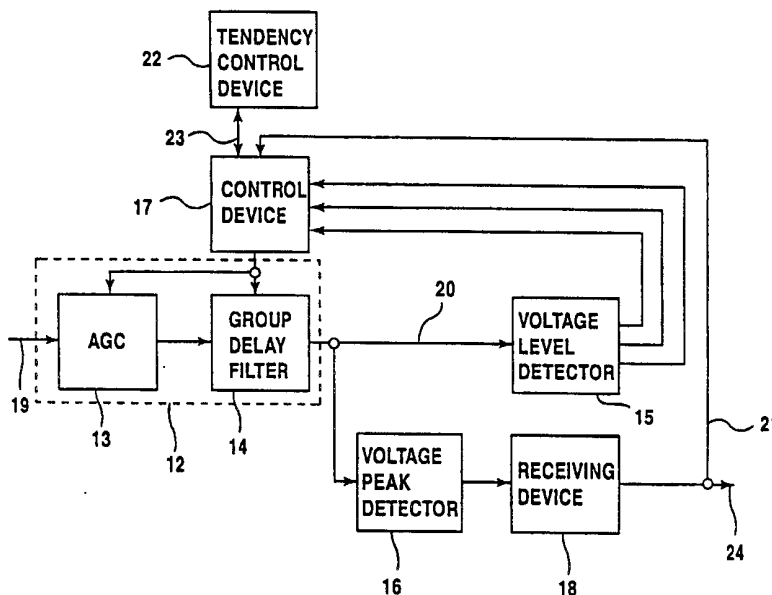




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(54) Title: METHOD AND DEVICE FOR DISTURBANCE SUPPRESSION IN A SIGNAL



(57) Abstract

Disturbances in bipolar signals transmitted on a transmission line are suppressed by adjusting the amplitude of the signal with coarse adjustment steps and with fine adjustment steps. The signals are compared to at least a lower reference level and an upper reference level for determining a first percentage of the signals violating the lower reference level and a second percentage of the signals violating the upper reference level. The coarse adjustment steps and the fine adjustment steps are selected in accordance with an adjustment characteristic which evaluates at least the first percentage and the second percentage. The adjustment characteristic is changed if a tendency of the signals is detected and a check is performed if the step of changing the adjustment characteristic improves the disturbance suppression. A device for suppressing disturbances is also provided.

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METHOD AND DEVICE FOR DISTURBANCE SUPPRESSIONIN A SIGNAL

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Background of the Invention:Field of the Invention:

The invention relates to a method and a device for suppressing disturbances in a signal transmitted on a transmission line, in particular for suppressing disturbances in a PCM (Pulse Code Modulation) signal.

PCM transmission lines are increasingly used for transmitting digital data over long distances. Increasing the distance over which data are transmitted makes it more difficult to meet the requirements regarding the suppression of interference, noise, jitter, and other disturbances on the transmission line. Long transmission lines may also result in an increased amplitude damping for the electronic components connected to the transmission line.

A reduced supply voltage results in a severe reduction of the available voltage modulation range. If, for example, the worst case supply voltage of a device is 4.75 V and the voltage modulation range is reduced by 1 V with respect to the supply potential and the ground potential, respectively, the

resulting voltage modulation range is 2.75 V. If, however, the worst case supply voltage is 3.15 V and the voltage modulation range is reduced by 0.8 V with respect to the supply potential and the ground potential, respectively, the

5 resulting voltage modulation range is only 1.55 V. The available voltage modulation range of an input signal at an input of an electronic component may furthermore be reduced by ESD (Electrostatic Discharge) protection diodes. The ESD protection diodes result in a reduced range for the signal

10 voltage amplitudes at the input of the electronic component. This leads to a low signal-to-noise ratio which even declines when long transmission lines are used, since the noise environment is kept to be the same due to required noise resistances in applications. In addition, the signal-to-noise

15 ratio deteriorates as the length of the transmission line increases.

The noise at an input of an electronic component includes the self-noise at the input of the electronic component and the

20 interference caused by disturbances picked up by transformers used for coupling signals into the transmission line. The disturbances picked up by the transformer are a major part of the NEXT (near-end cross talk) noise. The problem of cross talk is for example discussed in the article "Statistical

25 Behavior of Crosstalk Power Sum with Dominant Components" in Bell Syst. Tech J., Vol. 60, No. 7.

A further problem arises from the temperature requirements that a transmission system must meet. The electronic components and the transmission line must operate in a
5 temperature range of, for example, -40°C to $+85^{\circ}\text{C}$ without violating a maximum allowable bit error rate that the transmission system must be able to tolerate.

A transmission system furthermore has to be able to meet its
10 transmission requirements even when a sending unit varies the amplitude of the sending signal. The PCM24 standard, for example, allows a variation of the amplitude of the sending signal in the range of between 2.4 V and 3.6 V in a time period of $102 \times 650 \text{ nsec}$.

15

A prior art receiver receiving a signal is, for example, configured such that the signal amplitude is adjusted to have a desired value. A group delay equalization is performed in accordance with a fixed group delay equalization function that
20 is based on known transmission properties of the transmission line. The transmission line is for example a cable whose transmission properties have been measured. In other words, if the correct amplification, which results in the desired amplitude, has been found, then the group delay equalization
25 is also adjusted correctly. This adjustment method however presupposes that specific types of transmission lines or

cables with known properties, predetermined amplitudes at the transmitter, and specific external circuits are used.

After the adjustment of the amplitude and the group delay of
5 the received signal, the thus reconstructed signal fluctuates
within boundaries due to noise, due to an amplitude modulation
of the input signal, due to temperature changes, or due to any
other disturbances. If the reconstructed signal fluctuates
beyond these boundaries, then the signal may be adjusted
10 again. German Published, Non-Prosecuted Patent Application DE
197 35 752 A1 describes such a system that readjusts the
signals as soon as they leave their boundaries. A problem of
a system that works as described above is that the modulation
range may be severely restricted, as is described in the
15 article "A 284 kbit/s ISDN Burst Transceiver", in IEEE Journal
of Solid State Circuits, Vol. SC 22 No. 6, 1988.

Summary of the Invention:

It is accordingly an object of the invention to provide a
20 method and a device for disturbance suppression in a signal
which overcome the above mentioned disadvantages of the
heretofore-known methods and devices of this general type and
which effectively suppress modulation disturbances and at the
same time keep bit error rates low.

25

- With the foregoing and other objects in view there is provided, in accordance with the invention, a method for suppressing disturbances in signals transmitted on a transmission line, which comprises receiving signals having an amplitude; adjusting the amplitude of the signals with coarse adjustment steps and with fine adjustment steps for a disturbance suppression; subsequently comparing the signals to at least a lower reference level and an upper reference level for determining a first percentage of the signals violating the lower reference level and a second percentage of the signals violating the upper reference level; selecting the coarse adjustment steps and the fine adjustment steps in accordance with an adjustment characteristic which evaluates at least the first percentage and the second percentage; changing the adjustment characteristic if a tendency of the signals is detected; and checking if the step of changing the adjustment characteristic improves the disturbance suppression.
- 20 In accordance with another mode of the invention, the adjustment characteristic is changed such that an adjustment range of the fine adjustment steps is increased.

In accordance with a further mode of the invention, modulation disturbances are suppressed with the step of adjusting the amplitude of the signals.

In accordance with yet another mode of the invention, the coarse adjustment steps are performed with a filter selectively operating with a first filter stage suited for signals received from a transmission line having a first length and operating with a second filter stage suited for signals received from a transmission line having a second length, the second length being greater than the first length; the fine adjustment steps covering a first amplitude range, when the filter operates with the first filter stage and covering a second amplitude range when the filter operates with the second filter stage, the first amplitude range being greater than the second amplitude range.

In accordance with a further mode of the invention, the first amplitude range is extended by a range of at least 3dB beyond a lower boundary and beyond an upper boundary of the second amplitude range.

In accordance a yet a further mode of the invention, the range of the at least 3dB is divided into three fine adjustment steps each having a step width of at least 1dB.

In accordance a yet another mode of the invention, the adjustment characteristic is changed in dependence of the step of checking if the disturbance suppression has improved.

With the objects of the invention in view there is also provided a device for suppressing disturbances in signals transmitted on a transmission line, comprising an equalizer receiving signals having an amplitude and adjusting the

5 amplitude of the signals; the equalizer having a gain control device for providing fine adjustment steps and having a filter connected to the gain control device, the filter providing coarse adjustment steps; a level detector connected to the equalizer for comparing the signals to at least a lower
10 reference level and an upper reference level; a control device connected to the equalizer for determining a first percentage of the signals violating the lower reference level and a second percentage of the signals violating the upper reference level, for detecting a tendency of the signal, for controlling
15 the equalizer to perform the coarse adjustment steps and the fine adjustment steps in accordance with an adjustment characteristic which evaluates at least the first percentage and the second percentage, and for changing the adjustment characteristic on detection of the tendency of the signals.

20

In accordance with another feature of the invention, the control device changes the adjustment characteristic by increasing an adjustment range of the fine adjustment steps.

25 In accordance with yet another feature of the invention, the filter has a first filter stage suited for signals received

from a transmission line having a first length and has a second filter stage for signals received from a transmission line having a second length, the second length being greater than the first length, the fine adjustment steps covering a first amplitude range, when the filter operates with the first filter stage and covering a second amplitude range when the filter operates with the second filter stage, the first amplitude range being greater than the second amplitude range.

Although the invention is illustrated and described herein as embodied in a method and a device for disturbance suppression, it is nevertheless not intended to be limited to the details shown, since various modifications and structural changes may be made therein without departing from the spirit of the invention and within the scope and range of equivalents of the claims.

The construction and method of operation of the invention, however, together with additional objects and advantages thereof will be best understood from the following description of specific embodiments when read in connection with the accompanying drawings.

Brief Description of the Drawings:

Fig. 1 is a schematic configuration of a transmission system in which the method and device according to the invention can be used;

5

Fig. 2 is a schematic diagram of a transmission system with disturbances introduced into the transmission system;

Fig. 3 is a signal diagram of an undisturbed bipolar signal;

10

Fig. 4 is a signal diagram of a bipolar signal with disturbances;

Fig. 5 is a schematic block diagram of a device for

15 suppressing disturbances according to the invention;

Fig. 6 is a diagram indicating reference levels used in the device of Fig. 5;

20 Fig. 7 is a diagram of a bipolar signal which is sampled for detecting whether the reference levels are exceeded;

Fig. 8 is a diagram illustrating the tendency of a signal amplitude;

25

Figs. 9 and 10 are diagrams illustrating decision criteria used in the device of Fig. 5;

Fig. 11 is an eye pattern illustrating a variation in the
5 signal amplitude due to a variation in the data pulse density;

Fig. 12 illustrates equalizer stages for various cable lengths;

10 Fig. 13 is a diagram of the envelope of a modulated input signal and a reconstructed signal according to a fine adjustment with a tendency control;

Fig. 14 is an illustration of a switchover between two
15 different equalizer characteristics;

Fig. 15 is a diagrammatic view of an exemplary group delay filter; and

20 Fig. 16 is a graph illustrating the steps during a coarse adjustment and a fine adjustment.

Description of the Preferred Embodiments:

Referring now to the figures of the drawings in detail and
25 first, particularly, to Fig. 1 thereof, there is shown a schematic configuration of a transmission system which is used

for PCM ISDN (Integrated Services Digital Network) data transmission. The transmission system has a transmitter 1 and a receiver 2. The output of the transmitter 1 is connected to a first transformer 3. The first transformer 3 is also
5 connected to the transmitter side of a transmission line 4, which may for example be a 22 AWG cable. The receiver side of the transmission line 4 is connected to a second transformer 5, which has a transformer ratio of 1:N. The second transformer 5 is also connected to an input of the receiver 2.
10 The input of the receiver may also be connected to an external component 6, which is an interface circuit that performs functions such as transmission line adaption, impedance matching or component protection.

15 The transmitter 1 generates an output signal at its output. The output signal is coupled into the transmission line 4 via the first transformer 3. The signal is then received at the input of the receiver 2 via the second transformer 5.

20 Fig. 2 illustrates how disturbances that are coupled into the transmission system are simulated in a model. The output signal from the output of the transmitter 1 is modulated by a sine-wave modulation generator 7 that simulates noise generation in the transmission line. The far end crosstalk is
25 simulated by a FEXT (Far End Crosstalk) generator 8, the near

end crosstalk is simulated by a NEXT (Near End Crosstalk) generator 9.

The signal diagram of Fig. 3 shows an undisturbed bipolar
5 signal 10, after it has been output by the transmitter 1 and
after it has been transmitted through the first transformer 3.
At this point, the bipolar signal on the output side of the
first transformer 3 is not yet affected by any disturbance.
The pulses shown in Fig. 3 indicate digital values "1".

10

Fig. 4 illustrates a disturbed bipolar signal 11. A signal
transmitted over the transmission line 4 is affected by
various forms of disturbances such as noise, crosstalk, and
group delay. The undisturbed bipolar signal 10 shown in
15 Fig. 3 is thus transformed by those disturbances into the
disturbed bipolar signal illustrated in Fig. 4. The disturbed
signal needs to be reconstructed in order to reverse the
effects of the disturbances. The invention is described as
being used to suppress disturbances of bipolar signals, but
20 can of course be used in the same manner for suppressing
disturbances of unipolar signals.

Fig. 5 illustrates an exemplary embodiment of a device for
suppressing disturbances. An equalizer 12 for equalizing the
25 amplitude and the group delay of the received signals includes
an automatic gain control device 13, also called AGC, and a

group delay filter 14 connected to the automatic gain control device 13. A disturbed signal, such as the disturbed bipolar signal shown in Fig. 4, which is transmitted on a receive path 19 of a transmission line 4, is received at an input of the automatic gain control device 13. The output of the automatic gain control device 13 is connected to the group delay filter 14 and the disturbed bipolar signal is transferred from the automatic gain control device 13 to the group delay filter 14. The output of the group delay filter 14 is connected in parallel to an input of a voltage level detector 15 and to an input of a voltage peak detector 16.

An exemplary embodiment of the group delay filter 14 is described in further detail in Fig. 15. The group delay filter 14 preferably includes an anti-aliasing filter 29, a main equalizing filter 30, and a reconstruction filter 31 connected in series. The anti-aliasing filter 29 may be embodied as a low-pass filter for filtering out high frequency disturbances. The anti-aliasing filter may be described using a time-continuous transfer function. The actual equalization of the bipolar signal is performed in the main equalizing filter 30. The main equalizing filter may be characterized by a z-transfer function $H(z)$. The properties of the transmission line 4 determine the position of the poles and zero points of the z-transfer function $H(z)$ which is used to describe the main equalizing filter 30. A long length for the

transmission line 4 results in z-transfer function $H(z)$ having a high order. In such cases it is preferable to divide the main equalizing filter 30 into equalizer stages, each stage having a z-transfer function. The reconstruction filter 31 is
5 for example a low-pass filter that can be described in terms of a time-continuous transfer function $H(s)$.

The voltage level detector 15 may for example have three outputs for outputting signals indicating a violation of
10 voltage reference levels by the signals on line 20. These three outputs are connected to inputs of a control device 17 in order to form a feedback loop. An output of the control device 17 is connected to the automatic gain control 13 and to the group delay filter 14 for controlling these devices. The
15 output of the voltage peak detector 16 is connected to the input of a receiving device 18. The receiving device 18 receives the bipolar signals, processes them and outputs a data stream on an output line 24. The data stream may be further processed or transmitted on a further transmission
20 line. The output of the receiving device 18 is also connected to the control device 17 via a feeding back line 21 for feeding back the data stream. The control device 17 checks if there is a data stream present at the output of the receiving device 18 and evaluates the data stream for controlling the
25 equalizer 12. A tendency control device 22 is connected to the control device 17 via a bi-directional line 23.

The bipolar signals of the data stream on the receive path 19 are affected by the disturbances illustrated in Fig. 2. The automatic gain control 13 and the group delay filter 14
5 operate to reconstruct the amplitude of the bipolar signal to a constant nominal value. For this purpose, the voltage course on the output line 20 of the equalizer 12 is checked for violation of reference levels a, b, and c illustrated in Fig. 6. A lower reference level a and an upper reference
10 level c are used for a coarse adjustment of the voltage amplitude in the automatic gain control 13 and in the group delay filter 14. A middle reference level b for a fine adjustment is located at a distance D from the upper reference level c. The fine adjustment is achieved by setting the
15 automatic gain control 13 to a desired gain value. The coarse adjustment is achieved by setting the gain in the group delay filter 14. In the exemplary embodiment only five different settings for the gain are used in the group delay filter 14. These settings for the gains are also referred to as equalizer
20 stages. The range of modulation A of the automatic gain control 13 and the group delay filter 14 determines the distance between the upper reference level c and the lower reference level a. The distance D between the upper reference level c and the middle reference level b is determined by the
25 sum S of all disturbances to be taken into account and the range of modulation of the fine adjustment F of the automatic

gain control 13 and the group delay filter 14. In other words $D = F + S$ is determined by the relationship $F/2 + S = A/2$.

The invention furthermore uses so-called virtual reference levels a_0 and c_0 for a virtual expansion of the range of

5 modulation A.

Fig. 7 illustrates how a bipolar signal is sampled in order to detect whether the reference levels a, b, and c are exceeded.

In the example illustrated in Fig. 7 all three reference

10 levels a, b, and c are exceeded or violated. In response to output signals from the voltage level detector 15 which

indicate a violation of the reference levels a, b, or c and

the receiving device 18, the control device 17 adjusts the

setting of the automatic gain control 13 and the transfer

15 function of the group delay filter 14. The setting of the

automatic gain control 13 is adjusted in accordance with the

fine adjustment. The transfer function of the group delay

filter 14 is adjusted in accordance with the coarse

adjustment. The control device 17 further operates in

20 response to the tendency control device 22. The tendency

control 22 operates based on an evaluation of output signals

from the voltage level detector 15 and the receiving device

18. Fig. 8 illustrates the tendency of a voltage signal. The

tendency of a signal is the first derivative of the envelope

25 of the peaks of the signal amplitudes.

The bipolar signals are sampled in order to detect whether the reference levels a, b, and c are exceeded, in other words whether the reference levels are violated. Preferably a subsampling is used in which the sampling period is a multiple
5 of a zero-clock period of the signal to be sampled. A zero-clock period refers to a data stream having a minimum density of logical "1" signals, for instance a data stream having 15 "0" signals and three "1" signals. Due to the subsampling, a misinterpretation of the sampled levels may occur randomly.

10 These randomly occurring misinterpretations have to be taken into account when the sampled signal is evaluated. The violations of the voltage levels are preferably evaluated in the control device 17 in accordance with a random evaluation based on the principle of a median filtering. The evaluation
15 is based on decision criteria supplied by the tendency control device 22.

Figs. 9 and 10 illustrate in an exemplary manner the decision criteria for evaluating a violation of the upper reference
20 level c and the lower reference level a, respectively. Fig. 9 indicates a first decision threshold Z1 and second decision threshold Z2. If the upper level c is exceeded by a percentage value that is less than Z2, then the equalizer is adapted and no correction is necessary. If the upper level c
25 is exceeded by a percentage value between Z2 and Z1, then a fine adjustment is performed. If the upper level c is

exceeded by a percentage value larger than Z1, then a coarse adjustment is performed. Fig. 10 illustrates the behavior of the control device 17 in response to a violation of the lower reference level a. A third decision threshold Z3 and fourth
5 decision threshold Z4 are shown. If the lower level a is exceeded by a percentage value that is less than Z4, then a coarse adjustment is performed. If the lower level a is exceeded by a value between Z4 and Z3, then a fine adjustment is performed. If the lower level a is exceeded by a value
10 larger than Z3, then the equalizer is adapted and no correction is necessary. In Figs. 9 and 10, the decision thresholds Z1 and Z4 are a function of the symbol density of the bipolar signal whereas the decision thresholds Z2 and Z3 are constant with respect to the symbol density. Of course,
15 the decision thresholds Z2 and Z3 could also be provided as a function of the symbol density. The examples shown in Figs. 9 and 10 have decision thresholds Z1 to Z4 which are based on exceeding the upper and the lower threshold level by 50%.
This means that the number of bipolar signals in the data
20 stream whose amplitude exceeds the decision thresholds during the coarse adjustment or during the fine adjustment is 50%.

By detecting that violations of the reference levels occur occasionally it is possible to detect that an input bipolar
25 signal is modulated by a disturbance. Permitting violations of the reference levels up to a certain percentage value

causes an increased tolerance toward this modulation disturbance.

A coarse adjustment may for instance be necessary after
5 switching the transmission line or after a drastic jump in the signal amplitude. If, for instance, the lower reference level a is only exceeded by 5% of the pulses, a quick correction is necessary.

10 A fine adjustment may be necessary if the transmitter 1 leaves its specified operating range or if the voltage supply for the device shown in Fig. 5 leaves its specified operating range.
A fine adjustment may also be necessary due to random disturbances that are outside the frequency range in which the
15 median filtering suppresses disturbances. The fine adjustment is less likely to generate bit errors than the coarse adjustment.

Long-term effects such as temperature changes along the
20 transmission path or a drift of the voltage level of the transmitter are evaluated in a so called verification period.
The reconstructed signal amplitudes move within the voltage range bounded by the lower reference level a and the upper reference level c . In order to have a maximum modulation
25 range as well as long-term stability and a quick response time, the violation of the reference levels is determined by a

random evaluation, which takes into account the percentage of pulses that violate the reference levels. The evaluation thus distinguishes between a coarse adjustment and a fine adjustment. During the verification period, the middle
5 reference level is preferably no longer determined by an output signal of the voltage level detector 15 but is determined by the regenerated data stream that is supplied to the control device 17 via the feed back line 21.

10 In a preferred embodiment, an unnecessary termination of a fine adjustment or a termination of the verification period due to a drastic change in the statistic distribution of 0's and 1's in the data stream statistic is prevented by taking into account how the signal amplitude varies due to changes in
15 the pulse densities in the data stream. Fig. 11 is an eye pattern that shows such a variation in the signal amplitude of a reconstructed signal due to a variation in the data pulse density. S_1 indicates the path that results from a high pulse density, S_2 indicates the path that results from a low pulse
20 density or signal density. The variation in amplitude is indicated as ΔA . The evaluation of the violation of the levels a, b, and c is performed under the constraint that a certain amplitude variation ΔA is acceptable.

The tendency control device 22 operates such that the criteria for controlling the automatic gain control 13 and the group delay filter 14 can be changed during operation. In this case, the percentages of pulses violating the thresholds shown in Figs. 9 and 10 are changed during operation. Changing the criteria for controlling the automatic gain control 13 and the group delay filter 14 results in a system that has a high tolerance for severe random disturbances. In case the result of an evaluation of thresholds using such changed criteria is used in a further switching from one adjustment mode to another adjustment mode, then a closed loop control of the adjustment is achieved. In such a closed loop control of the adjustments, the fine adjustment may be varied in a predefined range and in a predefined manner. For the fine adjustment, the gain of the automatic gain control 13 may for example be switched step-wise up and down three times covering a total gain range of 4.2 dB. In case this variation in the fine adjustment does not result in a satisfactory result, the criteria for controlling the automatic gain control 13 and the group delay filter 14 can be changed again, in other words the percentages of pulses violating the thresholds shown in Figs. 9 and 10 are changed again. The described operation provides an extended range of adjustment by adding a so-called virtual adjustment range that extends beyond the lower reference level a and the upper reference level c. The added range of operation is indicated by the virtual reference levels a_0 and c_0 .

in Fig. 6. An equalizer stage represents a range of the length of the transmission line for which identical combinations of poles and zero points of the equalizer transfer function are used. In the exemplary example five
5 equalizer stages are used. In other words the group delay filter 14 can be set to any of those five equalizer stages. Each equalizer stage results in a specific gain by which the bipolar signal is amplified. Fig. 12 illustrates five
10 equalizer stages for cable lengths of 0 to 2000 m. The reconstructed course of the amplitudes for the five equalizer stages is also shown in Fig. 12. The equalizer stages overlap in their operating ranges such that there are no gaps in the coarse adjustment.

15 The graph in Fig. 16 illustrates the steps during a coarse adjustment and a fine adjustment. The bipolar signal is compared to three reference levels and is evaluated in the voltage level detector 15. If the evaluation detects that the upper or lower reference levels are violated in accordance
20 with criteria as shown in Figs. 9 and 10, then an adjustment of the equalizer setting may be necessary. If necessary, the group delay filter 14 is switched in a coarse adjustment step. The group delay filter 14 is switched upwards in coarse
adjustment steps by switching to increasingly higher equalizer
25 stages until the upper reference level c is violated. In a subsequent step, the coarse adjustment is switched back by one

equalizer stage. Then a fine adjustment is performed. During the fine adjustment, the gain of the automatic gain device 13 is switched up step-wise until the middle reference level b is violated. Fig. 16 illustrates the above-mentioned steps for a configuration using six stages for the automatic gain control (AGC stages) and using a bypass (stage 0) and four equalizer stages (stage 1-4) for the group delay filter 14. Except for the bypass each of the four equalizer stages uses as a default stage for the automatic gain control the third AGC stage.

The range of modulation A can be varied based on information on the set stage of the equalizer for a tendency control. In accordance with the invention, the first stage of the equalizer is adjusted to have a characteristic that is

different from the standard characteristic shown in Figs. 9 and 10. This has the effect that the limit frequency for suppressing periodic disturbances is lowered. Fig. 13 shows the envelope 25 of a modulated bipolar input signal and an envelope 26 of the reconstructed signal after changing the equalizer characteristic from the standard characteristic to a characteristic having a low limit frequency for periodic disturbances. Fig. 13 shows that the periodic disturbance or modulation indicated by 25 is suppressed in the reconstructed signal indicated by 26. The adjustment of the characteristic of the equalizer allows a fine adjustment that can follow an extended range of amplitudes without triggering a coarse

adjustment. The extended range of operation of the fine adjustment prevents a switching over to a coarse adjustment and thus prevents an increase in bit errors.

5 Fig. 14 illustrates a switchover of the equalizer characteristic from a standard characteristic 27 as shown in Fig. 9 to a characteristic 28 having a pronounced range for the fine adjustment. The switchover from the standard characteristic 27 to the characteristic 28 is performed in
10 dependence on the tendency of the voltage signal.

We claim:

1. A method for suppressing disturbances in signals transmitted on a transmission line, the method which comprises:

receiving signals having an amplitude;

adjusting the amplitude of the signals with coarse adjustment steps and with fine adjustment steps for a disturbance suppression;

subsequently comparing the signals to at least a lower reference level and an upper reference level for determining a first percentage of the signals violating the lower reference level and a second percentage of the signals violating the upper reference level;

selecting the coarse adjustment steps and the fine adjustment steps in accordance with an adjustment characteristic which evaluates at least the first percentage and the second percentage;

changing the adjustment characteristic if a tendency of the signals is detected; and

checking if the step of changing the adjustment characteristic improves the disturbance suppression.

2. The method according to claim 1, which comprises changing the adjustment characteristic such that an adjustment range of the fine adjustment steps is increased.

3. The method according to claim 1, which comprises suppressing modulation disturbances with the step of adjusting the amplitude of the signals.

4. The method according to claim 1, which comprises performing the coarse adjustment steps with a filter selectively operating with a first filter stage suited for signals received from a transmission line having a first length and operating with a second filter stage suited for signals received from a transmission line having a second length, the second length being greater than the first length;

the fine adjustment steps covering a first amplitude range, when the filter operates with the first filter stage and covering a second amplitude range when the filter operates with the second filter stage, the first amplitude range being greater than the second amplitude range.

5. The method according to claim 4, which comprises extending the first amplitude range by a range of at least 3dB beyond a lower boundary and beyond an upper boundary of the second amplitude range.

6. The method according to claim 5, which comprises dividing the range of the at least 3dB into three fine adjustment steps each having a step width of at least 1dB.

7. The method according to claim 1, which comprises further changing the adjustment characteristic in dependence of the step of checking if the disturbance suppression has improved.

8. The method according to claim 1, wherein the signals are bipolar signals.

9. A device for suppressing disturbances in signals transmitted on a transmission line, comprising:

an equalizer receiving signals having an amplitude and adjusting the amplitude of the signals;

said equalizer having a gain control device for providing fine adjustment steps and having a filter connected to said gain control device, said filter providing coarse adjustment steps;

a level detector connected to said equalizer for comparing the signals to at least a lower reference level and an upper reference level;

a control device connected to said equalizer for determining a first percentage of the signals violating said lower reference level and a second percentage of the signals violating said upper reference level, for detecting a tendency of the signals, for controlling said equalizer to perform said coarse adjustment steps and said fine adjustment steps in accordance with an adjustment characteristic which evaluates at least said first percentage and said second percentage, and for changing said adjustment characteristic on detection of the tendency of the signals.

10. The device according to claim 9, wherein said control device changes said adjustment characteristic by increasing an adjustment range of said fine adjustment steps.

11. The device according to claim 9, wherein said filter has a first filter stage suited for signals received from a transmission line having a first length and has a second filter stage for signals received from a transmission line having a second length, the second length being greater than the first length, said fine adjustment steps covering a first amplitude range, when said filter operates with said first

filter stage and covering a second amplitude range when said filter operates with said second filter stage, said first amplitude range being greater than said second amplitude range.

Fig.1

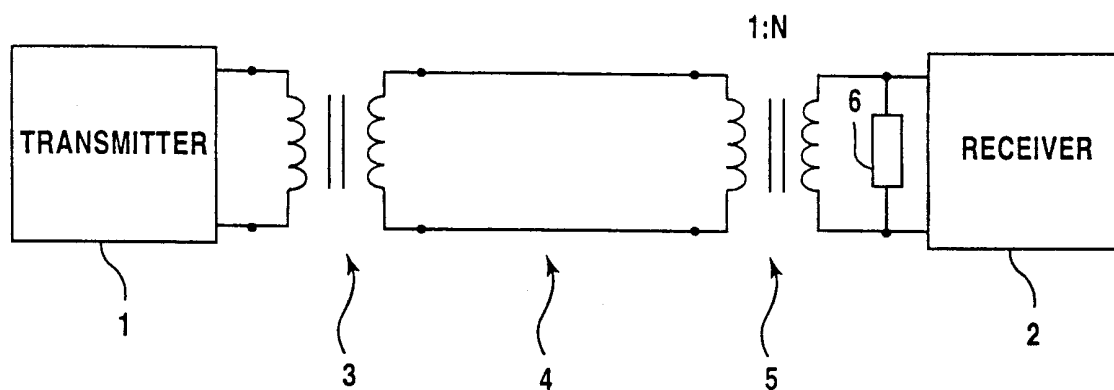


Fig.2

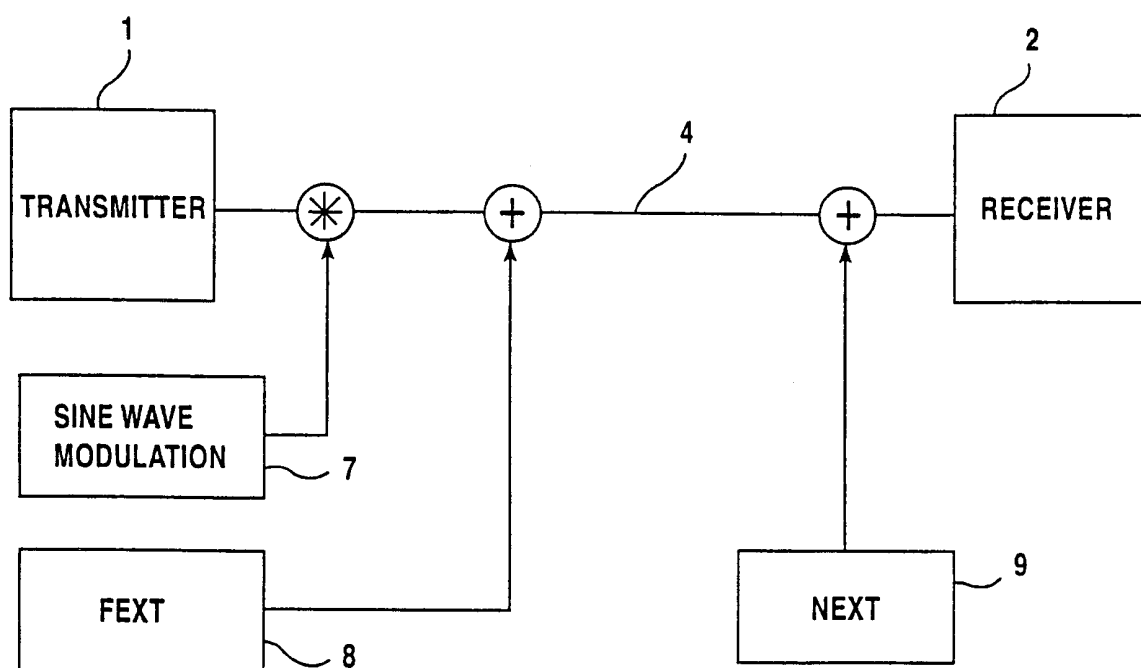


Fig.3

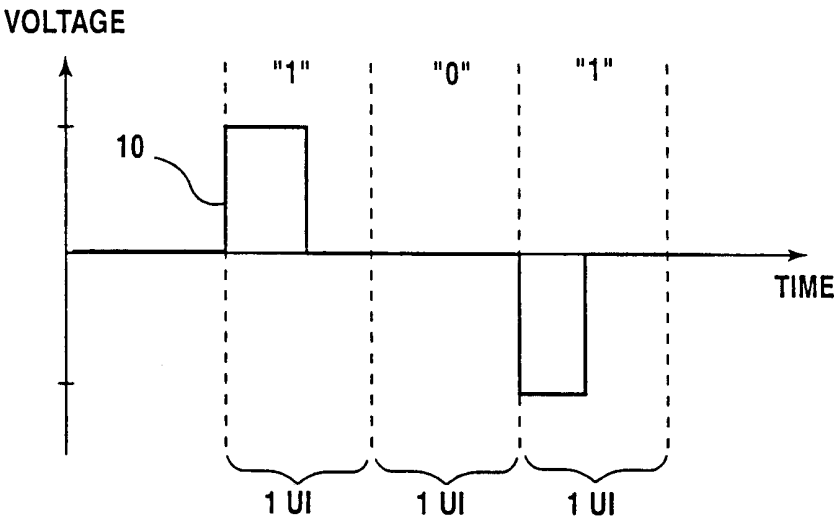


Fig.4

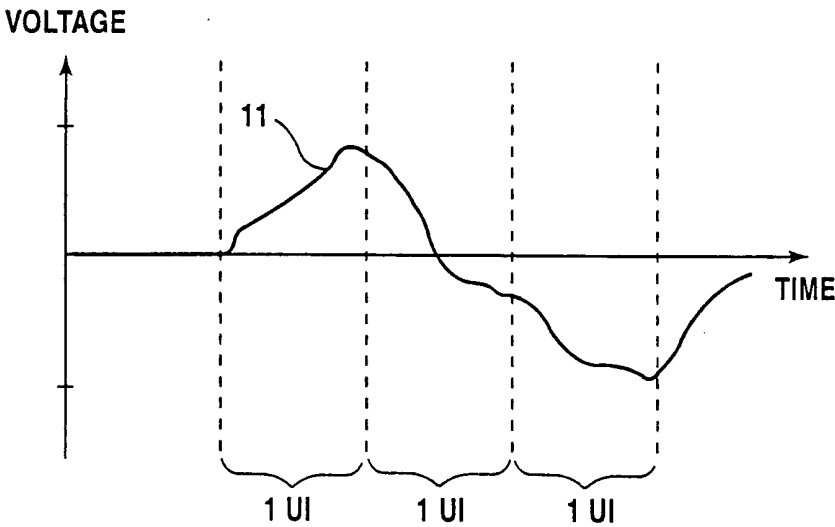


Fig.5

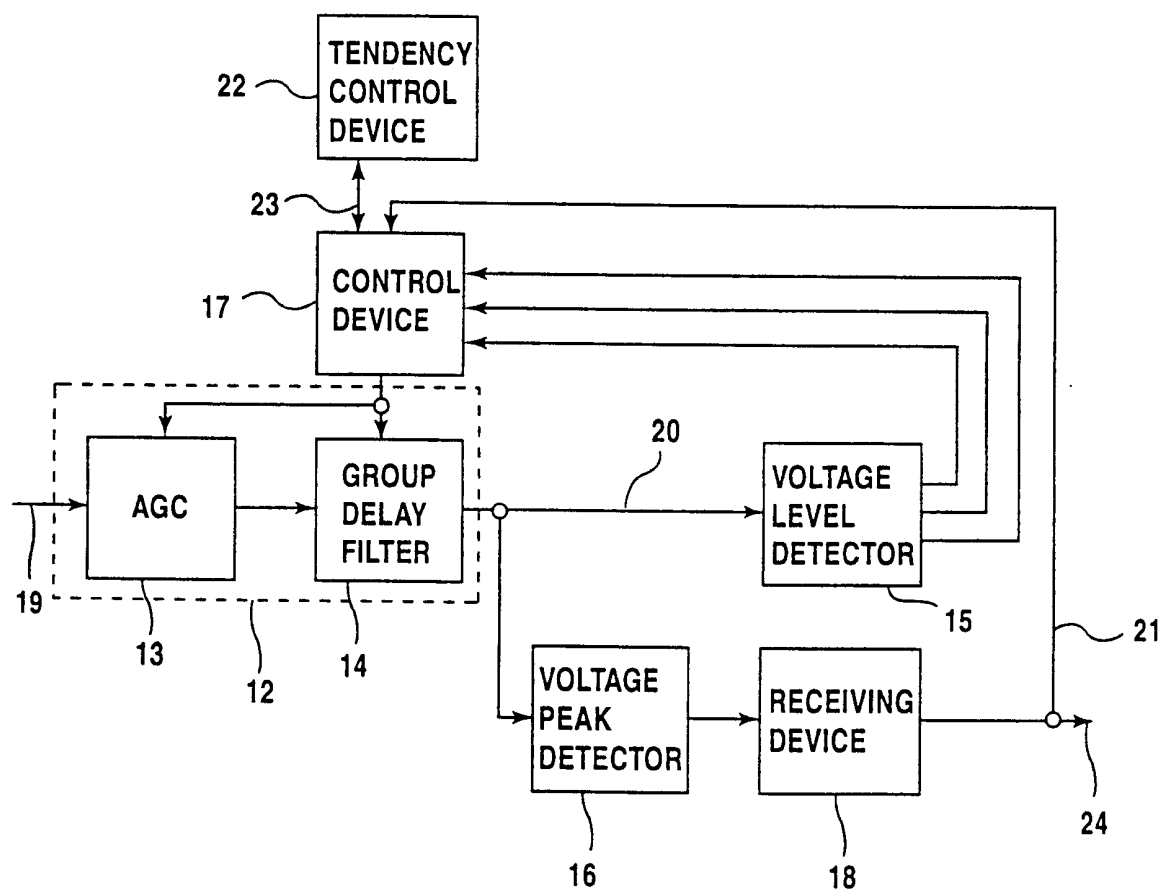


Fig.6

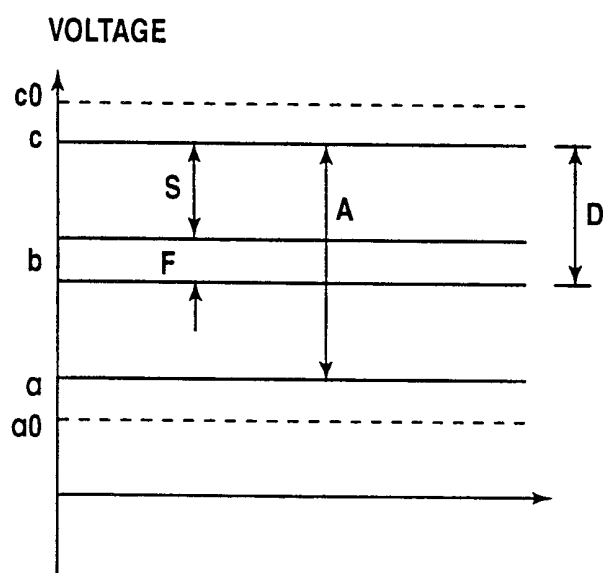


Fig.7

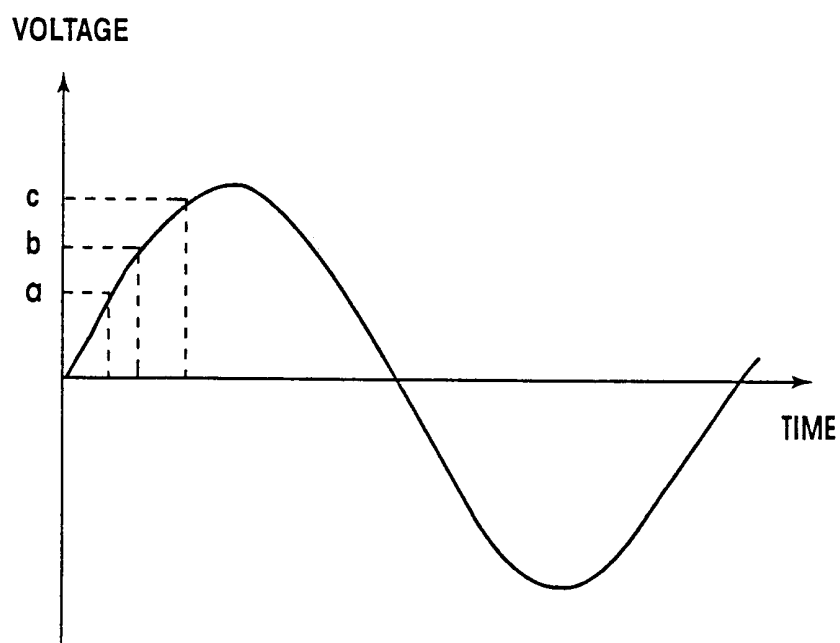


Fig.8

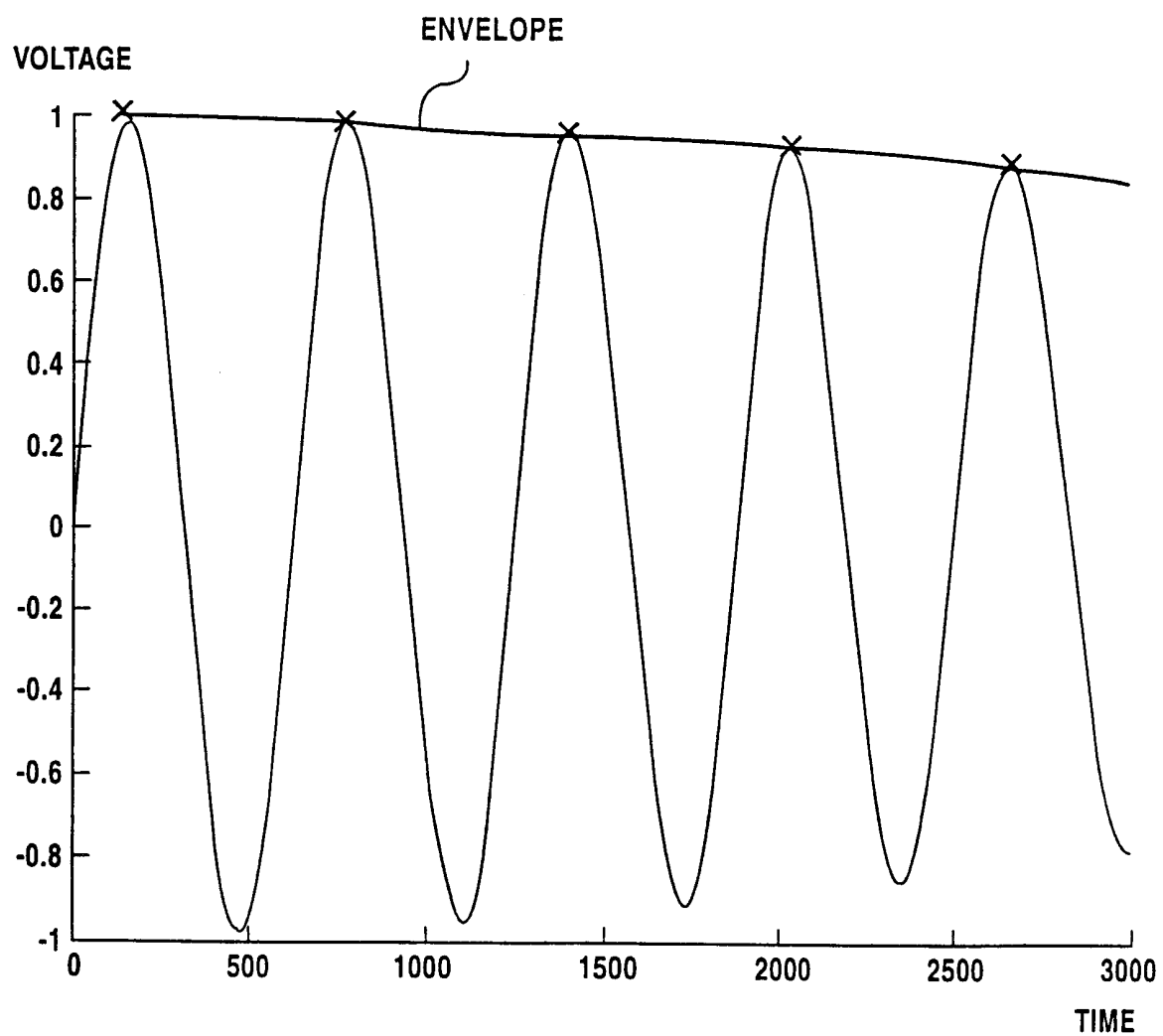


Fig.9

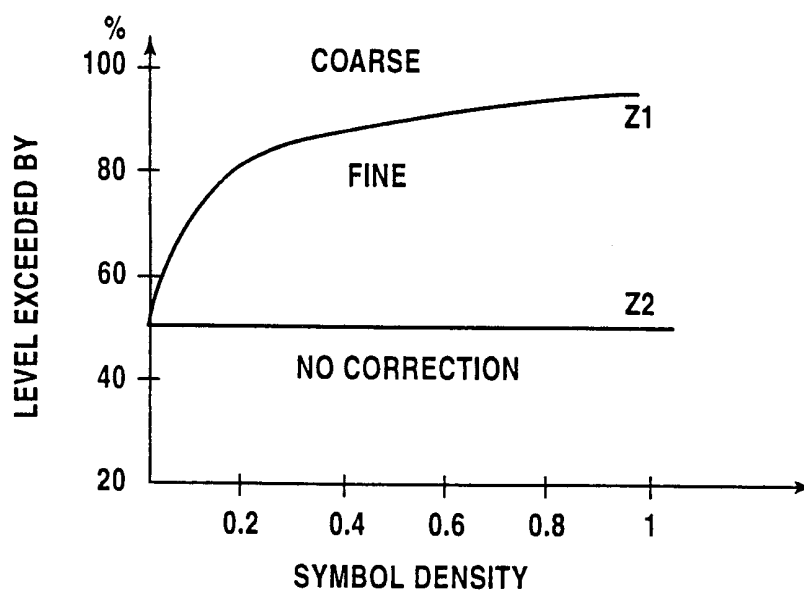


Fig.10

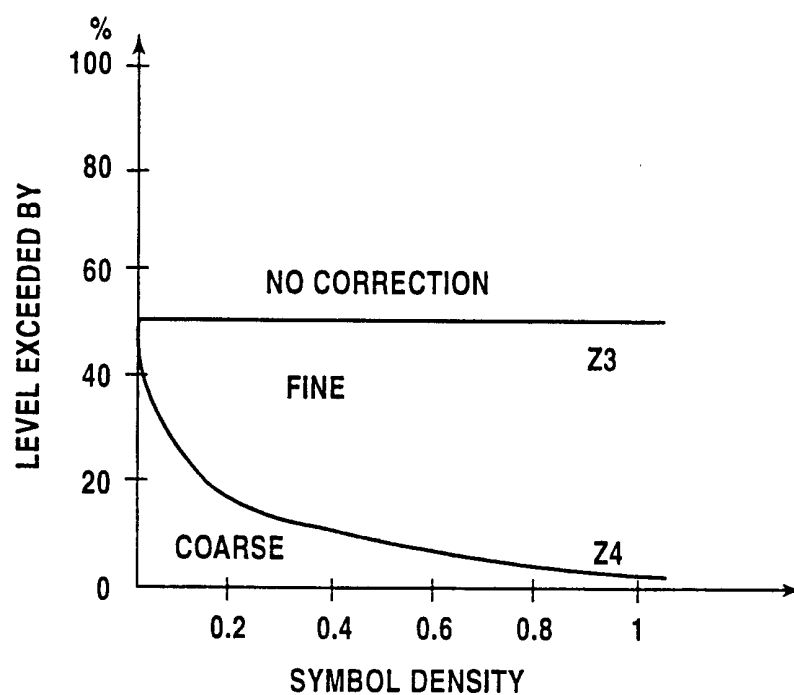


Fig.11

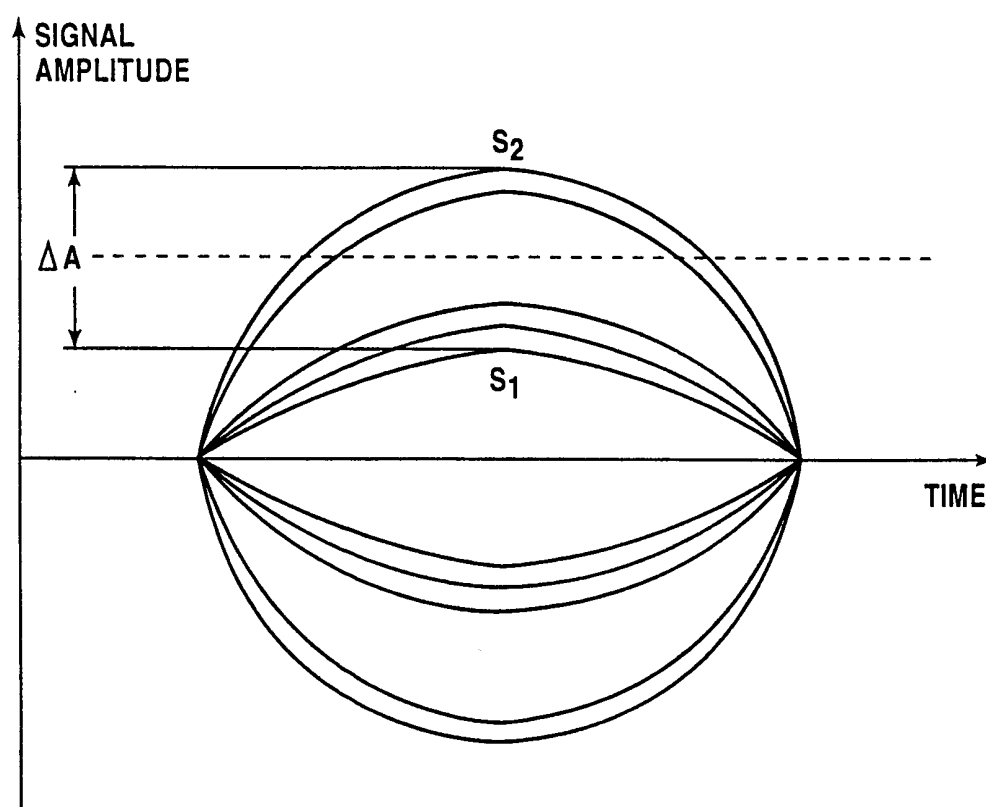


Fig.12

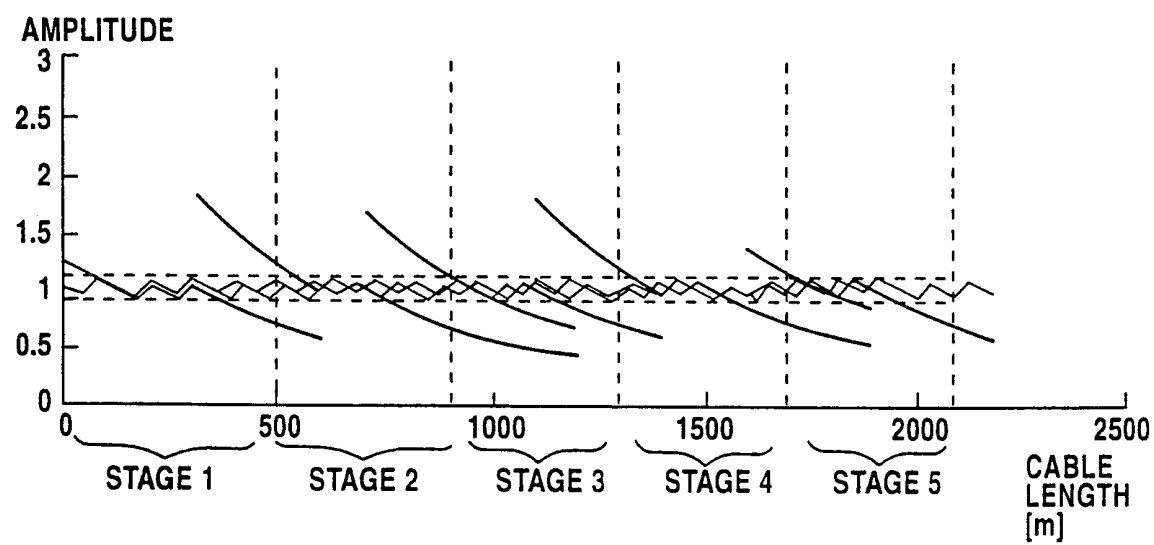


Fig.13

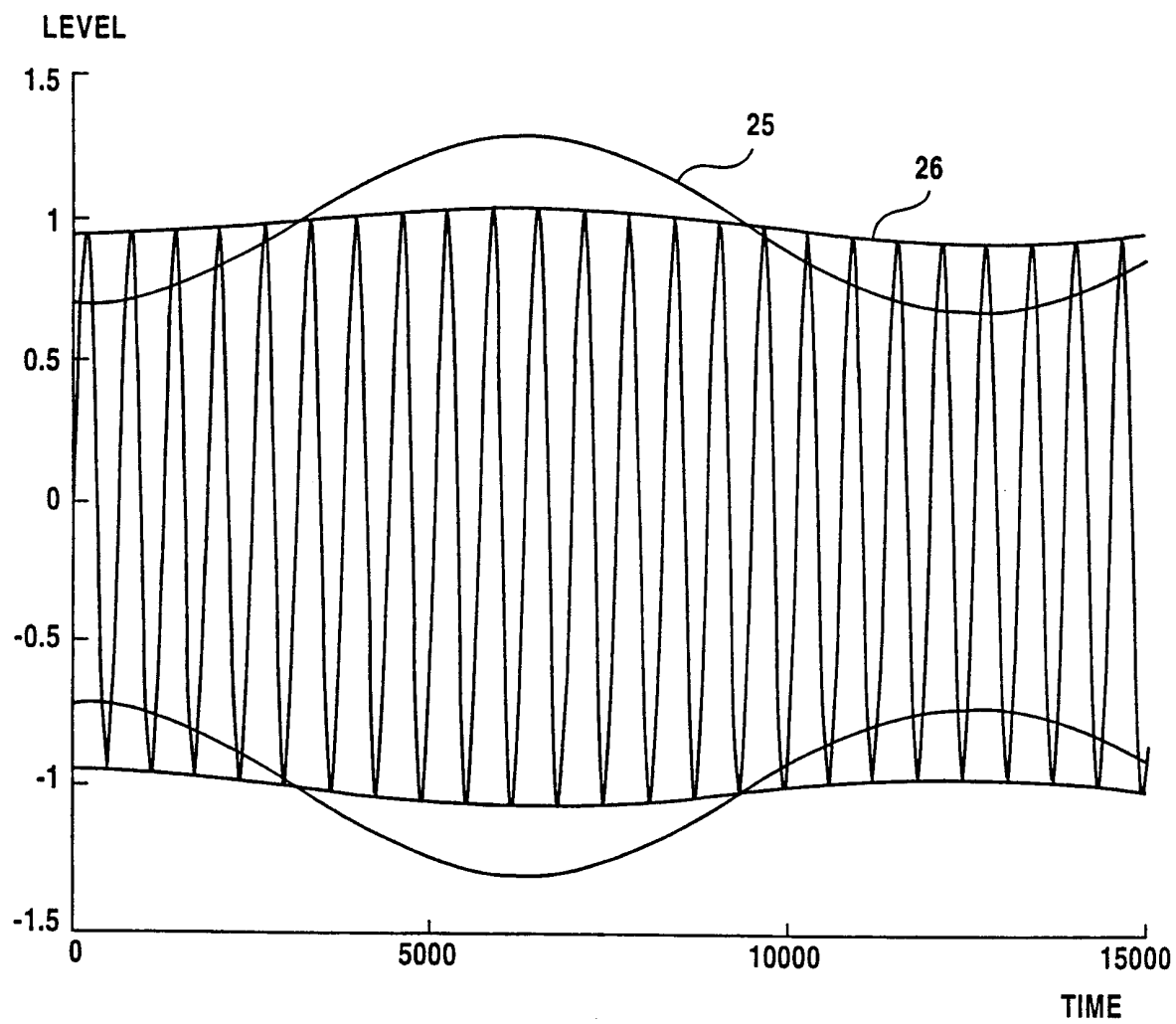
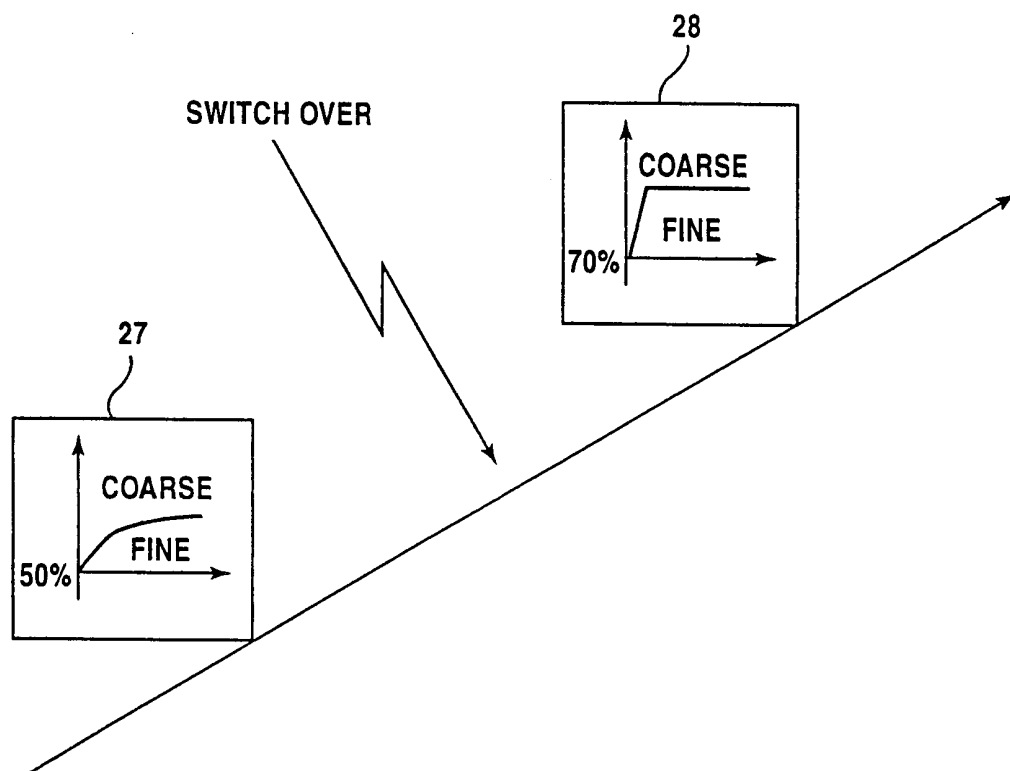


Fig.14



**FIGURE 15 AND FIGURE 16 WILL NOT BE TAKEN IN ACCOUNT
FOR THE PURPOSES OF INTERNATIONAL PROCESSING**

INTERNATIONAL SEARCH REPORT

International Application No
PCT/EP 00/03877

A. CLASSIFICATION OF SUBJECT MATTER IPC 7 H04B3/14 H04L25/02		
According to International Patent Classification (IPC) or to both national classification and IPC		
B. FIELDS SEARCHED Minimum documentation searched (classification system followed by classification symbols) IPC 7 H04B H04L		
Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched		
Electronic data base consulted during the international search (name of data base and, where practical, search terms used) EPO-Internal, INSPEC, WPI Data, PAJ		
C. DOCUMENTS CONSIDERED TO BE RELEVANT		
Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	DE 197 35 752 A (SIEMENS AG) 25 February 1999 (1999-02-25) cited in the application	1-3, 7-9
A	column 2, line 66 -column 5, line 39; claims 1-6; figures 1,2,4,5	4,11
A	EP 0 656 694 A (AT & T CORP) 7 June 1995 (1995-06-07) column 4, line 34 -column 5, line 26; figure 2 column 6, line 7 - line 36; claim 3; figure 3	4,11
<input type="checkbox"/> Further documents are listed in the continuation of box C. <input checked="" type="checkbox"/> Patent family members are listed in annex.		
* Special categories of cited documents : "A" document defining the general state of the art which is not considered to be of particular relevance "E" earlier document but published on or after the international filing date "L" document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified) "O" document referring to an oral disclosure, use, exhibition or other means "P" document published prior to the international filing date but later than the priority date claimed "T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention "X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone "Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art. "&" document member of the same patent family		
Date of the actual completion of the international search 24 August 2000		Date of mailing of the international search report 31/08/2000
Name and mailing address of the ISA European Patent Office, P.B. 5818 Patentlaan 2 NL - 2280 HV Rijswijk Tel. (+31-70) 340-2040, Tx. 31 651 epo nl, Fax: (+31-70) 340-3016		Authorized officer Nilsson, M

INTERNATIONAL SEARCH REPORT

information on patent family members

International Application No
PCT/EP 00/03877

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
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EP 0656694 A	07-06-1995	JP 2951555 B JP 7202949 A KR 191738 B US 5708703 A	20-09-1999 04-08-1995 15-06-1999 13-01-1998