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DESCRIPTION

[0001] The present invention relates to mobile communications systems, and in particular to digital mobile radio communications systems, such as the TETRA (TErrestrial TRunked RAdio) system.

[0002] Transmissions in mobile radio communications systems typically comprise a plurality of symbols. For example, transmissions in a multi-carrier system may comprise a plurality of symbols arranged on a time-frequency grid. The value of each symbol can be represented by a complex number. Various types of symbols are normally present, including for example, pilot symbols, data symbols and synchronisation symbols.

[0003] Conventionally, pilot symbols are included in transmissions to allow received signals to be corrected for the effects of distortion. Distortions may arise as transmissions pass through the propagation channel between a transmitter and a receiver. For example, in mobile communication systems the well known fading multipath distortion can arise, which may be represented as a time and frequency varying complex gain applied to each received symbol in the signal.

[0004] In these systems, pilot symbols with known amplitude and phase are typically inserted into the signal at the transmitter, and the receiver uses knowledge of the transmitted values to estimate the propagation channel distortion for each of the corresponding received symbols. Once the distortion is known at the receiver, a correction factor can be applied to remove it, and the whole signal can be demodulated successfully.

[0005] The use of pilot symbols for channel equalisation is well known and has been part of the TETRA2/TEDS standard since 2003. The properties of such symbols for equalisation purposes are also well known (see, for example, US 6424678).

[0006] Equally, synchronisation symbols are typically included in transmissions to facilitate frame and symbol timing recovery at the receiver, i.e. to allow a receiver to determine the correct timing of bursts, as is known in the art.

[0007] Another feature of modern mobile communications systems, such as the TETRA system, is that they typically support a number of functionalities beyond and/or in addition to the basic transmission of (e.g.) voice data. These can include, for example, a stealing mechanism, Direct Mode Operation (DMO), and the like (as is known in the art).

[0008] Stealing is used to allow a communication system protocol to temporarily replace a block of one type of information (e.g. user data or voice data) with another type of information (e.g. protocol control information). It is typically done to allow "traffic" data blocks, (a traffic channel) to be used for signalling data (as a "control" channel). This is done dynamically and randomly at the transmitter, and the change must be indicated to a receiver so that it may

demodulate the information correctly. Further details of stealing can be found, for example, in EN 300 392-2 sections 9.2.3.6 and 19.4.3.2.

[0009] A common method for indicating stealing (which method is used in conventional TETRA systems) is to have the transmitter flag stealing to a receiver using training sequences. Training sequences are short sequences of symbols whose primary purpose is to enable a receiver to determine burst timings. In the known method, a normal training sequence can be replaced by a different sequence to indicate stealing. The use of training sequences for indicating stealing is well known and has been part of the TETRA standard since the early 1990s (see, for example, EN 300 392-2, WO 2005/048551 and US 7133421).

[0010] It is becoming increasingly desirable in mobile communications systems to make more efficient use of the spectrum available for operation, since radio spectrum is a scarce resource. The Applicants believe therefore that there remains a need to be able to provide reduced bandwidth/higher spectral efficiency digital mobile radio communications systems.

[0011] The invention is defined and limited by the scope of appended claims. In the following description, any embodiment(s) referred to and not falling within the scope of the appended claims, is (are) merely example(s) useful to the understanding of the invention.

[0012] According to a first aspect of the present invention, there is provided a method of operating a digital mobile radio communications system as claimed in claim 1.

[0013] According to a second aspect of the present invention, there is provided a method of operating a digital mobile radio communications system as claimed in claim 3.

[0014] According to a third aspect of the present invention, there is provided a communications station for a digital mobile radio communications system as claimed in claim 4.

[0015] According to a fourth aspect of the present invention, there is provided a communications station for a digital mobile radio communications system as claimed in claim 7.

[0016] According to a fifth aspect of the present invention, there is provided a digital mobile radio communications system as claimed in claim 5.

[0017] According to a sixth aspect of the present invention, there is provided a digital mobile radio communications system as claimed in claim 8.

[0018] The present invention relates to a digital mobile communications system in which pilot symbols are transmitted for channel equalisation purposes and synchronisation symbols are transmitted to indicate burst timings. However, as well as the pilot symbols and synchronisation symbols being used for their conventional purposes, a particular sequence or sequences of pilot symbols or pilot and synchronisation symbols is also used to indicate channel stealing.

[0019] This then has the advantage that additional bits or symbols (and thereby channel bandwidth) over and above the (synchronisation, pilot and data) symbols already included in the transmission (burst), are not required in order to indicate the presence of stealing in the transmission. This thereby provides an efficient method of indicating stealing, that can accordingly reduce the radio bandwidth required for transmissions, thereby using available radio spectrum more efficiently (providing increased capacity).

[0020] For example, with respect to the TETRA system or TETRA-like systems (such as, for example, the Applicant's "Narrow Band Private Mobile Radio" (NB-PMR) system described in their earlier patent application GB 1302414.6), the present invention can avoid the need to use and/or provide training sequences for the purposes of indicating stealing. Indeed, in a preferred embodiment, the present invention does not use training sequences at all. In this way, the total bandwidth required for transmissions can be reduced and/or more of the available radio bandwidth can be used for the payload (data symbols) and/or other symbols (information).

[0021] Thus, the present invention can provide a higher capacity (more radio spectrum efficient) communications system, that can, e.g., operate in narrower frequency bands than existing systems.

[0022] In a preferred embodiment, stealing is indicated by the use of particular, selected pilot symbol or pilot and synchronisation symbol sequence or sequences in a transmitted burst, that have preferably (previously) been defined as indicating the presence of stealing. Thus, in a preferred embodiment of the present invention, the sequence of pilot symbols or pilot and synchronisation symbols that is used to indicate the presence of stealing is a particular, preferably selected, and preferably predefined, sequence of pilot symbols or pilot and synchronisation symbols.

[0023] The sequence of pilot symbols or pilot and synchronisation symbols may comprise any appropriate, preferably predefined, sequence of pilot symbols or pilot and synchronisation symbols. The sequence may comprise a sequence of pilot symbols, or a combined sequence of pilot symbols and synchronisation symbols.

[0024] The sequence(s) of pilot symbols or pilot and synchronisation symbols used to indicate the presence of stealing (the stealing sequence(s)) are preferably distinct from and preferably (relatively easily) distinguishable from the "normal" (non-stealing) sequence(s) of pilot symbols or pilot and synchronisation symbols that are preferably included in transmissions when stealing is not required. This facilitates the receiver identifying the presence of stealing.

[0025] In a preferred embodiment, when stealing is required, the appropriate (stealing) pilot symbol or pilot and synchronisation symbol sequence is accordingly preferably inserted into a transmitted burst, preferably in place of the "normal" pilot symbol or pilot and synchronisation symbol sequence.

[0026] Similarly, the receiver is preferably arranged to determine (and to be capable of determining) whether a "normal" or "stealing" pilot symbol or pilot and synchronisation symbol sequence is present in a received burst, and to indicate that (to pass that) to the upper protocol layers as appropriate. The receiver is then able to decode the received burst appropriately. (As will be understood, the step of decoding the received transmission (burst) preferably comprises the steps of equalisation (i.e. estimating and removing channel distortion), demodulation (i.e. removing the modulation or converting the received symbols into data), and error-correction decoding.)

Preferably, the pilot symbols or pilot and synchronisation symbols of both the normal and stealing pilot symbol or pilot and synchronisation symbol sequences are placed at positions in each burst that have been previously defined. Similarly, the receiver is preferably arranged to have knowledge of this positioning. This facilitates the receiver determining which of the normal or stealing pilot symbol or pilot and synchronisation symbol sequences has been transmitted/received.

[0027] Most preferably, the normal and stealing indicating pilot symbol or pilot and synchronisation symbol sequences can be, and preferably are, used for channel equalisation and/or synchronisation in the normal manner, but are distinct from and can be (relatively easily) distinguished from one another. Thus, in a preferred embodiment, the system of the present invention uses pilot symbols or pilot and synchronisation symbol sequences that have multiple purposes, equalisation and/or synchronisation, and as a "stealing" (or non-stealing) indicator.

[0028] In a preferred embodiment, the stealing pilot symbol or pilot and synchronisation symbol sequence(s) are arranged to be orthogonal to the normal, non-stealing indicating, pilot symbol or pilot and synchronisation symbol sequence(s).

[0029] In these embodiments, any appropriate orthogonal sets of pilot symbols or pilot and synchronisation symbols/sequences may be used. However, in a particularly preferred embodiment, the stealing pilot symbol or pilot and synchronisation symbol sequence(s) are arranged to be distinguishable from the normal pilot symbol or pilot and synchronisation symbol sequence(s) prior to equalisation.

[0030] In this regard, as well as the normal impairments to transmissions such as noise and fading multipath distortion, frequency and timing offsets can be introduced to transmissions between the transmitter and receiver that can give rise to additional types of distortion to the pilot or pilot and synchronisation symbols. The Applicants have recognised that these distortions can mean that pilot symbols or pilot and synchronisation symbols (sequence(s)) that are orthogonal in a simple sense under nominal conditions may not be orthogonal when these distortions are taken into account.

[0031] For example, the complex channel attenuation can effectively convert a transmitted pilot symbol to any value, and indeed even to the same value as a stealing pilot symbol. Thus, in embodiments in which the receiver is arranged to attempt to detect which of the two types of

symbol has been sent (i.e. which of a normal symbol or a stealing symbol has been sent) by comparing the received symbol with a list of known normal and stealing symbol values, it may determine incorrectly which symbol was sent.

[0032] Similar effects can be caused by frequency and timing offsets, which can introduce arbitrary phase changes to the received symbols (and thereby effectively convert a normal pilot or synchronisation symbol into a stealing symbol or *vice versa*).

[0033] The Applicants have furthermore recognised that the nature of the distortions (which include, for example, exponential terms, as will be discussed more fully below) means that a straightforward correlation may not normally exist between the received symbols and the known pilot or pilot and synchronisation symbols.

[0034] These problems are addressed in the preferred embodiment of the present invention by using stealing and normal (non-stealing) symbol sequences that can be distinguished from one another by the receiver prior to equalisation. In this way, the receiver is able to determine which symbol sequence has been transmitted before using the pilot symbols to carry out equalisation and demodulation, so that the receiver is able to use the correct symbol sequence for equalisation.

[0035] The receiver may be configured to determine which symbol sequence has been transmitted (before using the pilot or pilot and synchronisation symbols to carry out equalisation and demodulation) in any appropriate manner.

[0036] In a preferred embodiment this is achieved by using the phases of the symbols to distinguish between the stealing and normal (non-stealing) symbol sequences.

[0037] Thus, in a preferred embodiment, it is determined whether a stealing or normal (non-stealing) symbol sequence has been received (transmitted) by analysing the phases of the received symbols, with the symbol phases for the stealing and normal (non-stealing) stealing sequences being selected so as to facilitate such symbol phase analysis. Preferably the phases of the symbols in the received symbol sequence are analysed to generate a decision variable that can then be used to determine which of the two symbol sequences (normal or stealing) was transmitted.

[0038] In a particularly preferred embodiment, the receiver is configured to process the received signal in order to determine which (stealing or normal) symbol sequence has been transmitted/received, and then to carry out equalisation.

[0039] In a particularly preferred embodiment, when a symbol sequence for which it is to be determined whether it is a normal or stealing symbol sequence is received, the received symbol sequence is first "referenced" to a predetermined (nominal) reference normal symbol sequence, and, separately, to a predetermined (nominal) reference stealing symbol sequence, by removing the symbol phases for the respective reference symbol sequence from the

received symbol sequence.

[0040] The phases of the reference symbols may be removed from the received pilot or pilot and synchronisation symbols in any suitable manner, but in a preferred embodiment this is done by multiplying each received symbol by the conjugate of its reference symbol value.

[0041] In this arrangement, the phases of the symbols of the received symbols once the reference symbol phases have been removed are preferably then analysed to determine whether a stealing or normal (non-stealing) symbol sequence has been transmitted.

[0042] This is preferably done by first combining pairs of symbols (with each symbol having had its reference symbol phase removed) in the received symbol sequences, preferably to derive pair-wise first order differences of the so-referenced received symbols. This helps to reduce and/or eliminate timing and/or frequency offsets.

[0043] For example, in embodiments that use multi-carrier transmission schemes, pairs of symbols (with their reference symbol phases removed), taken from the same multi-carrier symbol may be combined. In alternative embodiments that use a TEDS-like multi-carrier scheme having pilot symbols on several different subcarriers, pairs of symbols (more) closely spaced in frequency (preferably as close as possible) are used. These arrangements have the advantage of minimising the channel attenuation phase difference between the symbols of the pair.

[0044] In other embodiments that use single carrier schemes, neighbouring pairs of symbols along the time axis may be combined. In these embodiments, the resulting quantity will have a phase related to a frequency offset rather than a timing offset, which may be reduced by combining symbol pairs in a corresponding manner.

[0045] In a preferred embodiment, pairs of symbols are combined by multiplying a first symbol (with its reference symbol phase removed) with the conjugate of a second symbol (with its reference symbol phase removed), to derive the pair-wise first order differences of the so-referenced received symbols. The Applicants have recognised that the conjugate can be used in this manner to eliminate the timing and/or frequency offset between the transmitted symbols and the received symbols.

[0046] Preferably these pair-wise first order differences are then summed for some or all (and preferably for all) of the received symbol sequence symbol pairs to provide an integral of the pair-wise first order differences of the received symbols when referenced to the reference stealing symbol sequence, and an integral of the pair-wise first order differences of the received symbols when referenced to the reference normal (non-stealing) symbol sequence, which integral (sum) values can then be used, e.g. as decision variables, to decide whether a stealing or normal (non-stealing) symbol sequence has been received.

[0047] In this regard, the Applicants have recognised that the noise terms in this summation

will tend towards zero provided that any differences in channel attenuation phase between adjacent pilot or pilot and synchronisation symbols are small. The Applicants have further recognised that this will generally be the case for pilot or pilot and synchronisation symbols in the same multi-carrier symbol, and/or where there is a small difference in frequency between the sub-carriers containing pilot or pilot and synchronisation symbols.

[0048] Thus, in a preferred embodiment, a first calculation is carried out in the manner described above using the reference symbols from the reference normal symbol sequence(s) to calculate a "normal symbol sequence" decision variable(s), and a second, corresponding calculation is preferably carried out using the reference symbols from the reference stealing symbol sequence(s) to calculate a "stealing symbol sequence" decision variable(s).

[0049] The normal and stealing symbol sequence decision variables are preferably then compared to determine which symbol sequence (set) was received (transmitted), preferably by comparing the magnitudes of the decision variables.

[0050] Preferably, the normal (non-stealing) and stealing-indicating symbol sequences are configured such that if the symbol sequence that the decision variable corresponds to has been received, then the value of the decision variable (of the integral) will be larger (e.g. maximised), but if the symbol sequence that the decision variable does not correspond to has been received, then the value of the decision variable (of the integral) will be smaller (e.g. minimised), or vice-versa, preferably by an appropriate choice of the pilot or pilot and synchronisation symbol phases (as will be discussed further below).

[0051] Then, in embodiments where the value of the decision variable is arranged to be maximised if the symbol sequence that the decision variable corresponds to has been received, then if the normal symbol sequence decision variable is larger than the stealing symbol sequence decision variable, the receiver may determine that the normal symbol sequence (set) was transmitted. Otherwise, it may determine that the stealing sequence (set) was transmitted (and vice-versa in the case where the value of the decision variable is arranged to be minimised if the symbol sequence that the decision variable corresponds to has been received).

[0052] Thus, according to a preferred embodiment, the step of determining whether a received transmission is indicative of stealing comprises: removing reference symbol phases from the received pilot or pilot and synchronisation symbols for each of the possible transmitted symbol sequences (sets); combining respective symbol pairs to reduce timing and/or frequency offsets; integrating the sets of resulting quantities to form decision variables; and comparing the decision variables to determine which symbol sequence (set) was transmitted.

[0053] As will be understood by those skilled in the art, this arrangement advantageously allows the stealing and normal symbol sequences to be distinguished from one another by the receiver prior to equalisation.

[0054] As discussed above, in a particularly preferred arrangement of these embodiments, the symbol phases for the stealing and normal (non-stealing) symbol sequences are selected so as to facilitate the above arrangement being able to distinguish between the stealing and normal (non-stealing) symbol sequences.

[0055] Thus, the symbol phases are preferably chosen so that one of: the integral of the pair-wise first order differences of the stealing symbol sequence after referencing with the reference normal (non-stealing) symbol sequence, and the integral of the pair-wise first order differences of the stealing symbol sequence after referencing with the reference stealing symbol sequence, will be a minimum, and the other integral will be a maximum (and correspondingly for the normal (non-stealing) symbol sequence).

[0056] Accordingly, in these embodiments if a stealing symbol sequence is received, the integral of the pair-wise first order differences of the received stealing symbol sequence after referencing with the reference stealing symbol sequence should be a maximum, but if the integral is calculated using the normal (non-stealing) reference symbol sequence, the integral should be a minimum (or vice-versa) (and correspondingly for the normal (non-stealing) symbol sequence).

[0057] Thus, where a stealing symbol sequence is received, the value of the integral calculated using the reference stealing symbol sequence should be greater than the value of the integral calculated using the reference normal (non-stealing) symbol sequence (or vice-versa), even in the presence of distortion (and correspondingly for the normal (non-stealing) symbol sequence).

[0058] (As discussed, the symbol phases may alternatively be chosen such that the above maximised values are instead minimised, and those values that are minimised are instead maximised.)

[0059] Thus, by using appropriate symbol phases, the orthogonality of the stealing indicating symbol sequence(s), and the normal, non-stealing indicating symbol sequence(s) can be effectively preserved, even in the presence of distortion.

[0060] Accordingly, the preferred method can advantageously be applied to signals which have not been equalised and where there is no knowledge of timing and frequency offsets between the transmitter and receiver. Accordingly, the effects of the propagation channel (distortion) on the signal, along with the effects of other system imperfections (such as frequency offsets), which would otherwise make it very difficult to determine which (normal or pilot) symbol sequence has been transmitted/received, can be addressed.

[0061] The communications station of the present invention may be a mobile station or a base station, as is known in the art. In one preferred embodiment it is a base station. In another preferred embodiment it is a mobile station.

[0062] Preferably, transmissions in the system use a multi-sub-carrier arrangement. Each channel may have any desired bandwidth, such as 12.5 kHz, 25 kHz, 50 kHz, 100 kHz or 150 kHz. The system may use only a single channel, but as will be appreciated by those skilled in the art, plural channels are preferably provided (supported) and used.

[0063] Each channel preferably contains plural sub-carriers. Any number of sub-carriers can be used, as will be appreciated by those skilled in the art. In a particularly preferred embodiment, four sub-carriers are used. In other embodiments, two, six, or eight sub-carriers can be used.

[0064] Each channel that is used in the present invention may be divided into the requisite multiple sub-carriers (sub-channels) in any desired and suitable manner. Most preferably, the sub-carriers (sub-channels) are equally spaced within the channel. For example, in embodiments using 12.5 kHz channels and four sub-carriers, a sub-carrier spacing of 2.7 kHz may be used.

[0065] In alternative embodiments, the system may use a single-carrier arrangement.

[0066] Each carrier or sub-carrier is preferably modulated with a modulation scheme that encodes one or more bits per symbol, and preferably at least 4 bits per symbol. The modulation scheme can be any desired and suitable modulation scheme. Suitable modulation schemes include QAM, PSK/DPSK, or OFDM schemes. Each carrier or sub-carrier should be and is preferably modulated using the same modulation scheme. In an embodiment, 16-QAM is used.

[0067] The present invention preferably uses particular (and identifiable) burst structures for its transmissions (as is conventional in mobile communications systems), so as to, e.g., allow particular forms (types) of transmissions to be readily identified and processed.

[0068] In general, the bursts preferably include pilot symbols spaced to facilitate equalisation at the receiver, synchronisation symbols to facilitate frame and symbol timing recovery at the receiver, and data symbols for carrying the useful payload (and any required header). In multi-carrier embodiments, the symbols are preferably ordered regularly across and along the time and frequency axes of each burst.

[0069] In an embodiment, the pilot symbols are placed along the time and frequency axes of the bursts at approximately regular intervals. This then allows for the correction factors used in channel equalisation to be determined in each dimension. This is particularly useful in a multi-carrier system, in which each symbol can experience a different distortion, so that correction factors are needed for each symbol in the time-frequency grid. In an alternative embodiment, the pilot symbols are placed along the time and frequency axes of the bursts at irregular intervals.

[0070] In an embodiment, the arrangement and spacing of the pilot symbols is not fixed, and in

an embodiment may be altered dynamically. In these embodiments, the transmitter and receiver are arranged to be able to communicate any change in the arrangement of pilot symbols. This then may be used, for example, to provide more pilot symbols to enable better equalisation when desired, for example in rapidly fading conditions.

[0071] In a preferred embodiment, pilot symbols are included (used) in the outer sub-carriers in the channel only (so in the outer two, but not the inner two, sub-carriers where four sub-carriers are being used). This helps to provide more symbols for data purposes.

[0072] In a preferred embodiment, synchronisation symbols are included (used) in all sub-carriers in the channel at the beginning of a burst. This helps to facilitate synchronisation.

[0073] In a preferred embodiment, the methods and apparatus of the present invention are used in a system that uses a four slot time-division multiple access scheme in 12.5 kHz channels, with each 12.5 kHz channel containing plural sub-carriers and each sub-carrier being modulated with a modulation scheme that encodes at least 4 bits per symbol, for transmission in the system.

[0074] Thus, in an embodiment, the method further comprises:
using a four slot time-division multiple access scheme in 12.5 kHz channels, with each 12.5 kHz channel containing plural sub-carriers and each sub-carrier being modulated with a modulation scheme that encodes at least 4 bits per symbol, for transmission in the system.

[0075] In an embodiment, the communications station further comprises:
means for making transmissions using a four slot time-division multiple access scheme in 12.5 kHz channels, with each 12.5 kHz channel containing plural sub-carriers and each sub-carrier being modulated with a modulation scheme that encodes at least 4 bits per symbol.

[0076] In an embodiment, the digital mobile radio communications system further comprises:
means for making transmissions using a four slot time-division multiple access scheme in 12.5 kHz channels, with each 12.5 kHz channel containing plural sub-carriers and each sub-carrier being modulated with a modulation scheme that encodes at least 4 bits per symbol.

[0077] The Applicants have furthermore recognised that the techniques of the present invention may be used to convey (transmit) other types of information between a transmitter and a receiver (e.g. information not necessarily related to stealing). In this regard, the Applicants have recognised that the techniques of the present invention essentially convey one (or more) additional bits of information (i.e. not stealing/stealing) using particular sequences of pilot symbols or pilot and synchronisation symbols. This information may be used for purposes other than stealing, for example to indicate different modulation levels or types (e.g. 4-QAM, 64-QAM, etc.) for the payload in the slot, to indicate different types of coding for the slot contents, to indicate different transmission bandwidths, etc..

[0078] Thus, in embodiments, additional (e.g. third and/or fourth and/or fifth (and so on))

symbol sequences (sets) are preferably defined and used to convey additional information.

[0079] Correspondingly, in an embodiment, the method further comprises:

including in a transmission a sequence of pilot symbols and/or synchronisation symbols to indicate to a receiver additional information in the transmission.

[0080] In an embodiment, the method further comprises:

determining whether a received transmission includes a sequence of pilot symbols and/or synchronisation symbols indicative of additional information in the transmission; and

if it is determined that a sequence of pilot symbols and/or synchronisation symbols indicative of additional information is present in the received transmission, identifying the additional information in the transmission.

[0081] In an embodiment, the communications station comprises:

a transmitter configured to include in a transmission a sequence of pilot symbols and/or synchronisation symbols to indicate to a receiver additional information in the transmission.

[0082] In an embodiment, the communications station comprises:

a receiver configured to determine whether a received transmission includes a sequence of pilot symbols and/or synchronisation symbols indicative of additional information in the transmission, and to, if it is determined that a sequence of pilot symbols and/or synchronisation symbols indicative of additional information is present in the received transmission identify the additional information in the transmission.

[0083] In an embodiment, the system comprises:

a transmitter configured to include in a transmission a sequence of pilot symbols and/or synchronisation symbols to indicate to a receiver additional information in the transmission.

[0084] In an embodiment, the system comprises:

a receiver configured to determine whether a received transmission includes a sequence of pilot symbols and/or synchronisation symbols indicative of additional information in the transmission, and to, if it is determined that a sequence of pilot symbols and/or synchronisation symbols indicative of additional information is present in the received transmissions, identify the additional information in the transmission.

[0085] As will be appreciated by those skilled in the art, these embodiments of the present invention can and preferably do include any one or more or all of the preferred and optional features of the invention described herein, as appropriate.

[0086] The additional information may comprise any suitable and desired additional information, over and above the information encoded by the symbols in the burst. The

additional information preferably relates to (and is about) the transmission. In various embodiments, the additional information may indicate, for example, stealing, modulation levels or types (e.g. 4-QAM, 64-QAM, etc.), types of coding, or a transmission bandwidth. Other uses for the additional information will occur to the skilled person.

[0087] The methods in accordance with the present invention may be implemented at least partially using software e.g. computer programs. It will thus be seen that when viewed from further aspects the present invention provides computer software specifically adapted to carry out the methods hereinabove described when installed on data processing means, and a computer program element comprising computer software code portions for performing the methods hereinabove described when the program element is run on data processing means. The invention also extends to a computer software carrier comprising such software which when used to operate a communications system and a communications station comprising data processing means causes in conjunction with said data processing means said system or station to carry out the steps of the method of the present invention. Such a computer software carrier could be a physical storage medium such as a ROM chip, CD ROM or disk, or could be a signal such as an electronic signal over wires, an optical signal or a radio signal such as to a satellite or the like.

[0088] It will further be appreciated that not all steps of the method of the invention need be carried out by computer software and thus from a further broad aspect the present invention provides computer software and such software installed on a computer software carrier for carrying out at least one of the steps of the methods set out hereinabove.

[0089] A number of preferred embodiments of the present invention will now be described by way of example only and with reference to the accompanying drawings, in which:

Figure 1 shows a schematic diagram of an exemplary hybrid mobile communications system in accordance with an embodiment of the present invention;

Figure 2 shows an exemplary spectrum of a single 12.5 kHz channel for a preferred embodiment of a narrow band mobile communications system in accordance with the present invention;

Figure 3 shows the symbol constellation diagram for a modulation scheme in accordance with a preferred embodiment of the present invention;

Figure 4 illustrates the structure of a frame for use in a preferred embodiment of the present invention;

Figure 5 shows an embodiment of the normal uplink burst structure to be used in preferred embodiments of the present invention;

Figure 6 shows a flow diagram illustrating a stealing process according to an embodiment of the present invention; and

Figure 7 shows the results of simulations of the (normalised) decision variables using a model of the NB-PMR communication system.

[0090] A preferred embodiment of the present invention will now be described with particular reference to a narrow band TETRA-compatible mobile communications system in which 12.5 kHz channels are used for transmission. However, as will be appreciated, the present invention is not exclusively limited to such an arrangement and will be applicable to other radio Standards and systems, including for example other multi-carrier systems such as the TEDS system.

[0091] For convenience, the narrow band communication scheme of the present embodiment will be referred to as NB-PMR.

[0092] Figure 1 shows a schematic diagram of an exemplary narrow band Private Mobile Radio (NB-PMR) TETRA-compatible mobile communications system in which 12.5 kHz channels are used for transmission in accordance with this preferred embodiment of the present invention. In Figure 1, the arrowed lines indicate 12.5 kHz channels. Also shown are vehicle and hand-held communications terminals (which are interchangeable). The system includes one or more base stations and infrastructure elements which form a 12.5 kHz trunked mode network 10, as is conventional in mobile communications systems.

[0093] Figure 1 also shows Direct Mode Operation, with 12.5 kHz direct mode networks 20, 30 which allow communications terminals to operate in direct mode with each other, and to communicate with the trunked mode network 10 via a trunked mode gateway 22 and/or a dual watch mobile station 32. One or more direct mode repeaters 34 may be provided in the system to extend the range of communication between communications terminals operating in direct mode.

[0094] Figure 2 shows an exemplary spectrum of a single 12.5 kHz channel that is used in this embodiment of the present invention. While Figure 2 illustrates a spectrum for one such channel, it will be appreciated that in practice preferably plural 12.5 kHz channels are provided (supported) and used.

[0095] According to this embodiment, and as shown in Figure 2, each 12.5 kHz channel carries (is divided into) four equally spaced sub-carriers, having a spacing of 2.7 kHz. The modulation scheme uses 2400 symbols per second, and an RRC modulation filter with $\alpha=0.2$. Accordingly, the nominal overall modulation bandwidth provided within the 12.5 kHz channel is 10.8 kHz.

[0096] A 16-QAM modulation scheme is used to modulate the symbols on each respective sub-carrier in the 12.5 kHz channel. Figure 3 shows the symbol constellation diagram for the 16-QAM modulation scheme which is used in this embodiment of the present invention. Circles

represent pilot symbols and stars represent data symbols. This modulation scheme encodes 4-bits per-symbol. Other modulation schemes could be used, if desired.

[0097] Figure 4 illustrates the timeslot and frame structure that is used in this preferred embodiment of the present invention. A four-slot TDMA arrangement, having the timing structure of TETRA 1 is used. Thus, each frame has four timeslots (shown as Slots 1 to 4), each time-slot is 14.17 ms long, and each frame is 56.7 ms long. Also, as shown in Figure 4, the arrangement is such that each timeslot contains 34 (NB-PMR) symbols per-sub-carrier. This facilitates compatibility with TETRA 1 systems whose time slot and frame structure is also shown in Figure 4 for comparison purposes.

[0098] In this arrangement, using the 16-QAM modulation scheme in a timeslot having the TETRA 1 timing structure and containing 34 symbols per sub-carrier, provides $4 \times 34 = 136$ bits per sub-carrier per timeslot, and so can be used to provide 544 bits (4×136) per timeslot where 4 sub-carriers (each carrying 34 16-QAM symbols) are being used. This bit capacity can be used to provide a system that is compatible with TETRA 1.

[0099] As well as the above modulation scheme and sub-carrier arrangement, etc., the present embodiment also uses particular burst structures for its transmissions when operating in the 12.5 kHz channels. Figure 5 shows an example of a burst structure that is used in the present embodiment, namely the normal uplink burst structure.

[0100] In Figure 5, the various symbols depicted comprise data symbols (labelled D) (from the 16-QAM symbol constellation), pilot symbols (labelled P) (from a pilot symbol constellation), and synchronisation symbols (labelled S) (from a synchronisation symbol constellation).

[0101] This burst structure is designed to allow the four sub-carrier modulation scheme having the symbol duration and slot duration described above (i.e. providing a total of 136 (4×34) symbols per timeslot) (and thus 544 bits per timeslot) (with four bits per symbol) of the present embodiment to provide sufficient capacity to support TETRA 1 compatible payloads, whilst still allocating a number of symbols in each slot as synchronisation symbols and as pilot symbols in order to support synchronisation and channel equalisation, etc..

[0102] Thus, the burst structures of this embodiment can be used to achieve payload capacities that are sufficient for compatibility with TETRA 1. Other burst structures would be possible, if desired.

[0103] As discussed above, pilot symbols are included in burst structures to allow received signals to be corrected for the effects of distortion. The pilot symbol values can be represented as a complex number, and can in principle take any value. In the present embodiment, the values of the pilot symbols are similar to values of the data symbols. Figure 5 shows an example of pilot symbol and data symbol values using a 16-QAM modulation scheme.

[0104] In the present embodiment, pilot symbols are placed on the outer two of the four

subcarriers and at time intervals of a few symbols. For example, in the normal uplink burst depicted in Figure 3 there are ten pilot symbols, with five on each subcarrier. In this embodiment, the pilot-carrying subcarrier frequency offset with respect to the centre of the channel is $\pm f_c$.

[0105] As discussed above, in the present embodiment, when stealing is required, the transmitter includes in a transmission a sequence of pilot symbols indicative of stealing, in place of the "normal" sequence of pilot symbols. This process is illustrated by Figure 6A. In step 40, it is determined whether stealing is required. If stealing is not required, the normal pilot symbol sequence is inserted into the burst in step 42, and the burst is then transmitted in step 46. If, on the other hand, stealing is required in step 40, the stealing pilot symbol sequence is inserted into the burst in step 44, and the burst is transmitted in step 46, in a corresponding manner. In embodiments where synchronisation symbols are additionally used to indicate stealing, stealing or normal synchronisation symbols are inserted into the burst in a corresponding manner.

[0106] The receiver is then able to determine whether the received transmission includes a sequence of symbols indicative of the presence of stealing (or not) (or a "normal" sequence of symbols), and if this is the case, to identify the presence of stealing in the transmission and to decode the transmission accordingly. This is illustrated by Figure 6B. In the present embodiment, decoding the received transmission (burst) comprises steps of equalisation (i.e. estimating and removing channel distortion), demodulation (i.e. removing the modulation or converting the received symbols into data), and error-correction decoding. Accordingly, in step 50, it is determined whether the received burst includes the normal symbol sequence or the stealing symbol sequence. If it is determined in step 52 that the burst includes the stealing symbol sequence, then the burst is equalised using the stealing pilot symbols in step 54, and is decoded (demodulated and error corrected) in step 56. If, on the other hand, the received burst does not include the stealing symbol sequence (i.e. instead includes the normal symbol sequence), then the burst is equalised using the normal pilot symbols in step 58, and is then decoded (demodulated and error corrected) in step 60. (Again, where synchronisation symbols are used to indicate stealing, the reception process will operate in a corresponding manner.)

[0107] As discussed above, transmissions between a transmitter and a receiver can be distorted due to effects of the transmissions passing through the propagation channel between the transmitter and the receiver (this distortion can be represented as a time and frequency varying complex gain applied to each received symbol in the signal). These problems are addressed in the present embodiment by using stealing and normal (non-stealing) symbol sequences that can be distinguished from one another by the receiver prior to equalisation. In the present embodiment, this is achieved by an appropriate choice of the stealing pilot symbol or pilot and synchronisation symbol phases.

[0108] The receiver is configured to determine which symbol sequence has been transmitted (before using the pilot or pilot and synchronisation symbols to carry out equalisation and demodulation) by processing the received signal in order to determine which (stealing or

normal) symbol sequence has been transmitted/received, and then to carry out equalisation. The manner of the processing used in the present embodiment to determine which (stealing or normal) symbol sequence has been transmitted/received will now be discussed.

[0109] In the present embodiment, a sequence (set) of k pilots $\mathbf{r} = (r_1 r_2 r_3 \dots r_k)$ is transmitted by the transmitter and subsequently received at the receiver as $\mathbf{p} = (p_1 p_2 p_3 \dots p_k)$, after passing through a propagation channel. A received pilot symbol p_i arriving at time t in the receiver will be of the form:

$$p_i = \alpha_i r_i \cdot e^{2\pi j f_c dt} \cdot e^{2\pi j f_0 t} + n_i \quad (1)$$

in the non-stealing case. In equation (1), $\alpha_i = (A_i, \psi_i)$ is the channel attenuation at symbol i ; $r_i = (R_i, \theta_i)$ is the i th transmitted pilot symbol of the pilot symbols in a burst; $e^{2\pi j f_c dt}$ is a phase rotation due to the timing offset dt between the receiver and the transmitter (in a perfectly synchronised receiver this would be zero, but in practical systems this is non-zero); $e^{2\pi j f_0 t}$ is a phase rotation due to a frequency offset f_0 between the receiver and the transmitter (in a perfectly frequency-locked receiver, this would be zero, but in practical systems it is non-zero); and $n_i = (N_i, \Xi_i)$, is a noise term at symbol i .

[0110] As discussed above, in conventional communications systems, pilot symbols are used to remove the channel distortion, α_i , using the known values of the pilot symbols. However, in the present embodiment, this is not done until it has been determined which of the two (normal or stealing) symbol sequences (sets) has been transmitted. In other words, a stealing detection step is carried out by the receiver prior to channel equalisation, so that the receiver can use the correct symbol sequence for equalisation. Accordingly, α_i is unknown when the transmission is received by the receiver.

[0111] Similarly, frequency and timing errors may be present in the received transmission, and a real-life receiver will not normally be able to fully correct them, so that the exponential terms in equation (1) will also be unknown when the transmission is received by the receiver. In the present embodiment, the presence of the unknown factor α_i and the unknown exponential terms prohibits a straightforward application of correlation between the received symbol sequence (set) and the known normal and stealing pilot symbol sequences (sets) for detecting stealing.

[0112] In the present embodiment a decision variable is defined which is used to minimise or eliminate the unknown terms in equation (1). The magnitude of the decision variable is used by the receiver to determine which of the two symbol sequences (sets) (normal or stealing) was transmitted.

[0113] In this embodiment, the reference symbol phases (reference to a "reference symbol" herein refers to the nominal value of a normal or stealing symbol in the absence of distortion or other system imperfections) are firstly removed from the received pilot symbols by multiplying each received symbol by the conjugate of its reference value:

$$\dots \dots \dots^* \left(\dots \dots \dots 2\pi j f_c dt \dots 2\pi j f_0 t \dots \dots \right)^* \dots$$

$$q_i = p_i r_i = (\alpha_i r_i \cdot e^{-j\psi_i} \cdot e^{-j\psi_i} + n_i) \cdot r_i^*$$

$$= \alpha_i \cdot e^{2\pi f_c dt} \cdot e^{2\pi f_0 t} + n_i r_i^*$$

Here, the reference pilot amplitudes have been assigned a value of unity. q_i are referred to as the pilot symbol referenced received symbols.

[0114] Next, pairs of pilot symbols (with their reference symbol phases removed), taken from the same multi-carrier symbol are combined:

$$d_j = q_i q_{i-1}^* = (\alpha_i \cdot e^{2\pi f_c dt} \cdot e^{2\pi f_0 t} + n_i r_i^*) (\alpha_{i-1} \cdot e^{-2\pi f_c dt} \cdot e^{-2\pi f_0 t} + n_{i-1} r_{i-1}^*)^*$$

$$= \left(\alpha_i \alpha_{i-1}^* \cdot e^{4\pi f_c dt} + \text{noise terms} \right) \quad (2)$$

for $i = 2, 4, \dots, n_p$, where n_p is the number of pilot symbols and $j = i/2$. d_j are referred to as the pair-wise first order differences of the pilot symbol referenced symbols.

[0115] In polar notation (R, θ), equation (2) can be re-written as:

$$d_j = q_i q_{i-1}^* = (A_i A_{i-1}, 4\pi f_c dt + \psi_i - \psi_{i-1}) + \text{noise terms}$$

The use of the conjugate

$$q_{i-1}^*$$

eliminates the frequency offset term between the transmitter and the receiver. The remaining exponential term is the same for all pairs of pilot symbols combined in this way, for a burst with a time offset dt . d_j is a complex vector of variable size, which depends on the channel attenuation, and has a phase related to the timing error plus the channel attenuation phase difference at time t .

[0116] The noise terms are:

$$n_i r_i^* \alpha_{i-1}^* \cdot e^{2\pi f_c dt} \cdot e^{-2\pi f_0 t} + n_{i-1}^* r_{i-1} \alpha_i \cdot e^{2\pi f_c dt} \cdot e^{2\pi f_0 t} + n_i r_i^* n_{i-1}^* r_{i-1} \quad (3)$$

[0117] In the present embodiment, a decision variable is defined as :

$$V_p = \sum_{j=1}^{n_p/2} d_j$$

The decision variable V_p is the integral of the pair-wise first order differences of the pilot symbol referenced received symbols.

[0118] In this summation the noise terms will tend to integrate towards zero, whereas the

$$\alpha_i \alpha_{i-1}^* \cdot e^{4\pi f_c dt}$$

term will integrate along the direction of the timing offset phase term. This holds true provided that the differences in the channel attenuation phase between pilots i and $i-1$ are small, which will generally be the case for pilots in the same multi-carrier symbol (i.e. multi-carrier symbols 1-31 defined in Figure 4), and where there is a small difference in frequency between the subcarriers containing the pilot symbols.

[0119] In the present embodiment, next, a similar analysis as above is carried out with respect

to the stealing pilot symbol sequence (set) $s_i = (S_i, \phi_i)$. The analysis is similar, except that the stealing symbol phases are removed in the first step:

$$q_i = p_i s_i^* = (\alpha_i r_i \cdot e^{2\pi i f_c dt} \cdot e^{2\pi i f_0 t} + n_i) \cdot s_i^* \\ = \alpha_i r_i s_i^* \cdot e^{2\pi i f_c dt} \cdot e^{2\pi i f_0 t} + n_i s_i^*$$

or in polar (R, θ) notation:

$$q_i = (A_i R_i S_i, \psi_i + \theta_i - \phi_i + 2\pi i f_c dt + 2\pi i f_0 dt) + (N_i S_i, \Xi_i - \phi_i) .$$

q_i are referred to as the stealing pilot symbol referenced received symbols.

[0120] Pairs of pilot symbols are then combined:

$$e_j = q_i q_{i-1}^* = (A_i A_{i-1}, \theta_i - \phi_i - \theta_{i-1} + \phi_{i-1} + \psi_i - \psi_{i-1} + 4\pi i f_c dt) + \text{noise terms} \quad (4)$$

for $i = 2, 4, \dots, n_p$, and $j = i/2$,

where a value of unity is assigned to the reference symbol amplitudes. The noise terms in this case are:

$$(A_i R_i S_i N_{i-1} S_{i-1}, \psi_i + \theta_i - \phi_i + 2\pi i f_c dt + 2\pi i f_0 dt - \Xi_{i-1} + \phi_{i-1}) \\ + (A_{i-1} R_{i-1} S_{i-1} N_i S_i, -\psi_{i-1} - \theta_{i-1} + \phi_{i-1} - 2\pi i f_c dt - 2\pi i f_0 dt + \Xi_i - \phi_i) \\ + (N_i N_{i-1} S_i S_{i-1}, \Xi_i - \phi_i - \Xi_{i-1} + \phi_{i-1})$$

or with unity reference amplitudes:

$$(A_i N_{i-1}, \psi_i + \theta_i - \phi_i + 2\pi i f_c dt + 2\pi i f_0 dt - \Xi_{i-1} + \phi_{i-1}) \\ + (A_{i-1} N_i, -\psi_{i-1} - \theta_{i-1} + \phi_{i-1} - 2\pi i f_c dt - 2\pi i f_0 dt + \Xi_i - \phi_i) \\ + (N_i N_{i-1}, \Xi_i - \phi_i - \Xi_{i-1} + \phi_{i-1})$$

This should again integrate towards zero in the decision variable summation.

[0121] As will be appreciated, equation (4) is essentially the same as equation (2) with the exception that there are additional phase difference terms $\theta_i - \phi_i$ present. These are the phase differences between the stealing pilot symbols and the normal pilot symbols. The e_j are a set of complex vectors, each with a component along the direction of the timing offset, and a component along the direction of the composite phase difference between pilot and stealing pilot symbols (i.e. the $\theta_i - \phi_i - \theta_{i-1} + \phi_{i-1}$ term, ignoring the channel phase difference for pilot symbols in the same multi-carrier symbol).

[0122] In the present embodiment, the decision variable formed from the correlation of the received pilot symbols with the stealing pilot reference symbols is defined as:

$$U_p = \sum_{j=1}^{n_p/2} e_j .$$

The decision variable U_p is the integral of the pair-wise first order differences of the stealing pilot symbol referenced received symbols.

[0123] In the present embodiment, this value is arranged to be zero (or at least minimised) by an appropriate choice of the stealing pilot symbol phases. These are accordingly chosen such that:

$$\sum_{j=1}^{n_p/2} \varepsilon_j = 0, 2\pi, 4\pi, \text{ etc.}, \quad (5)$$

where ε_j is the composite phase $\theta_j - \varphi_j - \theta_{j-1} + \varphi_{j-1}$ for the j th pair of symbols. That is, the integral of the pair-wise first order difference of the stealing pilot symbol referenced received symbols is minimised.

[0124] In the present embodiment, the magnitude of the decision variable is used by the receiver to determine which of the two symbol sequences (sets) (normal or stealing) was transmitted, i.e. if $|V_p| \geq |U_p|$, the receiver determines that the normal pilot symbol sequence (set) was transmitted, otherwise it determines the stealing sequence (set) was transmitted.

[0125] As an example, consider a sequence (set) of 10 pilot symbols in a NB-PMR burst:
 $-0.763925 - 0.645272i$, $0.997449 + 0.071229i$, $-0.995496 + 0.094606i$, $-0.240542 + 0.970593i$,
 $0.996472 - 0.083924i$, $0.654915 + 0.755685i$, $-0.458868 - 0.888500i$, $-0.315861 - 0.948803i$,
 $0.645272 - 0.763925i$, $0.071229 - 0.997449i$

[0126] In order to satisfy equation (5), a phase for pilot pair j is set as :

$$a = 2\pi j/5$$

for $j = 1, 2, \dots, 5$. Then φ is chosen such that $\theta_j - \varphi_j = a/2$ and $\theta_{j-1} - \varphi_{j-1} = -a/2$. The composite phase for this and each pair is a and the sum in equation (5) is:

$$\frac{2\pi}{5} \cdot \sum_{j=1}^5 j = 6\pi$$

[0127] The stealing pilot sequence (set) is then :

$-0.238744 - 0.971045i$, $-0.848808 + 0.528652i$, $0.217646 - 0.975993i$, $-0.997464 + 0.071178i$,
 $-0.996457 + 0.083923i$, $0.654905 + 0.755674i$, $-0.703205 + 0.710960i$, $-0.999957 + 0.007205i$,
 $0.073011 - 0.997295i$, $-0.643901 + 0.765074i$

[0128] As will be appreciated, an alternative sequence (set) of stealing pilots may be chosen using:

$$a = 2 \cdot \frac{2\pi j}{5}, \text{ etc. ,}$$

or in general, where there are n_p pairs of pilot symbols, the value a may be chosen as:

$$a = k \cdot 2\pi/n_p$$

where k is an integer.

[0129] It should be noted that other phase values satisfying equation (5) can be chosen, since it is not necessary to constrain the composite phase, ε_j , for each pair, as long as the sum of the composite phases satisfies equation (5).

[0130] Hence systems that deploy the pilot symbol design of the present invention may use normal, stealing and additional alternate symbol sequences (sets) to convey additional small amounts of information between the transmitter and the receiver, without additional payload capacity being used up.

[0131] The embodiment above is described in relation to the NB-PMR. However the methods of the present invention are applicable to other radio Standards and systems. In these embodiment, the basic receiver steps remain the same. Firstly, the reference symbol phases are removed for each of the possible symbol sequences (sets). Secondly, timing and/or frequency offsets are reduced by combining symbol pairs. The sets of resulting quantities are then integrated to form a decision variable. Finally, a comparison is made between the values of the resulting decision variables to determine which symbol sequence (set) was transmitted.

[0132] As will be appreciated, in other multi- or single-carrier systems, the second step may require different mathematical steps. For example, in embodiments using a single carrier scheme, it is not possible to combine pairs of symbols taken from the same multi-carrier symbol, as there is only one symbol present at any given time. In these embodiments, neighbouring pairs of symbols along the time axis are instead used, and the resulting phase term contains a phase related to frequency offset rather than timing offset.

[0133] In these embodiments, the subcarrier frequency offset f_c is effectively zero, so that the terms including f_c disappear. As will be appreciated, the frequency offset term does not then cancel out when the difference operation between pairs of symbols along the time axis is performed, but instead becomes the phase angle of the pilot referenced difference vectors. So the difference operation is in effect used to stabilize the phase of the complex variable used in the decision variable integration (i.e. the d_j) and could be applied to symbol pairs along the time axis and/or frequency axis, depending on the modulation.

[0134] In embodiments using a TEDS-like multi-carrier scheme in which pilot symbols are provided on several different subcarriers, elimination of the frequency offset terms (using equation 2) is preferably done using pairs of symbols as closely spaced in frequency as possible, in order to minimise the channel attenuation phase difference between the symbols of the pair.

[0135] More formally, the above detection steps can be described as the construction by the receiver of the decision variables:

$$V_j = \int \Delta_m(\mathbf{p} \circ \mathbf{r}_j^*) .$$

Where V_j is the decision variable derived at the receiver for each possible symbol set j ; $\mathbf{p}$ is the received symbol sequence (set) ;

$$\mathbf{r}_j^*$$

is the conjugate of the j th reference symbol sequence (set); $\circ$ is the Hadamard (element-wise) product operator; $\Delta_m(\)$ is a difference operator, typically operating on pairs of values from the Hadamard product (m signifies that the operator may be specific to a particular multi-carrier modulation system); and the integral, $\int$, is over the values from the symbol sequence (set). The symbol sequences (sets) $\mathbf{r}_j$ are designed such that V_j is a maximum when the transmitted sequence (set) is matched to the corresponding sequence (set) at the receiver and integrates to zero for the other sequences (sets), i.e., if

$$V_i = \int \Delta_m(\mathbf{r}_i \circ \mathbf{r}_j^*) ,$$

then:

$$V_{i-j} = MAX, \text{ and } V_{i+j} = 0.$$

[0136] The preferred embodiment as described above effectively conveys a single bit of information (i.e. not stealing/stealing). However, the methods of the present invention can be extended to alternatively and/or additionally convey (indicate) other additional information, as desired.

[0137] In an embodiment, third and/or fourth and/or fifth (and so on) symbol sequences (sets) are defined and used to convey additional information. The additional information may comprise any such desired information, for example (and in embodiments), the additional information may indicate different modulation levels or types (e.g., 4-QAM, 64-QAM, etc.) for the payload in the slot, different types of coding for the slot contents, different transmission bandwidths, and so on.

[0138] In this latter embodiment, a multi-carrier NB-PMR system that normally operates using (e.g.) 12.5 kHz channels and that has the capability to dynamically alter its transmissions to use (e.g.) 25 kHz channels, may indicate such a change to a receiver using the methods of the present invention (e.g. to indicate that additional subcarriers are present in a transmission).

[0139] In other embodiments, a dual/multi-purpose equalisation symbol sequence (set) is defined and used as a stealing flag, or for any other purposes as desired, e.g. as a protocol 'escape' flag.

[0140] The preferred embodiment described above refers to using sets (sequences) of pilot symbols, and to using all of the pilot symbols available in a burst. However, in other embodiments, any pre-defined subset of pilot symbols may be used. Equally, combined sequences of pilot symbols and synchronisation symbols may be used. As will be appreciated, in these embodiments, the transmitter and receiver should both have knowledge of how the sequence of pilot symbols or pilot and synchronisation symbols is defined.

[0141] The preferred method of the present invention was simulated using a model of the NB-PMR communication system, and the results are presented in Figure 7. Figure 7 shows the (normalised) decision variables V_p and U_p calculated for a static channel having 10dB signal-to-noise ratio, a frequency offset $f_0 = 100$ Hz, and a timing offset $dt = \sim 58 \mu s$. The values shown in first few frames (0 to 5) are due to simulation start-up delay and can therefore be ignored. As can be seen, $V_p > U_p$, and the correct pilot sequence is therefore detected.

[0142] It can be seen from the above, that the present invention, in its preferred embodiments at least, can provide an efficient narrower band mobile communications system. This is achieved in the preferred embodiments of the present invention at least by providing a digital mobile radio communications system that uses particular sequences of pilot symbols or pilot

and synchronisation symbols to indicate to the receiver the presence of stealing.

REFERENCES CITED IN THE DESCRIPTION

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Patentkrav

1. Fremgangsmåde til at drive et digitalt mobilradiokommunikationssystem hvor pilotsymboler anvendes til kanaludligning, og hvor synkroniseringssymboler

5 anvendes til at indikere burst timings, hvilken fremgangsmåde omfatter:

at anvende (44) en stealing-sekvens af pilotsymboler eller en stealing-sekvens af pilotsymboler og synkroniseringssymboler i en transmission for at indikere tilstedeværelsen af stealing i transmissionen til en transceiver; hvor pilotsymbolerne er placeret med intervaller langs tids- og

10 frekvensakserne af transmissionen;

hvilken fremgangsmåde yderligere omfatter:

at inkludere (42) en normal sekvens af pilotsymboler eller en normal sekvens af pilotsymboler og synkroniseringssymboler i en transmission, når stealing ikke er til stede; og

15 når stealing er til stede, at inkludere (44) stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniseringssymboler i transmissionen i stedet for den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringssymboler;

20 hvor den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringssymboler er forskellig fra stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniseringssymboler; og

25 hvor symbolfaserne af stealing-pilotsymbolerne eller stealing-pilot- og synkroniseringssymbolerne er indrettet således, at en af:

integralet eller summen af kombinerede par af symbolerne af stealing-symbolsekvensen efter multiplicering af hvert stealing-sekvenssymbol med konjugatet af dets

30 referencesymbol i en referencesekvens af normale symboler, og integralet eller summen af kombinerede par af symbolerne af stealing-symbolsekvensen efter multiplicering af hvert stealing-sekvenssymbol med konjugatet af dets

35 referencesymbol i en reference-stealingsymbolsekvens, vil være et minimum, og det andet integral eller den anden sum vil være et maksimum,

5 således at stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniseringssymboler kan skelnes fra den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringssymboler forud for udligning.

2. Fremgangsmåde ifølge krav 1, hvor symbolfaserne af de normale pilotsymboler eller de normale pilot- og synkroniseringssymboler er indrettet således, at en af: integralet eller summen af kombinerede par af symbolerne af den normale
10 symbolsekvens efter multiplicering af hvert normalt sekvenssymbol med konjugatet af dets referencesymbol i en referencesekvens af normale symboler, og integralet eller summen af kombinerede par af symbolerne af den normale symbolsekvens efter multiplicering af hvert normalt sekvenssymbol med konjugatet af dets referencesymbol i en reference- stealing-symbolsekvens, vil
15 være et maksimum, og det andet integral eller den anden sum vil være et minimum.

3. Fremgangsmåde til at drive et digitalt mobilradiokommunikationssystem, hvor pilotsymboler anvendes til kanaludligning, og hvor synkroniseringssymboler
20 anvendes til at indikere burst timings, hvilken fremgangsmåde omfatter:
at bestemme (50) om en modtaget transmission inkluderer en stealing-sekvens af pilotsymboler eller en stealing-sekvens af pilotsymboler og synkroniseringssymboler, som indikerer tilstedeværelsen af stealing i transmissionen, hvor pilotsymbolerne er placeret med intervaller langs tids- og frekvensakserne af transmissionen; og
25 hvis det bestemmes, at en stealing-sekvens af pilotsymboler eller en stealing-sekvens af pilotsymboler og synkroniseringssymboler, som indikerer tilstedeværelsen af stealing, er til stede i den modtagne transmission, at identificere tilstedeværelsen af stealing i transmissionen og at afkode transmissionen i overensstemmelse hermed;
30 hvilken fremgangsmåde yderligere omfatter:
at bestemme om en modtaget transmission inkluderer en normal sekvens af pilotsymboler eller en normal sekvens af pilotsymboler og synkroniseringssymboler; og

hvis det bestemmes, at en normal sekvens af pilotsymboler eller en normal sekvens af pilotsymboler og synkroniseringssymboler er til stede i den modtagne transmission, at afkode transmissionen i overensstemmelse hermed; hvor fremgangsmåden yderligere omfatter:

- 5 at bestemme hvilke af stealing- eller de normale sekvenserne er til stede i en modtaget transmission, før kanaludligning og/eller -synkronisering udføres;
hvor trinnet at bestemme, om en modtaget transmission indikerer tilstedeværelsen af stealing, omfatter:
- 10 at fjerne referencesymbolfaser fra de modtagne pilotsymboler eller pilot- og synkroniseringssymboler ved multiplicering af hvert symbol med konjugatet af dets referencesymbol i en referencesymbolsekvens for hver mulig transmitteret symbolsekvens;
- 15 at kombinere symbolpar for hver mulig transmitteret symbolsekvens; og derefter
at integrere eller summere sættene af resulterende kvantiteter og at bestemme, om integralet eller summen er et minimum eller et maksimum for at bestemme, om den
- 20 modtagne transmission inkluderer en sekvens af pilotsymboler eller en sekvens af pilotsymboler og synkroniseringssymboler, som indikerer tilstedeværelsen af stealing, eller ej.
- 25 **4.** Kommunikationsstation til et digitalt mobilradiokommunikationssystem, hvor pilotsymboler anvendes til kanaludligning, og hvor synkroniseringssymboler anvendes til at indikere burst timings, hvilken kommunikationsstation omfatter:
- en transmitter konfigureret til at inkludere (44) i en transmission en stealing-sekvens af pilotsymboler eller en stealing-sekvens af pilotsymboler
- 30 og synkroniseringssymboler for at indikere tilstedeværelsen af stealing i transmissionen til en receiver;
hvor pilotsymbolerne er placeret med intervaller langs tids- og frekvensakserne af transmissionen;
hvor transmitteren er konfigureret til:

- at inkludere (42) en normal sekvens af pilotsymboler eller en normal sekvens af pilotsymboler og synkroniseringssymboler i en transmission, når stealing ikke er til stede; og
- når stealing er til stede, at inkludere (44) stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniserings-symboler i transmissionen i stedet for den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringssymboler;
- hvor den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringssymboler er forskellig fra stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniseringssymboler; og
- hvor symbolfaserne af stealing-pilotsymbolerne eller stealing-pilot- og synkroniseringssymbolerne er indrettet således, at en af:
- integralet eller summen af kombinerede par af symbolerne af stealing- symbolsekvensen efter multiplicering af hvert stealing-sekvenssymbol med konjugatet af dets referencesymbol i en referencesekvens af normale symboler, og integralet eller summen af kombinerede par af symbolerne af stealing- symbolsekvensen efter multiplicering af hvert stealing-sekvenssymbol med konjugatet af dets referencesymbol i en reference-stealingsymbolsekvens, vil være et minimum, og det andet integral eller den anden sum vil være et maksimum
- således at stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniseringssymboler kan skelnes fra den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringssymboler forud for udligning.
- 5.** Digitalt mobilradiokommunikationssystem hvor pilotsymboler anvendes til kanaludligning, og hvor synkroniseringssymboler anvendes til at indikere burst timings, hvilket system omfatter:
- en transmitter konfigureret til at inkludere (44) i en transmission en stealing-sekvens af pilotsymboler eller en stealing-sekvens af pilotsymboler

og synkroniseringssymboler til at indikere tilstedeværelsen af stealing i transmissionen til en receiver;

hvor pilotsymbolerne er placeret med intervaller langs tids- og frekvensakserne af transmission;

5 hvor transmitteren er konfigureret til:

at inkludere (42) en normal sekvens af pilotsymboler eller en normal sekvens af pilotsymboler og synkroniseringssymboler i en transmission, når stealing ikke er til stede; og

10 når stealing er til stede, at inkludere (44) stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniseringsymboler i transmissionen i stedet for den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringsymboler;

15 hvor den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringssymboler er forskellig fra stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniseringssymboler; og

hvor symbolfaserne af stealing-pilotsymbolerne eller stealing-pilot- og synkroniseringssymbolerne er indrettet således, at en af:

20 integralet eller summen af kombinerede par af symbolerne af stealing-symbolsekvensen efter multiplicering af hvert stealing-sekvenssymbol med konjugatet af dets referencesymbol i en referencesekvens af normale symboler, og integralet eller summen af kombinerede par af symbolerne af stealing- symbolsekvensen

25 efter multiplicering af hvert stealing-sekvenssymbol med konjugatet af dets referencesymbol i en reference-stealingsymbolsekvens, vil være et minimum, og det andet integral eller den anden sum vil være et maksimum,

således at stealing-sekvensen af pilotsymboler eller stealing-sekvensen af pilotsymboler og synkroniseringssymboler kan skelnes

30 fra den normale sekvens af pilotsymboler eller den normale sekvens af pilotsymboler og synkroniseringssymboler forud for udligning.

6. Station eller system ifølge krav 4 eller 5, hvor symbolfaserne af de normale pilotsymboler eller de normale pilot- og synkroniseringssymboler er indrettet

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således, at en af: integralet eller summen af kombinerede par af symbolerne af den normale symbolsekvens efter multiplicering af hvert normalt sekvenssymbol med konjugatet af dets referencesymbol i en referencesekvens af normale symboler, og integralet eller summen af kombinerede par af symbolerne af den normale symbolsekvens efter multiplicering af hvert normalt sekvenssymbol med konjugatet af dets referencesymbol i en -reference- stealing-symbolsekvens, vil være et maksimum, og det andet integral eller den anden sum vil være et minimum.

- 10 **7.** Kommunikationsstation til et digitalt mobilradiokommunikationssystem, hvor pilotsymboler anvendes til kanaludligning, og hvor synkroniseringssymboler anvendes til at indikere burst timings, hvilken kommunikationsstation omfatter:
- 15 en receiver konfigureret til at bestemme (50) om en modtaget transmission inkluderer en stealing-sekvens af pilotsymboler eller en stealing-sekvens af pilotsymboler og synkroniseringssymboler, som indikerer tilstedeværelsen af stealing i transmissionen, hvor pilotsymbolerne er placeret med intervaller langs tids- og frekvensakserne af transmissionen, og til, hvis det bestemmes, at en stealing-sekvens af pilotsymboler eller en stealing-sekvens af pilotsymboler og synkroniseringssymboler, som indikerer tilstedeværelsen af stealing, er til stede i den modtagne transmission, at identificere tilstedeværelsen af stealing i transmissionen og at afkode transmissionen i overensstemmelse hermed;

20 hvor receiveren er konfigureret til:

 - 25 at bestemme om en modtaget transmission inkluderer en normal sekvens af pilotsymboler eller en normal sekvens af pilotsymboler og synkroniseringssymboler; og

hvis det bestemmes, at en normal sekvens af pilotsymboler eller en normal sekvens af pilotsymboler og synkroniseringssymboler er til stede i den modtagne transmission, at afkode transmissionen i overensstemmelse hermed;

30 hvor receiveren er konfigureret til at bestemme, hvilke stealing- eller de normale sekvenser er til stede i en modtaget transmission, før kanaludligning og/eller synkronisering udføres;

hvor receiveren er konfigureret til at bestemme, om en modtaget transmission indikerer tilstedeværelsen af stealing ved:

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5 at fjerne referencesymbolfaser fra de modtagne pilotsymboler
eller pilot- og synkroniseringssymboler ved multiplicering af
hvert med konjugatet af dets referencesymbol i en
referencesymbolsekvens for hver mulig transmitteret
10 symbolsekvens;
at kombinere symbolpar for hver mulig transmitteret
symbolsekvens; og derefter
at integrere eller summere sættene af resulterende
kvantiteter og at bestemme, om integralet eller summen er et
15 minimum eller et maksimum for at bestemme, om den
modtagne transmission inkluderer en sekvens af
pilotsymboler eller en sekvens af pilotsymboler og
synkroniseringssymboler, som indikerer tilstedeværelsen af
stealing, eller ej.

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8. Digitalt mobilradiokommunikationssystem hvor pilotsymboler anvendes til
kanaludligning, og hvor synkroniseringssymboler anvendes til at indikere burst
timings, hvilket system omfatter:

20 en receiver konfigureret til at bestemme (50) om en modtaget transmission
inkluderer en stealing-sekvens af pilotsymboler eller stealing-sekvens af
pilotsymboler og synkroniseringssymboler, som indikerer tilstedeværelsen
af stealing i transmissionen, hvor pilotsymbolerne er placeret med
intervaller langs tids- og frekvensakserne af transmissionen, og til, hvis det
bestemmes, at en stealing-sekvens af pilotsymboler eller en stealing-
25 sekvens af pilotsymboler og synkroniseringssymboler, som indikerer
tilstedeværelsen af stealing, er til stede i den modtagne transmission, at
identificere tilstedeværelsen af stealing i transmissionen og afkode
transmissionen i overensstemmelse hermed;
hvor receiveren er konfigureret til:

30 at bestemme om en modtaget transmission inkluderer en normal
sekvens af pilotsymboler eller en normal sekvens af pilotsymboler og
synkroniseringssymboler; og
hvis det bestemmes, at en normal sekvens af pilotsymboler eller en
normal sekvens af pilotsymboler og synkroniseringssymboler er til

- stede i den modtagne transmission, at afkode transmissionen i overensstemmelse hermed;
- hvor receiveren er konfigureret til at bestemme, hvilke af stealing- eller de normale sekvenser er til stede i en modtaget transmission,
- 5 før kanaludligning og/eller synkronisering udføres;
- hvor receiveren er konfigureret til at bestemme, om en modtaget transmission indikerer tilstedeværelsen af stealing ved:
- at fjerne referencesymbolfaser fra de modtagne pilotsymboler eller pilot- og synkroniseringssymboler ved multiplicering af
- 10 hvert symbol med konjugatet af dets referencesymbol i en referencesymbolsekvens for hver mulig transmitteret symbolsekvens;
- at kombinere symbolpar for hver mulig transmitteret symbolsekvens; og derefter
- 15 at integrere eller summere sættene af resulterende kvantiteter og at bestemme, om integralet eller summen er et minimum eller et maksimum for at bestemme, om den modtagne transmission inkluderer en sekvens af
- pilotsymboler eller en sekvens af pilotsymboler og
- 20 synkroniseringssymboler, som indikerer tilstedeværelsen af stealing, eller ej.

9. Computerprogramelement omfattende computersoftwarekodedele som, når programelementet køres på databehandlingsorgan, får databehandlingsorganet til

25 at udføre fremgangsmåden ifølge et hvilket som helst af kravene 1 til 3.

DRAWINGS

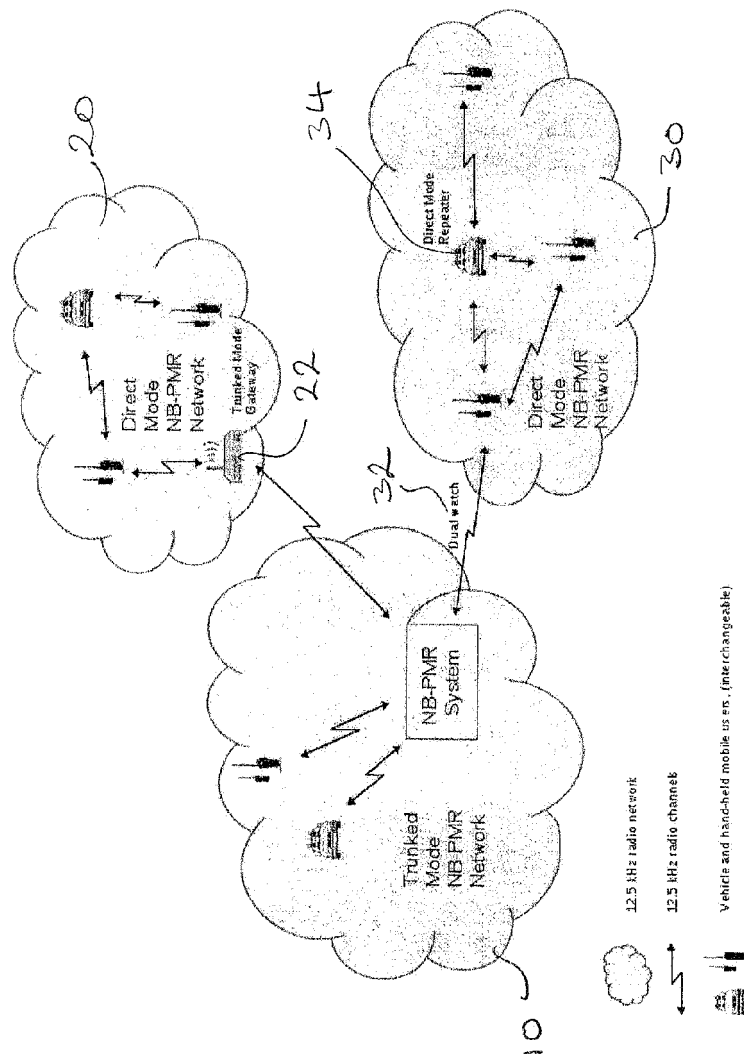


Fig. 1

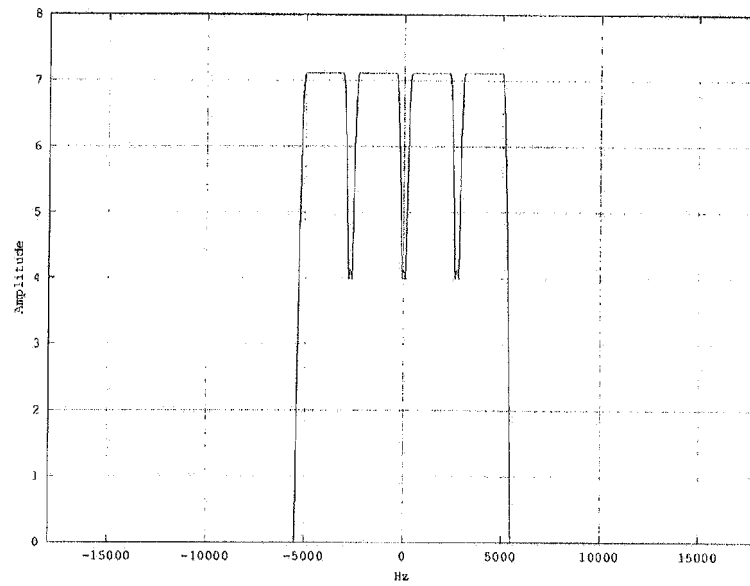


Fig. 2

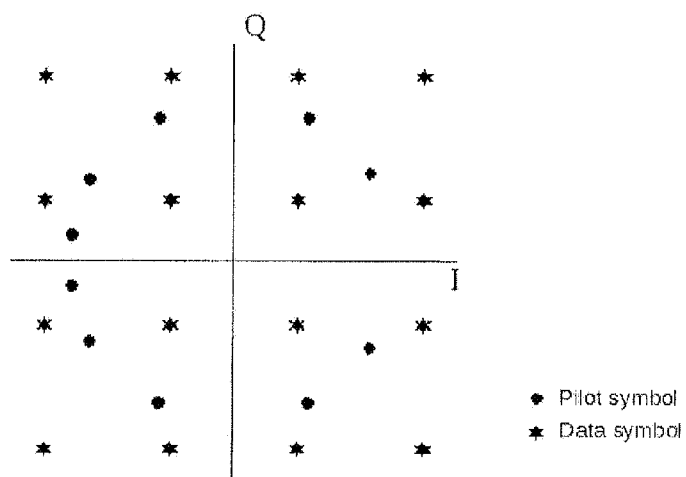


Fig. 3

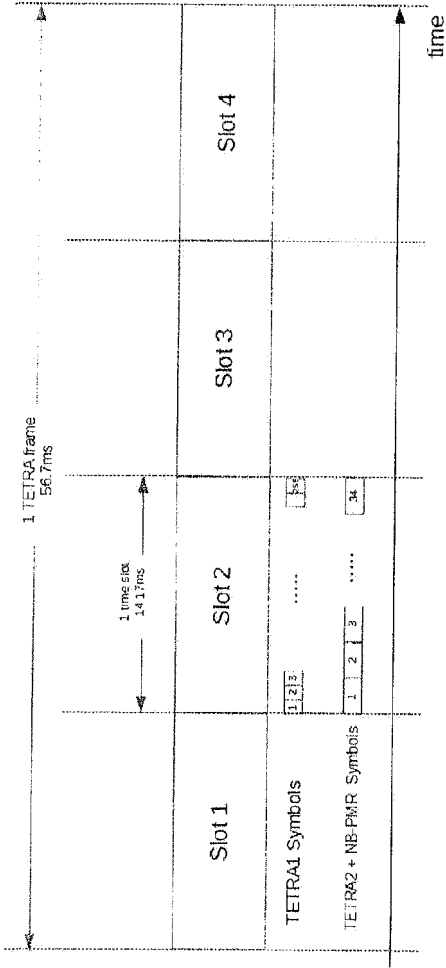


Fig. 4

0	S1	S5	D3	D7	D11	D15	D19	P1	D25	D29	D33	D37	D41	P3	D47	D51	D55	D59	D63	P5	D69	D73	D77	D81	D85	P7	D91	D95	D99	D103	P9
1	S2	D1	D4	D8	D12	D16	D20	D23	D26	D30	D34	D38	D42	D45	D48	D52	D56	D60	D64	D67	D70	D74	D78	D82	D86	D89	D92	D96	D100	D104	D107
2	S3	D2	D5	D9	D13	D17	D21	D24	D27	D31	D35	D39	D43	D46	D49	D53	D57	D61	D65	D68	D71	D75	D79	D83	D87	D90	D93	D97	D101	D105	D108
3	S4	S6	D6	D10	D14	D18	D22	P2	D28	D32	D36	D40	D44	P4	D50	D54	D58	D62	D66	P6	D72	D76	D80	D84	D88	P8	D94	D98	D102	D106	P10

Fig. 5

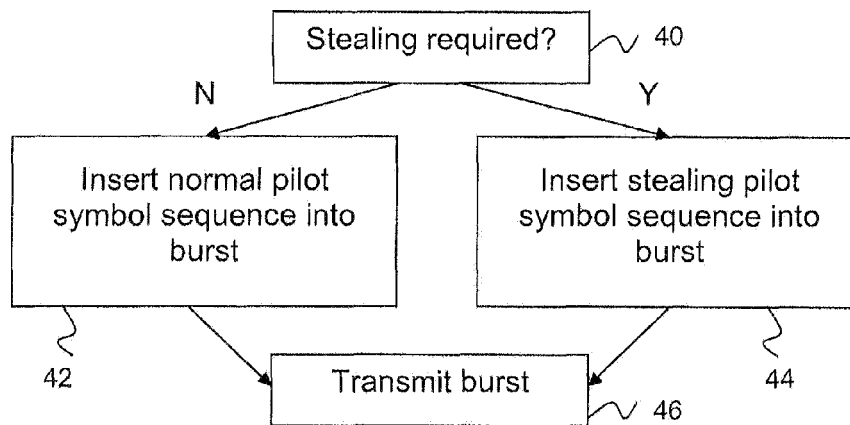


Fig. 6A

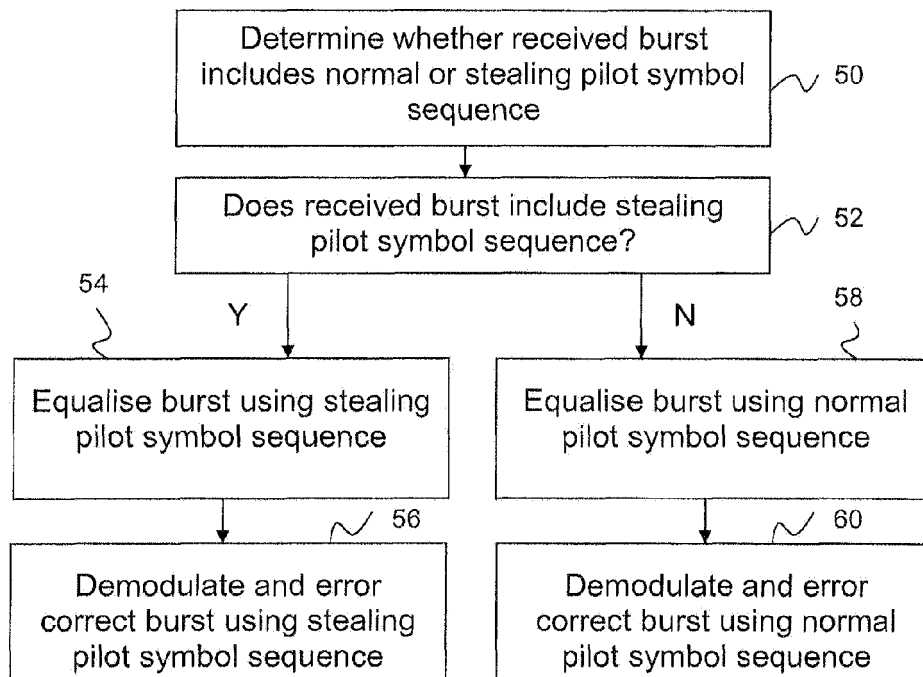


Fig. 6B

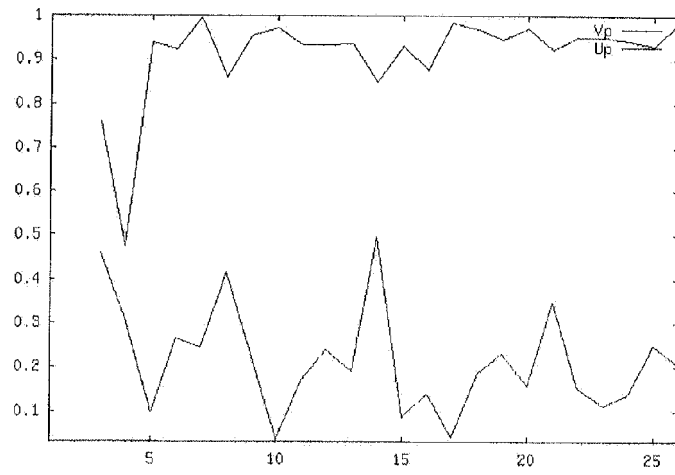


Fig. 7