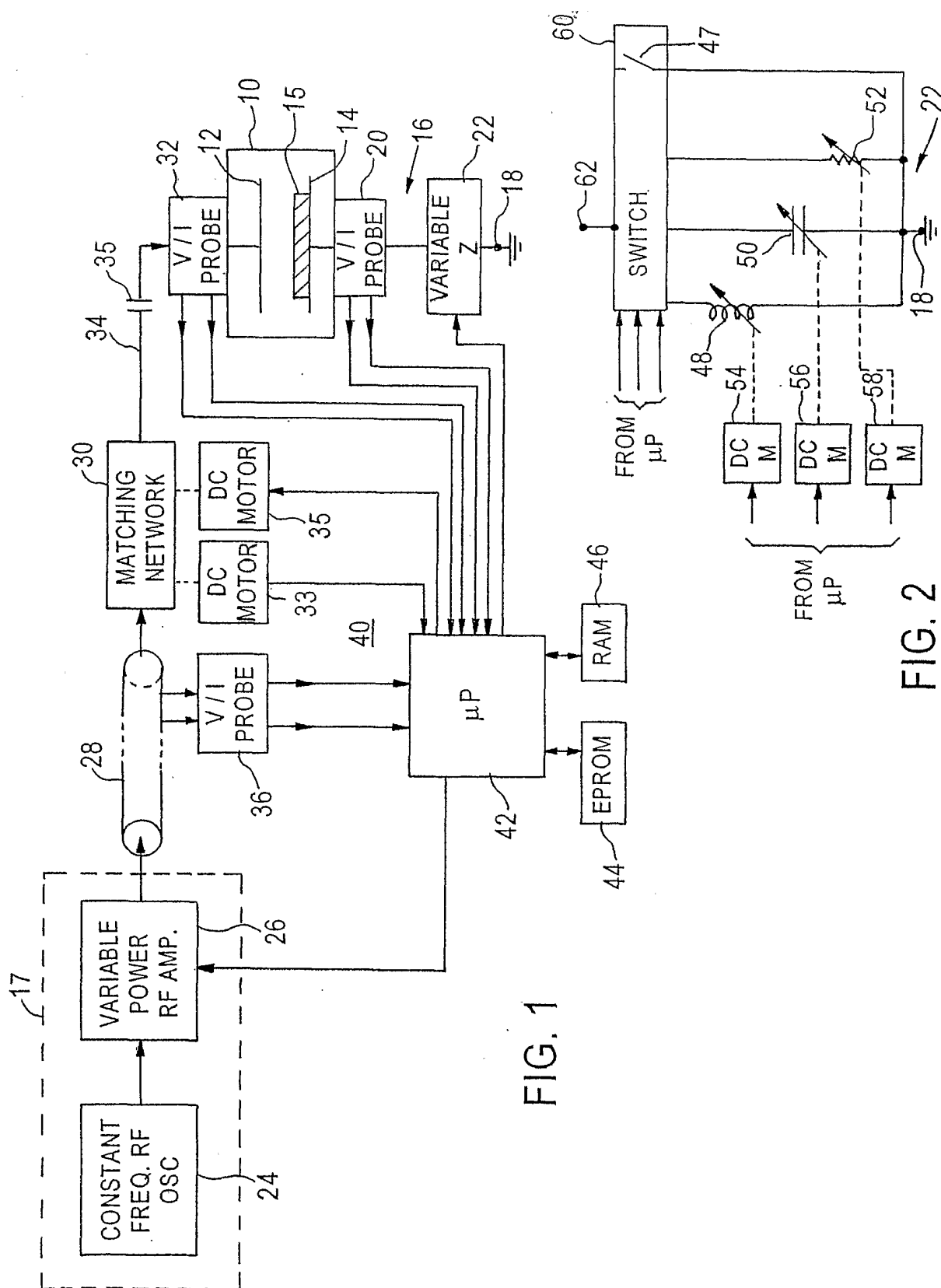
FIG. 1

FIG. 2

INTERNATIONAL SEARCH REPORT

A. CLASSIFICATION OF SUBJECT MATTER
IPC 7   H01J37/32

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)
IPC 7   H01J

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data, PAJ

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category°	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	US 4 622 094 A (OTSUBO TORU) 11 November 1986 (1986-11-11) column 7, line 1 -column 8, line 30 figure 7	1,2,8, 15,19
X	US 4 333 814 A (KUEYEL BIROL) 8 June 1982 (1982-06-08) column 5, line 1 - line 33 figure 7	1,2,8, 15,19
X	PATENT ABSTRACTS OF JAPAN vol. 007, no. 280 (E-216), 14 December 1983 (1983-12-14) -& JP 58 158929 A (KOKUSAI DENKI KK), 21 September 1983 (1983-09-21) abstract	1,2,8, 15,19

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| X | Further documents are listed in the continuation of box C. | X | Patent family members are listed in annex. |

° Special categories of cited documents :

"A" document defining the general state of the art which is not considered to be of particular relevance

"E" earlier document but published on or after the international filing date

"L" document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)

"O" document referring to an oral disclosure, use, exhibition or other means

"P" document published prior to the international filing date but later than the priority date claimed

"T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention

"X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

"Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such docu- ments, such combination being obvious to a person skilled in the art.

"&" document member of the same patent family

Date of the actual completion of the international search	Date of mailing of the international search report
20 September 2002	02/10/2002

Name and mailing address of the ISA	Authorized officer
European Patent Office, P.B. 5818 Patentlaan 2 NL – 2280 HV Rijswijk Tel. (+31-70) 340-2040, Tx. 31 651 epo nl, Fax: (+31-70) 340-3016	Aguilar, M.

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT		
Category °	Citation of document, with indication,where appropriate, of the relevant passages	Relevant to claim No.
X	US 5 605 576 A (AIHARA MASAMI ET AL) 25 February 1997 (1997-02-25) column 2, line 54 -column 3, line 41 figure 1	1,15
X	US 5 705 019 A (YAMADA MINORU ET AL) 6 January 1998 (1998-01-06) column 4, line 10 -column 7, line 45 figure 5	1,15

Patent document cited in search report		Publication date	Patent family member(s)		Publication date
US 4622094	A	11-11-1986	JP	1816401 C	18-01-1994
			JP	5027244 B	20-04-1993
			JP	60126832 A	06-07-1985
			DE	3479843 D1	26-10-1989
			EP	0145015 A2	19-06-1985
			KR	8904882 B1	30-11-1989
US 4333814	A	08-06-1982	CA	1157914 A1	29-11-1983
			DE	3071203 D1	28-11-1985
			EP	0031704 A2	08-07-1981
			GB	2067137 A ,B	22-07-1981
			JP	56100423 A	12-08-1981
JP 58158929	A	21-09-1983	JP	1019261 B	11-04-1989
			JP	1536399 C	21-12-1989
US 5605576	A	25-02-1997	JP	2592217 B2	19-03-1997
			JP	7147200 A	06-06-1995
			KR	137323 B1	01-06-1998
US 5705019	A	06-01-1998	JP	2956494 B2	04-10-1999
			JP	8124862 A	17-05-1996